Robust Adaptive Gain Higher Order Sliding Mode Observer Based Control-constrained Nonlinear Model Predictive Control for Spacecraft Formation Flying

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Abstract-This work deals with the development of a decentralized optimal control algorithm, along with a robust observer, for the relative motion control of spacecraft in leader-follower based formation. An adaptive gain higher order sliding mode observer has been proposed to estimate the velocity as well as unmeasured disturbances from the noisy position measurements. A differentiator structure containing the Lipschitz constant and Lebesgue measurable control input, is utilized for obtaining the estimates. Adaptive tuning algorithms are derived based on Lyapunov stability theory, for updating the observer gains, which will give enough flexibility in the choice of initial estimates. Moreover, it may help to cope with unexpected state jerks. The trajectory tracking problem is formulated as a finite horizon optimal control problem, which is solved online. The control constraints are incorporated by using a nonquadratic performance functional. An adaptive update law has been derived for tuning the step size in the optimization algorithm, which may help to improve the convergence speed. Moreover, it is an attractive alternative to the heuristic choice of step size for diverse operating conditions. The disturbance as well as state estimates from the higher order sliding mode observer are utilized by the plant output prediction model, which will improve the overall performance of the controller. The nonlinear dynamics defined in leader fixed Euler-Hill frame has been considered for the present work and the reference trajectories are generated using Hill-Clohessy-Wiltshire equations of unperturbed motion. The simulation results based on rigorous perturbation analysis are presented to confirm the robustness of the proposed approach.

Index Terms—Adaptive gain higher order sliding mode observer, leader-follower formation, nonlinear model predictive control, spacecraft formation flying, tracking control.

I. INTRODUCTION

THE past few decades have witnessed tremendous advances in space science and satellite technology. It has revolutionized our knowledge and understanding of Earth's diverse landscapes, with enormous scientific applications in

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various fields ranging from biological and geological studies to disaster management. The low Earth orbiting (LEO) satellites have already proven their efficacy in imaging and remote sensing applications in prior missions, with their high resolution imaging capability, which may help in educing pivotal information in various applications. The wider area of coverage, coupled with the added advantages of smaller spacecraft formation, make the constellation of miniaturized satellites more preferable over a single large platform, in many real missions.

The success of the constellation observing system for meteorology, ionosphere, and climate (COSMIC) mission, a collaborative project of the National Space Organization (NSPO) in Taiwan, China and the University Corporation for Atmospheric Research (UCAR) in the United States, making use of a constellation of six LEO micro satellites, is an example for this.

A decentralized optimal control segment, with good disturbance rejection capability, is a prerequisite for the success of a spacecraft formation flying (SFF) mission. Most of the SFF control strategies depicted in literature, are derived from multi-robotic systems. A brief survey on various spacecraft formation flying control and guidance strategies has been provided in [1]. In [2], Liu and Kumar deal with a digital control technique for the tracking control of spacecraft in formation, with network-induced communication delays and external disturbances. Sliding mode control is one of the robust techniques commonly used in formation control applications [3]-[5]. In [6], Nair et al. deal with the formation control of multiple satellites using artificial potential field based path planning scheme, combined with adaptive fuzzy sliding mode control. A Lyapunov-based, robust, nonlinear, adaptive control law for the relative position control of spacecraft is depicted in [7]. The design of consensus protocol for the time varying formation control of swarm systems is presented in [8] and [9]. But in all these strategies, the optimality in control input cannot be guaranteed. A comprehensive survey on the distributed attitude coordination control, provided in [10], points out optimal fuel consumption as one of the key features to be considered in formation control problem. A linear quadratic regulator based approach applied to inplane satellite formation keeping using tangential maneuvers, is presented in [11], in which a linearized model is used, where

Model predictive control (MPC) is a widely used planning based approach, in which the concept of prediction in control makes it unique from other optimal control strategies. An MPC technique, in which a distributed decision-making with systematic constraint handling, to operate a linear dynamic network, is proposed in [13]. In [14], Xia and Zhang have utilized an MPC approach for the operation efficiency improvement in energy systems. Reference [15] depicts an MPC approach applied to a linearized spacecraft formation dynamics with sensing noise. There are only very few researches done in the area of SFF, using nonlinear model predictive control (NMPC) approach. Most of the works are based on linear MPC approach. An NMPC technique applied to a nonlinear submarine system, is presented in [16], where the finite horizon open loop optimization problem is solved online. A similar approach is also utilized in [17] for the formation control of multiple unmanned aerial vehicles, in which inequality constraints for control inputs, collision avoidance and obstacle avoidance are included in the optimization routine. In that case, the step size has been chosen heuristically based on the operating conditions. But it is not recommendable for complex systems with diverse, unpredictable operating conditions. Moreover, Karush-Kuhn-Tucker variables are used for handling the constraints. It may further increase the complexity of the optimization problem. In [18], Muske and Badgwell deal with an offset-free linear model predictive control, in which the plant model is augmented with a state space disturbance model.

The precision in position and velocity measurements are also crucial for the efficient performance of decentralized control and guidance segment. The relative positions of the spacecraft are measurable using inertial navigation systems, viz. GPS. Velocity can either be obtained from the Doppler shift, which needs more complex post processing for compensating the Doppler noise, or by differentiating the position. The latter one using conventional differentiation, will give more erroneous result, if there is noise in the original position measurements. A robust observer can be utilized for obtaining the velocity as well as disturbance estimates from the noisy position measurements. Among the various estimation schemes available in literature, higher order sliding mode (HOSM) based approach is found to be highly robust to disturbances and insensitive to parametric variations. Moreover, chattering in HOSM is very less compared to that of conventional sliding mode observers.

Different types of higher order sliding mode differentiator structures are available in [19]–[22]. A second order super twisting sliding mode observer for state as well as disturbance estimation, along with robust distributed output feedback control scheme for an SFF system, are depicted in [23]. In this work, the authors have used only radial and tangential plane linearized dynamics based on Hill's equations, where the out-of-plane dynamics is decoupled. [22] deals with the

simultaneous fault and disturbance reconstruction in a multiinput multi-output (MIMO) nonlinear system using a network of interconnected sliding mode observers, where a complex HOSM structure has been used. HOSM observer for unknown input reconstruction and state estimation, for a MIMO system is described in [24], where the observer structure is independent of the control input and Lipschitz constant. Also, a system dynamics, with total vector relative degree less than the total number of states, has been considered. In [25], Iqbal *et al.* propose a robust feedback linearization technique, in which an HOSM differentiator, incorporating Lebesgue measurable control input and Lipschitz constant, is used for approximating the dynamics in a single-input single-output (SISO) system. Similarly, a continuous sliding mode controller and an HOSM based disturbance observer are designed for the tracking control of air-breathing hypersonic vehicles in [26]. In all the above works, the observer gains are precomputed offline, where trajectory convergence cannot be guaranteed, if the initial estimates are chosen arbitrarily, or if there are unexpected state jerks.

The objective of the present work is to develop a robust decentralized control scheme for the tracking control of LEO satellites in a leader-follower based formation. Here, we propose a robust NMPC technique with a real time finite horizon open loop optimization [16], [17] for achieving the same. A nonquadratic cost function [27] has been considered for incorporating the control constraints, which can reduce the additional burden of constraint handling in the optimization algorithm. The heuristic choice of step size in optimization algorithm is not recommendable in all operating conditions. Hence, an adaptive tuning algorithm has been derived for updating the step size, which may improve the convergence speed. We have applied the HOSM observer structure containing the Lipschitz constant and Lebesgue measurable control input [19], [25] for obtaining the disturbance as well as velocity estimates from noisy position measurements. The system dynamics, having a total vector relative degree equal to the total number of system states, has been considered. The system is transformed into canonical form using nonlinear coordinate transformation [24]. For the observers with precomputed fixed gains, the convergence can be guaranteed, only when initial estimates are chosen in the neighborhood of the actual values of the states [28]. Moreover, it may not cope with unexpected state jerks. To handle these issues, here, we have proposed adaptive tuning algorithms, derived on the basis of Lyapunov stability theory, for updating the observer gains. The disturbance as well as the velocity estimates from the HOSM disturbance observer have been utilized by the output prediction model, in the NMPC [16], [17]. The prescribed differentiator structure has good estimation accuracy [19]. The observer design is repeated with an alternate HOSM differentiator structure as well. Simulation studies are conducted for a two-spacecraft system, for different levels of perturbations, along with added random noise. The results are compared with that of two recent references [26] and [29] respectively.

The rest of this paper is organized as follows. In Section II, the adaptive gain HOSM based technique for state as well as disturbance estimation is presented. Section III deals with

the complete NMPC algorithm. The nonlinear model, and the design of proposed HOSM based predictive controller for the SFF system, are provided in Section IV. Simulation results are provided in Section V to validate the robustness of the proposed control scheme. Conclusions are drawn in Section VI.

II. ADAPTIVE GAIN HIGHER ORDER SLIDING MODE OBSERVER

For the completeness of the discussion, we begin with the HOSM based state estimation technique, depicted in [19], [22], [24]. Consider the following nonlinear system

$$\dot{p} = f(p) + g(p)u + w(p)d$$

$$Y = Q(p)$$
(1)

where $f(p) \in \mathbb{R}^n$, $g(p) = [g_1, g_2, \ldots, g_m] \in \mathbb{R}^{n \times m}$, $Q(p) = [Q_1, Q_2, \ldots, Q_m] \in \mathbb{R}^m$, $w(p) = [w_1, w_2, \ldots, w_m] \in \mathbb{R}^{n \times m}$, $p \in \mathbb{R}^n$, $Y \in \mathbb{R}^m$, $u \in \mathbb{R}^m$ is the control input, and $d \in \mathbb{R}^m$ is the disturbance, defined in an open set $\Upsilon \subset \mathbb{R}^n$. The system with motion dynamics given by (1), is locally detectable, if the following conditions [24], [30] are satisfied,

Condition 1: The outputs of the system are assumed to have a vector relative degree, $e = \{e_1, e_2, \ldots, e_m\}$, corresponding to w(p), if $L_{w_j}L_f^kQ_i(p) = 0$, $\forall i = 1, 2, \ldots, m$, $\forall j = 1, 2, \ldots, m$, $\forall k = 1, 2, \ldots, e_i - 2$, and $\sum_{i=1}^m e_i = n$. Condition 2: M(p) =

$$\begin{pmatrix} L_{w_1}(L_f^{e_1-1}Q_1) & L_{w_2}(L_f^{e_1-1}Q_1) & \dots & L_{w_m}(L_f^{e_1-1}Q_1) \\ L_{w_1}(L_f^{e_2-1}Q_2) & L_{w_2}(L_f^{e_2-1}Q_2) & \dots & L_{w_m}(L_f^{e_2-1}Q_2) \\ \vdots & \vdots & \ddots & \vdots \\ L_{w_1}(L_f^{e_m-1}Q_m) & L_{w_2}(L_f^{e_m-1}Q_m) & \dots & L_{w_m}(L_f^{e_m-1}Q_m) \end{pmatrix}$$

is nonsingular.

Condition 3: The distribution $\Lambda = span\{w_1, w_2, \cdots, w_m\}$ is involutive.

