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## RESEARCH ARTICLE

# Analysis and Suppression of Oscillation in Traction Network-EMUs Coupling System Under the Condition of Rising Pantograph

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**ABSTRACT** Low frequency oscillation (LFO) incidents may occur in coupled system composed of electric multiple units (EMUs) and traction network, which threaten the safety and stability of EMUs and traction power supply system. In order to solve this problem, a virtual impedance with band-pass filter link control strategy is proposed for the four quadrant converters (4QCs) of EMU. First, the impedance models of coupled system under the background of LFO are established based on the impedance matching. Next, this paper defines the worst scenario for coupled system, combined with the Nyquist curve and forbidden region criterion (FRC). The impact of the grid-side and EMU-side circuit parameters on system stability is analyzed. The proposed control strategy adjusts the input impedance of the EMUs equivalently by virtual impedance to meet FRC. Band-pass filter (BPF) is used to select a specific band to apply virtual impedance. This avoids passive impact on the dynamic performance of other frequency band. The parameters of the virtual impedance and BPF are designed according to the FRC and engineering requirements. The simulation results verify the effectiveness of the proposed control strategy for oscillation suppression. Finally, the coupled system model is built on the basis of hardware-in-the-loop experimental platform. The experimental results show that the proposed method can be used to suppress oscillations.

**INDEX TERMS** Long series EMUs-traction network system, impedance matching, forbidden region criterion, band-pass filter, virtual impedance control.

## I. INTRODUCTION

In recent years, rail transit has developed significantly in China. EMU is becoming the preferred mode of transportation for the public. Chinese standard EMU trains named “Fuxing” represents a spectrum of EMUs. They will develop in the direction of lightweight, main and auxiliary integration [1]. Source-network-train-storage integrated system has been proposed for electric railways. Among the integrated system, it is necessary to ensure the stability of traction network-EMUs coupling system. It is the most basic and important component of the traction power supply system

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(TPSS) [2]. However, it has been recorded that LFO accidents have occurred many times in the coupled system [3]. LFO seriously threatens the normal railway operation order and poses threat to the stability of the EMUs.

The traction network-EMUs coupling system can be regarded as a cascaded system composed of weak grid and converters [4]. Scholars have adopted different methods to analyze system stability, such as impedance analysis method in frequency-domain [5] and eigenvalue analysis method in time-domain [6]-[7]. For the impact factors, [8] proposes that it is more likely to occur in the state of rising pantograph than the state of fully operating; In [9], the influence of traction network parameters on the system stability are analyzed. When the inductance of the network-side increases, the

system is more likely to occur LFO incidents. It is manifested as the farther the distance between power supply station and EMUs, the more likely the system coupling LFO occurs; Short-series EMUs are taken as the analysis objects in [10]. The maximum number of EMUs allowed is six when they are lightly loaded at a closer distance simultaneously. However, since the input impedance model may be different, the conclusions on the impact of EMU number may not be applicable to the long-series EMUs [11] or vehicles which are fully operating. In [6], the eigenvalue analysis method is adopted. Combining the actual EMU design requirements with damping calculation, a basis for its subsequent passive controller design is provided. The impedance models of vehicle-traction network are established and analyzed in [4]. It is concluded that impedance mismatch is the main cause of instability. But the specific affecting factors are not stated.

For the improvement of cascaded system stability, most of the existing literatures studied from two perspectives, adjusting the relevant parameters [12]-[13] and imposing control [14]. Since the parameters are closely coupled in the actual system, the adjustment of the parameters will affect other settings and may require a large amount of engineering. So the applied control approach may be relatively easy to implement [15]-[16]. Specific control strategies can be classified as passive control [17], nonlinear control [18] and active control strategies [19]. In [20], based on the installation of static synchronous compensator (STATCOM) [21], it is proposed that the adoption of active power filtering devices can better suppress the occurrence of LFO and enhance system stability. However, the improvements for the network side often have the drawback of huge engineering effort. So the researchers prefer to improve control strategy for traction converters of EMUs, especially for the newly manufactured EMUs of the host plant. [22] adopts passivity based control to ensure the global stability of the system. Active disturbance rejection control (ADRC) is adopted in [23]. It uses the state feedback of the state observer to replace the output feedback of the original controller and achieves better control effect. Compared with the other control methods such as proportional integral control, It is more effective in suppressing LFO. By adopting this method, the conflict between overshoot and rapidity is resolved. However, ADRC strategy may be difficult to rectify due to the number of parameters it requires, which is not conducive to parametric design. A virtual inertia control strategy with adaptive sliding mode observer to suppress the occurrence of LFO is reported by the authors in [24]. But the design of its state observer parameters is vulnerable to the actual system error. A notch filter is designed to suppress LFO in [25], which is essentially an active control strategy. Reference [19] adopts proportional-derivative control strategy. It can equivalently changes vehicle input impedance to make the system satisfy the corresponding steady-state criterion. But the dynamic performance near the cut-off frequency of the proposed low-pass filter will be affected, and the performance of EMU may be affected while suppressing LFO.

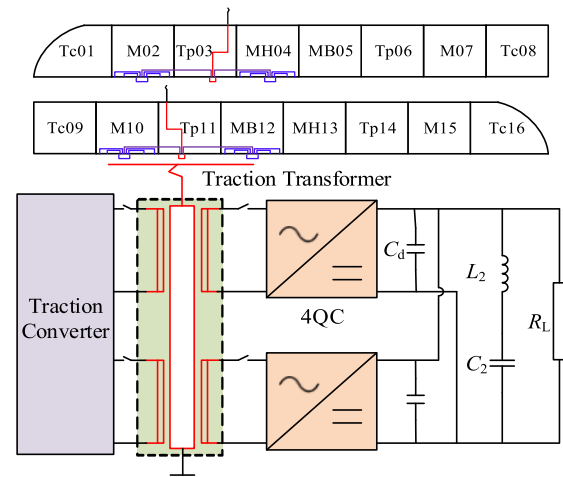


FIGURE 1. Main circuit structure of traction unit.

