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Robust Model Predictive Control for Takagi-Sugeno Model With Bounded Disturbances—Pólya Approach

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ABSTRACT This paper proposes a general robust model predictive control (MPC) approach for the constrained Takagi-Sugeno (T-S) fuzzy model with additive bounded disturbances. We adopt the homogeneous polynomially parameter-dependent (HPP) Lyapunov matrix with the arbitrary complexity degree and the corresponding HPP control law for the controller design. By applying the Pólya's theorem and the extended nonquadratic boundedness property, a systematic approach to construct a set of sufficient conditions for assessing robust stability described by parameter-dependent linear matrix inequalities (LMIs) is established. The proposed approach is an improvement over the existing approaches in terms of control performance and stabilizable model range. Numerical examples are provided to show the effectiveness of the proposed robust MPC approach.

INDEX TERMS Robust model predictive control, T-S fuzzy model, bounded disturbances, extended nonquadratic boundedness.

I. INTRODUCTION

Takagi-Sugeno (T-S) fuzzy model has been widely used to approximate the nonlinear systems, whose basic idea is representing the original system by a family of linear sub-models [1]–[3]. For the stability analysis of T-S model, many efforts are made based on the Lyapunov functions, such as the common quadratic Lyapunov function in [4], the parameter-dependent quadratic Lyapunov function in [5], the piecewise-quadratic Lyapunov function in [6], the nonquadratic Lyapunov function in [7], and barrier Lyapunov functions in [8], [9]. In order to further improve the performance and reduce the conservatism, a general nonquadratic stabilization conditions are presented by the multiple-parameterization approach in [10]. In [11], [12], other general forms of relaxed stabilization conditions are derived by means of affine parameter-dependent Lyapunov functions. More details are available in [2], [3].

Model predictive control (MPC), as a widespread control technique being implemented in a receding horizon fashion,

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has its advantages in constraints handling for multi-variable plants, e.g., the distributed MPC [14], the industrial hierarchical MPC [15], and the stochastic MPC [16]. Usually, at each sampling interval, MPC solves an optimization problem, with the performance index being associated with the system evolutions over a prediction horizon, subjected to the physical constraints; a sequence of control moves are treated as the decision variables, but only the first among this sequence is sent to the actuator. These actions are repeated in a receding horizon fashion. Since the future predictions for input/state/output are needed, which are obtained based on the system dynamic model, the accuracy of this model is crucial for the future prediction, which as a result, can influence the control performance [13]. The stability analysis and control synthesis for T-S fuzzy model by MPC approaches have been studied with variety. In [17], an interval type-2 fuzzy MPC approach is proposed for nonlinear networked systems. The model and controller are not required to share the same lower and upper membership functions. In [18], to improve the performance, a local stability approach is applied and an estimation of the domain of attraction is provided. The work of [19] investigates the robust fuzzy MPC,

which uses the nonlinear local models. More relaxed results are achieved, and on-line computational cost is significantly reduced. The authors in [20] propose a fuzzy generalized predictive control for T-S systems based on Kernel Ridge Regression strategy which learns the T-S fuzzy parameters from the input and output data. Reference [21] utilizes the zonotopic set and interval matrices to bound the membership function errors. The controller parameters are stored in an off-line table for searching and robust tubes can be time-varying. In [22], the nonlinear multivariable predictive control is proposed for the vehicle systems. For maintaining the robustness and stability, the controller design is based on LMI convex optimization. The cooperative fuzzy MPC is represented in [23] where the overall nonlinear plant consists of a group of parallel input-coupled T-S fuzzy models. For this cooperation, the convergence and stability are guaranteed.

In order to deal with unknown disturbances, a series of paradigms have been elaborated (see e.g., [26]–[31], [33]). In [26], the homothetic tube-based approach is proposed, which maintains the state predictions of the linear model in the presence of disturbance within an on-line scaling tube centered at the disturbance-free model trajectory. The work of [27] proposes a tube-based MPC for nonlinear continuous-time model, and the feedback control law is optimized off-line. The authors in [28] utilize an integral non-squared stage cost and a non-squared terminal cost, so that the robustness of the resultant MPC is ensured without additional stability constraints. In [29], the input-to-state stability property is utilized for the quasi-min-max MPC design, and the first control move from the control sequence can be optimized directly. In [30], the notion of quadratic boundedness (QB) is utilized, based on which, the system is guaranteed to be quadratically bounded in the presence of disturbance. Reference [31] proposes the full and partial dynamic output feedback MPC applying full Lyapunov matrix. The elliptical estimation error set is refreshed on-line based on optimized information of the last sampling instant. Reference [32] aims at the norm-bounded model parametric uncertainty. The estimated state feedback gain and state estimator matrix are optimized on-line while the state estimator gain is designed off-line. In [33], sufficient conditions for computing the positively invariant set for T-S fuzzy systems are derived, and the terminal constraint set for 0-step and N -step control strategies are obtained.

This paper characterizes MPC synthesis, based on improving the Lyapunov function, for constrained T-S fuzzy model with the bounded disturbance. Some general results for the positiveness of polynomials with matrix-valued coefficients (based on Pólya's theorem) is given in [35], where some complete characterization of the solution of parameter-dependent LMIs, usually arising in the robust stability analysis, is proposed. However, when the model parameters are time-varying uncertain, the results in [35] cannot be directly invoked. We deal with this issue in this paper, and contributions are summarized as follows.

- 1) The potentiality of Pólya's theorem is exploited. The general homogeneous polynomially parameter-dependent (HPP) Lyapunov matrix whose complexity degree is tunable, and corresponding HPP control law, are applied.
- 2) By a generalization of the methods based-on Pólya's theorem and parameter-dependent LMIs, a series of finite-dimensional LMI relaxations, as sufficient stability conditions, are developed to robustly stabilize the resultant closed-loop system.

In [34], a general robust MPC approach for linear parameter varying (LPV) systems in the absence of bounded disturbance has been proposed, which can include many existing approaches with common quadratic Lyapunov matrices and state feedback laws (e.g., [24], [25]) as special cases. As compared with [12], [34], this paper handles the unknown but bounded disturbance. Since the bounded disturbance is incorporated in the model, the characterization of the closed-loop stability is different from that in [34]. There are two differences as compared with [34].

- While [34] introduces a free control move (i.e., the control move is the immediate decision variable), this paper does not.
- While [34] utilizes the dilution parameter G , this paper does not.

The rationale of approximation is the same as in [34]: the complexity degree of HPP solutions is tunable for the proposed approach and, when it increases, the conservatism of the results reduces; the HPP Lyapunov matrix and HPP feedback gain matrix can asymptotically approximate any Lyapunov and feedback gain matrices which are continuous on the combining coefficient functions.

Notation: The symbol \star induces a symmetric structure in the matrix inequalities. A variable with superscript \star means the optimal solution to the optimization problem. $\mathbb{R}^{m \times n}$ denotes the $m \times n$ -dimensional real matrix set. \mathbb{N}_+ is the set of nonnegative integers. For the vector x and positive-definite matrix $P > 0$, $\|x\|_P^2 = x^T P x$. $M!$ denotes factorial of M . $x(i|k)$ is the value of x at the future interval $k + i$, predicted at interval k . I is the identity matrix with appropriate dimension. For the column vectors x and y , $[x; y] = [x^T, y^T]^T$. $\varepsilon_P = \{\xi | \xi^T P \xi \leq 1\}$ denotes the ellipsoid that is associated with the symmetric positive-definite matrix P . The time-dependence of the MPC decision variables is often omitted for brevity.

II. PROBLEM STATEMENT

Consider a class of T-S fuzzy systems, with its j th rule represented by

$$\begin{aligned} \text{Rule } j: & \text{ IF } \theta_1(k) \text{ is } H_1^{(j)}, \dots, \text{ and } \theta_\vartheta(k) \text{ is } H_\vartheta^{(j)}, \\ & \text{ THEN } x(k+1) = A_j x(k) + B_j u(k) + D_j w(k), \quad (1) \end{aligned}$$

where $j \in \{1, \dots, r\}$ with r rules. Let $\theta(k) = [\theta_1(k); \theta_2(k); \dots; \theta_\vartheta(k)]$ be the measurable premise variable. $H_1^{(j)}, H_2^{(j)}, \dots, H_\vartheta^{(j)}$ are the fuzzy sets. $x \in \mathbb{R}^n$, $u \in \mathbb{R}^m$, and $w(k)$ are measurable state, input, and bounded disturbance

vectors, respectively. The disturbance is persistent and satisfies $w(k) \in \varepsilon_{P_w}$. Then the acquisition of T-S model can be described as

$$x(k+1) = A(\theta(k))x(k) + B(\theta(k))u(k) + D(\theta(k))w(k),$$

$$\Lambda(\theta(k)) = \sum_{j=1}^r h_j(\theta(k))\Lambda_j, \quad \Lambda \in \{A, B, D\}. \quad (2)$$

$$h_j(\theta(k)) = \frac{\prod_{\tau=1}^{\vartheta} H_{\tau}^{(j)}(\theta_{\tau}(k))}{\sum_{j=1}^r \prod_{\tau=1}^{\vartheta} H_{\tau}^{(j)}(\theta_{\tau}(k))}, \quad (3)$$

where $H_{\tau}^{(j)}(\theta_{\tau}(k))$ denotes the grade of membership of $\theta_{\tau}(k)$ in $H_{\tau}^{(j)}$, and $h_j(\theta(k))$ is the normalized membership function (abbreviated as the membership function), with $h_j(\theta(k)) \geq 0$, $\sum_{j=1}^r h_j(\theta(k)) = 1$. For the sake of simplicity, h_j is short for $h_j(\theta(k))$ and h_{j+} is short for $h_j(\theta(k+1))$ for the rest of the paper.