These are necessary conditions for checking the observability of the nonlinear system in the presence of unknown inputs. The system given by (1) can be transformed into new coordinates as follows: $\forall i = 1, 2, ..., m$,

$$\gamma^{i} = \begin{pmatrix} \gamma_{1}^{i} \\ \gamma_{2}^{i} \\ \vdots \\ \gamma_{e_{i}}^{i} \end{pmatrix} = \psi(p) = \begin{pmatrix} \psi_{1}^{i}(p) \\ \psi_{2}^{i}(p) \\ \vdots \\ \psi_{e_{i}}^{i}(p) \end{pmatrix} = \begin{pmatrix} Q_{i}(p) \\ L_{f}Q_{i}(p) \\ \vdots \\ L_{f}^{e_{i}-1}Q_{i}(p) \end{pmatrix}$$

The transformation, $x = \psi^{-1}(\gamma)$, is a local diffeomorphism, since, it is described only in the neighborhood of any point, p, defined on an openset, $\Upsilon \subset \mathbb{R}^n$. Hence, the system given by (1) can be reformulated as

$$\dot{\gamma}_{j}^{i} = \gamma_{j+1}^{i}, \quad j = 1, 2, \dots, e_{i} - 1 \quad \forall i = 1, 2, \dots, m \quad (2)$$

$$\dot{\gamma}_{e_{i}}^{i} = \underbrace{L_{f}^{e_{i}}Q_{i}(\psi^{-1}(\gamma)) + L_{w}L_{f}^{e_{i}-1}Q_{i}(\psi^{-1}(\gamma))d}_{F^{i}(t,\gamma,d)} + \underbrace{L_{g}L_{f}^{e_{i}-1}Q_{i}(\psi^{-1}(\gamma))}_{P_{i}}u. \quad (3)$$

In this work, we have applied the HOSM differentiator structure [19], [25], incorporating the Lipschitz constant and Lebesgue measurable control input, to a complex multi-input multi-output system, to obtain the state as well as disturbance estimates from the noisy measurements. The observer gains are updated using adaptive tuning laws, derived based on Lyapunov stability theory. Hence the initial values of estimates can be chosen arbitrarily in the large compact subset of statespace. Moreover, it can handle unexpected state jerks. Using the system canonical form [24] given by (2) and (3), the HOSM observer can easily be designed. Here, the system dynamics under consideration has a total vector relative degree equal to n. The HOSM differentiator structure is defined as follows [19], [25],

$$\begin{split} \dot{\gamma}_{1}^{i} &= -\alpha_{1}^{i}\delta_{i}^{\frac{i}{e_{i}+1}}|\hat{\gamma}_{1}^{i} - Y_{i}|^{\frac{e_{i}}{e_{i}+1}}\mathrm{sign}(\hat{\gamma}_{1}^{i} - Y_{i}) + \hat{\gamma}_{2}^{i}\\ \dot{\hat{\gamma}}_{2}^{i} &= -\alpha_{2}^{i}\delta_{i}^{\frac{1}{e_{i}}}|\hat{\gamma}_{2}^{i} - \dot{\hat{\gamma}}_{1}^{i}|^{\frac{e_{i}-1}{e_{i}}}\mathrm{sign}(\hat{\gamma}_{2}^{i} - \dot{\hat{\gamma}}_{1}^{i}) + \hat{\gamma}_{3}^{i}\\ \vdots\\ \dot{\hat{\gamma}}_{e_{i}}^{i} &= \hat{F}^{i} + B^{i}u\\ \hat{F}^{i} &= -\alpha_{e_{i}}^{i}\delta_{i}^{\frac{1}{2}}|\hat{\gamma}_{e_{i}}^{i} - \dot{\hat{\gamma}}_{e_{i}-1}^{i}|^{\frac{1}{2}}\mathrm{sign}(\hat{\gamma}_{e_{i}}^{i} - \dot{\hat{\gamma}}_{e_{i}-1}^{i}) + \hat{\gamma}_{e_{i}+1}^{i}\\ \dot{\hat{\gamma}}_{e_{i}+1}^{i} &= -\alpha_{e_{i}+1}^{i}\delta_{i}\mathrm{sign}(\hat{\gamma}_{e_{i}+1}^{i} - \hat{F}^{i}) \end{split}$$
(4)

where $|\dot{F}^{i}(t,\gamma,d)| < \delta_{i}$, δ_{i} is the Lipschitz constant.

Theorem 1: Suppose that for the given nonlinear system, with the transformed dynamics as given in (2) and (3), and state observer structure as given in (4), the conditions (1)-(3) have been satisfied and the control input is Lebesgue measurable. Then, for any initial values of estimates chosen arbitrarily in the large compact subset of state space, the estimation error trajectories converge to origin along the sliding surface in finite time, if the observer gains are updated using the following adaptive tuning laws:

$$\dot{\hat{\alpha}}_{1}^{i} = \frac{1}{\beta_{1}^{i}} \delta^{\frac{1}{e_{i}+1}} |\eta_{1}^{i}|^{\frac{e_{i}}{e_{i}+1}+1}$$
(5)

$$\dot{\hat{\alpha}}_{2}^{i} = \frac{1}{\beta_{2}^{i}} \delta^{\frac{1}{e_{i}}} |\eta_{2}^{i}|^{\frac{e_{i}-1}{e_{i}}+1}$$
(6)

$$\dot{\hat{\alpha}}_{e_i}^i = \frac{1}{\beta_{e_i}^i} \delta^{\frac{1}{2}} |\eta_{e_i}^i|^{\frac{3}{2}}$$
(7)

$$\dot{\hat{\chi}}_{e_i+1}^i = \frac{1}{\beta_{e_i+1}^i} \delta |\eta_{e_i+1}^i|$$
(8)

where $\beta_j^i > 0, j = 1, 2, \dots, e_i + 1.$

Proof: The estimation error can be defined as

$$\eta_{1}^{i} = \hat{\gamma}_{1}^{i} - Y_{i} = \hat{\gamma}_{1}^{i} - \gamma_{1}^{i}$$

$$\eta_{2}^{i} = \hat{\gamma}_{2}^{i} - \gamma_{2}^{i}$$

$$\vdots$$

$$\eta_{e_{i}}^{i} = \hat{\gamma}_{e_{i}}^{i} - \gamma_{e_{i}}^{i}$$

$$\eta_{e_{i+1}}^{i} = \hat{\gamma}_{e_{i+1}}^{i} - F^{i}(t, \gamma, d).$$
(9)

Since,

$$\hat{\gamma}_{2}^{i} - \dot{\gamma}_{1}^{i} = \hat{\gamma}_{2}^{i} - \dot{\gamma}_{1}^{i} - \dot{\eta}_{1}^{i} = \eta_{2}^{i} - \dot{\eta}_{1}^{i} \hat{\gamma}_{3}^{i} - \dot{\gamma}_{2}^{i} = \hat{\gamma}_{3}^{i} - \dot{\gamma}_{2}^{i} - \dot{\eta}_{2}^{i} = \eta_{3}^{i} - \dot{\eta}_{2}^{i} \vdots$$

$$\begin{aligned} \hat{\gamma}_{e_{i}}^{i} - \dot{\hat{\gamma}}_{e_{i}-1}^{i} &= \hat{\gamma}_{e_{i}}^{i} - \dot{\gamma}_{e_{i}-1}^{i} - \dot{\eta}_{e_{i}-1}^{i} = \eta_{e_{i}}^{i} - \dot{\eta}_{e_{i}-1}^{i} \\ \hat{\gamma}_{e_{i}+1}^{i} - \hat{F}^{i} &= \hat{\gamma}_{e_{i}+1}^{i} - \dot{\hat{\gamma}}_{e_{i}}^{i} + B^{i}u = \hat{\gamma}_{e_{i}+1}^{i} - \dot{\hat{\gamma}}_{e_{i}}^{i} \\ &+ \dot{\gamma}_{e_{i}} - F^{i}(t, \gamma, d) = \eta_{e_{i}+1}^{i} - \dot{\eta}_{e_{i}}^{i}. \end{aligned}$$
(10)

The error dynamics can be written as follows [31]:

$$\begin{split} \dot{\eta}_{1}^{i} &= -\alpha_{1}^{i} \delta_{i}^{\frac{1}{e_{i}+1}} |\eta_{1}^{i}|^{\frac{e_{i}}{e_{i}+1}} \operatorname{sign}(\eta_{1}^{i}) + \eta_{2}^{i} \\ \dot{\eta}_{2}^{i} &= -\alpha_{2}^{i} \delta_{i}^{\frac{1}{e_{i}}} |\eta_{2}^{i} - \dot{\eta}_{1}^{i}|^{\frac{e_{i}-1}{e_{i}}} \operatorname{sign}(\eta_{2}^{i} - \dot{\eta}_{1}^{i}) + \eta_{3}^{i} \\ \vdots \\ \dot{\eta}_{e_{i}}^{i} &= -\alpha_{e_{i}}^{i} \delta_{i}^{\frac{1}{2}} |\eta_{e_{i}}^{i} - \dot{\eta}_{e_{i}-1}^{i}|^{\frac{1}{2}} \operatorname{sign}(\eta_{e_{i}}^{i} - \dot{\eta}_{e_{i}-1}^{i}) + \eta_{e_{i}+1}^{i} \\ \dot{\eta}_{e_{i}+1}^{i} \in -\alpha_{e_{i}+1}^{i} \delta_{i} \operatorname{sign}(\eta_{e_{i}+1}^{i} - \dot{\eta}_{e_{i}}^{i}) + [-\delta_{i}, +\delta_{i}]. \end{split}$$

Step 1: For the time $0 \le t \le t_1$, the Lyapunov candidate can be chosen as

$$V_1^i = \frac{(\eta_1^i)^2}{2}.$$
 (12)

Taking derivative of the Lyapunov function yields

$$\dot{V}_1^i = \eta_1^i \dot{\eta}_1^i.$$
 (13)

Substituting for $\dot{\eta}_1^i$

$$\begin{split} \dot{V}_{1}^{i} &= \eta_{1}^{i} \left(-\alpha_{1}^{i} \delta_{i}^{\frac{1}{e_{i}+1}} |\eta_{1}^{i}|^{\frac{e_{i}}{e_{i}+1}} \operatorname{sign}(\eta_{1}^{i}) + \eta_{2}^{i} \right) \\ &= -\alpha_{1}^{i} \delta_{i}^{\frac{1}{e_{i}+1}} |\eta_{1}^{i}|^{\frac{e_{i}}{e_{i}+1}} |\eta_{1}^{i}| + \eta_{1}^{i} \eta_{2}^{i} \\ &\leq -|\eta_{1}^{i}| \left(\alpha_{1}^{i} \delta_{i}^{\frac{1}{e_{i}+1}} |\eta_{1}^{i}|^{\frac{e_{i}}{e_{i}+1}} - |\eta_{2}^{i}| \right). \end{split}$$
(14)

If $\alpha_1^i > \frac{|\eta_2^i|_{\max}}{\delta_i^{\frac{1}{e_i+1}}(|\eta_1^i|_{\max})^{\frac{e_i}{e_i+1}}}$, the condition $\dot{V}_1^i \leq 0$ can be satisfied. But an observer with precomputed fixed gains, will

not guarantee convergence of estimation error to the origin, if the initial estimate is not chosen properly in the neighborhood of the actual state, or if there are unexpected state jerks. Hence, we are proposing an adaptive tuning rule for updating the gains. We have derived the adaptation rule for a given operating condition. Suppose that, there exists a finite, optimal value for gain, α_1^{i*} , at the given operating condition, which can ensure the convergence of the trajectories to the sliding manifolds, and it can be defined as

$$\alpha_1^{i*} = |\eta_2^i| \left(\delta_i^{\frac{1}{e_i+1}} |\eta_1^i|^{\frac{e_i}{e_i+1}} \right)^{-1} + \Theta_1^i \tag{15}$$

where $|\eta_1^i|_{\text{max}}$ represents the upper bound on the estimation error. Since the updation of gains and the convergence of the estimation error take place sequentially, η_2^i remains constant at the initial value, during the time step, $0 \le t \le t_1$, for the given operating condition. The value of constant Θ_1^i is chosen sufficiently high enough to ensure that the condition, $V_1^i \leq 0$, always holds. Hence $\dot{\alpha}_1^{i*} = 0$. The adaptation error can be defined as

$$\tilde{\alpha}_1^i = \hat{\alpha}_1^i - \alpha_1^{i*}. \tag{16}$$

The Lyapunov function can be modified as

$$V_1^i = \frac{1}{2} (\eta_1^i)^2 + \frac{\beta_1^i}{2} (\tilde{\alpha}_1^i)^2, \quad \beta_1^i > 0.$$