Considering the deficiencies in the above researches, this paper proposes a virtual resistive control strategy with the introduction of band-pass filtering link for 4QC of EMUs. The rest of this paper is organized as follows. In Section II, the structure of traction converter and existing 4QC control strategy are showed. In Section III, the input impedance model of EMU and the output impedance of the traction network are established. Based on the Nyquist curve of the open-loop transfer function and FRC [26], the impact of various electrical parameters, EMU operation states and other factors on the coupled system stability is analyzed. In Section IV, control strategy is proposed and the influence of virtual impedance and band-pass filter on system dynamic performance and stability is analyzed. And the corresponding parameters are designed. In Section V, a simulation model of the coupled system is built, and the simulation and experiment verify the correctness of the analysis results and the effectiveness of the proposed control strategy.

## II. INTRODUCTION OF EMU TRACTION CONVERTER

### A. STRUCTURE OF EMU TRACTION UNITS

Fig. 1 shows the main circuit structure of EMU traction units. It consists of the traction transformer secondary side, traction converters, traction motors and traction control unit (TCU). Among them, the traction converter consists of two parallel duplicated 4QCs, intermediate DC link, traction inverters and towing mode devices.

The traction system of the “Fuxing” long-series EMU consists of four traction units, each consisting of two motor units and two trailer units, and the four traction units adopt a symmetrical design. The load is composed of the traction inverters and auxiliary converters. In Fig. 1,  $C_d$  is the DC-side support capacitor of the traction converter,  $L_2$  and  $C_2$  are the DC-side second resonance circuit filter inductor and capacitor respectively. When traction converters begin to operate or only the auxiliary converter is fully operated, the load

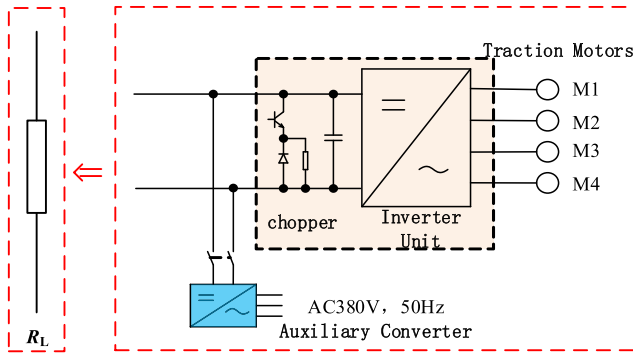


FIGURE 2. Schematic diagram of load equivalence.

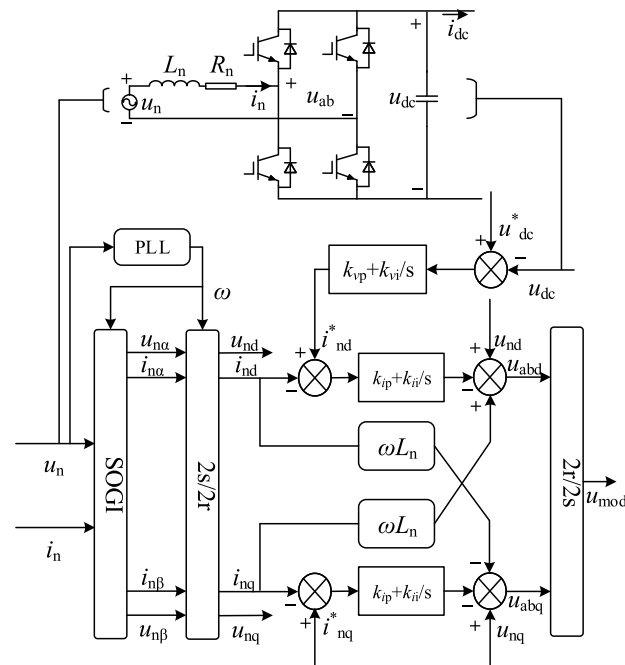


FIGURE 3. Diagram of dq decoupling current control for 4QC.

can be equated to the form of resistance  $R_L$ . The equivalent schematic diagram is shown in Fig. 2.

### B. EXSITING CONTROL STRATEGY

4QC mostly adopts instantaneous current control strategy or decoupling current control strategy, among which dq decoupling current control can meet the needs of higher power loads.

Fig. 3 shows the control strategy diagram, where  $u_n$  is the AC-side voltage of 4QC;  $i_n$  is the AC-side current of 4QC;  $L_n$  is the equivalent leakage inductance of the traction transformer;  $R_n$  is the equivalent leakage resistance of the traction transformer;  $u_{nd}$  and  $u_{nq}$  are the d-axis and q-axis components of  $u_n$ ;  $i_{nd}$  and  $i_{nq}$  are the d-axis and q-axis components of  $i_n$ ;  $u_{dc}^*$  is the DC-side reference voltage;  $i_{nd}^*$  and  $i_{nq}^*$  denote the d-axis and q-axis reference current, respectively;  $k_{vp}$  and  $k_{vi}$  denote the outer-loop proportional and integral coefficients of PI controller;  $k_{ip}$  and  $k_{ii}$  denote the proportional

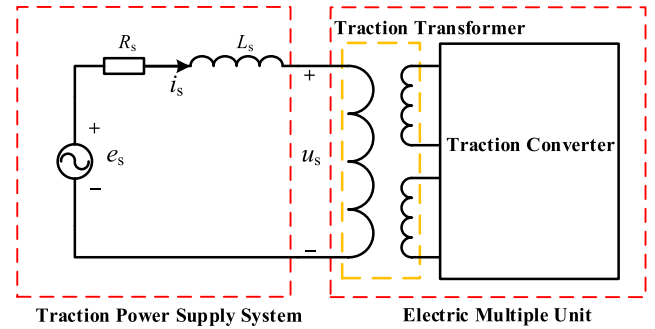


FIGURE 4. Schematic diagram of Traction- EMUs coupling.

and integral coefficients of inner-loop PI controller;  $u_{mod}$  is the generated voltage modulation signal;  $i_{dc}$  is the output current of 4QC. In this paper, 4QC adopts the dq decoupling current control based on Second-Order General Integrator (SOGI)-Phase Locked Loop (PLL).  $u_n$  and  $i_n$  can be transformed into dq frame components by SOGI-PLL and 2s/2r transformation.

