

Received September 12, 2021, accepted September 24, 2021, date of publication September 29, 2021, date of current version October 11, 2021.

Digital Object Identifier 10.1109/ACCESS.2021.3116706

Non-Cooperative Game Based Power Allocation for Energy and Spectrum Efficient Downlink NOMA HetNets

AHMED NASSER¹, (Member, IEEE), OSAMU MUTA², (Member, IEEE), HARIS GACANIN³, (Fellow, IEEE), AND MAHA ELSABROUTY⁴, (Senior Member, IEEE)

¹Department of Electronics and Communication, Faculty of Engineering, Suez Canal University, Ismailia 41522, Egypt

²Center for Japan-Egypt Cooperation in Science and Technology, Kyushu University, Fukuoka 812-0395, Japan

³Institute for Communication Technologies and Embedded Systems, RWTH Aachen University, 52062 Aachen, Germany

⁴Department of Electronics and Communication, Egypt-Japan University of Science and Technology, Alexandria 21934, Egypt

Corresponding author: Ahmed Nasser (ahmed.nasser@mobcom.ait.kyushu-u.ac.jp)

This work was supported in part by the Support Center for Advanced Telecommunications Technology Research Foundation, in part by the Japan Society for the Promotion of Science (JSPS) KAKENHI under Grant JP17K06427, in part by the Research Project "Super-HETs, Empowering 5G HetNets for better Performance" funded by the National Telecom Regulatory Authority (NTRA) Egypt, in part by Kyushu University Platform of Inter/Transdisciplinary Energy Research (Q-PIT), and in part by Suez Canal University.

ABSTRACT In this paper, the tradeoff between spectrum efficiency (SE) and energy efficiency (EE) is investigated in terms of interference management and power allocation for heterogeneous networks (HetNets) with non-orthogonal multiple access (NOMA). The EE and SE tradeoff is modeled as a multi-objective problem (MOP) under the maximum power and quality of service (QoS) constraints, which is non-convex. The MOP is relaxed into a convex single objective problem (SOP) by adopting a weighted sum strategy with the hypograph transformation. The SOP is solved in two steps. In the first step, we propose a power allocation technique based on non-cooperative (NC) game for EE and SE in NOMA HetNets. In the proposed NC game, the macro base station (MBS) and the small BSs (SBSs) compete with an equal priority in order to optimize their transmit powers towards maximizing the weighted sum of SE and EE. In the second step, a closed-form formula is proposed to control the power allocated to users while taking into account both QoS constraint and successive interference cancellation (SIC) condition. From simulations, the proposed technique can, in some dedicated settings, considerably improve the tradeoff between EE and SE over conventional techniques.

INDEX TERMS Heterogeneous networks (HetNets), energy efficiency (EE), spectrum efficiency (SE), interference mitigation, power control (PC), non-cooperative (NC) game.

I. INTRODUCTION

Towards more efficient communication systems, spectrum-efficient (SE) and energy-efficient (EE) cellular systems need to be maximized to meet the critical demand for high data rates while saving energy for the green communication objective. Two notable wireless technologies can be deployed to achieve this end. The first one is the heterogeneous networks (HetNets), where the SE can be achieved by deploying small-cell (SCs) tiers with a short-range small base station (SBS) under the coverage of a macro-cell (MC) tier with a powerful macro BS (MBS) [1]. The second technology

is the non-orthogonal multiple access (NOMA), where users are multiplexed on the same time/frequency/space resources while distinguishing them by allocating different power levels to users according to several criteria including quality of service (QoS) or relative channel gains [2], [3]. NOMA improves SE and EE at the cost of inter-user interference. However, due to the resource sharing among different tiers, HetNets with NOMA suffers from co-tier and cross-tier interference [4]–[7], while acquiring the advantages of NOMA HetNets depends on mitigating these types of interference.