Hence,

$$\begin{split} \dot{V}_{1}^{i} &= -\eta_{1}\dot{\eta}_{1} + \beta_{1}^{i}\tilde{\alpha}_{1}\dot{\dot{\alpha}}_{1} \\ &= -\hat{\alpha}_{1}^{i}\delta_{i}^{\frac{1}{e_{i}+1}}|\eta_{1}^{i}|^{\frac{e_{i}}{e_{i}+1}}|\eta_{1}^{i}| + \eta_{1}^{i}\eta_{2}^{i} + \beta_{1}^{i}\tilde{\alpha}_{1}^{i}\dot{\dot{\alpha}}_{1}^{i} \\ &\leq -|\eta_{1}^{i}|\left(\hat{\alpha}_{1}^{i}\delta_{i}^{\frac{1}{e_{i}+1}}|\eta_{1}^{i}|^{\frac{e_{i}}{e_{i}+1}} - |\eta_{2}^{i}|\right) + \beta_{1}^{i}\tilde{\alpha}_{1}^{i}\dot{\dot{\alpha}}_{1} \\ &= -|\eta_{1}^{i}|\left(\hat{\alpha}_{1}^{i}\delta_{i}^{\frac{1}{e_{i}+1}}|\eta_{1}^{i}|^{\frac{e_{i}}{e_{i}+1}} - |\eta_{2}^{i}| \\ &+ \alpha_{1}^{i*}\delta_{i}^{\frac{1}{e_{i}+1}}|\eta_{1}^{i}|^{\frac{e_{i}}{e_{i}+1}} - \alpha_{1}^{i*}\delta_{i}^{\frac{1}{e_{i}+1}}|\eta_{1}^{i}|^{\frac{e_{i}}{e_{i}+1}}\right) + \beta_{1}^{i}\tilde{\alpha}_{1}^{i}\dot{\dot{\alpha}}_{1}^{i} \\ &= -|\eta_{1}^{i}|\left((\hat{\alpha}_{1}^{i} - \alpha_{1}^{i*})\delta_{i}^{\frac{1}{e_{i}+1}}|\eta_{1}^{i}|^{\frac{e_{i}}{e_{i}+1}}\right) + \beta_{1}^{i}\tilde{\alpha}_{1}^{i}\dot{\dot{\alpha}}_{1}^{i} \\ &- |\eta_{2}^{i}| + \alpha_{1}^{i*}\delta_{i}^{\frac{1}{e_{i}+1}}|\eta_{1}^{i}|^{\frac{e_{i}}{e_{i}+1}}\right) + \beta_{1}^{i}\tilde{\alpha}_{1}^{i}\dot{\dot{\alpha}}_{1}^{i}. \end{split} \tag{17}$$

Substituting for α_1^{i*} from (15),

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$$\begin{split} \dot{V}_{1}^{i} &= - |\eta_{1}^{i}| \left(\left(\hat{\alpha}_{1}^{i} - \alpha_{1}^{i*} \right) \delta_{i}^{\frac{1}{e_{i}+1}} |\eta_{1}^{i}|^{\frac{e_{i}}{e_{i}+1}} - |\eta_{2}^{i}| \right. \\ &+ |\eta_{2}^{i}| \left(\delta_{i}^{\frac{1}{e_{i}+1}} |\eta_{1}^{i}|^{\frac{e_{i}}{e_{i}+1}} \right)^{-1} \delta_{i}^{\frac{1}{e_{i}+1}} |\eta_{1}^{i}|^{\frac{e_{i}}{e_{i}+1}} \\ &+ \Theta_{1}^{i} \delta_{i}^{\frac{1}{e_{i}+1}} |\eta_{1}^{i}|^{\frac{e_{i}}{e_{i}+1}} \right) + \beta_{1}^{i} \tilde{\alpha}_{1}^{i} \dot{\tilde{\alpha}}_{1}^{i}. \end{split}$$

Since $(-|\eta_2^i| + |\eta_2^i| (\delta_i^{\frac{-i}{e_i+1}} |\eta_1^i|_{\max}^{\frac{-i}{e_i+1}})^{-1} \delta_i^{\frac{-i}{e_i+1}} |\eta_1^i|^{\frac{e_i}{e_i+1}}) \leq 0$, we can write

$$\dot{V}_{1}^{i} \leq - |\eta_{1}^{i}| \left(\tilde{\alpha}_{1}^{i} \delta_{i}^{\frac{1}{e_{i}+1}} |\eta_{1}^{i}|^{\frac{e_{i}}{e_{i}+1}} + \Theta_{1}^{i} \delta_{i}^{\frac{1}{e_{i}+1}} |\eta_{1}^{i}|^{\frac{e_{i}}{e_{i}+1}} \right) + \beta_{1}^{i} \tilde{\alpha}_{1}^{i} \dot{\hat{\alpha}}_{1}^{i}.$$

$$(18)$$

If
$$\dot{\alpha}_{1}^{i} = \frac{1}{\beta_{1}^{i}} \delta^{\frac{1}{e_{i}+1}} |\eta_{1}^{i}|^{\frac{e_{i}}{e_{i}+1}+1}$$
 (19)

then
$$\dot{V}_1^i \le -|\eta_1^i|\Theta_1^i\delta_i^{\frac{1}{e_i+1}}|\eta_1^i|^{\frac{e_i}{e_i+1}}.$$
 (20)

Since Θ_1^i , $\delta_i > 0$, $\dot{V}_1^i \le 0$, and hence, η_1^i and $\tilde{\alpha}_1^i$ are bounded. The initial values of the estimates as well as the observer gains are chosen such that the estimation error dynamics stay uniformly bounded [22]. Moreover, the updation of the observer gains as well as convergence of the estimation error to the origin, take place sequentially. Hence, η_2^i remains bounded during this time step. Since, η_1^i and η_2^i are bounded, from (15) and (16), it is clear that $\hat{\alpha}_1^{i*}$, and $\hat{\alpha}_1^{i}$ are bounded. From this, using (11) and (20), we found that V_1^i is bounded. Hence, V_1^i is uniformly continuous, and by Barbalat's Lemma [32], system is asymptotically stable. The observer gain, α_1^i is bounded, where α_1^{i*} is the bound at the given operating condition.

From (11), using Filippov theory [19], [31], we can find that the differential inclusion is consistent with the dilation, $t \rightarrow \tau t, \eta^i_i \rightarrow \tau^{e_i-j+1}, j = 0, 1, \dots, e_i, \forall i.$ Hence it is homogeneous with a homogeneity degree of -1. Since the homogeneity degree is negative, the finite time convergence [19], [21], [25], [33] of η_1^i to zero can be guaranteed. The same has already been proved by Levant in [34], using the property of contractivity. Moreover, by Orlov's theorem [33], [35], the finite time stability of the system can be guaranteed, if the system is asymptotically stable and the degree of homogeneity is negative. Since \dot{V}_1^i is uniformly continuous, the system is asymptotically stable. Hence, using Orlov's theorem, the finite time stability can also be guaranteed, and the estimation error converges to zero in finite time say t_1 , and the remaining error dynamics will converge to a hyper ball in finite time [32].

Step 2: For the time $t_1 \leq t \leq t_2$, the remaining error dynamics becomes

$$0 = -\alpha_{1}^{i} \delta_{i}^{\frac{1}{e_{i}+1}} |\eta_{1}^{i}|^{\frac{e_{i}}{e_{i}+1}} \operatorname{sign}(\eta_{1}^{i}) + \eta_{2}^{i}$$

$$\dot{\eta}_{2}^{i} = -\alpha_{2}^{i} \delta_{i}^{\frac{1}{e_{i}}} |\eta_{2}^{i}|^{\frac{e_{i}-1}{e_{i}}} \operatorname{sign}(\eta_{2}^{i}) + \eta_{3}^{i}$$

$$\vdots$$

$$\dot{\eta}_{e_{i}}^{i} = -\alpha_{e_{i}}^{i} \delta_{i}^{\frac{1}{2}} |\eta_{e_{i}}^{i} - \dot{\eta}_{e_{i}-1}^{i}|^{\frac{1}{2}} \operatorname{sign}(\eta_{e_{i}}^{i} - \dot{\eta}_{e_{i}-1}^{i}) + \eta_{e_{i}+1}^{i}$$

$$\dot{\eta}_{e_{i}+1}^{i} \in -\alpha_{e_{i}+1}^{i} \delta_{i} \operatorname{sign}(\eta_{e_{i}+1}^{i} - \dot{\eta}_{e_{i}}^{i}) + [-\delta_{i}, +\delta_{i}]. \quad (21)$$

The Lyapunov candidate function can be chosen as

$$V_2^i = \frac{(\eta_2^i)^2}{2}.$$
 (22)

Taking derivative of the Lyapunov function and, substituting for $\dot{\eta}_2^i$, as in Step 1 yields

$$\dot{V}_{2}^{i} \leq -|\eta_{2}^{i}| \left(\alpha_{2}^{i} \delta_{i}^{\frac{1}{e_{i}}} |\eta_{2}^{i}|^{\frac{e_{i}-1}{e_{i}}} - |\eta_{3}^{i}| \right).$$
(23)

If $\alpha_2^i > \frac{|\eta_3^i|_{\max}}{\delta_i^{\frac{1}{e_i}}(|\eta_2^i|_{\max})^{\frac{e_i-1}{e_i}}}$, the condition $\dot{V}_1^i \leq 0$ can be satisfied. Similar to that in the previous step, the nominal value

satisfied. Similar to that in the previous step, the nominal value of the gain α_2^{i*} can be defined as

$$\alpha_2^{i*} = |\eta_3^i| \left(\delta_i^{\frac{1}{e_i}}(|\eta_2^i|)_{\max}^{\frac{e_i-1}{e_i}} \right)^{-1} + \Theta_2^i$$
(24)

where $\Theta_2^i > 0$. The adaptation error can be defined as

$$\tilde{\alpha}_2^i = \hat{\alpha}_2^i - \alpha_2^{i*}.$$
(25)

The Lyapunov function can be modified as

$$V_2^i = \frac{1}{2}(\eta_2^i)^2 + \frac{\beta_2^i}{2}(\tilde{\alpha}_2^i)^2, \quad \beta_2^i > 0.$$

Taking the derivative of V_2^i , and proceeding in the same way as in Step 1 yields,

$$\dot{V}_{2}^{i} \leq -|\eta_{2}^{i}| \left(\tilde{\alpha}_{2}^{i} \delta_{i}^{\frac{1}{e_{i}}} |\eta_{2}^{i}|^{\frac{e_{i}-1}{e_{i}}} + \Theta_{2}^{i} \delta_{i}^{\frac{1}{e_{i}}} |\eta_{2}^{i}|^{\frac{e_{i}-1}{e_{i}}} \right) + \beta_{2}^{i} \tilde{\alpha}_{2}^{i} \dot{\hat{\alpha}}_{2}^{i}.$$
(26)

If
$$\dot{\hat{\alpha}}_{2}^{i} = \frac{1}{\beta_{2}^{i}} \delta^{\frac{1}{e_{i}}} |\eta_{2}^{i}|^{\frac{e_{i}-1}{e_{i}}+1}$$
 (27)

then
$$\dot{V}_{2}^{i} \leq -|\eta_{2}^{i}|\Theta_{2}^{i}\delta_{i}^{\frac{1}{e_{i}}}|\eta_{2}^{i}|^{\frac{e_{i}-1}{e_{i}}}.$$
 (28)

Just as in the previous step, we can prove that \dot{V}_2^i is uniformly continuous, and the finite time convergence of the trajectories to the sliding manifolds can be guaranteed; $\eta_2^i \rightarrow$ 0. Hence, after t_2 , the estimation error stays equal to zero. Step 3: For the time $t_2 \leq t \leq t_n$, the remaining error dynamics becomes

$$0 = -\alpha_{1}^{i} \delta_{i}^{\frac{1}{e_{i}+1}} |\eta_{1}^{i}|^{\frac{e_{i}}{e_{i}+1}} \operatorname{sign}(\eta_{1}^{i}) + \eta_{2}^{i}$$

$$0 = -\alpha_{2}^{i} \delta_{i}^{\frac{1}{e_{i}}} |\eta_{2}^{i}|^{\frac{e_{i}-1}{e_{i}}} \operatorname{sign}(\eta_{2}^{i}) + \eta_{3}^{i}$$

$$\vdots$$

$$0 = -\alpha_{e_{i}-1}^{i} \delta_{i}^{\frac{1}{3}} |\eta_{e_{i}-1}^{i}|^{\frac{2}{3}} \operatorname{sign}(\eta_{e_{i}-1}^{i}) + \eta_{e_{i}}^{i}$$

$$\dot{\eta}^{i} = -\alpha_{i}^{i} \delta_{2}^{\frac{1}{2}} |\eta^{i}|^{\frac{1}{2}} \operatorname{sign}(\eta^{i}) + \eta^{i}$$

$$(29)$$

$$\dot{\eta}_{e_i}^{i} = -\alpha_{e_i \circ i}^{i} + \eta_{e_i} + \delta_i \operatorname{Sign}(\eta_{e_i+1}^{i} - \dot{\eta}_{e_i}^{i}) + [-\delta_i, +\delta_i].$$
(30)