## III. IMPEDANCE MODELING AND STABILITY ANALYSIS

### A. OUTPUT IMPEDANCE MODEL OF NETWORK

For the LFO of the coupled system, the traction network output impedance model can be equated to a series resistance-inductance circuit of the form  $Z_s = R_s + j\omega L_s$  [15].  $Z_s$  is the traction network equivalent output impedance,  $R_s$  is the traction network equivalent resistance, and  $L_s$  is the traction network equivalent inductance. Fig. 4 shows the traction network-EMU coupling model, where  $u_s$  is the voltage at the coupling point and  $i_s$  is the current through the coupling point. It shows that the degree of dd-axis impedance matching in the system open-loop transfer function has the most obvious effect on the system stability in [19]. The feasibility of researching dd-axis impedance matching for high power factor cascade systems to achieve stability analysis of AC systems has been clarified in [27]. Therefore, this paper focuses on analyzing the effects of different factors on the dd-axis component based on the system open-loop transfer function. The output impedance dd-axis component  $Z_{sdd}(s)$  of the traction network side is shown in (1).

$$Z_{sdd}(s) = R_s + sL_s \quad (1)$$

### B. INPUT IMPEDANCE MODEL OF EMU

To establish the input impedance model of 4QC, the mathematical expression can be obtained according to Fig. 3, combined with the Laplace transformation as

$$\begin{cases} u_n = (j\omega L_n + R_n)i_n + u_{ab} \\ u_{nd}(s) = (sL_n + R_n)i_{nd}(s) + u_{abd}(s) - \omega L_n i_{nq}(s) \\ u_{nq}(s) = (sL_n + R_n)i_{nq}(s) + u_{abq}(s) + \omega L_n i_{nd}(s) \end{cases} \quad (2)$$

where  $u_{ab}$  is the input voltage of the network side rectifier,  $u_{abd}$  and  $u_{abq}$  are the d and q axis components of  $u_{ab}$ . Since

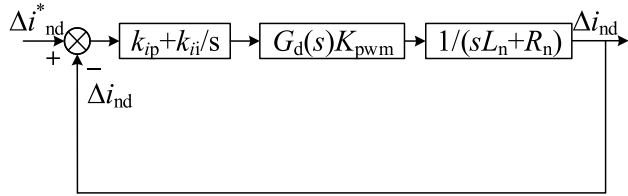


FIGURE 5. Small-signal model of the current inner-loop.

the existing control strategy is a double closed-loop proportional integral control method, combined with dq current decoupling control, the following equation can be obtained as

$$\begin{cases} i_{nd}(s) = \frac{(k_{ip} + k_{ii}/s)(i_{nd}^*(s) - i_{nd}(s))G_d(s)}{(sL_n + R_n)} \\ i_{nq}(s) = \frac{(k_{ip} + k_{ii}/s)(i_{nq}^*(s) - i_{nq}(s))G_d(s)}{(sL_n + R_n)} \end{cases} \quad (3)$$

where  $G_d(s)$  is the system delay link and the small-signal model of the current inner-loop can be plotted as shown in Fig. 5. Since LFO of the coupled system of the traction network rolling stock are analyzed and suppressed by the dd-axis component represented in this paper, only the model of the d-axis current inner-loop is plotted.

Further, combining with power conservation, the input-output power expression can be deduced as

$$\frac{1}{2}(u_{nd}i_{nd} + u_{nq}i_{nq}) = u_{dc}C_d \frac{du_{dc}}{dt} + u_{dc}i_{dc} \quad (4)$$

The physical quantities can be expressed as their corresponding steady-state quantities summed with their small disturbance components, such as

$$\begin{cases} u_{nd} = U_{nd} + \Delta u_{nd} \\ u_{nq} = U_{nq} + \Delta u_{nq} \\ i_{nd} = I_{nd} + \Delta i_{nd} \\ i_{nq} = I_{nq} + \Delta i_{nq} \\ u_{dc} = U_{dc} + \Delta u_{dc} \\ i_{dc} = I_{dc} + \Delta i_{dc} \end{cases} \quad (5)$$

where the upper-case physical quantities in (5) are the steady-state quantities. Combining with (5), (4) can be obtained as follows.

$$\begin{aligned} & \frac{1}{2}(U_{nd}\Delta i_{nd} + \Delta u_{nd}I_{nd} + U_{nq}\Delta i_{nq} + \Delta u_{nq}I_{nq}) \\ & = U_{dc}C_d \frac{d\Delta u_{dc}}{dt} + U_{dc}\Delta i_{dc} + \Delta u_{dc}I_{dc} \end{aligned} \quad (6)$$

Since 4QC of EMUs should ensure the approximate unit power factor operation as much as possible,  $U_{nq}$  and  $I_{nq}$  can be assumed to be zero. Utilize Laplace transformation, (6) can be expressed as

$$\begin{aligned} \Delta u_{dc}(s) &= \frac{I_{nd}}{2(U_{dc}C_d s + I_{dc})} \Delta u_{nd}(s) \\ &+ \frac{U_{nd}}{2(U_{dc}C_d s + I_{dc})} \Delta i_{nd}(s) \end{aligned}$$

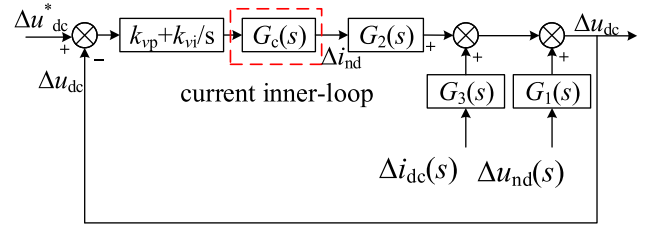


FIGURE 6. Small-signal model of the voltage outer-loop.

$$- \frac{U_{dc}}{U_{dc}C_d s + I_{dc}} \Delta i_{dc}(s) \quad (7)$$

where

$$\begin{cases} G_1(s) = \frac{I_{nd}}{2(U_{dc}C_d s + I_{dc})} \\ G_2(s) = \frac{U_{nd}}{2(U_{dc}C_d s + I_{dc})} \\ G_3(s) = - \frac{U_{dc}}{U_{dc}C_d s + I_{dc}} \end{cases} \quad (8)$$