One way to manage both co-tier and cross-tier interferences is to control the allocated power to the adopted BSs in NOMA HetNets such that each BS transmits

The associate editor coordinating the review of this manuscript and approving it for publication was Jiafeng Xie.

a proper power level to sustain its users' QoS without imposing excessive interference to users of other cells [8], [9]. However, the power control problem can itself be a dilemma since adjusting the transmit power has a contradictory effect on both SE and EE [10], [11]. On one hand, increasing the transmit power from one of the BSs has a positive impact on its spectral efficiency (i.e., SE) as it reduces the probability of outage and allows for higher-order modulation to be utilized. However, interference on the other cells caused by increasing the transmit power has a negative impact on energy efficiency (i.e., EE), which needs to be improved to conform to the green communication objective. Consequently, a tradeoff exists such that sacrificing SE can be reflected as an EE gain or vice versa [12]. In other words, the power allocated to the system cannot be optimized to improve both EE and SE simultaneously. A Pareto optimal solution for the power control problem can be attained by solving a multi-objective problem (MOP) that maximizes both SE and EE [10], [12]. However, the MOP that represents SE-EE tradeoff is non-convex and nonlinear problem, which is very computationally costly to be solved in the MOP form. One way to solve the EE-SE tradeoff problem is to relax the original MOP into a single objective problem (SOP) using the weighted sum strategy, through which the direction of the optimization problem is dynamically changed according to the application demands or surrounding circumstances. Among different techniques for PC, utilizing the non-cooperative (NC) game can reduce the complexity of the PC without significantly increasing the required signaling sharing overhead among different cells. The NC game based PC has been proposed to maximize only EE in [8] or only SE in [9]. However, the NC game has not been adopted before to jointly maximize both EE and SE.

In this paper, we propose a low-complex and a fast convergence power control (PC) technique based on the NC game to maximize both SE and EE (NC-EE-SE) in HetNets with NOMA. Unlike the existing works, the proposed technique considers a general NOMA-HetNets model, where both co-tier and cross-tier interference can be managed by jointly optimizing the allocated power to MBS and a general number of SBSs with a general number of NOMA users. The proposed algorithm is able to find a better tradeoff point that improves EE and SE in comparison with the state-of-the-art PC based NC game techniques. The main contributions of this paper are summarized as follows:

- We formulate the tradeoff between SE and EE in terms of MOP, where SE and EE are jointly maximized in HetNets with NOMA. In the formulated MOP, the allocated power to the MBS and all SBSs are optimized under the maximum power and minimum QoS constraints, which is non-convex. The non-convexity of the MOP is relaxed into a convex SOP by applying the weighted sum strategy with hypograph transformation.
- The formulated SOP problem is solved in two steps. In the first step, a PC technique is proposed based on the non-cooperative (NC) game. In the proposed NC game, SBSs and MBS fairly compete for

optimizing the price of their transmit power to maximize the weighted sum of SE and EE simultaneously and independently. In the second step, a closed-form formula is proposed to control the power allocated to NOMA users at each BS independently. The proposed formula considers both QoS constraint and successive interference cancellation (SIC) condition simultaneously.

- Through the numerical results, the tradeoff is confirmed such that the EE improvement is achieved at the cost of SE or vice versa. The proposed algorithm is compared with the upper and lower limits of the NC algorithms in [8], [9]. Our results show that the proposed technique can, in some dedicated settings, considerably improve the tradeoff between EE and SE over conventional techniques.

II. RELATED WORKS

The interference problem has been widely studied in HetNets in terms of maximizing the SE by properly controlling the power allocated to the SBSs and MBS as in [5], [13]. Authors in [5] propose a distributed PC algorithm with a user scheduling scheme, while authors in [13] propose a PC algorithm based on the compressive sensing theory to improve the SE of HetNets. Game theory is also utilized for SE-based PC in HetNets as in [4], [7], [9], where the many-to-one matching game for PC is proposed in [4], and the leader-follower Stackelberg game is proposed in [7]. In addition, a non-cooperative game-based PC is proposed in [9], where the game is performed only between the MBS and one SBS (not all SBSs) that has the worst channel condition user. In [14], a joint transmission coordinated multi-point (JT-CoMP) scheme is designed for NOMA HetNets, where SE is maximized by allowing users to benefit from multi-connectivity of CoMP. Then, a mixed-integer monotonic optimization and sequential programming is proposed to solve the PC problem. In [15], the joint optimization of user association and power control is formulated as a mixed integer programming problem for SE maximization, where Lagrange duality theory is applied to solve the formulated problem. However, the literature [4], [7], [9], [14], [15] investigate the PC only for SE maximization and do not include EE in the problem formulation. In other words, although these PC techniques can guarantee the minimum QoS for all users, they are not energy-efficient.