The Lyapunov candidate function can be chosen as

$$V_{e_i}^i = \frac{(\eta_{e_i}^i)^2}{2}.$$
(31)

Taking derivative of the Lyapunov function and substituting for $\dot{\eta}_{e_i}^i$ as in Step 1 yields,

$$\dot{V}_{e_i}^i \le -|\eta_{e_i}^i| \left(\alpha_{e_i}^i \delta_i^{\frac{1}{2}} |\eta_{e_i}^i|^{\frac{1}{2}} - |\eta_{e_i+1}^i| \right).$$
(32)

If $\alpha_{e_i}^i > \frac{|\eta_{e_i+1}^i|_{\max}}{\delta_i^{\frac{1}{2}}(|\eta_{e_i}^i|_{\max})^{\frac{1}{2}}}$, then $\dot{V}_{e_i}^i \leq 0$. As in the previous

ep, the nominal value of the gain
$$\alpha_{e_i}^{**}$$
 can be defined as

$$\chi_{e_{i}}^{i*} = |\eta_{e_{i}+1}^{i}| \left(\delta_{i}^{\frac{1}{2}} (|\eta_{e_{i}})^{i}|_{\max}^{\frac{1}{2}} \right)^{-1} + \Theta_{e_{i}}^{i}$$
(33)

where $\Theta_{e_i}^i > 0$. The adaptation error can be defined as

$$\tilde{\alpha}_{e_i}^i = \hat{\alpha}_{e_i}^i - \alpha_{e_i}^{i*}.$$
(34)

The Lyapunov function can be modified as

$$V_{e_i}^i = \frac{1}{2} (\eta_{e_i}^i)^2 + \frac{\beta_{e_i}^i}{2} (\tilde{\alpha}_{e_i}^i)^2, \quad \beta_{e_i}^i > 0.$$

Taking the derivative of $V_{e_i}^i$, and proceeding in the same way as in Step 1 yields,

$$\dot{V}_{e_{i}}^{i} \leq -|\eta_{e_{i}}^{i}| \left(\tilde{\alpha}_{e_{i}}^{i} \delta_{i}^{\frac{1}{2}} |\eta_{e_{i}}^{i}|^{\frac{1}{2}} + \Theta_{e_{i}}^{i} \delta_{i}^{\frac{1}{2}} |\eta_{e_{i}}^{i}|^{\frac{1}{2}} \right) + \beta_{e_{i}}^{i} \tilde{\alpha}_{e_{i}}^{i} \dot{\hat{\alpha}}_{e_{i}}^{i}.$$
(35)

If
$$\hat{\alpha}_{e_i}^i = \frac{1}{\beta_{e_i}^i} \delta^{\frac{1}{2}} |\eta_{e_i}^i|^{\frac{3}{2}}$$
 (36)

then
$$\dot{V}_{e_i}^i \le -|\eta_{e_i}^i|\Theta_{e_i}^i\delta_i^{\frac{1}{2}}|\eta_{e_i}^i|^{\frac{1}{2}}.$$
 (37)

Similar to Step 1, we can find $V_{e_i}^i$ as uniformly continuous, and the finite time convergence of the trajectories to the sliding manifolds can be guaranteed; $\eta_{e_i}^i \rightarrow 0$. Hence, after t_{e_i} , the estimation error stays equal to zero.

Step 4: For the time $t_{e_i} \leq t \leq t_{e_i+1}$, the remaining error dynamics becomes

$$0 = -\alpha_{1}^{i} \delta_{i}^{\frac{1}{e_{i}+1}} |\eta_{1}^{i}|^{\frac{e_{i}}{e_{i}+1}} \operatorname{sign}(\eta_{1}^{i}) + \eta_{2}^{i}$$

$$0 = -\alpha_{2}^{i} \delta_{i}^{\frac{1}{e_{i}}} |\eta_{2}^{i}|^{\frac{e_{i}-1}{e_{i}}} \operatorname{sign}(\eta_{2}^{i}) + \eta_{3}^{i}$$

$$\vdots$$

$$0 = -\alpha_{e_{i}-1}^{i} \delta_{i}^{\frac{1}{3}} |\eta_{e_{i}-1}^{i}|^{\frac{2}{3}} \operatorname{sign}(\eta_{e_{i}-1}^{i}) + \eta_{e_{i}}^{i}$$

$$0 = -\alpha_{e_{i}}^{i} \delta_{i}^{\frac{1}{2}} |\eta_{e_{i}}^{i}|^{\frac{1}{2}} \operatorname{sign}(\eta_{e_{i}}^{i}) + \eta_{e_{i}+1}^{i}$$

$$\dot{\eta}_{e_{i}+1}^{i} \in -\alpha_{e_{i}+1}^{i} \delta_{i} \operatorname{sign}(\eta_{e_{i}+1}^{i}) + [-\delta_{i}, +\delta_{i}]. \quad (38)$$

The Lyapunov candidate function can be chosen as

$$V_{e_i+1}^i = \frac{(\eta_{e_i+1}^i)^2}{2}.$$
(39)

Taking derivative of the Lyapunov function yields

$$V_{e_i+1}^i = \eta_{e_i+1}^i \dot{\eta}_{e_i+1}^i.$$
(40)

Substituting for $\dot{\eta}_{e_i+1}^i$

$$\dot{V}_{e_{i}+1}^{i} = V_{i}^{'} \in \eta_{e_{i}+1}^{i}(-\alpha_{e_{i}+1}^{i}\delta_{i}\mathrm{sign}(\eta_{e_{i}+1}^{i}) + [-\delta_{i}, +\delta_{i}]) \\
= V_{i}^{'} \in -\alpha_{e_{i}+1}^{i}\delta_{i}|\eta_{e_{i}+1}^{i}| + \eta_{e_{i}+1}^{i}[-\delta_{i}, +\delta_{i}] \\
\leq V_{i}^{'} \in -|\eta_{e_{i}+1}^{i}|(\alpha_{e_{i}+1}^{i}\delta_{i} - |[-\delta_{i}, +\delta_{i}]|)$$
(41)

where $|[-\delta_i, +\delta_i]|$ represents the absolute values of the elements of the closed set. The nominal value of $\alpha_{e_i+1}^i$ can be chosen as

$$\alpha_{e_i+1}^{i*} \in |[-\delta_i, +\delta_i]| \delta_i^{-1} + \Theta_{e_i+1}$$
(42)

where $\Theta_{e_i+1} > 0$. The adaptation error can be defined as

$$\tilde{\alpha}_{e_i+1}^i = \hat{\alpha}_{e_i+1}^i - \alpha_{e_i+1}^{i*}.$$

The Lyapunov function can be modified as

$$V_{e_i+1}^i = \frac{1}{2} (\eta_{e_i+1}^i)^2 + \frac{\beta_{e_i+1}^i}{2} (\tilde{\alpha}_{e_i+1}^i)^2, \quad \beta_{e_i+1}^i > 0.$$

Taking the derivative,

$$\begin{split} \dot{V}_{e_{i}+1}^{i} &= V_{i}^{''} \in \eta_{e_{i}+1}^{i}(-\hat{\alpha}_{e_{i}+1}^{i}\delta_{i}\mathrm{sign}(\eta_{e_{i}+1}^{i}) + [-\delta_{i}, +\delta_{i}]) \\ &+ \beta_{e_{i}+1}^{i}\tilde{\alpha}_{e_{i}+1}^{i}\dot{\alpha}_{e_{i}+1}^{i} \\ &\leq V_{i}^{''} \in -|\eta_{e_{i}+1}^{i}|(\hat{\alpha}_{e_{i}+1}^{i}\delta_{i} - |[-\delta_{i}, +\delta_{i}]|) \\ &+ \beta_{e_{i}+1}^{i}\tilde{\alpha}_{e_{i}+1}^{i}\dot{\alpha}_{e_{i}+1}^{i} \\ &\leq V_{i}^{''} \in -|\eta_{e_{i}+1}^{i}|(\hat{\alpha}_{e_{i}+1}^{i}\delta_{i} - |[-\delta_{i}, +\delta_{i}]| \\ &+ \alpha_{e_{i}+1}^{i*}\delta_{i} - \alpha_{e_{i}+1}^{i*}\delta_{i}) + \beta_{e_{i}+1}^{i}\tilde{\alpha}_{e_{i}+1}^{i}\dot{\alpha}_{e_{i}+1}^{i}. \end{split}$$

Using (42),

$$\dot{V}_{e_i+1}^i \le -|\eta_{e_i+1}^i|(\tilde{\alpha}_{e_i+1}^i\delta_i + \Theta_{e_i+1}\delta_i) + \beta_{e_i+1}^i\tilde{\alpha}_{e_i+1}^i\dot{\alpha}_{e_i+1}^i.$$
(43)

$$\begin{split} \text{If} & \dot{\alpha}_{e_{i}+1}^{i} = \frac{1}{\beta_{e_{i}+1}^{i}} \delta_{i} |\eta_{e_{i}+1}^{i}| \\ \text{then} & \dot{V}_{2}^{i} \leq -|\eta_{e_{i}+1}^{i}| \Theta_{e_{i}+1}^{i} \delta_{i}. \end{split}$$

Hence, as in the previous steps, the finite time convergence of the trajectories to the sliding manifolds can be guaranteed. From (41), it is clear that $\hat{\alpha}_{e_i+1}^i > 1$ is a simple and sufficient condition to ensure the same.

Theorem 2: Upon satisfaction of Conditions 1-3, for any nonlinear system, with the transformed motion dynamics given by (2) and (3), and the control input is Lebesgue measurable, the disturbance inputs can be estimated from the observer structure, represented by (4), as

$$\hat{d} = M^{-1}(\psi^{-1}(\hat{\gamma})) \left[\begin{cases} \hat{\gamma}_{e_{1}+1}^{1} \\ \hat{\gamma}_{e_{2}+1}^{2} \\ \vdots \\ \hat{\gamma}_{e_{m}+1}^{m} \end{cases} - \begin{cases} L_{f}^{e_{1}}Q_{1}(\psi^{-1}(\hat{\gamma})) \\ L_{f}^{e_{2}}Q_{2}(\psi^{-1}(\hat{\gamma})) \\ \vdots \\ L_{f}^{e_{m}}Q_{m}(\psi^{-1}(\hat{\gamma})) \end{cases} \right] \right].$$

Proof: When $\eta_1^i \to 0$, from (11), it is clear that $\dot{\eta}_1^i = \eta_2^i$, since sign $(\eta_1^i) = 0$. When $\dot{\eta}_1^i = \eta_2^i$, sign $(\eta_2^i - \dot{\eta}_1^i) = 0$, hence $\dot{\eta}_2^i = \eta_3^i$. Proceeding in this way, the following results can be obtained from (11):

$$\dot{\eta}_{1}^{i} = \eta_{2}^{i} \\
\dot{\eta}_{2}^{i} = \eta_{3}^{i} \\
\vdots \\
\dot{\eta}_{e_{i}}^{i} = \eta_{e_{i}+1}^{i}.$$
(44)

From (10) and (44), $\hat{\gamma}_{e_i+1}^i - \hat{F}^i(t, \gamma, d) = 0$. Hence, using (3), the disturbance estimates can be obtained as

$$\hat{d} = M^{-1}(\psi^{-1}(\hat{\gamma})) \begin{bmatrix} \left\{ \begin{array}{c} \hat{\gamma}_{e_{1}+1}^{l} \\ \hat{\gamma}_{e_{2}+1}^{2} \\ \vdots \\ \hat{\gamma}_{e_{m}+1}^{m} \end{array} \right\} - \left\{ \begin{array}{c} L_{f}^{e_{1}}Q_{1}(\psi^{-1}(\hat{\gamma})) \\ L_{f}^{e_{2}}Q_{2}(\psi^{-1}(\hat{\gamma})) \\ \vdots \\ L_{f}^{e_{m}}Q_{m}(\psi^{-1}(\hat{\gamma})) \end{array} \right\} \end{bmatrix}$$
(45)

where $\hat{\gamma}_1, \hat{\gamma}_2, \ldots, \hat{\gamma}_{e_i}$ give the respective state estimates.

The observer design is repeated with an alternate HOSM differentiator structure as well.

