

Radio Frequency Coil for Dual-Nuclei MR Muscle Energetics Investigation Based on Two Capacitively Coupled Periodic Wire Arrays

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Abstract—In this letter, we describe a dedicated radio frequency (RF) coil for magnetic resonance imaging and spectroscopy at 4.7 T optimized for investigation of muscle energetics in a human forearm. The coil operating at the Larmor frequencies of protons ^1H (200.1 MHz) and phosphorous ^{31}P (81 MHz) is based on two cylindrical periodic structures made of thin metal wires with tips interconnected through structural capacities of printed overlapping patches and sliding electric contacts. By independently exciting two orthogonal eigenmodes of the structure, the coil can be tuned and matched at both frequencies without lumped capacitors and cover the target region of interest for the forearm flexor muscles.

Index Terms—Magnetic resonance imaging, magnetic resonance spectroscopy (MRS), radio frequency (RF) coil, wire array.

I. INTRODUCTION

IN ORDER to investigate energy production in muscle tissues, ^{31}P phosphorous (^{31}P) magnetic resonance spectroscopy (MRS) has been largely used over the past decades [1]–[4]. More particularly, a large number of studies have been devoted to the investigation of changes in adenosine triphosphate, phosphocreatine (PCr), and pH, which are key players in muscle energetics [5]–[7]. The low sensitivity of ^{31}P -MRS has been largely recognized so that measurements have been mostly performed using small radio frequency (RF) surface coils [8].

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These compact magnetic loop antennas attached to investigated regions of interest (ROIs), offer an improved signal-to-noise ratio as compared to volume coils surrounding the ROI in many biomedical studies [3], [6], [9], [10]. In the context of muscle energetics studies, investigations have been performed for superficial muscles, such as fingers flexor, quadriceps femoris, and gastrocnemius muscles [6], [9]. For these applications, surface coils have been tailored to cover the ROI. In this context, ^{31}P coils have to be combined to proton coils for shimming and/or localization procedures. Dual-nuclei magnetic resonance (MR) studies are also of interest given that metabolites can be detected at both frequencies, e.g., ^1H and ^{31}P . In such case, RF coils are typically dual-tuned transceiver antennas independently operating at two Larmor frequencies of the given nuclei. Different designs for dual-tuned surface coils have been proposed so far. A single loop operating at two frequencies with independent tuning and matching circuits has been designed [11]. Alternatively, two separate loops could be combined as long as they are decoupled [12], containing a capacitive network for dual-band tuning and matching $50\ \Omega$. A limitation of conventional loop coils is the depth of the ROI, which is inversely proportional to the loop area, thus coil sensitivity gets maximal at the coil center rapidly decreasing with distance increasing from this point [8]. However, in the muscle energetics MR studies on a human arm, it is important to localize the ROI nearby the area of the flexor muscles, as described in Fig. 1(a). To be exact, the ROI for ^1H imaging should cover roughly a half of an arm cross section to be able to perform single-voxel spectroscopy and anatomical imaging with the same coil. In contrast, at ^{31}P , the ROI should coincide with the investigated muscles' cross section. To remove this limitation, it has been recently proposed to employ hybridized eigenmodes [13], [14] excited in periodic arrays of thin TEM resonators [15], resonant wires or strips [16], [17] and in the open "bird-cage" structure [18].

In this letter, we report on the design, numerical simulation, and experimental demonstration of a genuine dual-tuned coil based on two capacitively connected periodic wire arrays. The coil provides the unique combination of a distributed structural capacity together with the advantage of independent geometrical tuning at two Larmor frequencies 81.0 and 200.1 MHz ($^1\text{H}/^{31}\text{P}$ nuclei at 4.7 T) covering the target ROIs for muscle energetics studies.