The physical constraints to be considered is as follows,

$$-\underline{u} \leq u(i|k) \leq \bar{u}, \quad i \geq 0, \quad (4)$$

$$-\underline{\psi} \leq \Psi x(i+1|k) \leq \bar{\psi}, \quad i \geq 0, \quad (5)$$

where $\underline{u} = [\underline{u}_1; \underline{u}_2; \dots; \underline{u}_m]$, $\bar{u} = [\bar{u}_1; \bar{u}_2; \dots; \bar{u}_m]$, $\underline{\psi} = [\underline{\psi}_1; \underline{\psi}_2; \dots; \underline{\psi}_r]$, $\bar{\psi} = [\bar{\psi}_1; \bar{\psi}_2; \dots; \bar{\psi}_r]$ with $\underline{u}_s > 0$, $\bar{u}_s > 0$, $s = 1, 2, \dots, m$ and $\underline{\psi}_s > 0$, $\bar{\psi}_s > 0$, $s = 1, 2, \dots, r$. $\Psi \in \mathfrak{R}^{r \times n}$ can be any pre-specified forms of constraints on state x .

The corresponding disturbance-free (reference) model is expressed as

$$\tilde{x}(k+1) = A(\theta(k))\tilde{x}(k) + B(\theta(k))\tilde{u}(k), \quad (6)$$

where \tilde{x} and \tilde{u} are the disturbance-free state and input signals.

The control objective is to design an MPC controller which steers $x(k)$, with the increase of sampling interval k , to converge to a neighborhood of the origin, while satisfying the constraints (4)–(5). Accordingly, the cost function is given as

$$J_{\infty}(k) = \sum_{i=0}^{\infty} \left[\|\tilde{x}(i|k)\|_Q^2 + \|\tilde{u}(i|k)\|_R^2 \right], \quad (7)$$

where $Q > 0$ and $R > 0$ are the given symmetric weighting matrices. $\tilde{x}(0|k) = x(k)$. Note that the penalized signals are \tilde{x} and \tilde{u} from the disturbance-free model (6), any other signals linearly dependent on \tilde{x} and \tilde{u} are also allowed.

III. MODEL PREDICTIVE CONTROL SYNTHESIS

In this section, we propose the main result. Firstly, the standard MPC synthesis approach with the nonquadratic Lyapunov function is proposed with guaranteed recursive feasibility and closed-loop stability. Then, we apply the Pólya's theorem to extend the result such that the LMI conditions in the MPC optimization problem is relaxed.

A. FORMULATION OF OPTIMIZATION PROBLEM

This paper utilizes a g -degree HPP control law in the form of

$$u(k) = F_g x(k), \quad (8)$$

where $F_g = Y_g S_g^{-1}$ is the parameter-dependent feedback gain, which is define as follows.

Firstly, introduce some definitions in order to be consistent with [35]. Let the HPP matrix with tunable complexity degree g be

$$G_g(\eta) = \sum_{p \in \mathcal{K}(g)} \eta_1^{p_1} \eta_2^{p_2} \dots \eta_r^{p_r} G_p, \quad p = p_1 p_2 \dots p_r, \quad (9)$$

where $\eta \in \Omega_r$, $p_i \in \mathbb{N}_+$, $i = 1, 2, \dots, r$, each $\eta_1^{p_1} \eta_2^{p_2} \dots \eta_r^{p_r}$ is a monomial. For all $p \in \mathcal{K}(g)$, $G_p \in \mathfrak{R}^{n \times n}$ are matrices. $\mathcal{K}(g)$ is the family of r -tuples, which is comprised of all the terms $p_1 p_2 \dots p_r$, $p_i \geq 0$, $i = 1, 2, \dots, r$, with $p_1 + p_2 + \dots + p_r = g$. $\mathcal{K}(g)$ includes $J(g)$ number of elements with

$$J(g) = \frac{(r+g-1)!}{g!(r-1)!}.$$

For example, for the case with $g = 2$ and $r = 2$, $\mathcal{K}(g) = \{02, 11, 20\}$ and $J(2) = 3$, subject to the form $G_2(\eta) = \eta_2^2 G_{02} + \eta_1 \eta_2 G_{11} + \eta_1^2 G_{20}$.

Hence, in light of (9), we have parameter-dependent HPP Lyapunov matrices $Y_g = \sum_{p \in \mathcal{K}(g)} h_1^{p_1} h_2^{p_2} \dots h_r^{p_r} Y_p$, $S_g = \sum_{p \in \mathcal{K}(g)} h_1^{p_1} h_2^{p_2} \dots h_r^{p_r} S_p$, with $p = p_1 p_2 \dots p_r$.

Lemma 1: Consider a parameter-dependant LMI

$$P(\mu, \eta) \triangleq P_0(\eta) + \mu_1 P_1(\eta) + \dots + \mu_M P_M(\eta) > 0,$$

where $P_0(\cdot), P_1(\cdot), \dots, P_r(\cdot)$ are continuous functions with respect to parameters $\eta = [\eta_1, \eta_2, \dots, \eta_r]^T \in \Omega_r$ and unknown $\mu \in \mathbb{R}^M$. If there exists $\mu(\eta)$ ensuring $P(\mu(\eta), \eta) > 0$ for all η , then a homogenous polynomial function $\hat{\mu}(\eta)$ exists such that $P(\hat{\mu}(\eta), \eta) > 0$ holds.

Proof: See [35].

Remark 1: Lemma 1 implies that a homogeneous polynomial form of solutions, whose parameters lie in the unit simplex, is a very general form, i.e., can be readily transformed into any other continuous solutions of parameter-dependent LMIs. By considering the famous Weierstrass approximation theorem and applying Lemma 1, it infers that P_g and corresponding F_g can represent any Lyapunov matrix and feedback gain matrix that are parameterized by h_j , $j \in \{1, 2 \dots r\}$ as g increases.

By applying (8) on (2), the closed-loop system is obtained as

$$x(k+1) = (A(\theta(k)) + B(\theta(k))Y_g S_g^{-1})x(k) + D(k)w(k). \quad (10)$$

In order to guarantee the stability of closed-loop system (10), the technique of quadratic boundedness (QB) ([37]), which is primarily utilized for the state estimation problem and can be particularly useful for handling the system with bounded disturbance, is utilized to ensure that the state will stay in a quadratically bounded set. The definition and theorem of QB according to [37] are reviewed as follows.

Definition 1: For all allowable $w(k) \in \varepsilon_{P_w}$, $k \geq 0$, the autonomous linear system $x(k+1) = Ax(k) + Dw(k)$ is quadratically bounded with a common Lyapunov matrix P ,

if $x(k)^T P_x(k) \geq 1$ implies that $x(k+1)^T P_x(k+1) \leq x(k)^T P_x(k)$.

The QB condition is applied to many MPC problems with common quadratic Lyapunov functions. However, the resulting control performance can be conservative due to the utilization of the form of Lyapunov function. Since it is known that the extended nonquadratic Lyapunov method outperforms the common quadratic one, the extended nonquadratic boundedness is applied to characterize the closed-loop property in this paper.

Let P_g be the nonquadratic Lyapunov matrix. At interval $k+1$, define $P_{g+} = \sum_{p \in \mathcal{K}(g)} h_1(k+1)^{p_1} h_2(k+1)^{p_2} \dots h_r(k+1)^{p_r} P_p = \sum_{q \in \mathcal{K}(g)} h_{1+}^{q_1} h_{2+}^{q_2} \dots h_{r+}^{q_r} P_q$, with $q = q_1 q_2 \dots q_r$, satisfying $q_1 + q_2 + \dots + q_r = g$, $\sum_{j=1}^r h_{j+} = 1$.

Definition 2: System (10) is strictly nonquadratically bounded with an extended nonquadratic Lyapunov matrix $P_g > 0$, if $x(k)^T P_g x(k) \geq 1$ implies that $x(k+1)^T P_{g+x}(k+1) \leq x(k)^T P_g x(k)$ for any $w(k) \in \varepsilon_{P_w}$, $k \geq 0$.

By inheriting the results in [37] and extending to the case with the extended nonquadratic boundedness, the following conclusion can be obtained

Lemma 2: For all allowable $w(k+i)$, $i \geq 0$, the following statements are equivalent.

- a) System (2) is nonquadratically bounded with an extended nonquadratic Lyapunov matrix P_g .
- b) The ellipsoid $\varepsilon_{S_g^{-1}}$ is a positively invariant set for (2).
- c) $x(i|k)^T P_g x(i|k) \geq 1$ implies that $x(i+1|k)^T P_{g+x}(i+1|k) \leq x(i|k)^T P_g x(i|k)$.