Corollary 1: Upon satisfaction of Conditions 1-3, for any nonlinear system, with the transformed motion dynamics given by (2) and (3), and the control input is Lebesgue measurable, choosing an alternate structure for an HOSM differentiator [19], [24] given by,

$$\begin{split} \dot{\hat{\gamma}}_{1}^{i} &= -\alpha_{1}^{i} |\hat{\gamma}_{1}^{i} - Y_{i}|^{\frac{e_{i}}{e_{i}+1}} \operatorname{sign}(\hat{\gamma}_{1}^{i} - Y_{i}) + \hat{\gamma}_{2}^{i} \\ \dot{\hat{\gamma}}_{2}^{i} &= -\alpha_{2}^{i} |\hat{\gamma}_{2}^{i} - \dot{\hat{\gamma}}_{1}^{i}|^{\frac{e_{i}-1}{e_{i}}} \operatorname{sign}(\hat{\gamma}_{2}^{i} - \dot{\hat{\gamma}}_{1}^{i}) + \hat{\gamma}_{3}^{i} \\ \vdots \\ \dot{\hat{\gamma}}_{e_{i}}^{i} &= \hat{F}^{i} \\ \hat{F}^{i} &= -\alpha_{e_{i}}^{i} |\hat{\gamma}_{e_{i}}^{i} - \dot{\hat{\gamma}}_{e_{i}-1}^{i}|^{\frac{1}{2}} \operatorname{sign}(\hat{\gamma}_{e_{i}}^{i} - \dot{\hat{\gamma}}_{e_{i}-1}^{i}) + \hat{\gamma}_{e_{i}+1}^{i} \\ \dot{\hat{\gamma}}_{e_{i}+1}^{i} &= -\alpha_{e_{i}+1}^{i} \operatorname{sign}(\hat{\gamma}_{e_{i}+1}^{i} - \hat{F}^{i}) \end{split}$$
(46)

the disturbance inputs can be estimated as

$$\hat{d} = M^{-1}(\psi^{-1}(\hat{\gamma})) \left[\begin{cases} \hat{\gamma}_{e_{1}+1}^{e_{1}} \\ \hat{\gamma}_{e_{2}+1}^{e_{2}} \\ \vdots \\ \hat{\gamma}_{e_{m}+1}^{m} \end{cases} - \begin{cases} L_{f}^{e_{1}}Q_{1}(\psi^{-1}(\hat{\gamma})) \\ L_{f}^{e_{2}}Q_{2}(\psi^{-1}(\hat{\gamma})) \\ \vdots \\ L_{f}^{e_{m}}Q_{m}(\psi^{-1}(\hat{\gamma})) \end{cases} \right] - \left\{ \begin{array}{c} L_{g}L_{f}^{e_{1}-1}Q_{1}(\psi^{-1}(\gamma))u \\ L_{g}L_{f}^{e_{2}-1}Q_{2}(\psi^{-1}(\gamma))u \\ \vdots \\ L_{g}L_{f}^{e_{m}-1}Q_{m}(\psi^{-1}(\gamma))u \end{array} \right\} \right].$$
(47)

Proof: Unlike Theorems 1 and 2, in this case, $F^i(t, \gamma, d)$ is defined as

$$F^{i}(t,\gamma,d) = L_{f}^{e_{i}}Q_{i}(\psi^{-1}(\gamma)) + L_{w}L_{f}^{e_{i}-1}Q_{i}(\psi^{-1}(\gamma))d + L_{g}L_{f}^{e_{i}-1}Q_{i}(\psi^{-1}(\gamma))u.$$
(48)

The proof will be similar to that of Theorem 2. The update laws for the observer gains also remain the same, provided the value of δ_i is to be taken as 1. Finally, as in previous case, we will arrive at the condition, $\hat{\gamma}^i_{e_i+1} - \hat{F}^i(t, \gamma, d) = 0$. From



Fig. 1. Block schematic of the proposed scheme.

this, making use of (48), the required expression for disturbance observer can be obtained.

The complete NMPC algorithm, incorporating the disturbance estimates from an HOSM disturbance observer, is given in next section.

III. NONLINEAR MODEL PREDICTIVE CONTROL

The block schematic of the proposed scheme is shown in Fig. 1. The fundamentals of NMPC problem and the optimization algorithm used [16], [17], can be detailed as follows. The objective is to find the control input sequence $u_0, u_1, \ldots, u_{N-1}$, that minimizes the following nonquadratic cost function:

$$J = \chi(\tilde{p}_N) + \sum_{k=0}^{N-1} L(\tilde{p}_k, u_k)$$
(49)

$$\chi(\tilde{p}_N) = \frac{1}{2} \tilde{p}_N^T S \tilde{p}_N \tag{50}$$

$$L(\tilde{p}_k, u_k) = \frac{1}{2} \tilde{p}_k^T P \tilde{p}_k + A(u_k)$$
(51)

where $A : \mathbb{R}^n \to \mathbb{R}$, is utilized to incorporate the control constraints, and it can be defined in terms of continuous one-to-one real-analytic integrable inverse hyperbolic tangent function [27] as

$$A(u_k) = \int_0^{u_k} \tanh^{-1} \left(\frac{u_k}{u_{\max}}\right)^T R du_k$$
 (52)

where u_{max} is the bound on control input. The discretized dynamic constraints to be satisfied, is given by

$$p(k+1) = p(k) + T\dot{p} = f_k$$
 (53)

where f_k is a function of d_k , which can be estimated using HOSM disturbance observer; $\tilde{p} = p_d - p$; and p_d is the reference trajectory.

The Hamiltonian can be written as

$$H_k = L(\tilde{p}_k, u_k) + \lambda_{k+1}^T f_k$$

For driving the derivatives of the cost function to zero, the Lagrange multipliers can be chosen as

$$\lambda_N^T = -\tilde{p}_N^T S \tag{54}$$

$$\lambda_k^T = \lambda_{k+1}^T \frac{\partial f_k}{\partial p_k} - \tilde{p}_k^T P.$$
(55)

The gradient of the Hamiltonian is given by

$$\left(\frac{\partial H_k}{\partial u_k}\right)^T = \tanh^{-1} \left(\frac{u_k}{u_{\max}}\right)^T R + \lambda_{k+1}^T \frac{\partial f_k}{\partial u_k}.$$
 (56)

The control update law [16] can be written as

$$u_k^{j+1} = u_k^j + \xi_k \frac{\partial H_k}{\partial u_k}, \quad k = 0, 1, \dots, N-1$$
 (57)

where ξ_k is the step size.

Theorem 3: For the given nonquadratic cost function, the Hamiltonian can be minimized, and convergence to the optimal value function can be guaranteed, if the step size is updated using the adaptive update law.

$$\dot{\hat{\xi}}_{k} = -\frac{1}{\Omega T} \left(\frac{\partial H}{\partial u}\right)^{T} \left(\frac{\partial H}{\partial u}\right) H$$
(58)

where T is the sampling time.

Proof: Consider the energy function

$$V(H_k) = \frac{1}{2}H_k^2.$$
 (59)

Taking the derivative,

$$\dot{V}(H_k) = H_k \left[\left(\frac{\partial H_k}{\partial u_k} \right)^T \frac{du_k}{dt} + \left(\frac{\partial H_k}{\partial p_k} \right)^T \frac{dp_k}{dt} + \left(\frac{\partial H_k}{\partial \tilde{p}_k} \right)^T \frac{d\tilde{p}_k}{dt} + \left(\frac{\partial H_k}{\partial d_k} \right)^T \frac{dd_k}{dt} \right].$$

From (57),

$$\frac{\partial u_k}{\partial t} = \frac{1}{T} \xi_k \frac{\partial H_k}{\partial u_k}.$$
(60)

Hence,

$$\dot{V}(H_k) = H_k \left[\frac{1}{T} \xi_k \left(\frac{\partial H_k}{\partial u_k} \right)^T \left(\frac{\partial H_k}{\partial u_k} \right) + \left(\frac{\partial H_k}{\partial p_k} \right)^T \frac{dp_k}{dt} + \left(\frac{\partial H_k}{\partial \tilde{p}_k} \right)^T \frac{d\tilde{p}_k}{dt} + \left(\frac{\partial H_k}{\partial d_k} \right)^T \frac{dd_k}{dt} \right].$$

Suppose that, there exists a finite optimal value for step size, ξ_k^* , at the given operating condition, such that the Hamiltonian can be minimized, and convergence to the optimal value function can be guaranteed.

$$\xi_k^* = T \left[\left(\frac{\partial H_k}{\partial u_k} \right)^T \left(\frac{\partial H_k}{\partial u_k} \right) \right]^{-1} \left\{ - \left(\frac{\partial H_k}{\partial p_k} \right)^T \frac{dp_k}{dt} - \left(\frac{\partial H_k}{\partial \tilde{p}_k} \right)^T \frac{d\tilde{p}_k}{dt} - \left(\frac{\partial H_k}{\partial d_k} \right)^T \frac{dd_k}{dt} - \mu_1 H_k \right\}$$

where $\mu_1 > 0$. Since the computation of ξ_k^* involves the unknown parameters such as derivative of disturbance vector, and the gradient of Hamiltonian with respect to the disturbance vector, it is not possible to compute ξ_k^* from the above equation. Hence we are proposing an adaptive tuning algorithm for updating the step size for the given operating condition. For a given operating condition, the optimal value of step size, should be a fixed value. From the above equation, the optimal value of step size, ξ_k^* , is assumed to be slowly varying with respect to time. Hence it is reasonable to assume that $\hat{\xi}_k^* = 0$ at the given operating condition. The response of $\hat{\xi}_k$ shown in Section V confirms this, i.e., $\hat{\xi}_k$ is settling to different fixed values of ξ_k^* , at different operating conditions. If $\hat{\xi}_k$ represents the estimated value of the step size, the adaptation error can be defined as

$$\tilde{\xi}_k = \hat{\xi}_k - \xi_k^*. \tag{61}$$

The energy function can be modified as

$$\begin{split} V(H_k, \tilde{\xi}_k) &= \frac{1}{2} H_k^2 + \frac{1}{2} \Omega \tilde{\xi}_k^2, \quad \Omega > 0 \\ \dot{V}(H_k, \tilde{\xi}_k) &= H_k \left[\frac{1}{T} \hat{\xi}_k \left(\frac{\partial H_k}{\partial u_k} \right)^T \left(\frac{\partial H_k}{\partial u_k} \right) \\ &+ \left(\frac{\partial H_k}{\partial p_k} \right)^T \frac{dp_k}{dt} + \left(\frac{\partial H_k}{\partial \tilde{p}_k} \right)^T \frac{d\tilde{p}_k}{dt} \\ &+ \left(\frac{\partial H_k}{\partial d_k} \right)^T \frac{dd_k}{dt} \right] + \Omega \tilde{\xi}_k \dot{\tilde{\xi}}_k \\ &= H_k \left[\frac{1}{T} \hat{\xi}_k \left(\frac{\partial H_k}{\partial u_k} \right)^T \frac{\partial H_k}{\partial u_k} + \left(\frac{\partial H_k}{\partial u_k} \right)^T \frac{dp_k}{dt} \\ &+ \left(\frac{\partial H_k}{\partial \tilde{p}_k} \right)^T \frac{d\tilde{p}_k}{dt} + \left(\frac{\partial H_k}{\partial u_k} \right)^T \frac{\partial H_k}{du_k} \right] \\ &+ \Omega \tilde{\xi}_k \dot{\tilde{\xi}}_k \\ &= H_k \left[\frac{1}{T} \tilde{\xi}_k \left(\frac{\partial H_k}{\partial u_k} \right)^T \frac{\partial H_k}{\partial u_k} - \frac{\xi_k^*}{T} \left(\frac{\partial H_k}{\partial u_k} \right)^T \frac{\partial H_k}{dt} \right] \\ &+ \Omega \tilde{\xi}_k \dot{\tilde{\xi}}_k \\ &= H_k \left[\frac{1}{T} \tilde{\xi}_k \left(\frac{\partial H_k}{\partial u_k} \right)^T \frac{\partial H_k}{\partial u_k} + \left(\frac{\partial H_k}{\partial p_k} \right)^T \frac{dp_k}{dt} \\ &+ \left(\frac{\partial H_k}{\partial \tilde{p}_k} \right)^T \frac{d\tilde{p}_k}{dt} + \left(\frac{\partial H_k}{\partial u_k} \right)^T \frac{dd_k}{dt} \\ &+ \left(\frac{\partial H_k}{\partial \tilde{p}_k} \right)^T \frac{d\tilde{p}_k}{dt} + \left(\frac{\partial H_k}{\partial u_k} \right)^T \frac{dd_k}{dt} \\ &+ \frac{\xi_k^*}{T} \left(\frac{\partial H_k}{\partial u_k} \right)^T \frac{\partial H_k}{\partial u_k} \right] + \Omega \tilde{\xi}_k \dot{\tilde{\xi}}_k. \end{split}$$