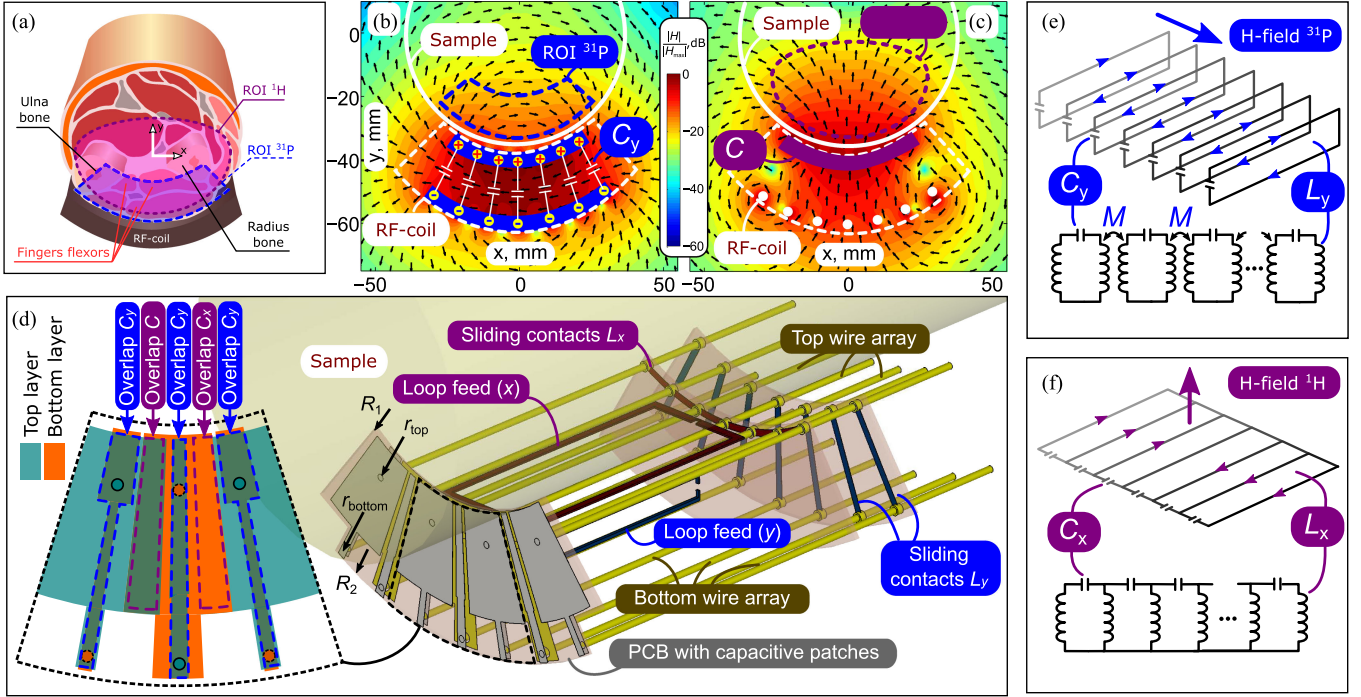


Fig. 1. Application and practical realization of proposed dual-nuclei coil. (a) Target ROIs at ^{31}P (81 MHz) and ^1H (200.1 MHz) covering flexor muscles of a human arm; capacitive interconnection of wire ends and corresponding orthogonal-polarized eigenmodes of dual-stage wire array. (b) Volume mode for ^{31}P . (c) Surface mode at ^1H . (d) Practical realization of the proposed coil (inset on the left demonstrates the layout of printed-circuit board (PCB) with capacitive patches for interconnection of wires). (e) Wire currents of volume mode and its equivalent circuit. (f) Wire currents of surface mode and its equivalent circuit.