From Lemma 2, we can obtain that the system (2) is nonquadratically bounded if at each sampling interval k , the following condition are satisfied:

$$\begin{aligned} x(i|k)^T P_g x(i|k) &\geq 1 \\ \Rightarrow \|x(i+1|k)\|_{P_{g+}}^2 - \|x(i|k)\|_{P_g}^2 \\ &\leq -x(i|k)^T Q x(i|k) - u(i|k)^T R u(i|k), \quad i \geq 0. \end{aligned} \quad (11)$$

Since $w(k+i) \in \varepsilon_{P_w}$, $i \geq 0$, $x(i|k)^T P_g x(i|k) \geq 1$ is equivalent to $w(k+i)^T P_w w(k+i) \leq x(i|k)^T P_g x(i|k)$. Hence, the condition (11) is equivalent to

$$\begin{aligned} w(k+i)^T P_w w(k+i) &\leq x(i|k)^T P_g x(i|k) \\ \Rightarrow \|x(i+1|k)\|_{P_{g+}}^2 - \|x(i|k)\|_{P_g}^2 \\ &\leq -x(i|k)^T Q x(i|k) - u(i|k)^T R u(i|k), \quad i \geq 0. \end{aligned} \quad (12)$$

According to (10), (12) can be expressed in quadratic form as

$$\begin{aligned} \begin{bmatrix} x(i|k) \\ w(k+i) \end{bmatrix}^T \begin{bmatrix} P_g & 0 \\ 0 & -P_w \end{bmatrix} \begin{bmatrix} x(i|k) \\ w(k+i) \end{bmatrix} &\geq 0 \\ \Rightarrow \begin{bmatrix} x(i|k) \\ w(k+i) \end{bmatrix}^T \\ &\times \begin{bmatrix} \Delta & \star \\ -D(k+i)^T P_{g+} \mathcal{A} & -D(k+i)^T P_{g+} D(k+i) \end{bmatrix} \\ &\times \begin{bmatrix} x(i|k) \\ w(k+i) \end{bmatrix} \leq 0, \end{aligned} \quad (13)$$

where $\Delta := P_g - \mathcal{A}^T P_{g+} \mathcal{A} - Q - S_g^{-T} Y_g^T R Y_g S_g^{-1}$, and $\mathcal{A} = A(\theta(k+i)) + B(\theta(k+i)) Y_g S_g^{-1}$.

By eliminating the variables $[x(i|k)^T w(k+i)^T]^T$ and invoking the S-procedure, it is shown that (12) is satisfied if and only if there exists a scalar $\alpha > 0$ such that (see [37])

$$\begin{bmatrix} \Delta & \star \\ -D(k+i)^T P_{g+} \mathcal{A} & -D(k+i)^T P_{g+} D(k+i) \end{bmatrix} \geq \alpha \begin{bmatrix} P_g & 0 \\ 0 & -P_w \end{bmatrix}. \quad (14)$$

By substitute $P_g = \gamma S_g^{-1}$, $P_{g+} = \gamma S_{g+}^{-1}$, pre- and post-multiplying (14) by $\text{diag}\{S_g, I\}$ (which leaves the inequality unaffected), and applying the Schur complement, it is shown that (14) is guaranteed by

$$\begin{bmatrix} (1-\alpha)S_g & \star & \star & \star & \star \\ 0 & \alpha P_w & \star & \star & \star \\ A_j S_g + B_j Y_g & D_j & S_{g+} & \star & \star \\ Q^{1/2} S_g & 0 & 0 & \gamma I & \star \\ R^{1/2} Y_g & 0 & 0 & 0 & \gamma I \end{bmatrix} \geq 0, \quad j \in \{1, 2, \dots, r\}. \quad (15)$$

Proposition 1: Reference model (6) is quadratically stable with the Lyapunov matrix P_g .

Proof: By neglecting the disturbance, (15) is reduced to $(1-\alpha)\gamma S_g^{-1} - \mathcal{A}^T \gamma S_{g+}^{-1} \mathcal{A} \geq Q + S_g^{-T} Y_g^T R Y_g S_g^{-1}$ which guarantees that $\gamma S_g^{-1} - \mathcal{A}^T \gamma S_{g+}^{-1} \mathcal{A} \geq Q + S_g^{-T} Y_g^T R Y_g S_g^{-1}$. Thus, by defining the HPP quadratic function $V(\tilde{x}(i|k)) = \tilde{x}(i|k)^T P_g \tilde{x}(i|k)$, and substituting $V(\tilde{x}(i+1|k)) = \tilde{x}(i+1|k)^T P_{g+\tilde{x}}(i+1|k)$, we have

$$\begin{aligned} V(\tilde{x}(i+1|k)) - V(\tilde{x}(i|k)) &\leq -\tilde{x}(i|k)^T Q \tilde{x}(i|k) \\ &\quad - \tilde{u}(i|k)^T R \tilde{u}(i|k), \quad i \geq 0. \end{aligned} \quad (16)$$

According to Lemma 2, since the condition (16) is satisfied, (2) is quadratically bounded with the Lyapunov matrix P_g .

Thus, the conclusion holds.

Remark 2: The HPP quadratic function $V(\cdot)$ is in a very general form, which implies that it covers many existing Lyapunov functions. For example, by taking $g = 0$, the Lyapunov function in [24] is recovered; by taking $g = 1$, [25] is recovered.

Based on Proposition 1, the state and input prediction of reference model (6) will converge to the origin, i.e., $\lim_{i \rightarrow \infty} \tilde{x}(i|k) = 0$ and $\lim_{i \rightarrow \infty} \tilde{u}(i|k) = 0$. Hence, summing (16) from $i = 0$ to ∞ yields

$$J_\infty \leq V(\tilde{x}(0|k)) = \tilde{x}(0|k)^T P_g \tilde{x}(0|k). \quad (17)$$

Since $P_g = \gamma S_g^{-1}$ and $\tilde{x}(0|k) = x(k)$, let γ be the upper bound of (17), i.e., $V(\tilde{x}(0|k)) \leq \gamma$, then the following holds:

$$\begin{bmatrix} 1 & \star \\ x(k) & S_g \end{bmatrix} \geq 0. \quad (18)$$

Similar to the procedure in [24], the input and state constraints in (4) and (5) are guaranteed by

$$\begin{bmatrix} S_g & \star \\ Y_g & Z \end{bmatrix} \geq 0, \quad Z_{ss} \leq z_{s,\text{inf}}^2, \quad s \in \{1, 2, \dots, m\}, \quad (19)$$

$$\begin{bmatrix} S_g & \star \\ \Psi(A_j S_g + B_j Y_g) & \Gamma \end{bmatrix} \geq 0, \quad j \in \{1, 2, \dots, r\},$$

$$\Gamma_{ss} \leq \psi_{s,\text{inf}}^2, \quad s \in \{1, 2, \dots, r\}, \quad (20)$$

where $z_{s,\text{inf}} = \min\{\underline{z}_s, \bar{z}_s\}$, $\psi_{s,\text{inf}} = \min\{\underline{\psi}_s, \bar{\psi}_s\}$ and Z_{ss} (Γ_{ss}) is the s th diagonal element of Z (Γ).

As the usual practice in robust MPC, the optimization problem is formulated as the following min-max form:

$$\min_{S_g, Y_g, \gamma, Z, \Gamma} \gamma \quad \text{s.t.} \quad (15), (18), (19), (20). \quad (21)$$

Not that by specifying α in the interval $(0, 1)$, problem (21) will become a convex optimization which can be solved efficiently by the interior point method. The MPC optimization (21) is simply an extension of the approach in [24] to the case with nonquadratic Lyapunov function. We will further extend this approach by the utilization of the Pólya's theorem in the next section.

Theorem 1: For (2), once there exists a feasible solution to optimization problem (21) at any interval k , then it will be feasible at $k + 1$, and x will converge to a neighborhood of the origin.

Proof: The proof contains two steps.

1) RECURSIVE FEASIBILITY

Supposed there exists feasible solution to (21) at sampling interval k . At next interval $k + 1$, we need to check that whether the constraints in (21) can still be satisfied. It is noted the optimization problems at k and $k + 1$ are different only in constraint (18) which contains the state $x(k)$.

At interval $k + 1$, let feasible bound $\bar{\gamma}(k + 1) = \gamma^*(k)$. Constraints (15) and (18) at k imply $V(\bar{x}(i + 1|k)) \leq V(\bar{x}(i|k)) \leq \gamma^*(k) = \bar{\gamma}(k + 1)$ which, by applying the Schur complement, is shown to be equivalent to

$$\begin{bmatrix} 1 & \star \\ x(k + 1) & S_{g+}^* \end{bmatrix} \geq 0. \quad (22)$$

Thus, the optimal solution at sampling interval k still satisfies constraint (18) at interval $k + 1$. Other constraints can be naturally satisfied at $k + 1$ by substituting the optimal solution at k . Hence, the constraints of the optimization problem are satisfied at $k + 1$.

2) STABILITY

Define the optimal upper bound $\gamma^*(k)$ as the candidate Lyapunov function. Denote by $\gamma^*(k + 1)$ the corresponding Lyapunov function at $k + 1$. According to the previous proof 1), it implies that $\gamma^*(k + 1) \leq \bar{\gamma}(k + 1) = \gamma^*(k)$ which means that $\bar{x} \rightarrow 0$ as $k \rightarrow \infty$. Since optimization problem (21) is feasible at sampling interval $k \geq 0$, constraint (15) is always satisfied, x will converge to a neighborhood of the origin.