Combined with (2), the small signal model of 4QC can be drawn as shown in Fig. 6, and the dd-axis input impedance of 4QC can be obtained as

$$Z_{indd} = \frac{\Delta u_{nd}(s)}{\Delta i_{nd}(s)} = \frac{sL_n + R_n + G_4(s) + G_5(s)}{1 - G_6(s)} \quad (9)$$

where

$$\begin{cases} G_4(s) = \frac{k_{ip} + k_{ii}/s}{0.5T_{ds} + 1} \\ G_5(s) = \frac{(k_{vp} + k_{vi}/s)(k_{ip} + k_{ii}/s)}{0.5T_{ds} + 1} G_2(s) \\ G_6(s) = \frac{1 + (k_{vp} + k_{vi}/s)(k_{ip} + k_{ii}/s)}{0.5T_{ds} + 1} G_1(s) \end{cases} \quad (10)$$

As the long-series EMU needs double pantographs (Tp03 and Tp11 or Tp06 and Tp14), single pantograph corresponds to two traction units, and each traction unit can be regarded as two 4QCs in parallel, the total input impedance of the train should be 1/8 times the input impedance of a single 4QC. And because of the need to adopt the impedance matching method to analyze the causes of instability and provide a basis for improvement, the input impedance obtained from the traction converter system combined with the traction transformer is imputed to the network side, the traction transformer ratio is 25000:1900, and the imputed dd-axis input impedance  $Z_{in}(s)$  of the EMU can be obtained as

$$Z_{in}(s) = \frac{1}{8} \frac{25000^2}{1900^2} \frac{sL_n + R_n + G_4(s) + G_5(s)}{1 - G_6(s)} \quad (11)$$

### C. STABILITY ANALYSIS BASED ON FRC

Based on the principle of impedance matching and FRC, the open-loop transfer function  $T_m(s)$  of the coupled system in the worst case (the system structure schematic is shown in Fig. 4) can be expressed as

$$T_m(s) = n \frac{Z_{sdd}(s)}{Z_{in}(s)} \quad (12)$$

TABLE 1. Parameters of coupled system.

Symbol	Quantity	Values
$R_s$	equivalent resistance of traction network	1.173 $\Omega$
$L_s$	equivalent inductance of traction network	3.37 mH
$u_{dc}^*$	reference voltage	2800 V
$R_n$	equivalent resistance of 4QC	0.0145 $\Omega$
$L_n$	equivalent inductance of 4QC	3.2 mH
$C_d$	DC link capacitor	3 mF
$k_{vp}$	proportional coefficient of voltage controller	0.6
$k_{vi}$	integral coefficient of voltage controller	2
$k_{ip}$	proportional coefficient of current controller	1
$k_{ii}$	integral coefficient of current controller	0.1

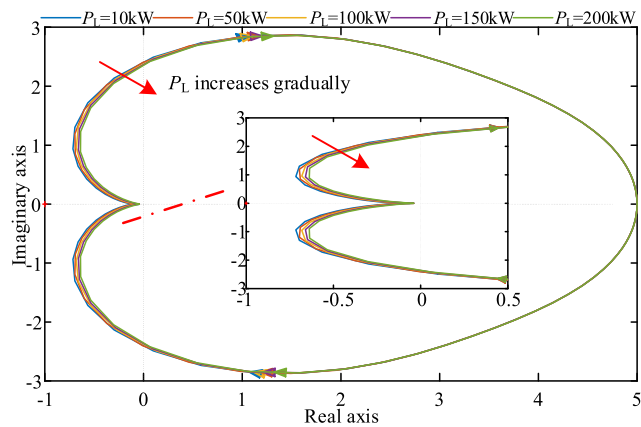


FIGURE 7. Nyquist plots of  $T_m(s)$  with different  $P_L$ .

Fig. 7 plots the Nyquist curve with various load power when only one EMU operates. The higher the load input power, the more stable the system, which shows that LFO is more likely to occur when the EMU operates under light load. It should be noted that although the stability margin decreases when single EMU begins to operate compared with the full power operation, the system is still stable at this time according to the stability criterion of FRC.

Most of the existing literatures have taken CRH3, CRH5 and other short-series vehicles as the research objects, and proposes that coupled LFO may occur when six or more vehicles start to operate simultaneously. But in existing literature, it is not considered the possible LFO caused by long series EMUs. To further investigate the influence of EMUs number on the system stability, the Nyquist curve of  $T_m(s)$  is plotted in Fig. 8. When the number of trains is 3, part of the curve enters the forbidden zone delineated by the red dashed line (the solid part is equal to -1). The system may become unstable at this time. When the number of trains is 2, the system is in a critical stable state.

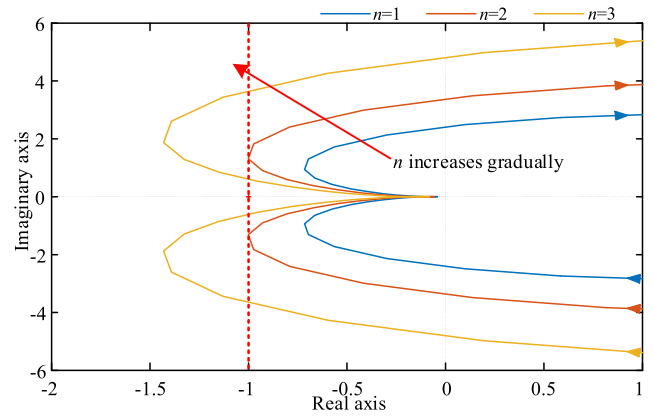


FIGURE 8. Nyquist plots of  $T_m(s)$  with different number.

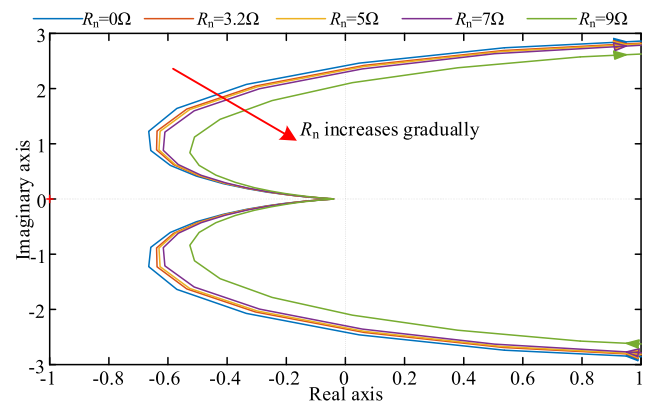


FIGURE 9. Nyquist plots of  $T_m(s)$  with different  $R_n$ .