On the other hand, maximizing EE by controlling the allocated power for interference management in HetNets has been studied in literature [8], [16]–[24]. A Dinkelbach based method is utilized to solve the HetNets EE maximization problem in [16]–[18], where the Dinkelbach is combined with PC time switching control in [16] and combined with the Lagrange dual decomposition (LDD) method to obtain a closed form expression for the optimal PC in [17]. Moreover, Dinkelbach is joined with a PC non-cooperative game in [18]. Also, the PC-based non-cooperative game is proposed in [8], where the MBS and SBS compete to maximize their EE. Moreover, in [19], the Stackelberg game is utilized for EE maximization with frequency allocation optimization.

Authors in [20] maximize the EE by jointly considering PC with interference alignment (IA). Moreover, a particle swarm optimization (PSO) is proposed in [21], while a convolution neural network-based scheme is proposed in [22] for EE-based PC in HetNets. In [23], sequential quadratic programming (SQP) is utilized to estimate the optimal power while maximizing EE with QoS in NOMA HetNets. In [24], the joint problem of PC and user association to maximize the EE is formulated as a fractional programming problem. However, for [8], [16]–[24], the objective function of the maximization problem considers only the EE while the QoS is considered as a constraint.

To jointly optimize SE and EE, multi-objective optimization algorithms are needed through which we can find the optimal tradeoff points, i.e., Pareto-optimal solution. Since the priority for SE and EE is the same, the weighted sum is an appropriate strategy to convert the MOP into an SOP as in [10]–[12], [25], [26]. Authors in [25] propose a Dinkelbach-based iterative approach to maximize the weighted sum EE-SE problem in single-cell OFDM systems. Authors in [10] propose a dual Lagrangian-based PC algorithm to maximize the weighted-sum EE-SE problem among SCs only for co-tier interference management. However, [10] does not consider the cross-tier interference. In [11], the dual Lagrangian algorithm is utilized in HetNet operating in reverse time division duplex (RTDD) to avoid the cross-tier interference. Thus, in [11], the cross-tier interference does not contribute to the PC problem. In [12], the authors propose a Levenberg-Marquardt-based PC algorithm to maximize the weighted sum problem while a fractional frequency reuse is utilized. However, in [12], the co-tier interference is not considered while optimizing the power allocated to the BSs. Authors in [26] divide the weighted-sum problem into several subproblems that can be solved separately using the concave-convex procedure (CCCP). However, [26] does not consider co-tier interference. In [27], the stochastic geometry is employed to model and analyze a spectrum-aware energy efficiency cognitive D2D communication within HetNets. However, NOMA is not included in the analysis carried out in [27]. From the above literature, we can conclude that the tradeoff between SE and EE in NOMA HetNets has not been investigated sufficiently, where both co-tier and cross-tier interference contribute to the PC problem. Besides, the complexity of the above EE-SE algorithms is still high. Thus, more investigation is needed.

III. HetNets SYSTEM MODEL

Fig. 1 shows the downlink two-tier NOMA HetNets considered in this paper, where a set of N_{SC} SCs tiers, denoted by $SC \triangleq \{1, \dots, N_{SC}\}$, each with a single-antenna SBS are uniformly distributed under the coverage of a single MC of MBS with single-antenna. Regarding the resource allocation per time/frequency slot, the MBS and each SBS serve a set of N_{SU} small users (SUs) and N_{MU} macro users (MUs) denoted by $SU \triangleq \{1, \dots, N_{SU}\}$ and $MU \triangleq \{1, \dots, N_{MU}\}$, respectively. NOMA is adopted in both MC and SCs, where users



FIGURE 1. The proposed NOMA HetNets system model with an illustration of the considered types of interference.

with the worst channel conditions are decoded first and then sequentially subtracted from the received signal¹ [7], [28]. For system simplicity, we assume that all users in a given BS are grouped into one NOMA cluster.² All BSs are proposed to reuse the same time/frequency resources. Also, the channel state information (CSI) between the user and their BS, and that between the user and its interfering BSs are assumed to be shared with a central control unit (CCU) that allocates the power to BSs. A predetermined steps of fixed user association and user pairing are assumed. Notations used in this paper are summarized in Table 1.