Substituting for ξ_k^* , yields

$$\dot{V}(H_k, \tilde{\xi}_k) = \frac{\tilde{\xi}_k}{T} H_k \left(\frac{\partial H_k}{\partial u_k}\right)^T \left(\frac{\partial H_k}{\partial u_k}\right) - \mu_1 H_k^2 + \Omega \tilde{\xi}_k \dot{\hat{\xi}}_k.$$

If the update law is chosen such that

$$\dot{\hat{\xi}}_{k} = -\frac{1}{\Omega T} H_{k} \left(\frac{\partial H_{k}}{\partial u_{k}} \right)^{T} \left(\frac{\partial H_{k}}{\partial u_{k}} \right)$$
(62)

 $\dot{V}(H_k, \tilde{\xi}_k) = -\mu_1 H_k^2$. Hence $\dot{V}(H_k, \tilde{\xi}_k)$ is negative semidefinite, this implies that H_k and $\tilde{\xi}_k$ are bounded. Taking the derivative of $\dot{V}(H_k, \tilde{\xi}_k)$

$$\begin{split} \ddot{V}(H_k, \tilde{\xi}_k) &= -\mu_1 H_k \left[\frac{1}{T} \hat{\xi}_k \left(\frac{\partial H_k}{\partial u_k} \right)^T \left(\frac{\partial H_k}{\partial u_k} \right) \right. \\ &+ \left(\frac{\partial H_k}{\partial p_k} \right)^T \frac{dp_k}{dt} + \left(\frac{\partial H_k}{\partial \tilde{p}_k} \right)^T \frac{d\tilde{p}_k}{dt} \\ &+ \left(\frac{\partial H_k}{\partial d_k} \right)^T \frac{dd_k}{dt} \right] + \Omega \tilde{\xi}_k \dot{\tilde{\xi}}_k. \end{split}$$

Replacing $\hat{\xi}_k$ by $(\tilde{\xi}_k + \xi_k^*)$, and substituting for ξ_k^* , yields

$$\ddot{V}(H_k, \tilde{\xi}_k) = -\mu_1 H_k \left(\frac{\partial H_k}{\partial u_k}\right)^T \left(\frac{\partial H_k}{\partial u_k}\right) \tilde{\xi}_k + T\mu_1^2 H_k^2.$$

The optimal value of gain, ξ_k^* is a bounded value. Since $\tilde{\xi}_k$ is bounded, $\hat{\xi}_k$ is also bounded. Since u is assumed to be Lebesgue measurable, bounded and Lipschitz, from (60), we can find that $\frac{\partial H_k}{\partial u_k}$ is bounded. Since, we have already proved that H_k is bounded, from the above equation, it is clear that $\ddot{V}(H_k, \tilde{\xi}_k)$ is bounded. Hence, $\dot{V}(H_k, \tilde{\xi}_k)$ is uniformly continuous, and by Barbalat's Lemma [32], it has been found that $H_k, \tilde{\xi}_k \to 0$ as $t \to \infty$.

The complete NMPC algorithm [16], [17], with the real time finite horizon optimization technique, is provided in the next section.

IV. TRACKING CONTROL OF SFF USING THE PROPOSED SCHEME

A schematic of the Earth orbiting SFF system in a leaderfollower based framework is shown in Fig. 2.



Fig. 2. Schematic representation of the satellite formation flying system.

The chief (leader) is assumed to be in circular orbit. The equations of motion of the deputy (follower) spacecraft [2], [36] are defined in local-vertical-local-horizontal (LVLH) frame, fixed at the center of the chief. Let $r = [r, 0, 0]^T$ denotes the position vector of the chief in Earth centered inertial reference frame, and $\rho = [x, y, z]^T$ denotes the relative position vector of the deputy with respect to the chief in LVLH frame. The nonlinear SFF model can be written as follows:

$$\ddot{r} - r\dot{\theta}^2 + \frac{\mu}{r^2} = 0$$

$$\ddot{r}\ddot{\theta} + 2\dot{\theta}\dot{r} = 0$$

$$\ddot{x} - 2\dot{\theta}\dot{y} - \ddot{\theta}y - \dot{\theta}^2x + \frac{\mu(r+x)}{l^3} - \frac{\mu}{r^2} = u_x + d_x \quad (63)$$

$$\ddot{y} + 2\dot{\theta}\dot{x} + \ddot{\theta}x - \dot{\theta}^2y + \frac{\mu y}{l^3} = u_y + d_y$$

$$\ddot{z} + \frac{\mu z}{l^3} = u_z + d_z$$

where $l = \sqrt{(r+x)^2 + y^2 + z^2}$; $d = [d_x, d_y, d_z]^T$, and $u = [u_x, u_y, u_z]^T$, represent the bounded external disturbance and the thrust acceleration (control input) respectively; μ denotes the gravitational constant of the Earth; and $\dot{\theta}$ is the angular velocity of the circular orbit. The reference trajectories for the relative position and velocity of the deputy with respect to chief, have been derived based on Hill's linearized equations of relative unperturbed motion [2], [37], and that is having the following form:

$$\ddot{x} - 2\omega \dot{y} - 3\omega^2 x = 0 \tag{64}$$

$$\ddot{u} + 2\omega \dot{x} = 0 \tag{65}$$

$$\ddot{z} + \omega^2 z = 0 \tag{66}$$

where $\omega = \dot{\theta} = \sqrt{\frac{\mu}{r^3}}$. If $p = [x, \dot{x}, y, \dot{y}, z, \dot{z}]^T$, the dynamics given by (64)–(66) can be expressed as $\dot{p} = Ap$, where

$$A = \begin{pmatrix} 0 & 1 & 0 & 0 & 0 & 0 \\ 3\omega^2 & 0 & 0 & 2\omega & 0 & 0 \\ 0 & 0 & 0 & 1 & 0 & 0 \\ 0 & -2\omega & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 & 1 \\ 0 & 0 & 0 & 0 & -\omega^2 & 0 \end{pmatrix}.$$

The reference trajectory is chosen such that it satisfies:

$$\dot{p_d} = Ap_d \tag{67}$$

where p_d represents the desired trajectory. Our objective is to design an optimal control law for controlling the relative positions and velocities of the deputy with respect to chief. It is assumed that deputy has the onboard arrangements to acquire the information regarding its relative position with respect to chief. For the system model given by (63),

$$M(p) = \begin{pmatrix} 1 & 0 & 0\\ 0 & 1 & 0\\ 0 & 0 & 1 \end{pmatrix}$$
(68)

is nonsingular; hence, the relative degree of the system is $\{2, 2, 2\}$. The distribution, $\Lambda = span\{w_1, w_2, w_3\}$, is involutive. Therefore, the system is locally detectable in the domain Υ . The structure of the adaptive gain HOSM can be defined as follows:

$$\begin{aligned} \dot{\hat{\gamma}}_{1}^{i} &= -\alpha_{1}^{i} \delta_{i}^{\frac{1}{3}} |\hat{\gamma}_{1}^{i} - Y_{i}|^{\frac{2}{3}} \operatorname{sign}(\hat{\gamma}_{1}^{i} - Y_{i}) + \hat{\gamma}_{2}^{i} \\ \dot{\hat{\gamma}}_{2}^{i} &= \hat{F}^{i} + B^{i} u \\ \hat{F}^{i} &= -\alpha_{2}^{i} \delta_{i}^{\frac{1}{2}} |\hat{\gamma}_{2}^{i} - \dot{\hat{\gamma}}_{1}^{i}|^{\frac{1}{2}} \operatorname{sign}(\hat{\gamma}_{2}^{i} - \dot{\hat{\gamma}}_{1}^{i}) + \hat{\gamma}_{3}^{i} \\ \dot{\hat{\gamma}}_{3}^{i} &= -\alpha_{3}^{i} \delta_{i} \operatorname{sign}(\hat{\gamma}_{3}^{i} - \hat{F}^{i}) \end{aligned}$$
(69)

where $Y = [x, y, z]^T$. Using Theorem 2, the disturbance estimates can be obtained as

$$\begin{aligned} \hat{d}_x &= \hat{\gamma}_3^1 - \sqrt{\frac{4\mu}{r^3}} \hat{\gamma}_2^2 - \frac{\mu \hat{\gamma}_1^1}{r^3} \\ &+ \frac{\mu (\hat{\gamma}_1^1 + r)}{((\hat{\gamma}_1^1 + r)^2 + (\hat{\gamma}_1^2)^2 + (\hat{\gamma}_1^3)^2)^{\frac{3}{2}}} - \frac{\mu}{r^2} \\ \hat{d}_y &= \hat{\gamma}_3^2 + \sqrt{\frac{4\mu}{r^3}} \hat{\gamma}_2^1 - \frac{\mu \hat{\gamma}_1^2}{r^3} + \frac{\mu \hat{\gamma}_1^2}{((\hat{\gamma}_1^1 + r)^2 + (\hat{\gamma}_1^2)^2 + (\hat{\gamma}_1^3)^2)^{\frac{3}{2}}} \\ \hat{d}_z &= \hat{\gamma}_3^3 + \frac{\mu \hat{\gamma}_1^3}{((\hat{\gamma}_1^1 + r)^2 + (\hat{\gamma}_1^2)^2 + (\hat{\gamma}_1^3)^2)^{\frac{3}{2}}}. \end{aligned}$$
(70)

The velocity estimates can be obtained as $\dot{x} = \hat{\gamma}_2^1$, $\dot{y} = \hat{\gamma}_2^2$ and $\dot{z} = \hat{\gamma}_2^3$. These estimates are utilized by the prediction model in the NMPC. Using the alternate observer structure given by (46), the disturbances can be estimated as follows:

$$\begin{aligned} \hat{d}_x &= \hat{\gamma}_3^1 - \sqrt{\frac{4\mu}{r^3}} \hat{\gamma}_2^2 - \frac{\mu \hat{\gamma}_1^1}{r^3} \\ &+ \frac{\mu (\hat{\gamma}_1^1 + r)}{((\hat{\gamma}_1^1 + r)^2 + (\hat{\gamma}_1^2)^2 + (\hat{\gamma}_1^3)^2)^{\frac{3}{2}}} - \frac{\mu}{r^2} - u_x \\ \hat{d}_y &= \hat{\gamma}_3^2 + \sqrt{\frac{4\mu}{r^3}} \hat{\gamma}_2^1 - \frac{\mu \hat{\gamma}_1^2}{r^3} \\ &+ \frac{\mu \hat{\gamma}_1^2}{((\hat{\gamma}_1^1 + r)^2 + (\hat{\gamma}_1^2)^2 + (\hat{\gamma}_1^3)^2)^{\frac{3}{2}}} - u_y \\ \hat{d}_z &= \hat{\gamma}_3^3 + \frac{\mu \hat{\gamma}_1^3}{((\hat{\gamma}_1^1 + r)^2 + (\hat{\gamma}_1^2)^2 + (\hat{\gamma}_1^3)^2)^{\frac{3}{2}}} - u_z. \end{aligned}$$
(71)

The complete tracking scheme is detailed in Algorithm 1. The simulation results, given in the next section, confirm the efficacy of the proposed scheme.

V. RESULTS AND DISCUSSION

The simulation studies have been done for a two-spacecraft formation flying model given by (63). We have considered a projected circular formation, in which the chief is moving in a circular orbit with an angular velocity of $\omega = \sqrt{\frac{\mu}{r^3}}$. The reference trajectory for relative position [2] has been taken as, $\rho_d(t)$ $= (\frac{r_f}{2} \sin(\omega t + \phi), r_f \cos(\omega t + \phi), r_f \sin(\omega t + \phi))^T$, where ϕ is the in-plane phase angle between the chief and deputy, and r_f is the formation size. The simulation parameters are shown in Table I. The reference trajectories have been chosen such that, they satisfy (67). For improving the computational efficiency, the length of the control horizon, N, is taken as 3. To validate the proposed strategy, simulations are done for different levels and types of perturbations. Random noise has been added to the output measurements in all the cases.