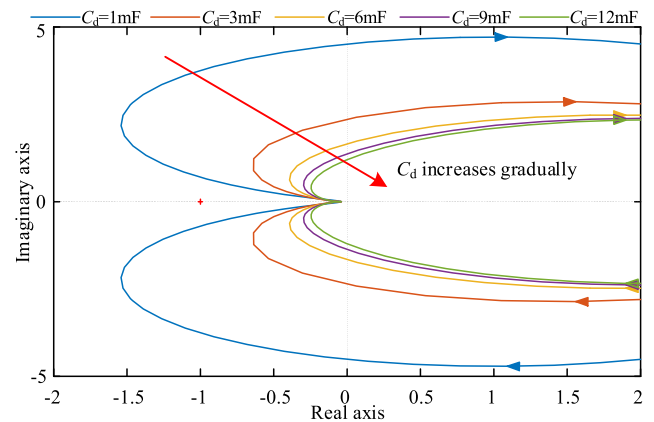


FIGURE 10. Nyquist plots of  $T_m(s)$  with different  $C_d$ .

To analyze the influence of EMU-side parameters on system stability, the impact of various  $R_n$  and  $C_d$  on system stability are analyzed based on Nyquist curves, as shown in Fig. 9 and Fig. 10. It can be seen that the increase of both  $R_n$  and  $C_d$  will enhance the system stability and improve the system stability margin, where the effect is more apparent when  $C_d$  increases. The effect of traction transformer leakage inductance on the stability of coupled system is not analyzed in this paper, please refer to the Appendix for details.

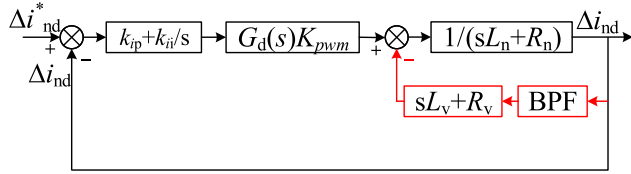


FIGURE 11. Small-signal model of the current with improvement.

Combined with the above analysis, this paper defines the most adverse case for the stability of the coupled system: multiple EMUs operate at the end of the same power supply arm with the condition of rising pantograph. In this case, the existing maximum number of EMUs allowed for stability is two.

IV. VIRTUAL IMPEDANCE WITH BANDPASS FILTER LINK CONTROL STRATEGY

A. MECHANISM OF PROPOSED CONTROL STRATEGY

Although the increase of  $R_n$  and  $C_d$  will theoretically enhance the system stability, the increase of  $C_d$  will affect the actual parameters selection of the second resonant circuit [28]. In addition, there will be errors in the measurement of  $R_n$ , which makes it difficult to accurately design parameters. The increase of  $R_n$  and  $C_d$  will occupy more space, which is not conducive to the lightweight design of EMU.

This paper proposes a virtual impedance with band-pass filtering link control strategy, considering above-mentioned application problems. The small-signal control block diagram of the proposed control strategy is shown in Fig. 11.

In Fig. 11,  $(sL_v + R_v)$  is virtual impedance link, BPF represents band-pass Filter, transfer function  $G_F(s)$  of which is expressed as

$$G_F(s) = \frac{s/(Q\omega_0)}{(s/\omega_0)^2 + s/(Q\omega_0) + 1} \tag{13}$$

Virtual impedance is used to change the original input impedance of EMU without increasing the actual resistive element and make  $T_m(s)$  meet FRC. The band-pass filter link is used to select the specific signal as modified and apply the virtual impedance to the characteristic band of the LFO. It is used to avoid the impact of virtual impedance link on other frequency bands signals, so that the dynamic performance of traction converters can be ensured.

B. DESIGN OF RELEVANT PARAMETERS

The parameters of the band-pass filter need to be designed first such as bandwidth value and center frequency. According to the spectrum of LFO measured in engineering [19], the current harmonics are mainly distributed in 40-60 Hz. So the bandwidth of the band-pass filter link is designed to 20 Hz and the center frequency is set to 50 Hz. Therefore, in (13),  $\omega_0$  is set to 314 rad/s and  $Q$  is set to 2.5.

Fig. 12 shows the impact of different virtual inductance values on Nyquist curve of  $T_m(s)$ .  $k$  indicates that the virtual resistive value is  $k$  times greater than the original resistive

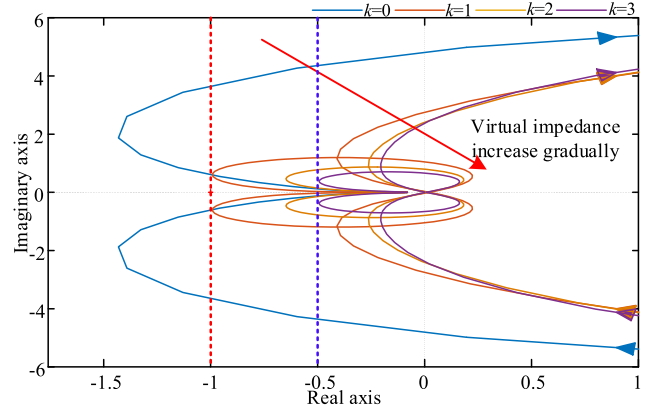


FIGURE 12. Nyquist plots of  $T_m(s)$  with different virtual resistance value and inductance value.

value of the system, i.e.  $k = (sL_v + R_v) / (sL_n + R_n)$ . It is clear that the stability margin of the system gradually increases, as the virtual resistance value increases. When the virtual resistance value equals to the original resistance value, system Nyquist curve meets FRC. However, in order to ensure sufficient stability margin in engineering, the forbidden zone boundary is mostly set to -0.5 in the real part, as shown by the blue dashed line in Fig. 12. To meet the margin requirement in engineering, the virtual resistance value should be selected as three times or more of the existing network side resistance value.

Further, Fig. 13 depicts the effect of different virtual impedance values on the voltage bandwidth after setting the parameters of the BPF link. The increase of virtual resistance and virtual inductance will both reduce the system bandwidth, so the increase of virtual resistance and inductance in the time domain system will slow down the system dynamic response process. Therefore, the value of the virtual resistance and inductance can not be too large to ensure system dynamic performance, but should be a compromise between the consideration of system dynamic performance and stability.