TABLE 1. Definition of notations.

Notation	definition
\mathbf{X}	Vectors are denoted by boldface letters.
SC, N_{SC}	The set of SCs and the number of SCs.
SC_i, SBS_i	The i^{th} SC and the i^{th} SBS.
MU, SU	The sets of MC's users and SC's users.
N_{MU}, N_{SU}	Number of MUs, and SUs per SC.
$MU_n, SU_{i,n}$	The n^{th} user in the MC and SC_i .
$\alpha_n^{[M]}, \alpha_{i,n}^{[S]}$	Power allocation coefficients for MU_n and $SU_{i,n}$, respectively.
$(\cdot)^T, (\cdot)^H, \#(\cdot)$	Transpose, conjugate transpose, and cardinality operators.
$ \cdot , a \setminus b$	Absolute operator, and excluding 'b' from 'a'
$\mathbb{R}^{a \times b}, \mathbb{C}^{a \times b}$	Real, and complex fields of dimension $a \times b$.

A. MATHEMATICAL SIGNAL MODELLING

Let us consider $x^{[M]} = \sum_{n=1}^{N_{MU}} x_n^{[M]}$ and $x_i^{[S]} = \sum_{n=1}^{N_{SU}} x_{i,n}^{[S]}$ are the transmit superimposed NOMA signal from the MBS and the SBS_i , respectively, where $x_n^{[M]} = \alpha_n^{[M]} p_n^{[M]} s_n^{[M]}$ and $x_{i,n}^{[S]} = \alpha_{i,n}^{[S]} p_i^{[S]} s_{i,n}^{[S]}$ are the transmitted signal to MU_n and

¹In this work, we assume that a perfect SIC detection is carried out at the receiver sides, which provides an upper bound in terms of the achieved data rates.

²Single-cluster NOMA provides a benchmark performance of the proposed algorithm for the EE-SE tradeoff. Also, the proposed algorithm is straightforwardly compatible with any clustering technique [29]–[31].

SU_{*i,n*}, respectively. A fractions of $\alpha_n^{[M]}$ and $\alpha_{i,n}^{[S]}$ from the MBS's power and SBS's power, $p^{[M]}$ and $p_i^{[S]}$, are assigned to MU_{*n*} and SU_{*i,n*}, respectively, while $s_n^{[M]}$ and $s_{i,n}^{[S]}$ are the message signals to MU_{*n*} and SU_{*i,n*}, respectively.

By considering $i \in \mathcal{SC}$, and $n \in \mathcal{SU}$, the received signal at the SU_{*i,n*}, $y_{i,n}^{[S]}$, can be written as:

$$y_{i,n}^{[S]} = \underbrace{h_{i,n}^{[S]} x_{i,n}^{[S]}}_{\text{Desired signal}} + \underbrace{h_{i,n}^{[S]} \sum_{l=1, l \neq n}^{N_{SU}} x_{i,l}^{[S]}}_{\text{Inter-user interference}} + \underbrace{\sum_{j=1, j \neq i}^{N_{SC}} f_{j,i,n}^{[S]} x_j^{[S]}}_{\text{Co-tier interference}} + \underbrace{g_{i,n}^{[S]} \sum_{k=1}^{N_{MU}} x_k^{[M]}}_{\text{Cross-tier interference}} + \underbrace{z_{i,n}^{[S]}}_{\text{Noise}}, \quad (1)$$

where $h_{i,n}^{[S]}$, $f_{j,i,n}^{[S]}$, and $g_{i,n}^{[S]}$ are the channel coefficients that between SBS_{*i*} and SU_{*i,n*}, the co-channel coefficients that between SBS_{*j*} and SU_{*i,n*}, and the cross-channel coefficients that between MBS and SU_{*i,n*}, respectively. $z_{i,n}^{[S]}$ is the additive white Gaussian noise (AWGN) at SU_{*i,n*} with variance σ^2 . Similarly, by assuming $n \in \mathcal{MU}$, the received signal at the MU_{*i,n*}, $y_n^{[M]}$, can be written as:

$$y_n^{[M]} = \underbrace{h_n^{[M]} x_n^{[M]}}_{\text{Desired signal}} + \underbrace{h_n^{[M]} \sum_{l=1, l \neq n}^{N_{MU}} x_l^{[M]}}_{\text{Inter-user interference}} + \underbrace{\sum_{i=1}^{N_{SC}} g_{i,n}^{[M]} x_i^{[S]}}_{\text{Cross-tier interference}} + \underbrace{z_n^{[M]}}_{\text{Noise}}, \quad (2)$$

where $h_n^{[M]}$, and $g_{i,n}^{[M]}$ are the channel coefficients that between MBS and its MU_{*n*}, and the cross-channel coefficients that between SBS_{*i*} and MU_{*n*}, respectively. $z_n^{[M]}$ is the AWGN at MU_{*n*}. From equations (1) and (2), since the MC and the SCs share the same resources, three distinct kinds of interference exist. Inter-user interference occurs among users in the same cell due to the non-orthogonal multiplexing of NOMA. SUs experience co-tier interference from other SBSs, while both SUs and MUs are affected by a cross-tier interference from MBS and SBSs, respectively.

For efficient NOMA signal detection under the presence of interference, the decoding order has to be in the ascending order of the users' normalized channel gain, as explained in [4], [5]. The normalized channel gain is defined as the channel gain-to-the noise, cross-tier, and co-tier interference, and can be expressed as

$$\kappa_{i,n}^{[S]} = \frac{|h_{i,n}^{[S]}|^2}{\underbrace{\sum_{j=1, j \neq i}^{N_{SC}} |f_{j,i,n}^{[S]}|^2 p_j^{[S]}}_{\text{Co-tier interference}} + \underbrace{|g_{i,n}^{[S]}|^2 p^{[M]}}_{\text{Cross-tier interference}} + \sigma^2}. \quad (3)$$

and

$$\kappa_n^{[M]} = \frac{|h_n^{[M]}|^2}{\underbrace{\sum_{j=1}^{N_{SC}} |g_{j,n}^{[M]}|^2 p_j^{[S]}}_{\text{Cross-tier interference}} + \sigma^2}, \quad (4)$$

where $\kappa_{i,n}^{[S]}$ and $\kappa_n^{[M]}$ are the normalized channel gain for SU_{*i,n*} and MU_{*n*}, respectively. In this work, we assume that the normalized channel gain order for MC and SC_{*i*} are $\kappa_1^{[M]} \geq \kappa_2^{[M]} \geq \dots \geq \kappa_{N_{MU}}^{[M]}$ and $\kappa_{i,1}^{[S]} \geq \kappa_{i,2}^{[S]} \geq \dots \geq \kappa_{i,N_{SU}}^{[S]}$, respectively, where $n = N_{MU}$ and $n = N_{SU}$ correspond to users with the worst channel condition in MC and SC, respectively.

Based on $\kappa_{i,n}^{[S]}$ and $\kappa_n^{[M]}$, the signal to interference plus noise power ratio (SINR) for SU_{*i,n*} and MU_{*n*} can be, respectively, expressed as

$$\gamma_{i,n}^{[S]} = \frac{p_{i,n}^{[S]} \kappa_{i,n}^{[S]}}{\sum_{l=1}^{n-1} p_{i,l}^{[S]} \kappa_{i,n}^{[S]} + 1}, \quad (5)$$

and

$$\gamma_n^{[M]} = \frac{p_n^{[M]} \kappa_n^{[M]}}{\sum_{l=1}^{n-1} p_l^{[M]} \kappa_n^{[M]} + 1}, \quad (6)$$

where $\gamma_{i,n}^{[S]}$ and $\gamma_n^{[M]}$ are the SINR at SU_{*i,n*} and MU_{*n*} respectively. Consequently, the sum rates of SC_{*i*} and MC can be, respectively, calculated from

$$R_{SC_i} = \sum_{n=1}^{N_{SU}} r_{i,n}^{[S]} = \sum_{n=1}^{N_{SU}} \log_2(1 + \gamma_{i,n}^{[S]}), \quad (7)$$

and

$$R_{MC} = \sum_{n=1}^{N_{MU}} r_n^{[M]} = \sum_{n=1}^{N_{MU}} \log_2(1 + \gamma_n^{[M]}), \quad (8)$$

respectively, where $r_{i,n}^{[S]}$ and $r_n^{[M]}$ are the individual data rates for SU_{*i,n*} and MU_{*n*}, respectively.