Random intermediate perturbations are applied, with an initial offset of $(\boldsymbol{\rho} - \boldsymbol{\rho}_d) = [400, 400, 300]^T$ m, and $(\dot{\boldsymbol{\rho}} - \dot{\boldsymbol{\rho}}_d) = [0, 0, 0]^T$ m. The Lipschitz constants are chosen as $\delta_1 = 1.7 \times 10^{-4}$, $\delta_2 = 8 \times 10^{-4}$ and $\delta_3 = 6 \times 10^{-6}$ respectively. The values of $\boldsymbol{\beta}$ are chosen as $[\beta_1^1, \beta_2^1, \beta_3^1]^T = [0.1, 0.1, 5E-5]^T$, $[\beta_1^2, \beta_2^2, \beta_3^2]^T = [5E-3, 1E-2, 2E-4]^T$, and $[\beta_1^3, \beta_2^3, \beta_3^3]^T = [6E-3, 3E-2, 1E-6]^T$ respectively. The value of Ω used in the update law for step size is taken as 0.1. The control constraint, u_{max}^i , is chosen as 80 mm/s^2 , $\forall i$, since we have considered a

Algorithm 1. NMPC algorithm

- 1: Initialize the control sequence $u_0, u_1, \ldots, u_{N-1}$; obtain the position measurements.
- 2: Estimate the velocities and disturbance inputs using HOSM observer, given by (4). The observer gains can be updated using the adaptive tuning rules given in (5)-(8). The velocity estimates can be obtained as $\dot{x} = \hat{\gamma}_2^1$, $\dot{y} = \hat{\gamma}_2^2$, and $\dot{z} =$ $\hat{\gamma}_2^3$, respectively, and the disturbance estimates can be computed using (70).

3: while
$$|\triangle J| > \epsilon$$
 do

4:	for	k	~	0,	N	_	1	do
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5:	Obtain p_{k+1} and d_{k+1}
6:	end for
7:	for $k \leftarrow N, 1$ do
8:	Find λ_k using (54) and (55)
9:	end for
10:	for $k \leftarrow 0, N-1$ do
11:	Compute $\frac{\partial H_k}{\partial u_k}$ using (56)
12:	end for
13:	for $k \leftarrow 0, N-1$ do
14:	Update ξ_k using the adaptive tuning law given in (62)
15:	end for

- Determine $\triangle J = J^i J^{j-1}$ 16:
- if $\triangle J < 0$ then 17:
- Update the control sequence $u_0, u_1, \ldots, u_{N-1}$ using (57) 18: 19: else
- 20:
- Continue 21: end if
- 22: J + +
- end while 23:
- 24: Apply u_0 to the system and reinitialize the control sequence as $u_0, u_1, \ldots, u_{N-1}$ for improving the computational performance [17].

TABLE I SIMULATION PARAMETERS

Parameters	Values			
The Earth's gravitational constant	$\mu = 3.985 \times 10^{14} {\rm m}^3/{\rm s}^2$			
Orbital radius of the circular reference orbit	6878.137 km			
Formation size	$r_f = 1 \mathrm{km}$			
In-plane phase angle between chief and deputy	$\phi = 0$			
ϵ	1E-4			

very high initial distance separation of 640.3 m, for the desired formation size of 1000 m. All the observer gains are initialized as zero. The other parameters such as R = (1E-3)I(3), and S = (2E - 2)I(6), are chosen heuristically, where I is identity matrix. The phase portrait given by Fig. 3, and the relative position trajectories of the deputy given by Fig. 4, confirm the robustness of the proposed approach. A projected circular formation [2] is maintained along the y-z plane. Fig. 5 shows the applied intermediate disturbances and the reconstructed signal using the proposed HOSM observer. Different types of perturbations have been applied along x, y and z directions. The perturbing accelerations are expressed as functions of angular velocity of the circular reference orbit (ω). The HOSM disturbance observer is found to be capable of rebuilding even the worst case disturbances. The estimated relative velocity trajectories, and the corresponding error histories are shown in Figs. 6 and 7 respectively. The rotated magnified version of the initial region of tracking result is shown in cubical box. Irrespective of the initial values, the estimated trajectories are found to be converging to the actual one within $100 \, \text{s}$. To show the convergence speed, the velocity error histories are provided in Fig. 8.



Fig. 3. Phase portrait: (a) X - Y - Z coordinates, (b) X - Y coordinates, (c) Y - YZ coordinates and, (d) X-Z coordinates.



Fig. 4. Relative position trajectories of the deputy.



Fig. 5. Disturbance trajectories of the deputy.



Fig. 6. Estimated relative velocity trajectories of the deputy.



Fig. 7. Relative velocity estimation error trajectories of the deputy.



Fig. 8. Relative velocity error trajectories of the deputy.

The control histories given by Fig. 9, show that the required thrust acceleration inputs are staying within the available control authority. The maximum control effort is found to be less than 60 mm/s², even though the relative position has been perturbed by 64 %, when compared to the required formation size. The control accuracy can be further improved by increasing the length of control horizon, but it can affect the computational performance. The computation time is a key factor in determining the performance of an optimal control technique. Fig. 10 shows the computational effort per sampling instant, and the average computational effort per choice of ϵ can also considerably save the computation time. Fig. 11 (a) provides the cost function trajectory and Fig. 11 (b) shows the net distance separation error.

In order to show the tuning drift, the time responses of the various adapted parameters such as observer gains and step size are presented in Fig. 12. Fig. 13 verifies the input const-



Fig. 9. Control inputs to the deputy.



Fig. 10. Computation time.



Fig. 11. (a) Cost function trajectory and (b) Net relative distance error.

raint handling capability of the proposed NMPC scheme. The simulations are repeated for different values of u_{max} , with the same level of initial and intermediate disturbances applied as in the previous case. Even though the trajectories are found to be converging with a reasonable speed in all the cases, comparatively more smooth and fast convergence is observed for $u_{\text{max}} = 70 \text{ mm/s}^2$. From Fig. 14, it is clear that the robustness and convergence speed of proposed adaptive gain HOSM are preserved in all the cases, where different initial estimates are chosen in different cases. The performance of the proposed observer, in case of unexpected state jerks, can be validated using Fig. 15. The magnified versions of the respective regions of interest confirm the efficiency of the proposed adaptive observer, in dealing with unpredicted state jerks. In order to show the step size adaptation, simulations are repeated for different initial distance separations, and the corresponding tuning responses are provided in Fig. 16. From this, it is clear that a heuristic choice of step size may not be appropriate at all operating conditions.

The estimation responses show that the proposed HOSM will work satisfactorily for locally Lipschitz case as well, provided the Lipschitz nonlinearity is bounded. Fig. 17 shows the disturbance estimates obtained using the alternate HOSM



Fig. 12. Tuning response: (a) gain α^1 , (b) gain α^2 , (c) gain α^3 and, (d) step size ξ_k .



Fig. 13. Control histories and net distance separation error for different input constraints (a) and (b) $u_{\text{max}} = 20 \text{ mm/s}^2$, (c) and (d) $u_{\text{max}} = 50 \text{ mm/s}^2$, (e) and (f) $u_{\text{max}} = 70 \text{ mm/s}^2$.



Fig. 14. Relative velocity trajectories for different initial estimates.

observer structure, given by (47). The same levels of random perturbations as in the previous case, have been applied. The results are compared with that of the estimates obtained in Fig. 5. The performance of this observer is found to deteriorate, for very high levels of initial perturbations. This reaffirms the efficacy of the proposed observer structure.

We have compared the results with that of two recent relevant literature, [26] and [29]. In [26], an HOSM based disturbance observer has been used, for improving the distur-



Fig. 15. Relative velocity estimation trajectories in the case of induced state jerks.



Fig. 16. Tuning response of ξ_k for different initial conditions.



Fig. 17. Disturbance trajectories using alternate HOSM differentiator structure.

bance rejection capability of continuous sliding mode controller. There are only slight differences in the differentiator structure, when compared to that of the proposed one. The structure is given by

$$\begin{split} \hat{\Gamma}_{1}^{i} &= \hat{F}^{i} + B^{i}u \\ \hat{F}^{i} &= -\kappa_{1}^{i}\ell_{i}^{\frac{1}{e_{i}+1}} |\hat{\Gamma}_{1}^{i} - \gamma_{e_{i}}^{i}|^{\frac{e_{i}}{e_{i}+1}} \mathrm{sign}(\hat{\Gamma}_{1}^{i} - \gamma_{e_{i}}^{i}) + \hat{\Gamma}_{2}^{i} \\ \hat{\Gamma}_{2}^{i} &= -\kappa_{2}^{i}\ell_{i}^{\frac{1}{e_{i}}} |\hat{\Gamma}_{2}^{i} - \dot{\hat{\Gamma}}_{1}^{i}|^{\frac{e_{i}-1}{e_{i}}} \mathrm{sign}(\hat{\Gamma}_{2}^{i} - \dot{\hat{\Gamma}}_{1}^{i}) + \hat{\Gamma}_{3}^{i} \\ &\vdots \end{split}$$

$$\dot{\hat{\Gamma}}_{e_{i}}^{i} = -\kappa_{e_{i}}^{i} \ell_{i}^{\frac{1}{2}} |\hat{\Gamma}_{e_{i}}^{i} - \dot{\hat{\Gamma}}_{e_{i}-1}^{i}|^{\frac{1}{2}} \operatorname{sign}(\hat{\Gamma}_{e_{i}}^{i} - \dot{\hat{\Gamma}}_{e_{i}-1}^{i}) + \hat{\Gamma}_{e_{i}+1}^{i} \\ \dot{\hat{\Gamma}}_{e_{i}+1}^{i} = -\kappa_{e_{i}+1}^{i} \ell_{i} \operatorname{sign}(\hat{\Gamma}_{e_{i}+1}^{i} - \dot{\hat{\Gamma}}_{e_{i}}^{i}).$$
(72)

This HOSM based observer, when it is applied to SFF dynamics, utilizes velocity measurements to obtain the disturbance estimates. $\hat{\Gamma}_2^i$ directly gives the disturbance estimate. We have combined this fixed gain disturbance observer with the proposed NMPC scheme. When the initial estimates are chosen in the neighborhood of the actual value, the results are somewhat satisfactory. But, when those values are perturbed as in the case of previous simulations, the results are not at all satisfactory, and the corresponding trajectories are shown in Fig. 18. Since we have retained the same constraint value for the control input, the control requirement remains same. But the speed of convergence is less.



Fig. 18. Disturbance estimates with the estimation scheme proposed in [29].

We have also simulated and compared the spacecraft formation flying scheme, which is proposed in [29]. In this paper, an HOSM based observer is used for the feedforward compensation of a nonsingular terminal sliding mode controller. The results are given in Figs. 19-21. For the ease of design, we can reformulate the dynamics given by (63), in the form, $\ddot{\rho} + f(\rho, \dot{\rho}) = u + d$. In this case, there is a slight difference in the differentiator structure, when compared to (72), where $\hat{\Gamma}_1^i = \hat{F}^i - f^i(\rho, \dot{\rho}) - \ddot{\rho}_d^i + B^i u$, $\hat{F}^i = -\kappa_1^i \ell_i^{\frac{1}{e_i+1}} |\hat{\Gamma}_1^i - \dot{e}^i|^{\frac{e_i}{e_i+1}} \operatorname{sign}(\hat{\Gamma}_1^i - \dot{e}^i) + \hat{\Gamma}_2^i.$ Instead of velocity, it uses the tracking error to estimate the unknown disturbances. $\hat{\Gamma}_2^i$ directly gives the disturbance estimate. The nonsingular terminal sliding surface is chosen as S = e+ $\kappa \operatorname{sig}^{\eta}(\dot{e})$, where $e = \rho - \rho_d$; $\dot{e} = \dot{\rho} - \dot{\rho}_d$; $\operatorname{sig}^{\eta}(\dot{e}) =$ $[|\dot{e}_1|^{\eta} \operatorname{sign}(\dot{e}_1), |\dot{e}_2|^{\eta} \operatorname{sign}(\dot{e}_2), \dots, |\dot{e}_n|^{\eta} \operatorname{sign}(\dot{e}_n)]^T; 1 < \eta < 2,$ $\kappa = \text{diag}\{\kappa_1, \kappa_2, \dots, \kappa_n\}; \text{ and } \kappa_i > 0, \forall i. \text{ The control input}$ can be designed as $u = [\ddot{\rho}_d - (1/\kappa\eta) \operatorname{sig}^{2-\eta}(\dot{e}) + f(\rho,\dot{\rho}) - \dot{\rho}_d$ $K \tanh(S) - \hat{d}$, where \hat{d} is the disturbance estimate. The observer structure is more or less model dependent. The results show that the control requirement is very high, and the trajectory convergence speed is not even comparable with the proposed scheme. It is found to be more than 3500 s. The magnified regions indicate that the disturbance estimation error is also high.