II. RF COIL DESIGN AND CHARACTERIZATION

The aim of the proposed coil is to fit to the ROI corresponding to the location of flexor muscles in a human arm, as shown in Fig. 1(a). The surface area to be covered is around 60 mm along its bottom semicircumference and 100 mm along the arm axis z . The arm is placed in the scanner along B_0 vector and together with the coil must fit to the space available in a narrow preclinical bore of a 4.7-T MR system (30 cm diameter). To meet the requirements, the proposed coil operates combining the principles of the surface wire resonator, i.e., a single periodic array of resonant coupled wires [16], [19] and the volume wire resonator composed of two separate periodic wire arrays with connected ends in the same structure [20]. In such resonators, a number of hybridized eigenmodes can be excited [13]. These eigenmodes differ by their near-field patterns (in particular, by the field confinement nearby the wires) and resonant frequencies. In the proposed coil, two arrays of $N = 7$ wires oriented in the z -direction of the length 150 mm each, as shown in Fig. 1(d), the top and the bottom ones were both placed on coaxial cylindrical surfaces with a constant angular periodicity $\alpha = 10^\circ$ to repeat the shape of a human arm. To make the structure resonant at 81 and 200.1 MHz, the tips of all wires on one side of the structure were connected by structural capacities [21]. On the other side of the structure, the wires were connected by sliding short-circuiting contacts, as shown in Fig. 1(d). Corresponding wires from the top and bottom arrays are connected vertically through structural capacities C_y , as schematically shown in Fig. 1(b) in addition to the figure of short-circuiting sliding contacts. Due to

these connections at the resonant frequency of the fundamental mode, the two arrays were excited out of phase confining the magnetic field in between. However, the magnetic field of this mode is also generated outside of the space between the arrays and penetrates into a sample at the distance that depends on the periodicity of the wires. In this regime, the resonator can be considered as an array of seven inductively coupled loops, each loaded to capacity C_y . This effect was used to localize the ROI at 81 MHz. The magnetic field \mathbf{H} of this volume mode in the sample has a horizontal polarization and its normalized magnitude pattern in the axial plane is shown in Fig. 1(b). The wire currents directions and the equivalent circuit corresponding to the volume mode are depicted in Fig. 1(e). The equivalent circuit of the volume mode is represented by seven inductively coupled LC circuits based on the structural fixed capacitance C_y and inductance L_y of an individual rectangular loop of current consisting of one wire from the top and one from the bottom layer as well as the short-circuit slider. In the same wire structure, another mode with a vertical polarization can be excited with the magnetic field pattern depicted in Fig. 1(c). To enable this surface mode at 200.1 MHz, each pair of neighboring wire ends in the top array was connected horizontally by capacities C_x , as shown in Fig. 1(c). Current directions on the top layer of wires and the corresponding equivalent circuit of the surface mode are shown in Fig. 1(f). The equivalent circuit in this case is represented by a chain of six high-pass LC circuits, where in addition to the fixed structural capacitance C_x in each unit cell, there is a wire inductance depending on the position of the sliding short circuit. Both the capacities C_x and C_y were fixed

and realized using overlapped patches placed on the opposite sides of a PCB with a 1.5 mm thick Rogers RT/Duroid 6010 substrate ($\epsilon_r = 10.7$ and $\tan \delta = 0.0023$). The overlap regions on the PCB are responsible for both types of capacity and are shown in the inset on the left-hand side of Fig. 1(d).

Every RF coil for biomedical research requires the possibility to adjust tuning and matching to compensate in the experiment for the sample properties' variation. In the proposed coil, fine frequency tuning of both modes is achieved by moving the horizontal and vertical sliding contacts shown in Fig. 1(d) to mechanically adjust the structural inductances L_x and L_y in the equivalent circuits of Fig. 1(e) and (f), correspondingly. The position of each slider changes the distributed inductance and controls the mode frequencies (L_x individually for the surface mode while L_y individually for the volume mode) with the tunability of 1 MHz/mm. Thanks to the wide range of distributed inductance values the coil is appropriate for biomedical applications allowing frequency tuning for the large range of sample volumes. Moreover, independent tuning for the selected modes is possible because L_x short circuits only top array wires while L_y pair the corresponding wires, which belong to different arrays. The resonant frequencies of the volume and surface fundamental modes can be estimated from the equivalent circuits shown in Fig. 1(e) and (f). The circuit estimations differ from numerical simulations by 20%–30% due to the presence of the sample and parasitic reactances in the structure. Finally, the exact positions of both sliding contacts were chosen by numerical optimization.