Thus, the proof is complete. \square

Remark 3: For simplifying the presentation, we only consider the case when switching horizon $N = 0$ which is consistent with the benchmark work [24]. When $N > 0$, free control moves are added before the control law (8), which can be achieved easily by generalization, being omitted here for brevity.

B. OPTIMIZATION PROBLEM VIA HPP SOLUTIONS

Theorem 2: Suppose Ω_r be the simplex set satisfies $\Omega_r = \{\eta \in \mathbb{R}^r \mid \sum_{j=1}^r \eta_j = 1, \eta_j \geq 0\}$. If $f \in \mathbb{R}^r$ is homogeneous and positive on Ω_r , then for a sufficiently large scalar d , all the coefficients of $(\eta_1 + \eta_2 + \dots + \eta_r)^d f(\eta_1, \eta_2, \dots, \eta_r)$ are positive.

In this paper, we consider the homogeneous polynomial matrix X with degree $g \times g$ in the following form

$$X = \sum_{q \in \mathcal{K}(g)} \eta_{1+}^{q_1} \eta_{2+}^{q_2} \dots \eta_{r+}^{q_r} \sum_{p \in \mathcal{K}(g)} \eta_1^{p_1} \eta_2^{p_2} \dots \eta_r^{p_r} X_{p,q}, \quad (23)$$

where each $X_{p,q}$ is a matrix. Also, we have $\sum_{j=1}^l \eta_j = 1$, $\sum_{j=1}^l \eta_{j+} = 1$. $p = p_1 p_2 \dots p_r$, $p_1 + p_2 + \dots + p_r = g$, and $q = q_1 q_2 \dots q_r$, $q_1 + q_2 + \dots + q_r = g$.

Proposition 2: Suppose (23) is positive, then it is guaranteed that a set of sufficiently large d and d_+ exist which ensure the positiveness of all the coefficients of $(\eta_1 + \dots + \eta_r)^d (\eta_{1+} + \dots + \eta_{r+})^{d_+} X$.

Proof: This is referred to Theorem 2. Let us consider the following matrix-valued function:

$$f(\zeta_1, \zeta_2 \dots \zeta_{2r}) = \sum_{\rho \in \mathcal{K}(2g)} \zeta_1^{\rho_1} \zeta_2^{\rho_2} \dots \zeta_{2r}^{\rho_{2r}} X_\rho, \quad (24)$$

where $\rho = \rho_1 \rho_2 \dots \rho_{2r}$ such that $\rho_1 + \rho_2 + \dots + \rho_{2r} = 2\rho$. Note that each $\zeta_1^{\rho_1} \zeta_2^{\rho_2} \dots \zeta_{2r}^{\rho_{2r}}$ is a monomial and $\mathcal{K}(2g)$ is the set of $2r$ -tuples. If we rewrite

$$\zeta_j = \frac{\eta_j}{2}, \quad \zeta_{j+r} = \frac{\eta_{j+}}{2}, \quad j \in \{1, 2, \dots, r\}$$

$$X_\rho = 4^g X_{p,q}, \quad p = \rho_1 \rho_2 \dots \rho_r, \quad q = \rho_{r+1} \rho_{r+2} \dots \rho_{2r}.$$

then $X = f(\zeta_1, \zeta_2 \dots \zeta_{2r})$, $\sum_{j=1}^r \rho_j = g$, $\sum_{j=1}^r \rho_{j+r} = g$.

By the extension of scalar-valued function case in [36, Th1], we can obtain that a set of sufficiently large d and d_+ exist which ensures the positiveness of all the coefficients of $(\eta_1 + \dots + \eta_r + \eta_{1+} + \dots + \eta_{r+})^{d+d_+} X$. Since $(\eta_1 + \dots + \eta_r)^d (\eta_{1+} + \dots + \eta_{r+})^{d_+} X$ is a composition of part of terms in $(\eta_1 + \dots + \eta_r + \eta_{1+} + \dots + \eta_{r+})^{d+d_+} X$, all the coefficient of $(\eta_1 + \dots + \eta_r)^d (\eta_{1+} + \dots + \eta_{r+})^{d_+} X$ are positive. \square

We utilize the procedure in [35] based-on Pólya's theorem to further handle (15), (18), (19), and (20) in order to obtain a non-conservativeness result.

For r -tuples p and p' , we represent $p \succeq p'$ if $p_j \geq p'_j, j = 1, 2, \dots, r$. The definition for the r -tuple $e_j \in \mathcal{K}(1)$ is $e_j = 0 \dots 0 \underbrace{1}_{j\text{th}} 0 \dots 0$. Let $\pi(p) = (p_1!)(p_2!) \dots (p_r!)$.

Theorem 3: For sufficiently large scalars $d \geq 0, d_+ \geq 0$, if there exist matrices Y_p and symmetric matrices

$S_p, p \in \mathcal{K}(g)$ satisfying inequalities (25)-(28), as shown at the bottom of the page, then constraints (15), (18), (19), (20) are guaranteed.

Proof: Consider the constraint (26) firstly. Let $\Theta = \begin{bmatrix} 1 & \star \\ x(k) & S_g \end{bmatrix}$. Since $\sum_{j=1}^r h_j = 1$ and $\Theta = (h_1 + h_2 + \dots + h_r)^d \Theta$, by applying the equality

$$\sum_{p \in \mathcal{K}(g+d)} h_1^{p_1} h_2^{p_2} \dots h_r^{p_r} \frac{(g+d)!}{\pi(p)} = 1, \quad p = p_1 p_2 \dots p_r,$$

and utilizing the procedure in [35], Θ is rewritten as

$$\Theta = \sum_{p \in \mathcal{K}(g+d)} h_1^{p_1} h_2^{p_2} \dots h_r^{p_r} \begin{bmatrix} \frac{(g+d)!}{\pi(p)} & \star \\ \frac{(g+d)!}{\pi(p)} x(k) & \sum_{\substack{p' \in \mathcal{K}(d), \\ p \geq p'}} \frac{(d)!}{\pi(p')} S_{p-p'} \end{bmatrix}, \quad \forall p \in \mathcal{K}(g+d). \quad (29)$$

It is clearly shown that $\Theta \geq 0$ is guaranteed by (26).

According to Proposition 2, the proof of (25) contains two steps.

1) Handling S_g, Y_g : Let

$$\Upsilon = \begin{bmatrix} (1-\alpha)S_g & \star & \star & \star & \star \\ 0 & \alpha P_w & \star & \star & \star \\ A(k+i)S_g + B(k+i)Y_g & D(k+i) & S_{g+} & \star & \star \\ Q^{1/2}S_g & 0 & 0 & \gamma I & \star \\ R^{1/2}Y_g & 0 & 0 & 0 & \gamma I \end{bmatrix}.$$

Since $\sum_{j=1}^r h_j = 1$, $\Upsilon = (h_1 + h_2 + \dots + h_r)^d \Upsilon$. By the fact that

$$\sum_{p \in \mathcal{K}(g+d+1)} h_1^{p_1} h_2^{p_2} \dots h_r^{p_r} \frac{(g+d+1)!}{\pi(p)} = 1, \quad p = p_1 p_2 \dots p_r,$$

Υ is further rewritten as

$$\Upsilon = \sum_{p \in \mathcal{K}(g+d+1)} h_1^{p_1} h_2^{p_2} \dots h_r^{p_r} X_p, \quad p = p_1 p_2 \dots p_r,$$

$$X_p = \sum_{\substack{p' \in \mathcal{K}(d), \\ p \geq p'}} \sum_{\substack{j \in \{1, \dots, r\}, \\ p_j > p'_j}} \frac{d!}{\pi(p')}$$

$$\begin{aligned} & \frac{(g+d+1)!}{\pi(q)} \sum_{\substack{p' \in \mathcal{K}(d), \\ p \geq p'}} \sum_{\substack{j \in \{1, \dots, r\}, \\ p_j > p'_j}} \frac{d!}{\pi(p')} \begin{bmatrix} (1-\alpha)S_{p-p'-e_j} & \star & \star & \star & \star \\ 0 & 0 & \star & \star & \star \\ A_j S_{p-p'-e_j} + B_j Y_{p-p'-e_j} & D_j & 0 & \star & \star \\ Q^{1/2}S_{p-p'-e_j} & 0 & 0 & 0 & \star \\ R^{1/2}Y_{p-p'-e_j} & 0 & 0 & 0 & 0 \end{bmatrix} \\ & + \frac{(g+d+1)!}{\pi(p)} \begin{bmatrix} 0 & \star & \star & \star & \star \\ 0 & \frac{(g+d+1)!}{\pi(q)} \alpha P_w & \star & \star & \star \\ 0 & 0 & \sum_{\substack{q' \in \mathcal{K}(d_+), \\ q \geq q'}} \frac{(d_+)!}{\pi(q')} (S_{q-q'}) & \star & \star \\ 0 & 0 & 0 & \frac{(g+d+1)!}{\pi(q)} \gamma I & \star \\ 0 & 0 & 0 & 0 & \frac{(g+d+1)!}{\pi(q)} \gamma I \end{bmatrix} \geq 0, \\ & \forall p \in \mathcal{K}(g+d+1), \quad \forall q \in \mathcal{K}(g+d_+), \end{aligned} \quad (25)$$