In this paper, the principle for the selection of the virtual impedance value is: under the premise of ensuring the system stability margin required in engineering, the smaller virtual impedance value is selected as much as possible to fully ensure the dynamic performance of traction converters. Therefore, the virtual impedance value is selected as three times of the original resistance value, that is,  $L_v = 9.6$  mH,  $R_v = 0.0435 \Omega$ , and the bandwidth is 46.8951 Hz. With the proposed control strategy, the maximum number of EMUs allowed for stability increases from three to six, as shown in Fig. 14.

V. SIMULATIONS AND EXPERIMENTS

A. SIMULATIONS

To verify the correctness and effectiveness of the proposed control strategy, a simulation platform of the traction network-EMUs coupled system is built in the environment

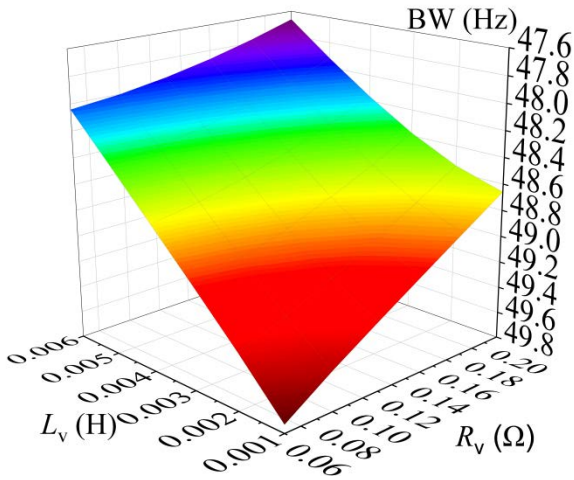


FIGURE 13. The relationship between the bandwidth and different virtual impedance value.

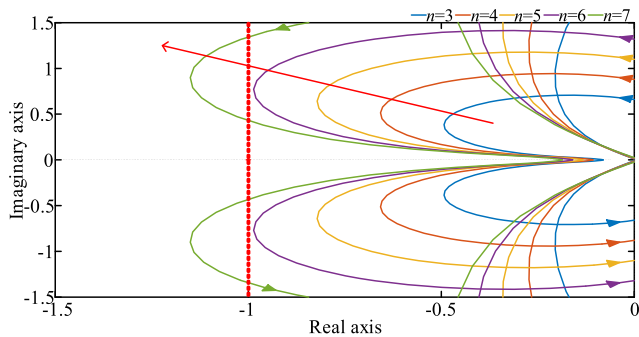


FIGURE 14. Nyquist plots of  $T_m(s)$  with different number after adopting proposed control strategy.

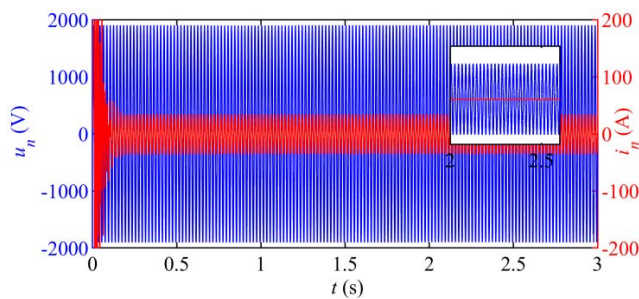


FIGURE 15. Simulation results of  $u_n$  and  $i_n$  when single EMU operates.

of Matlab/Simulink. Parameters of the traction network and EMU are shown in Table 1.

Fig. 15 verifies that the original control parameters and system electrical parameters proposed in this paper are effective, when single EMU operates. 4QC can work normally and guarantee the approximate unit power factor operation.

Fig. 16 illustrates the waveforms of  $u_s$  and  $i_s$  when three EMUs operate without improvement. When the number of EMUs is three, LFO occurs and oscillation frequency is about 7 Hz under the worst case scenario. It verifies the

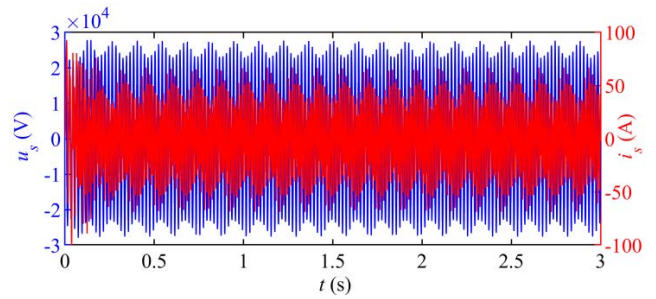


FIGURE 16. Simulation results of  $u_s$  and  $i_s$  with original control strategy when  $n = 3$ .

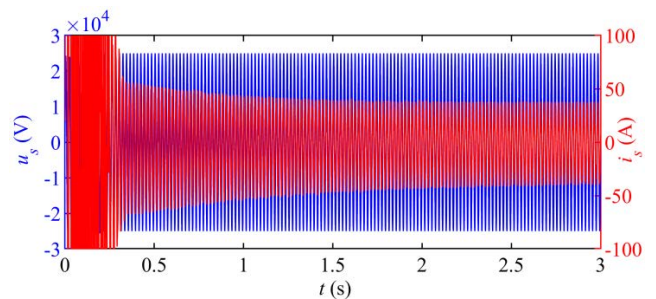


FIGURE 17. Simulation results of  $u_s$  and  $i_s$  with improvement when  $n = 3$ , and  $k = 2$ .

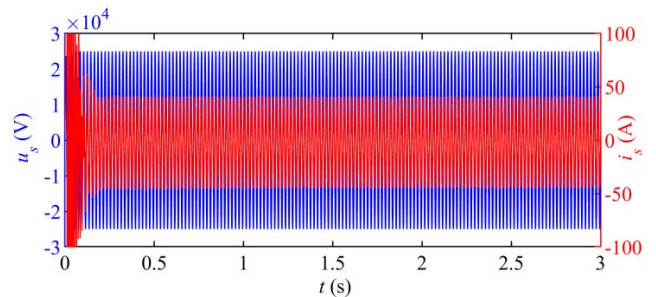


FIGURE 18. Simulation results of  $u_s$  and  $i_s$  with improvement when  $n = 3$ , and  $k = 3$ .

conclusion that the simultaneous raising of the three EMUs deteriorates the stability of the network coupling system as described in the previous section.