Both modes are excited by two mutually orthogonal feeding loops, areas of which affect the input impedance of two feeding points. The vertical loop (Y) excites the volumetric mode, whereas the horizontal loop (X) couples to the surface mode. The loops were made expandable in the z -direction thanks to telescopic sliding tubes to adjust the coil matching to 50 Ω at both frequencies. By numerical optimization in CST Studio Suite, the following structural parameters of the PCB and sliders were chosen: the radii of the top and bottom wire arrays $r_{\text{top}} = 60$ mm and $r_{\text{bottom}} = 90$ mm and the PCB internal and external radii $R_1 = 50$ mm and $R_2 = 93$ mm. The distances from the sliding contacts to the PCB plane were $L_x = 98$ mm and $L_y = 109$ mm. All wires were made of brass and had the diameter of a circular cross section of 2 mm. The coil was considered in the simulation inside a cylindrical MR-system bore of the diameter 30 cm with a homogeneous cylindrical phantom ($\epsilon_p = 34$ and $\sigma_p = 0.4$ S/m) having the radius $R_p = 50$ mm and length $L_p = 250$ mm, representing a human arm. For the aforementioned numerically optimized geometric parameters of the PCB and wires of the coil, one can estimate the inductance of a loop between one top and one bottom wire as $L_y = 188$ nH, and the structural overlap capacitance $C_y = 7.8$ pF predicting for $M = 0.95L_y$, the fundamental volume mode resonance at 95 MHz. By taking the structural patch overlap capacity $C_x = 10.1$ pF and the inductance of the top-layer wire of $L_x = 48$ nH, one estimates the resonant frequency of the surface mode equal to 260 MHz. However, the precise resonant frequencies calculated in CST equal 80.1 and 200 MHz. Indeed, the simulated S -parameters represented in Fig. 2 show that the coil at both 50- Ω ports connected to the loops (X) and (Y) with the chosen parameters is

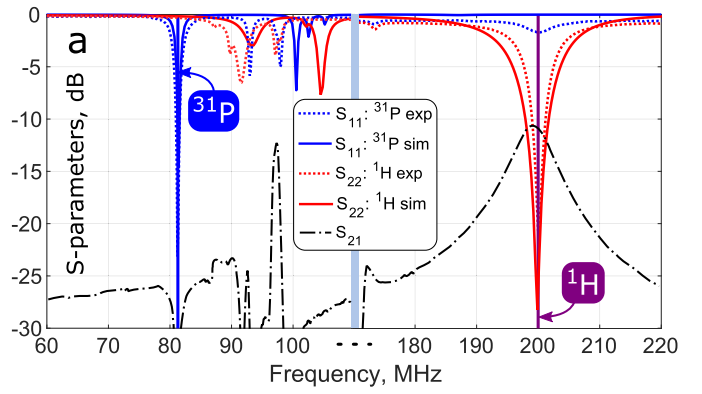


Fig. 2. Simulated and measured S_{11} (reflection coefficient at ^{31}P port) and S_{22} (reflection coefficient at ^1H port) and measured S_{21} (transmission coefficient between two ports) of the proposed coil in two frequency ranges: from 60 to 120 and from 170 to 220 MHz.