Fig. 19. Disturbance estimation error with the estimation scheme proposed in [29].



Fig. 20. Control inputs with the control scheme proposed in [29].



Fig. 21. Relative positions with the control scheme proposed in [29].

From the results, it is clear that, the robustness of the NMPC has been enhanced considerably with the HOSM based disturbance observer. The prescribed structure is found to have good estimation accuracy. The adaptation of the observer gains give enough flexibility in the choice of initial values of estimates. Moreover, it is able to handle unexpected state jerks. The constraint handling has been made easy, and thereby, complexity of the optimization scheme has been reduced, with the choice of nonquadratic cost function, in the NMPC scheme. The adaptation in step size has further improved the efficiency of the optimization algorithm.

VI. CONCLUSION

This paper has focused on the tracking control of spacecraft formation flying system consisting of two LEO satellites, in a leader-follower based framework. In this work, we have applied the HOSM differentiator structure, containing the Lipschitz constant and Lebesgue measurable control input [19], [25], to a complex multi-input multi-output SFF system, to estimate the unmeasurable disturbances and velocities from the noisy position measurements. The observer gains are updated using adaptive tuning algorithms, derived based on Lyapunov stability theory. The prescribed differentiator structure has good estimation accuracy [19]. The system is transformed into canonical form using nonlinear coordinate transformation [24]. SFF dynamics, having a total vector relative degree equal to n, has been considered. The observer design is repeated with an alternate HOSM differentiator structure as well for comparative study.

A robust NMPC technique with a nonquadratic cost function [27] has been considered for incorporating the control constraints, which has reduced the additional burden of constraint handling in real time optimization algorithms. The heuristic choice of step size in optimization algorithm is not recommendable at all operating conditions. Hence, an adaptive tuning algorithm has been derived for updating the step size. The estimates from the HOSM disturbance observer has been utilized by the output prediction model in NMPC. For simulation studies, the nonlinear SFF model defined in leader fixed Euler-Hill frame, has been considered, and the reference trajectories are generated using Hill-Clohessy-Wiltshire equations of unperturbed motion. Simulation studies are conducted for different levels of perturbations, along with added random measurement noise. The feasibility of the algorithm based on NMPC, combined with the adaptive gain HOSM based disturbance observer and a state estimator, has been evaluated in sufficient detail, and the following conclusions are drawn:

1) The prescribed adaptive gain HOSM differentiator is found to have good estimation accuracy. It fits well as state as well as disturbance estimator for such complex systems like SFF system.

2) The performance of the adaptive gain HOSM is less affected by the extremely high level of initial perturbations applied.

3) The adaptation in observer gain has added more flexibility to the choice of initial estimates. Moreover, the observer is found to be robust to unexpected state jerks.

4) The results indicate that the proposed scheme will work satisfactorily for locally Lipschitz systems with bounded Lipschitz nonlinearity as well.

5) NMPC technique, with the real time open loop finite horizon optimization algorithm is found to be computationally efficient and is best suited for satellite formation control applications.

6) The non-quadratic performance functional has considerably reduced the burden of input constraint handling.

7) The adaptation in step size, is found to be an attractive alternative to the heuristic choice of it, well suited for systems with diverse operating conditions, which may also improve the convergence speed.

8) The inclusion of disturbance estimate in the model prediction is found to add more robustness to the scheme, therefore, eliminates the necessity of exact modeling.

The inclusion of actuator dynamics, can be viewed as a further extension of this work.

REFERENCES

- D. P. Scharf, F. Y. Hadaegh, and S. R. Ploen, "A survey of spacecraft formation flying guidance and control. Part II: control," in *Proc. 2004 American Control Conf.*, Boston, Massachusetts, 2004, pp. 2976–2985.
- [2] X. Liu and K. D. Kumar, "Network-based tracking control of spacecraft formation flying with communication delays," *IEEE Trans. Aerosp. Electronic Syst.*, vol. 48, no. 3, pp. 2302–2314, Jul. 2012.
- [3] L. Hui and J. F. Li, "Terminal sliding mode control for spacecraft formation flying," *IEEE Trans. Aerosp. Electronic Syst.* vol. 45, no. 3, pp. 835–846, Jul. 2009.
- [4] R. R. Nair and L. Behera, "Swarm aggregation using artificial potential field and fuzzy sliding mode control with adaptive tuning technique," in *Proc. 2012 American Control Conf.*, Montreal, Canada, 2012, pp. 6184 –6189.
- [5] T. E. Massey and Y. B. Shtessel, "Continuous traditional and highorder sliding modes for satellite formation control," J. Guid. Control Dynamics, vol. 28, no. 4, pp. 826–831, Jul–Aug. 2005.
- [6] R. R. Nair, L. Behera, V. Kumar, and M. Jamshidi, "Multisatellite formation control for remote sensing applications using artificial potential field and adaptive fuzzy sliding mode control," *IEEE Syst. J.*, vol. 9, no. 2, pp. 508–518, Jun. 2015.
- [7] M. S. de Queiroz, V. Kapila, and Q. G. Yan, "Adaptive nonlinear control of multiple spacecraft formation flying," *J. Guid. Control Dynamics*, vol. 23, no. 3, pp. 385–390, May 2000.
- [8] X. W. Dong, B. C. Yu, Z. Y. Shi, and Y. S. Zhong, "Time-varying formation control for unmanned aerial vehicles: theories and applications," *IEEE Trans. Control Syst. Technol.*, vol. 23, no. 1, pp. 340–348, Jan. 2015.
- [9] X. W. Dong, Y. Zhou, Z. Ren, and Y. S. Zhong, "Time-varying formation control for unmanned aerial vehicles with switching interaction topologies," *Control Eng. Pract.*, vol. 46, pp. 26–36, Jan. 2016.
- [10] L. Ma, H. B. Min, S. C. Wang, Y. Liu, and S. Y. Liao, "An overview of research in distributed attitude coordination control," *IEEE/CAA J. of Autom. Sinica*, vol. 2, no. 2, pp. 121–133, Apr. 2015.
- [11] Y. Ulybyshev, "Long-term formation keeping of satellite constellation using linear-quadratic controller," J. Guid. Control Dynamics, vol. 21, no. 1, pp. 109–115, Jan.–Feb. 1998.
- [12] S. B. McCamish, M. Romano, and X. P. Yun, "Autonomous distributed control of simultaneous multiple spacecraft proximity maneuvers," *IEEE Trans. Automat. Sci. Eng.*, vol. 7, no. 3, pp. 630–644, Jul. 2010.
- [13] E. Camponogara and H. F. Scherer, "Distributed optimization for model predictive control of linear dynamic networks with control-input and output constraints," *IEEE Trans. Automat. Sci. Eng.*, vol. 8, no. 1, pp. 233–242, Jan. 2011.
- [14] X. H. Xia and J. F. Zhang, "Operation efficiency optimisation modelling and application of model predictive control," *IEEE/CAA J. of Autom. Sinica*, vol. 2, no. 2, pp. 166–172, Apr. 2015.
- [15] L. Breger, J. P. How, and A. Richards, "Model predictive control of spacecraft formations with sensing noise," in *Proc. 2005 American Control Conf.*, Portland, OR, USA, 2005, pp. 2385–2390.
- [16] G. J. Sutton and R. R. Bitmead, "Performance and computational implementation of nonlinear model predictive control on a submarine," in *Nonlinear Model Predictive Control*, F. Allgöower and A. Zheng, Eds., Birkhäuser Basel, 2000, pp. 461–471.
- [17] J. Shin and H. J. Kim, "Nonlinear model predictive formation flight," *IEEE Trans. Syst. Man Cybernet.- A: Syst. Humans*, vol. 39, no. 5, pp. 1116–1125, Sep. 2009.
- [18] K. R. Muske and T. A. Badgwell, "Disturbance modeling for offset-free linear model predictive control," *J. Process Control*, vol. 12, no. 5, pp. 617–632, Aug. 2002.

- [19] A. Levant, "Higher-order sliding modes, differentiation and outputfeedback control," *Int. J. Control*, vol. 76, no. 9–10, pp. 924–941, Sept. 2003.
- [20] A. Benallegue, A. Mokhtari, and L. Fridman, "High-order sliding-mode observer for a quadrotor UAV," *Int. J. Robust Nonlin. Control*, vol. 18, no. 4–5, pp. 427–440, Mar. 2008.
- [21] A. Levant, "Robust exact differentiation via sliding mode technique," *Automatica*, vol. 34, no. 3, pp. 379–384, Mar. 1998.
- [22] R. Sharma and M. Aldeen, "Fault and disturbance reconstruction in non-linear systems using a network of interconnected sliding mode observers," *IET Control Theory Appl.*, vol. 5, no. 6, pp. 751–763, Apr. 2011.
- [23] P. P. Menon and C. Edwards, "An observer based distributed controller for formation flying of satellites," in *Proc. 2011 American Control Conf.*, San Francisco, CA, USA, 2011, pp. 196–201.
- [24] L. Fridman, Y. Shtessel, C. Edwards, and X.-G. Yan, "Higher-order sliding-mode observer for state estimation and input reconstruction in nonlinear systems," *Int. J. Robust Nonlin. Control*, vol. 18, no. 4–5, pp. 399–412, Mar. 2008.
- [25] S. Iqbal, C. Edwards, and A. I. Bhatti, "Robust feedback linearization using higher order sliding mode observer," in *Proc. 50th IEEE Conf. Decision and Control and European Control Conf. (CDC-ECC)*, Orlando, FL, USA, 2011, pp. 7968–7973.
- [26] C. X. Mu, Q. Zong, B. L. Tian, and W. Xu, "Continuous sliding mode controller with disturbance observer for hypersonic vehicles," *IEEE/CAA J. of Autom. Sinica*, vol. 2, no. 1, pp. 45–55, Jan. 2015.
- [27] A. Heydari and S. N. Balakrishnan, "Finite-horizon control-constrained nonlinear optimal control using single network adaptive critics," *IEEE Trans. Neural Netw. Learning Syst.*, vol. 24, no. 1, pp. 145–157, Jan. 2013.
- [28] N. Boizot, E. Busvelle, and J.-P. Gauthier, "An adaptive high-gain observer for nonlinear systems," *Automatica*, vol. 46, no. 9, pp. 1483– 1488, Sep. 2010.
- [29] Q. X. Lan, J. Yang, S. H. Li, and H. B. Sun, "Finite-time control for 6dof spacecraft formation flying systems," *J. Aerosp. Eng.*, vol. 28, no. 5, Article no. 04014137, Sep. 2015.
- [30] A. Isidori, Nonlinear Control Systems (Third edition). Springer: Berlin, 1995.
- [31] A. F. Filippov, *Differential Equations with Discontinuous Righthand Sides*. Dordrecht: Kluwer Academic, 1988.
- [32] J.-J. Slotine and W. P. Li, *Applied Nonlinear Control.* Englewood Cliffs, NJ: Prentice-Hall, 1991.
- [33] S. P. Bhat and D. S. Bernstein, "Geometric homogeneity with applications to finite-time stability," *Math. Control Signals Syst.*, vol. 17, no. 2, pp. 101–127, Jun. 2005.

- [34] A. Levant, "Homogeneity approach to high-order sliding mode design," *Automatica*, vol. 41, no. 5, pp. 823–830, May 2005.
- [35] Y. Orlov, "Finite time stability and robust control synthesis of uncertain switched systems," SIAM J. Control Optim., vol. 43, no. 4, pp. 1253– 1271, Apr. 2004.
- [36] K. T. Alfriend, S. R. Vadali, P. Gurfil, J. P. How, and L. S. Breger, *Spacecraft Formation Flying: Dynamics, Control and Navigation*. Amsterdam: Elsevier, 2010.
- [37] D. Vallado, Fundamentals of Astrodynamics and Applications (Second edition). New York: Springer, 2001.



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