$$\begin{bmatrix} \frac{(g+d)!}{\pi(p)} & \star \\ \frac{(g+d)!}{\pi(p)} x(k) & \sum_{p' \in \mathcal{K}(d), p \geq p'} \frac{(d)!}{\pi(p')} S_{p-p'} \end{bmatrix} \geq 0, \quad \forall p \in \mathcal{K}(g+d), \quad (26)$$

$$\begin{bmatrix} \sum_{p' \in \mathcal{K}(d), p \geq p'} \frac{(d)!}{\pi(p')} S_{p-p'} & \star \\ \sum_{p' \in \mathcal{K}(d), p \geq p'} \frac{(d)!}{\pi(p')} Y_{p-p'} & \frac{(g+d)!}{\pi(p)} Z \end{bmatrix} \geq 0, \quad Z_{ss} \leq z_{s,\text{inf}}^2, \quad s \in \{1, 2, \dots, m\}, \quad \forall p \in \mathcal{K}(g+d), \quad (27)$$

$$\begin{aligned} & \begin{bmatrix} \sum_{\substack{p' \in \mathcal{K}(d), \\ p \geq p'}} \sum_{\substack{j \in \{1, \dots, r\}, \\ p_j > p'_j}} \frac{d!}{\pi(p')} S_{p-p'-e_j} & \star \\ \sum_{p' \in \mathcal{K}(d), p \geq p'} \sum_{\substack{j \in \{1, \dots, r\}, \\ p_j > p'_j}} \frac{d!}{\pi(p')} \Psi(A_j S_{p-p'-e_j} + B_j Y_{p-p'-e_j}) & \frac{(g+d+1)!}{\pi(p)} \Gamma \end{bmatrix} \geq 0, \\ & \Gamma_{ss} \leq \psi_{s,\text{inf}}^2, \quad s \in \{1, 2, \dots, r\}, \quad p \in \mathcal{K}(g+d+1). \end{aligned} \quad (28)$$

$$\times \begin{bmatrix} (1-\alpha)S_{p-p'-e_j} & \star & \star & \star & \star \\ 0 & 0 & \star & \star & \star \\ A_j S_{p-p'-e_j} + B_j Y_{p-p'-e_j} & D_j & 0 & \star & \star \\ Q^{1/2} S_{p-p'-e_j} & 0 & 0 & 0 & \star \\ R^{1/2} Y_{p-p'-e_j} & 0 & 0 & 0 & 0 \end{bmatrix} + \frac{(g+d+1)!}{\pi(p)} \begin{bmatrix} 0 & \star & \star & \star & \star \\ 0 & \alpha P_w & \star & \star & \star \\ 0 & 0 & S_{g+} & \star & \star \\ 0 & 0 & 0 & \gamma I & \star \\ 0 & 0 & 0 & 0 & \gamma I \end{bmatrix}.$$

Then we can conclude that $\Upsilon \geq 0$, i.e., constraint (18) is guaranteed by $X_p \geq 0$.

2) Handling S_{g+} : Since $\sum_{j=1}^r h_{j+} = 1$, by the fact

$$\sum_{q \in \mathcal{K}(g+d_+)} h_{1+}^{q_1} h_{2+}^{q_2} \dots h_{r+}^{q_r} \frac{(g+d_+)!}{\pi(q)} = 1, \quad q = q_1 q_2 \dots q_r,$$

we can obtain that

$$\begin{aligned} & (h_{1+} + h_{2+} + \dots + h_{r+})^{d_+} X_p \\ &= \sum_{q \in \mathcal{K}(g+d_+)} h_{1+}^{q_1} h_{2+}^{q_2} \dots h_{r+}^{q_r} X_{p,q}, \\ X_{p,q} &= \frac{(g+d_+)!}{\pi(q)} \sum_{\substack{p' \in \mathcal{K}(d), \\ p \geq p'}} \sum_{\substack{j \in \{1, \dots, r\}, \\ p_i > p'_i}} \frac{d!}{\pi(p')} \Phi_1 \\ &+ \frac{(g+d+1)!}{\pi(p)} \Phi_2, \end{aligned}$$

where Φ_1 and Φ_2 are shown at the bottom of the page.

We can find that $X_{p,q} \geq 0$ ensures $X_p \geq 0$. Analogously, it is easy to prove that (19), (20) are satisfied if (27), (28) hold, respectively.

Thus, the proof is completed. \square

In summary, the optimization problem (21) is reexpressed as

$$\min_{S_p, Y_p, Z, \Gamma, \gamma} \gamma \quad \text{s.t.} \quad (25) - (28). \quad (30)$$

Similarly to (21), problem (30) can be solved via convex optimization if α is pre-specified. However, note that the computational burden of (30) can be much heavier but the result is non-conservative.

Corollary 3.1: If the optimization problem (30) has a feasible solution for a particular set of $\{g_0, d_0, d_{0+}\}$, then it also has for $g > g_0, d > d_0, d_+ > d_{0+}$.

Proof: see [35].

Remark 4: An alternative methodology to calculate the HPP Lyapunov solution to (30) is by only increasing g and choosing $d = 0, d_+ = 0$. However, more decision variables are emerged by increasing g while the increase in $\{d, d_+\}$ brings the larger number of LMIs. If the computational efficiency is a crucial factor, one can simply reduce g, d and d_+ to a satisfactory level.

IV. ILLUSTRATIVE EXAMPLE

Example 1: Consider the following discrete-time nonlinear model:

$$\begin{aligned} x_1(k+1) &= x_1(k) - x_1(k)x_2(k) + (5 + x_1(k))u(k) \\ &\quad + 0.5x_1(k)w(k) \\ x_2(k+1) &= -x_1(k) - 0.5x_2(k) + 2x_1(k)u(k), \end{aligned} \quad (31)$$

where $x_1(k) \in [-\beta, \beta]$ with $\beta > 0$ and disturbance $|w(k)| \leq 0.5$. Let the membership functions $h_1 = (\beta + x_1(k))/(2\beta)$ and $h_2 = (\beta - x_1(k))/(2\beta)$ be the combination coefficients. Then, the nonlinear system (31) can be represented by the T-S model in the form of (2) with

$$\begin{aligned} A_1 &= \begin{bmatrix} 1 & -\beta \\ -1 & -0.5 \end{bmatrix}, \quad A_2 = \begin{bmatrix} 1 & \beta \\ -1 & -0.5 \end{bmatrix}, \\ B_1 &= \begin{bmatrix} 5 + \beta \\ 2\beta \end{bmatrix}, \quad B_2 = \begin{bmatrix} 5 - \beta \\ -2\beta \end{bmatrix}, \\ D_1 &= \begin{bmatrix} 0.5\beta \\ 0 \end{bmatrix}, \quad D_2 = \begin{bmatrix} -0.5\beta \\ 0 \end{bmatrix}. \end{aligned}$$

The T-S model can be different by simply changing β . Larger β implies the model in a larger region, which is more difficult to control.

Choose $Q = I, R = I, P_w = 2, \alpha = 0.3, w(k) = 0.5\sin(k)$. In order to show the effectiveness of the proposed approach, (30) is solved for a variety of pairs $\{g, d, d_+\}$. For each different $\{g, d, d_+\}$, there exists a crucial β_0 that if $\beta \leq \beta_0$, then (30) becomes feasible, i.e., the T-S model can be stabilizable.

$$\begin{aligned} \Phi_1 &= \begin{bmatrix} (1-\alpha)S_{p-p'-e_j} & \star & \star & \star & \star \\ 0 & 0 & \star & \star & \star \\ A_j S_{p-p'-e_j} + B_j Y_{p-p'-e_j} & D_j & 0 & \star & \star \\ Q^{1/2} S_{p-p'-e_j} & 0 & 0 & 0 & \star \\ R^{1/2} Y_{p-p'-e_j} & 0 & 0 & 0 & 0 \end{bmatrix}, \\ \Phi_2 &= \begin{bmatrix} 0 & \star & \star & \star & \star \\ 0 & \frac{(g+d_+)!}{\pi(q)} \alpha P_w & \star & \star & \star \\ 0 & 0 & \sum_{q' \in \mathcal{K}(d_+), q \geq q'} \frac{(d_+)!}{\pi(q')} (S_{q-q'}) & \star & \star \\ 0 & 0 & 0 & \frac{(g+d_+)!}{\pi(q)} \gamma I & \star \\ 0 & 0 & 0 & 0 & \frac{(g+d_+)!}{\pi(q)} \gamma I \end{bmatrix}, \quad \forall p \in \mathcal{K}(g+d+1), \quad \forall q \in \mathcal{K}(g+d_+). \end{aligned}$$

TABLE 1. Feasible values of β_0 (β_0 is feasible when $\beta \leq \beta_0$).

g	0	1	1	1	2	2	2	3	3
d	0	0	1	3	0	1	2	0	1
d_+	0	0	1	1	0	0	2	0	1
β_0	0.73	1.28	1.47	1.52	1.48	1.52	1.55	1.57	1.57

Different pairs of $\{g, d, d_+\}$ and β_0 are chosen for the simulation and they are listed in Table 1. It can be inferred that as the values of $\{g, d, d_+\}$ increase, the stabilizable range of T-S model is enlarged by applying the approach in this paper. Moreover, in order to get the same β_0 , one can increase g while maintaining d and d_+ to be zero, or increase the set $\{g, d, d_+\}$ as a whole.