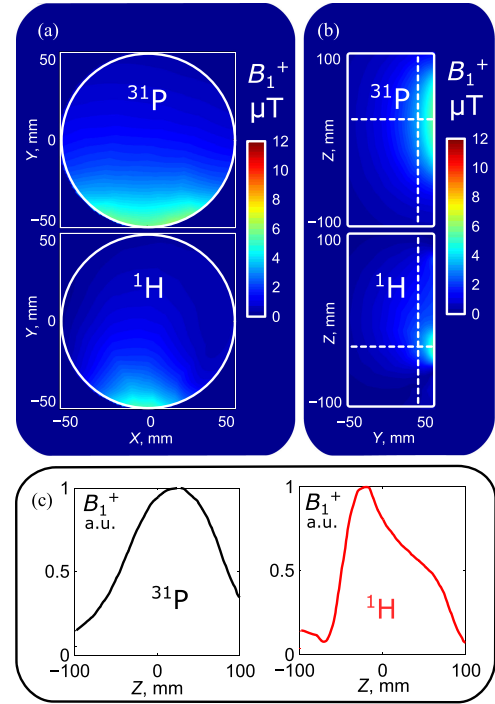


Fig. 3. Simulation results of the proposed coil at two Larmor frequencies: B_1^+ patterns for 1 W of accepted power in (a) transverse and (b) sagittal planes; (c) B_1^+ profiles for 1 W of accepted power in the z -direction at the depth 25 mm in the phantom ($y = -25$ mm).

simultaneously tuned and matched at two Larmor frequencies. The calculated coefficient S_{12} of parasitic coupling between the two ports was negligible due to orthogonality of the excited eigenmodes and, therefore, it is not presented in Fig. 2. As can be seen from Fig. 2, the relative bandwidth of matching for ^1H nucleus is much higher than for ^{31}P one.

The calculated magnitude patterns of B_1^+ field, i.e., a right-handed circularly polarized component of magnetic RF field, normalized by 1 W of accepted power, created by the coil at the Larmor frequencies, are shown in Fig. 3(a) and (b) in the transverse (XY) and sagittal (YZ) planes of the phantom

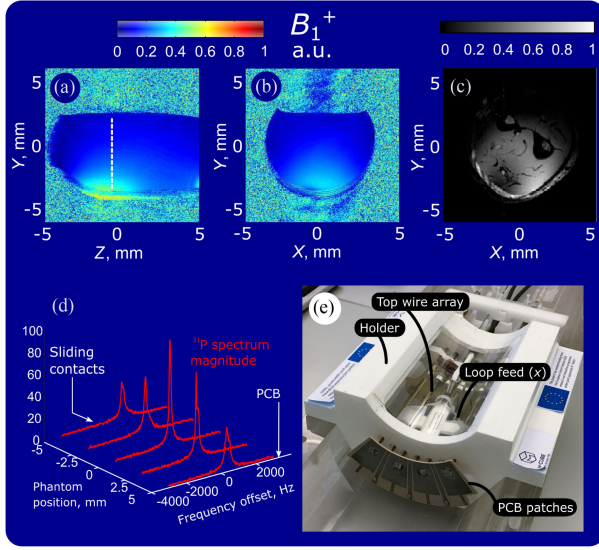


Fig. 4. Experiment: B_1^+ field maps for ^1H in (a) sagittal and (b) transverse planes. (c) *In vivo* images of a human forearm in the transverse plane. (d) ^{31}P spectra of a test-tube phantom depending on its position in the z -direction. (e) Prototype of the proposed coil.

correspondingly. Normalized profiles of B_1^+ field along the z -direction are given in Fig. 3(c) for the depth 25 mm in the phantom. The corresponding coordinate line is shown by vertical dashed lines in Fig. 3(b).

Due to the calculated transmit field patterns and profiles, one can conclude that the proposed coil has the required coverage area with the length of 100–120 mm in the z -direction and around 60 mm in the arm circumference direction. The coverage area is defined as the area where B_1^+ magnitude holds higher than 50% of the maximum of the whole pattern. In other words, the proposed coil creates relatively homogeneous field in the half of the phantom's axial cross section while producing the magnetic field in the ROI with the required shape closer to the bottom surface at ^{31}P .