Fig. 17 shows the the waveforms of  $u_s$  and  $i_s$  after adopting proposed control when  $k = 2$ .  $i_s$  oscillates at a larger amplitude than  $u_s$  in about first 0.2 s, indicating that  $R_n$  has a more obvious impact on  $i_s$  when LFO occurs under the condition of rising pantograph. After the system starts for 2-2.5 s, the amplitude of  $i_s$  gradually decays until it reaches relative stability, which proves that the proposed control strategy in this paper can suppress the low-frequency oscillation, but the suppression effect is limited due to the parameters selection factor at this time.

Fig. 18 depicts the the waveforms of  $u_s$  and  $i_s$  after adopting proposed control when  $k = 3$ . At this time the system reaches relative stability after 0.2 s,  $u_s$  and  $i_s$  do not occur low-frequency oscillations, and the time to reach relative stability is shorter than that in Fig. 17, achieving the desired goal.

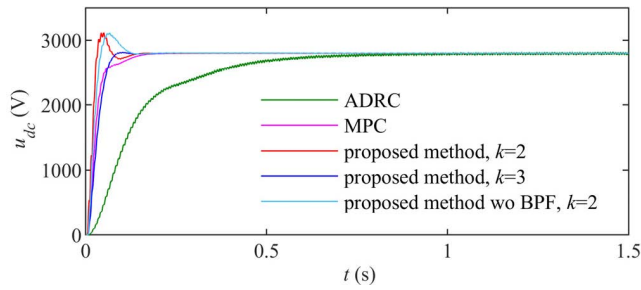


FIGURE 19. Simulation results of  $u_{dc}$  with different methods.

TABLE 2. Performance Indexes of Different Methods.

methods	Overshoot(%)	Adjustment Time(s)	Voltage Fluctuation(V)
$k = 2$	3.57	0.23	$\pm 2.62$
$k = 3$	0.36	0.14	$\pm 0.27$
wo BPF, $k = 2$	3.56	0.31	$\pm 2.80$
MPC	none	0.27	$\pm 1.48$
ADRC	none	0.87	$\pm 13.11$

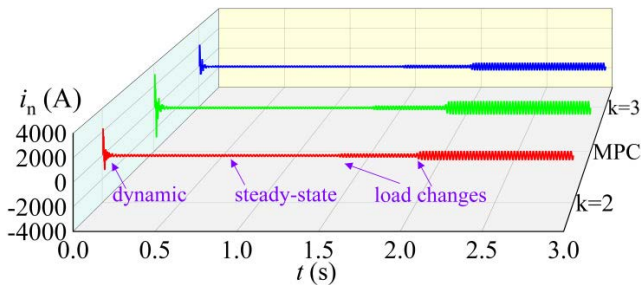


FIGURE 20. Simulation results of  $i_n$  with different control strategies.

Further, the proposed method is compared with four different methods, i.e. i) ADRC [24], ii) MPC and iii) virtual impedance control without BPF.

As shown in Fig. 19 and TABLE 2, virtual impedance control strategy has a shorter adjustment time. Dynamic performance of proposed method is better than that of ADRC and MPC. Compared with the condition of  $k = 2$ , the condition of  $k = 3$  takes longer to adjust, but the overshoot is significantly reduced. In addition, the dynamic performance of the method containing BPF is better than that without BPF.

Fig. 20 depicts the waveform of  $i_n$  when three different methods are used. Proposed method has the lowest peak current and MPC has the highest peak current. It shows that the proposed control strategy is more suitable for practical application.

The above simulation verifies the effectiveness under the condition of rising pantograph. When EMU eases the braking and gradually starts the loads after the preparation is completed, the simulation analyzes the voltage change of  $u_{dc}$  when 4QC adopts the proposed control strategy, as shown in Fig. 21.

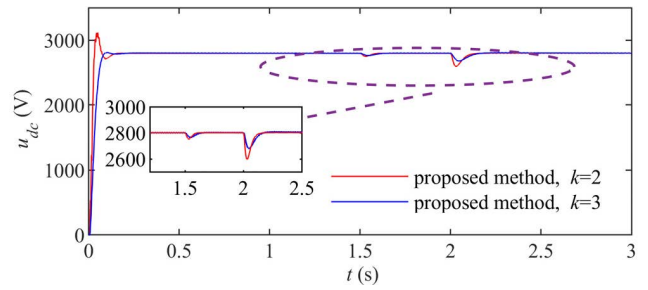


FIGURE 21. Simulation results of  $u_{dc}$  with different  $k$  when loads operates gradually.

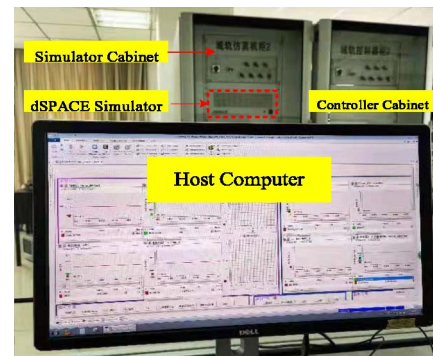


FIGURE 22. Nyquist plots of  $T_m(s)$  with different virtual resistance value and inductance value.

At 1.5 s, 225 kW load is put in to simulate the auxiliary converter start-up and the consumption of auxiliary devices. At 2 s, 1200 kW load is applied to simulate the traction inverter unit and traction motors start-up. When the load is applied, the DC link voltage drop decreases with the increase of virtual impedance and the recovery time to the relative stability increases slightly, which verifies that the proposed control strategy can ensure the normal operation of EMU traction mode when  $k = 3$ .

B. EXPERIMENTS

To further verify the correctness of the theoretical analysis and the proposed control strategy, a simulation model of traction network-EMUs system is built on the hardware-in-the-loop simulation platform. The existing control strategy and proposed control strategy for 4QC are implemented by TI DSP TMS320F28335 control board. The hardware-in-the-loop experimental platform is shown in Fig. 22.  $u_{pwm}$  is the input voltage of traction motor;  $i_{mon}$  is the input current of traction motor. In this experiment, it is assumed that all EMUs are operating under the condition of rising pantograph.