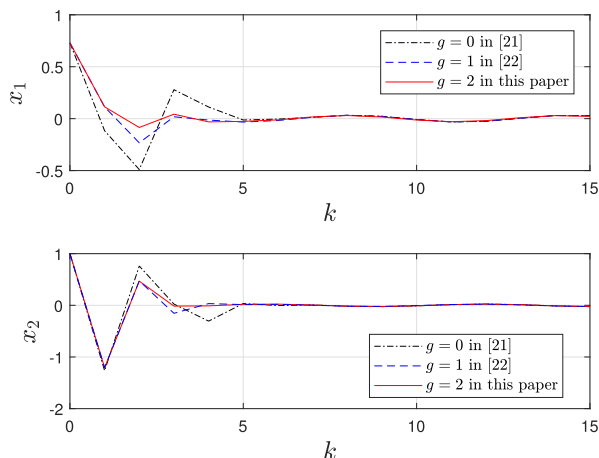


FIGURE 1. Trajectories of state via different g .

To illustrate the effectiveness of different g , three special cases with $\{d = 0, d_+ = 0\}$ for the same T-S model are considered, i.e., the case $g = 0$ that is applied in benchmark work [24], the case $g = 1$ that is applied in another work [25], and case $g = 2$ that is tuned in this paper. Choose $\beta_0 = 0.73$, $x(0) = [0.73; 1]$. The resultant trajectories of states and inputs are shown in Figures 1 and 2, respectively. Define the cumulated cost $J_{sum} = \sum_{k=1}^k [\|x(k_1)\|_Q^2 + \|u(k_1)\|_R^2]$ to assess the control performance. Calculate J_{sum} for $g = 0, g = 1, g = 2$, respectively. As can be seen in Figure 3, the value of J_{sum} tends to be smaller as g increases. Hence, it is concluded that a larger value of g brings the performance improvement.

To validate the effectiveness of $\{d, d_+\}$, we consider three different cases with fixed $g = 1$, i.e., $\{d = 0, d_+ = 0\}$ in [25], $\{d = 3, d_+ = 0\}$, and $\{d = 6, d_+ = 3\}$. Choose $\beta_0 = 0.73$, $x(0) = [0.73; 1]$. The resultant trajectories of states and inputs are shown in Figures 4 and 5, respectively. Compute J_{sum} for $\{d = 0, d_+ = 0\}, \{d = 3, d_+ = 0\}$ and $\{d = 6, d_+ = 3\}$, respectively, and the results are shown in Figure 6. It can be observed that, the control performance is improved with $\{d, d_+\}$ increasing.

Example 2: Consider another benchmark example, i.e., a continuous stirred tank reactor (CSTR) whose continuous-

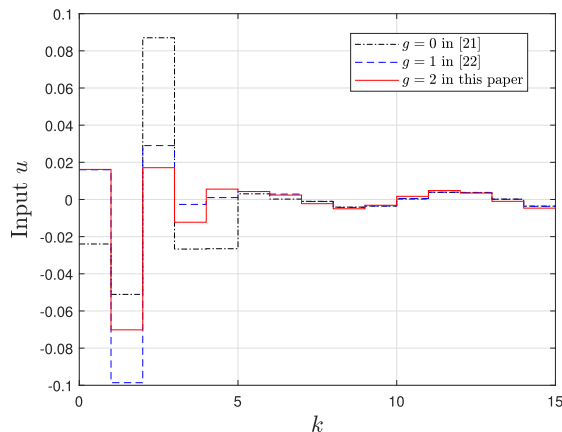


FIGURE 2. The control input signal via different g .

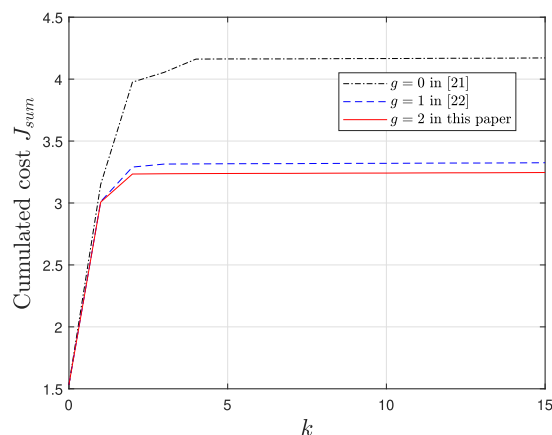


FIGURE 3. Performances index J_{sum} via different g .

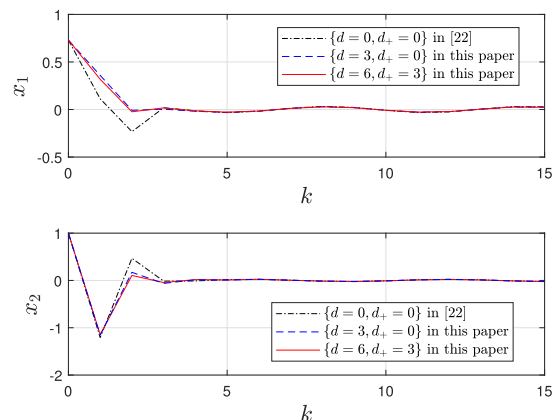


FIGURE 4. Trajectories of state via different pairs $\{d, d_+\}$.

time dynamics is

$$\begin{aligned}
 \dot{C}_A &= \frac{q}{V}(C_{Af} - C_A) - k_0 \exp\left(-\frac{E}{RT}\right) C_A + Dw \\
 \dot{T} &= \frac{q}{V}(T_f - T) + \frac{(-\Delta H)}{\rho C_p} k_0 \exp\left(-\frac{E}{RT}\right) C_A \\
 &\quad + \frac{UA}{V\rho C_p}(T_c - T)
 \end{aligned} \tag{32}$$

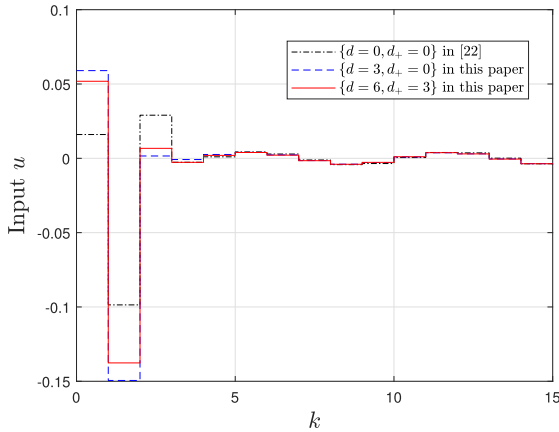


FIGURE 5. The control input signal via different pairs $\{d, d_+\}$.

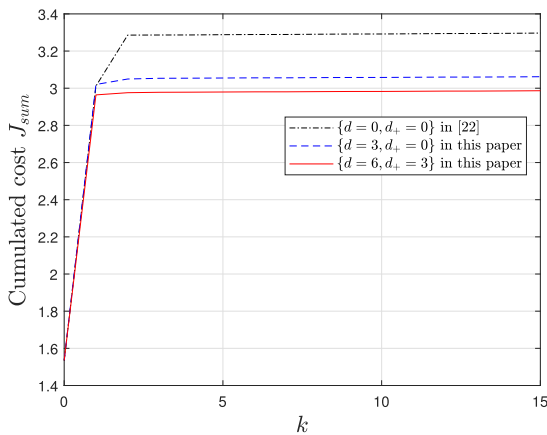


FIGURE 6. Performances index J_{sum} via different pairs $\{d, d_+\}$.

where C_A is the concentration of irreversible exothermic reaction A in the reactor, T is the measurable reactor temperature, T_c is the temperature of the coolant stream. The disturbance $|w| \leq 1$. The objective is to regulate C_A and T by manipulating T_c . Corresponding parameters used are summarized in Table 2.

TABLE 2. System parameters.

Parameter	Value	Units
q	100	l/min
V	100	l
C_{Af}	0.9	mol/l
k_0	3.456×10^{10}	min^{-1}
E/R	8750	K
T_f	350	K
$-\Delta H$	2.5×10^4	J/mol
ρ	1000	g/l
C_p	0.239	J/gK
UA	5×10^4	J/minK
D	0.01	-

The constraints are $328\text{K} \leq T_c \leq 348\text{K}$, $340\text{K} \leq T \leq 360\text{K}$, $0 \leq C_A \leq 1\text{mol/l}$. Denote the non-zero equilibrium as $\{C_A^{eq}, T^{eq}, T_c^{eq}\}$ where $C_A^{eq} = 0.5\text{mol/l}$, $T^{eq} = 350\text{K}$, $T_c^{eq} = 338\text{K}$. Define the state vector $x = [C_A - C_A^{eq}, T - T^{eq}]^T$ and the input $u = T_c - T_c^{eq}$. Denote the constraint

on x_2 as $\underline{x}_2 \leq x_2 \leq \bar{x}_2$. Moreover, define $\varphi_1(x_2) = k_0 \exp(-(E/R)/(x_2 + T^{eq}))$, $\varphi_2(x_2) = k_0 [\exp(-(E/R)/(x_2 + T^{eq})) - \exp(-(E/R)/T^{eq})] C_A^{eq} \frac{1}{x_2}$, $H_1 = \frac{1}{2}(\varphi_1(x_2) - \varphi_1(\underline{x}_2))/(\varphi_1(\bar{x}_2) - \varphi_1(\underline{x}_2))$, $H_2 = \frac{1}{2}(\varphi_1(\bar{x}_2) - \varphi_1(x_2))/(\varphi_1(\bar{x}_2) - \varphi_1(\underline{x}_2))$, $H_3 = \frac{1}{2}(\varphi_2(x_2) - \varphi_2(\underline{x}_2))/(\varphi_2(\bar{x}_2) - \varphi_2(\underline{x}_2))$, $H_4 = \frac{1}{2}(\varphi_2(\bar{x}_2) - \varphi_2(x_2))/(\varphi_2(\bar{x}_2) - \varphi_2(\underline{x}_2))$. By discretizing the continuous-time model with sampling period $\Delta t_s = 0.2\text{minute}$, the nonlinear system (32) can be approximated by the following four rules of T-S fuzzy model ($h_1 = H_1$, $h_2 = H_2$, $h_3 = H_3$, and $h_4 = H_4$):