To experimentally characterize the proposed coil, its prototype has been manufactured and the results of the measurements are presented in Fig. 4. The prototype shown in Fig. 4(e) was built of 14 brass wires with the capacitive PCB soldered to their ends, two sliding contacts for precise dual-band tuning and two feeding loops soldered to two coaxial cables for connecting to two frequency channels of the MR-systems' transceiver. All the parts were mounted on a 3-D printed holder with a shape conformal to a human arm. On-bench measurements of S -parameters were done for each channel with a vector network analyzer (VNA MS2036 C, Anritsu). $|S_{11}|$ and $|S_{22}|$ parameters represent the reflection coefficients from the vertical and horizontal feed loop ports, respectively, whereas $|S_{21}|$ characterizes the isolation of the two ports. In the measurements, both sliding contacts were adjusted for proper dual-band tuning. The coil was loaded by a 500 mL glass jar containing distilled water with a $0.1 \text{ mol}\cdot\text{L}^{-1}$ concentration of phosphorus nuclei from dissolved polyphosphoric acid. Comparison of the measured S -parameters to the calculated ones shown in Fig. 2 is very good and confirms the

possibility of simultaneous tuning to 81 and 200.1 MHz and matching to 50Ω at both frequencies without using lumped capacitors. Moreover, the measured $|S_{21}|$ holds below -11 dB at any frequency, showing that the two ports are sufficiently isolated due to the orthogonality of the excited modes.

To map the RF-field patterns created by the proposed coil and test its imaging and spectroscopy capabilities, a set of MR acquisitions were performed at 4.7 T on 47/30 Biospec Avance MR system (Bruker, Karlsruhe, Germany). Fast low-angle shot (FLASH) gradient echo sequences were performed in sagittal and axial orientations of the phantom with T_R/T_E of 4.3/500 ms, 5 mm slice thickness, and $12 \times 12 \text{ cm}^2$ field of view (FOV) with 0.625 mm^2 pixels. This sequence was repeated with nine different nominal flip angles ranging from 1.5° to 30° . These series of images allowed us to retrieve the relative B_1^+ amplitude, i.e., the transmit efficiency of the coil (proportional to the ratio of B_1^+ to the square root of input power in the impedance-matched regime). This method was already described in [22] and successfully implemented for B_1^+ maps retrieval [20]. The measured relative B_1^+ patterns at ^1H frequency are depicted in Fig. 4(a) for the axial and (b) for sagittal slices crossing the maximum. The measured field patterns are in good agreement with the numerical results. In order to illustrate how the increased FOV depth affects an anatomical hydrogen MR image of a human arm, an *in vivo* scan using a FLASH sequence with T_R/T_E of 500/10 ms, a voxel size of $0.54 \times 0.54 \text{ mm}^2$, slice thickness of 5 mm, and an FOV of $103.8 \times 103.8 \text{ mm}^2$ was performed. As expected, the proposed coil creates an FOV covering almost half of the arm transverse cross section, which is illustrated in Fig. 4(c). To characterize the proposed coil at ^{31}P , single-voxel spectroscopy has been made using a small 5 mL test tube filled with the same liquid, as mentioned earlier. A large coverage area provided by the coil can be clearly seen in Fig. 4(d) where obtained single-voxel spectra of a small test tube are presented depending on the position of the latter in the z -direction with respect to the coil. Indeed, for the proposed coil, the spectrum holds higher than 50% of the maximum value within a range of 100 mm in the z -direction.

III. CONCLUSION

In conclusion, a new dual-frequency RF-coil operating at 81.0 and 200.1 MHz for muscle energetics MR investigation has been designed, and the corresponding performance has been characterized. Due to the orthogonality of its excited surface and volumetric modes, independent tuning and matching of its ^{31}P and ^1H channels was shown. This possibility along with a structural capacity of printed overlap patches is a unique combination among the known dual-tuned surface coils for MRI/MRS. The proposed coil precisely covers the region of flexor muscles in a human arm at the Larmor frequency of ^{31}P due to a relatively confined RF field of the volume mode. This helps in spectroscopy methods to selectively receive RF signals from those investigated muscles. At the same time, it homogeneously covers the whole arm cross section at the Larmor frequency of protons, thus providing the possibility to obtain reference anatomical maps of a subject.

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