Fig. 23 shows the waveforms of the system when the original control strategy is adopted and the number of EMUs is three. At this point, the amplitudes of  $u_n$  and  $i_n$  changes significantly, and oscillation occurs. However, there is no obvious distortion of  $i_{mon}$ , indicating that the support capacitance of the existing traction inverters can ensure the normal operation of traction motors when  $u_{dc}$  oscillates.



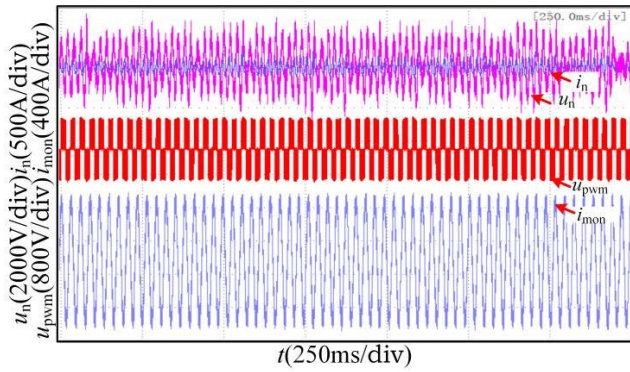


FIGURE 23. Simulation results without proposed control strategy.

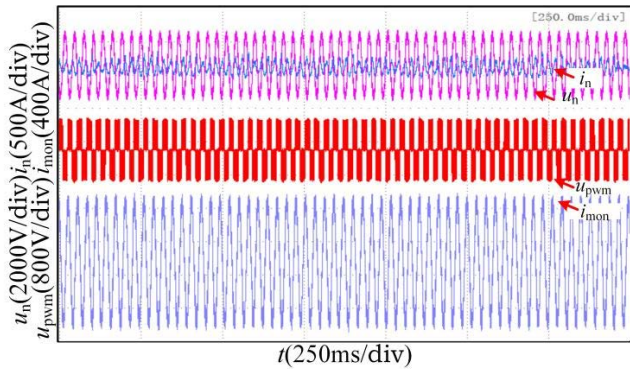


FIGURE 24. Simulation results with proposed strategy when  $k = 3$ .

Fig. 24 depicts the waveforms of the system when the proposed control strategy is adopted and  $k = 3$ . It is clear that the amplitudes of  $u_n$  and  $i_n$  changes slightly and the oscillation is suppressed. The effectiveness of the proposed control strategy and parameters selection are proved. In addition, the comparison of  $i_{mon}$  between Fig. 22 and Fig. 23 shows that  $i_{mon}$  does not have obvious distortion, indicating that the proposed control strategy will not affect the normal operation of traction converters.

### VI. CONCLUSION

Low-frequency oscillation occurs when three or more EMUs rise pantograph under the worst case scenario. To suppress LFO accidents and ensure dynamic performance of traction converter, a virtual impedance with band-pass filter link control strategy is proposed in this paper. The following conclusions are drawn.

(1)  $P_L$ ,  $n$ ,  $R_n$  and  $C_d$  all affect the coupling system stability. The system stability can be enhanced by increasing  $P_L$ ,  $R_n$  and  $C_d$ . Reducing  $n$  can enhance the system stability.

(2) Virtual impedance adjusts the input impedance of EMUs equivalently to make  $T_m(s)$  meet FRC. The proposed method can improve the system stability margin and suppress the occurrence of LFO in coupled system. The maximum number of EMUs allowed for stability increases.

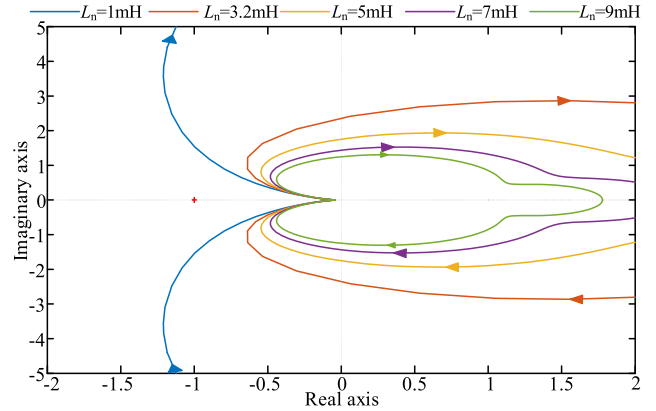


FIGURE 25. Nyquist plots of  $T_m(s)$  with different  $L_n$ .

(3) The proposed control strategy will increase the magnitude of  $i_n$  when EMU starts to operate and decrease system bandwidth. So parameters selection requires a compromise between the stability of the coupled system and the dynamic performance of EMU itself. The proposed control strategy still needs further improvement.

### APPENDIX

To analyze the effect of  $L_n$  on the coupled system stability, the Nyquist curves of different  $L_n$  are depicted, as shown in Fig. 25. With the increase of  $L_n$ , system stability margin gradually increases. But according to the existing studies, it cannot yet be qualitatively assumed that the increase of  $L_n$  enhances or decreases the coupled system stability [3], [8], [26], [29]-[30]. [29] proposes that the increase of  $L_n$  when 4QC adopts instantaneous current control strategy can make the system more stable. [30] proposes that the increase of the equivalent leakage inductance when 4QC adopts dq current decoupling control can make the system more stable, and the same conclusion is not only applicable to the PI controller, but also to the sliding-mode variable structure controller proposed in their paper. The influence of the EMU electrical parameters on the system stability is analyzed by Bode diagram, and the conclusion that the increase of  $L_n$  may reduce the system stability is obtained in the literature [8], [26].

According to the conclusions of the above literatures, especially for the comparative analysis of the Bode diagram and Nyquist curve showed in the above literature, it can be seen that the impact of  $L_n$  on the coupled system stability may be influenced by the system switching frequency, controller parameters and other electrical parameters at the EMU side. When the EMU-side parameters are different, the trend of the Bode curve may be affected, which in turn will affect the judgment of the impact on the system stability when analyzing different  $L_n$  [3] (which can be obtained from a clear comparison of Bode plots in [26], [30]). The impact of different  $L_n$  on the coupled system adopting different control and electrical parameters is still to be systematically quantified.

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