Rule 1: IF $x_2(k)$ is H_1 , THEN

$$x(k+1) = \sigma \begin{bmatrix} 0.5347 & -0.0073 \\ 27.7559 & 0.7640 \end{bmatrix} x(k) + \sigma \begin{bmatrix} 0 \\ 0.4184 \end{bmatrix} u(k) + \sigma \begin{bmatrix} 0.01 \\ 0 \end{bmatrix} w(k)$$

Rule 2: IF $x_2(k)$ is H_2 , THEN

$$x(k+1) = \sigma \begin{bmatrix} 0.8271 & -0.0073 \\ -2.8341 & 0.7640 \end{bmatrix} x(k) + \sigma \begin{bmatrix} 0 \\ 0.4184 \end{bmatrix} u(k) + \sigma \begin{bmatrix} 0.01 \\ 0 \end{bmatrix} w(k)$$

Rule 3: IF $x_2(k)$ is H_3 , THEN

$$x(k+1) = \sigma \begin{bmatrix} 0.6809 & -0.0060 \\ 12.4609 & 1.0059 \end{bmatrix} x(k) + \sigma \begin{bmatrix} 0 \\ 0.4184 \end{bmatrix} u(k) + \sigma \begin{bmatrix} 0.01 \\ 0 \end{bmatrix} w(k)$$

Rule 4: IF $x_2(k)$ is H_4 , THEN

$$x(k+1) = \sigma \begin{bmatrix} 0.6809 & -0.0013 \\ 12.4609 & 0.5220 \end{bmatrix} x(k) + \sigma \begin{bmatrix} 0 \\ 0.4184 \end{bmatrix} u(k) + \sigma \begin{bmatrix} 0.01 \\ 0 \end{bmatrix} w(k)$$

Then choose $Q = I$, $R = I$, $w(k) = \sin(k)$.

We use another way to modify the model through multiplying submodels by σ simultaneously. The maximal values of σ for the stabilizable T-S models are listed in Table 3. From the Table 3, it is shown that with values of $\{g, d, d_+\}$ increasing, σ becomes larger. Thus, the stabilizable model range is enlarged.

TABLE 3. Feasible values of maximal σ .

g	0	1	1	2	2
d	0	0	1	0	1
d_+	0	0	0	0	1
σ_{max}	1.28	1.45	1.5	1.73	1.75

To illustrate the effectiveness via different complexity degrees g , we consider three cases for the same T-S model, i.e., $\{g = 0, d = 0, d_+ = 0\}$, $\{g = 1, d = 0, d_+ = 0\}$, $\{g = 2, d = 0, d_+ = 0\}$. Choose $\sigma = 1$, $x(0) = [0.2, 4]^T$. The trajectories of states and input are depicted in Figures 7–8. Figures 7–8 show that the state and input evolve to the neighborhood of the origin without the constraints violation over the whole simulation horizon, so the closed-loop system is nonquadratically bounded.

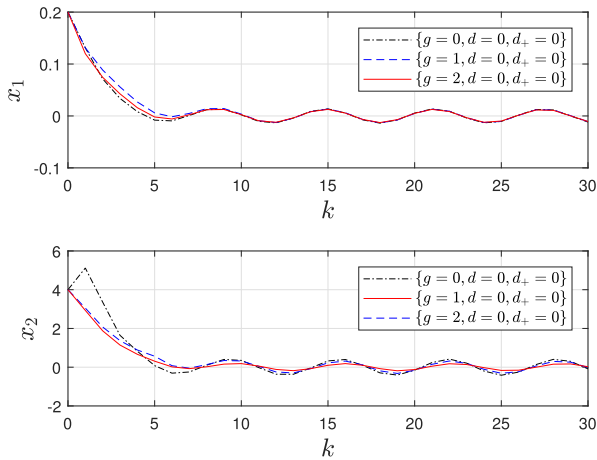


FIGURE 7. Responses of states via different g .

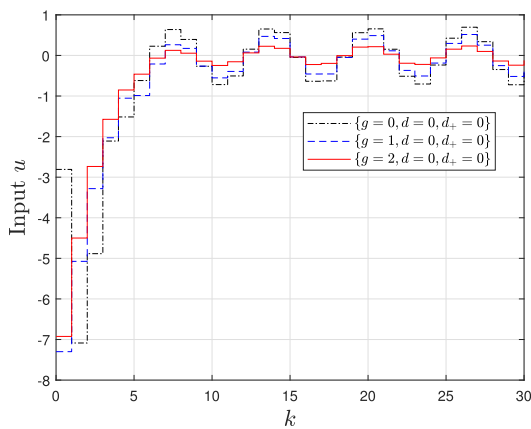


FIGURE 8. The control input signal via different g .

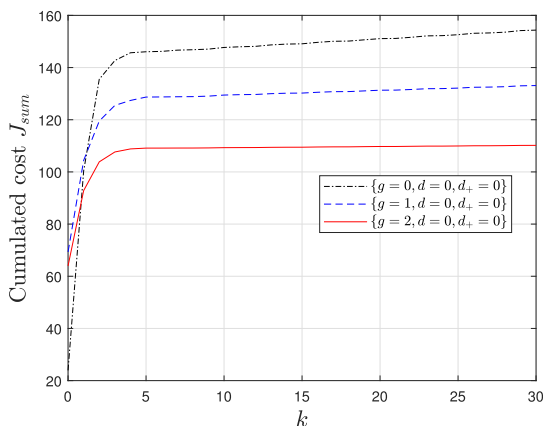


FIGURE 9. Performance index J_{sum} via different g .

Choose $J_g = \sum_{k=0}^{30} [\|x(k)\|_Q^2 + \|u(k)\|_R^2]$ as the performance criterion. Calculate J_g via $g = 0, g = 1, g = 2$, respectively, and we obtain that $J_{(g=0)} = 154.3521, J_{(g=1)} = 133.0989, J_{(g=2)} = 110.1759$ (see Figure 9). As being shown, J_g reduces as the value of g increases. It implies that the control performance is improved with increasing the value of g .

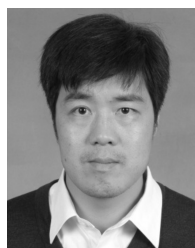
V. CONCLUSION

In this paper, a less conservative MPC approach for T-S fuzzy model with bounded disturbance is proposed. A general form of HPP Lyapunov function and the corresponding HPP control law are adopted to extend the previous approaches which are taken as special cases in this paper. The complexity degree is allowed to be tuned in order to balance the control performance and the computational efficiency. The proposed technique brings less conservatism as well as enlarged stabilizable model range. The systems subject to measurement noises widely exist in practice, so our future attention is on extending the proposed method to other systems, such as Markov systems [38], [39], the linear parameter-varying systems [40], and the networked control systems [41].

REFERENCES

- [1] T. Takagi and M. Sugeno, “Fuzzy identification of systems and its applications to modeling and control,” *IEEE Trans. Syst., Man, Cybern.*, vol. SMC-15, no. 1, pp. 116–132, Jan. 1985.
- [2] H. K. Lam, “A review on stability analysis of continuous-time fuzzy-model-based control systems: From membership-function-independent to membership-function-dependent analysis,” *Eng. Appl. Artif. Intell.*, vol. 67, pp. 390–408, Jan. 2018.
- [3] F. R. López-Estrada, D. Rotondo, and G. Valencia-Palomo, “A review of convex approaches for control, observation and safety of linear parameter varying and Takagi–Sugeno systems,” *Processes*, vol. 7, no. 11, pp. 814–853, 2019.
- [4] C.-H. Fang, Y.-S. Liu, S.-W. Kau, L. Hong, and C.-H. Lee, “A new LMI-based approach to relaxed quadratic stabilization of T-S fuzzy control systems,” *IEEE Trans. Fuzzy Syst.*, vol. 14, no. 3, pp. 386–397, Jun. 2006.
- [5] K. Tanaka, T. Hori, and H. O. Wang, “A multiple Lyapunov function approach to stabilization of fuzzy control systems,” *IEEE Trans. Fuzzy Syst.*, vol. 11, no. 4, pp. 582–589, Aug. 2003.
- [6] G. Feng, C. L. Chen, D. Sun, and Y. Zhu, “ H_∞ controller synthesis of fuzzy dynamic systems based on piecewise Lyapunov functions and bilinear matrix inequalities,” *IEEE Trans. Fuzzy Syst.*, vol. 13, no. 1, pp. 94–103, Feb. 2005.
- [7] T. M. Guerra and L. Vermeiren, “LMI-based relaxed nonquadratic stabilization conditions for nonlinear systems in the Takagi–Sugeno’s form,” *Automatica*, vol. 40, no. 5, pp. 823–829, May 2004.
- [8] L. Kong, W. He, C. Yang, Z. Li, and C. Sun, “Adaptive fuzzy control for coordinated multiple robots with constraint using impedance learning,” *IEEE Trans. Cybern.*, vol. 49, no. 8, pp. 3052–3063, Aug. 2019.
- [9] X. Yu, W. He, H. Li, and J. Sun, “Adaptive fuzzy full-state and output-feedback control for uncertain robots with output constraint,” *IEEE Trans. Syst., Man, Cybern., Syst.*, vol. 51, no. 11, pp. 6994–7007, Nov. 2021.
- [10] D. H. Lee, J. B. Park, and Y. H. Joo, “Improvement on nonquadratic stabilization of discrete-time Takagi–Sugeno fuzzy systems: Multiple-parameterization approach,” *IEEE Trans. Fuzzy Syst.*, vol. 18, no. 2, pp. 425–429, Apr. 2010.
- [11] X. Xie, H. Ma, Y. Zhao, D.-W. Ding, and Y. Wang, “Control synthesis of discrete-time T-S fuzzy systems based on a novel non-PDC control scheme,” *IEEE Trans. Fuzzy Syst.*, vol. 21, no. 1, pp. 147–157, Feb. 2013.
- [12] M. Jungers, R. C. L. F. Oliveira, and P. L. D. Peres, “MPC for LPV systems with bounded parameter variations,” *Int. J. Control*, vol. 84, no. 1, pp. 24–36, 2011.
- [13] D. Q. Mayne, “Model predictive control: Recent developments and future promise,” *Automatica*, vol. 50, no. 12, pp. 2967–2986, 2014.
- [14] B. Jin, H. Li, W. Yan, and M. Cao, “Distributed model predictive control and optimization for linear systems with global constraints and time-varying communication,” *IEEE Trans. Autom. Control*, vol. 66, no. 7, pp. 3393–3400, Jul. 2021.
- [15] Y. Yang, Y. Zou, and S. Li, “Economic model predictive control of enhanced operation performance for industrial hierarchical systems,” *IEEE Trans. Ind. Electron.*, early access, Jun. 16, 2021, doi: 10.1109/TIE.2021.3088334.

- [16] J. Lu, Y. Xi, and D. Li, "Stochastic model predictive control for probabilistically constrained Markovian jump linear systems with additive disturbance," *Int. J. Robust Nonlinear Control*, vol. 29, no. 15, pp. 5002–5016, Oct. 2019.
- [17] Q. Lu, P. Shi, H.-K. Lam, and Y. Zhao, "Interval type-2 fuzzy model predictive control of nonlinear networked control systems," *IEEE Trans. Fuzzy Syst.*, vol. 23, no. 6, pp. 2317–2328, Dec. 2015.
- [18] D. Lee and J. Hu, "Local model predictive control for T-S fuzzy systems," *IEEE Trans. Cybern.*, vol. 47, no. 9, pp. 2556–2567, Sep. 2017.
- [19] L. Teng, Y. Wang, W. Cai, and H. Li, "Robust fuzzy model predictive control of discrete-time Takagi–Sugeno systems with nonlinear local models," *IEEE Trans. Fuzzy Syst.*, vol. 26, no. 5, pp. 2915–2925, Oct. 2018.
- [20] I. Boukkaibet, K. Belarbi, S. Bououden, T. Marwala, and M. Chadli, "A new T-S fuzzy model predictive control for nonlinear processes," *Expert Syst. Appl.*, vol. 88, pp. 132–151, Dec. 2017.
- [21] X. Ping, J. Yao, B.-C. Ding, P. Wang, and Z. Li, "Time-varying tube-based output feedback robust MPC for T-S fuzzy systems," *IEEE Trans. Fuzzy Syst.*, early access, Feb. 25, 2021, doi: [10.1109/TFUZZ.2021.3062149](https://doi.org/10.1109/TFUZZ.2021.3062149).
- [22] S. Bououden, M. Chadli, and H. R. Karimi, "A robust predictive control design for nonlinear active suspension systems," *Asian J. Control*, vol. 18, no. 1, pp. 122–132, Jan. 2016.
- [23] M. Killian, B. Mayer, A. Schirrer, and M. Kozek, "Cooperative fuzzy model-predictive control," *IEEE Trans. Fuzzy Syst.*, vol. 24, no. 2, pp. 471–482, Apr. 2016.
- [24] M. V. Kothare, V. Balakrishnan, and M. Morari, "Robust constrained model predictive control using linear matrix inequalities," *Automatica*, vol. 32, no. 10, pp. 1361–1379, 1996.
- [25] E. Garone and A. Casavola, "Receding horizon control strategies for constrained LPV systems based on a class of nonlinearly parameterized Lyapunov functions," *IEEE Trans. Autom. Control*, vol. 3, no. 1, pp. 2354–2360, Sep. 2012.
- [26] S. V. Raković, B. Kouvaritakis, R. Findeisen, and M. Cannon, "Homothetic tube model predictive control," *Automatica*, vol. 48, no. 8, pp. 1631–1638, Aug. 2012.
- [27] S. Yu, C. Maier, H. Chen, and F. Allgöwer, "Tube MPC scheme based on robust control invariant set with application to Lipschitz nonlinear systems," *Syst. Control Lett.*, vol. 62, no. 2, pp. 194–200, Feb. 2013.
- [28] X. Liu, D. Constantinescu, and Y. Shi, "Robust model predictive control of constrained non-linear systems: Adopting the non-squared integrand objective function," *IET Control Theory Appl.*, vol. 9, no. 5, pp. 649–658, Mar. 2015.
- [29] D.-F. He, H. Huang, and Q.-X. Chen, "Quasi-min–max MPC for constrained nonlinear systems with guaranteed input-to-state stability," *J. Franklin Inst.*, vol. 351, no. 6, pp. 3405–3423, Jun. 2014.
- [30] T. Zou and S. Li, "Stabilization via extended nonquadratic boundedness for constrained nonlinear systems in Takagi–Sugeno's form," *J. Franklin Inst.*, vol. 348, no. 10, pp. 2849–2862, Dec. 2011.
- [31] B. Ding and H. Pan, "Output feedback robust MPC for LPV system with polytopic model parametric uncertainty and bounded disturbance," *Int. J. Control*, vol. 89, no. 8, pp. 1554–1571, 2016.
- [32] J. Hu and B. Ding, "Output feedback robust MPC for linear systems with norm-bounded model uncertainty and disturbance," *Automatica*, vol. 108, Oct. 2019, Art. no. 108489.
- [33] W. Yang, G. Feng, and T. Zhang, "Robust model predictive control for discrete-time Takagi–Sugeno fuzzy systems with structured uncertainties and persistent disturbances," *IEEE Trans. Fuzzy Syst.*, vol. 22, no. 5, pp. 1213–1228, Oct. 2014.
- [34] Y. Yang and B. Ding, "Model predictive control for LPV models with maximal stabilizable model range," *Asian J. Control*, vol. 22, no. 5, pp. 1940–1950, Sep. 2020.
- [35] R. C. L. F. Oliveira and P. L. D. Peres, "Parameter-dependent LMIs in robust analysis: Characterization of homogeneous polynomially parameter-dependent solutions via LMI relaxations," *IEEE Trans. Autom. Control*, vol. 52, no. 7, pp. 1334–1340, Jul. 2007.
- [36] V. Powers and B. Reznick, "A new bound for Pólya's theorem with applications to polynomials positive on polyhedra," *J. Pure Appl. Algebra*, vol. 164, nos. 1–2, pp. 221–229, Oct. 2001.
- [37] A. Alessandri, M. Baglietto, and G. Battistelli, "On estimation error bounds for receding-horizon filters using quadratic boundedness," *IEEE Trans. Autom. Control*, vol. 49, no. 8, pp. 1350–1355, Aug. 2004.
- [38] G. Zong, W. Qi, and H. R. Karimi, " L_1 control of positive semi-Markov jump systems with state delay," *IEEE Trans. Syst., Man, Cybern., Syst.*, vol. 51, no. 12, pp. 7569–7578, Dec. 2020, doi: [10.1109/TSMC.2020.2980034](https://doi.org/10.1109/TSMC.2020.2980034).
- [39] G. Zong, Y. Li, and H. Sun, "Composite anti-disturbance resilient control for Markovian jump nonlinear systems with general uncertain transition rate," *Sci. China Inf. Sci.*, vol. 62, no. 2, pp. 309–317, Feb. 2019.
- [40] D. Yang, G. Zong, S. K. Nguang, and X. Zhao, "Bumpless transfer H_∞ anti-disturbance control of switching Markovian LPV systems under the hybrid switching," *IEEE Trans. Cybern.*, early access, Oct. 15, 2020, doi: [10.1109/TCYB.2020.3024988](https://doi.org/10.1109/TCYB.2020.3024988).
- [41] G. Zong, H. Ren, and H. R. Karimi, "Event-triggered communication and annular finite-time H_∞ filtering for networked switched systems," *IEEE Trans. Cybern.*, vol. 51, no. 1, pp. 309–317, Jan. 2021.



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