B. PROBLEM FORMULATION

In NOMA HetNets, although the increase in the transmit power from the BS of a dedicated cell will increase its sum-rate (i.e., SE), it affects the other cells negatively through interference as long as increasing the consumed energy (i.e., EE). In contrast, the decrease in the transmit power from the BS will save its consumed energy. However, users may fail to reach their required QoS. The utilized problem

in this work investigates the trade-off that exists between SE and EE in NOMA HetNets. Our goal is to maximize both SE and EE by adequately allocating power to SBSs and MBS, while considering the maximum power and the QoS constraints. Due to the contradiction between SE and EE, MOP is formulated to maximize both as in [10].

SE reflects how the available spectrum is efficiently utilized in terms of the achieved data-rate over an assigned bandwidth, B . Thus, the SE for the SC_i and the MC can be given, respectively, as

$$SE_{SC_i} = \frac{R_{SC_i}}{B}, \quad \forall i \in \mathcal{SC} \quad (9)$$

and

$$SE_{MC} = \frac{R_{MC}}{B}. \quad (10)$$

On the other hand, the EE of a BS indicates the amount of data transferred per unit energy consumed by this BS. Thus, the EE for SC_i and the MC can be given, respectively, as

$$EE_{SC_i} = \frac{R_{SC_i}}{p_i^{[S]} + p_{C_i}^{[S]}}, \quad \forall i \in \mathcal{SC} \quad (11)$$

and

$$EE_{MC} = \frac{R_{MC}}{p^{[M]} + p_c^{[M]}}, \quad (12)$$

where $p_c^{[M]}$ and $p_{C_i}^{[S]}$ are the circuit power consumption for MC and SC_i , respectively. Consequently, the investigated MOP for $SC_i, \forall i \in \mathcal{SC}$, and MC can be, respectively, formulated as

$$\max_{p_i^{[S]}} EE_{SC_i}, SE_{SC_i}, \quad (13)$$

$$s.t. C_1^{[S]} : \sum_{n=1}^{N_{SU}} p_{i,n}^{[S]} \leq p_{\max}^{[S]}, \quad (13a)$$

$$C_2^{[S]} : r_{i,n}^{[S]} \geq r_{th}^{[S]}, \quad \forall n, \quad (13b)$$

and

$$\max_{p^{[M]}} EE_{MC}, SE_{MC}, \quad (14)$$

$$s.t. C_1^{[M]} : \sum_{n=1}^{N_{MU}} p_n^{[M]} \leq p_{\max}^{[M]}, \quad (14a)$$

$$C_2^{[M]} : r_n^{[M]} \geq r_{th}^{[M]}, \quad \forall n, \quad (14b)$$

where the constraints $C_1^{[S]}$ and $C_1^{[M]}$ are introduced to guarantee that the power allocated to SBS_i and the MBS do not exceed the maximum transmitting power, $p_{\max}^{[S]}$ and $p_{\max}^{[M]}$, respectively. The constraints $C_2^{[S]}$ and $C_2^{[M]}$ ensure that the minimum data rate for $SU_{i,n}$ and MU_n do not fall below a predefined threshold values $r_{th}^{[S]}$ and $r_{th}^{[M]}$, respectively.

IV. PROPOSED NC GAME-BASED ALGORITHM

The MOPs in (13) and (14) are non-convex and nonlinear problems, which are very computationally costly to be solved in this form. In this section, we try to find a sub-optimum solution for the power allocation problem represented by the

MOPs in (13) and (14). First, the MOPs in (13) and (14) is reformulated into an SOP using the weighted sum strategy. Then, the SOP problem of power allocation is solved in two stages. In the first stage, the power is allocated to Het-Net SBSs and MBs through a non-cooperative game based technique for a near-optimum solution. In the second stage, the power is allocated to NOMA users independently at each cell considering both QoS constraint and SIC condition.

A. PROPOSED WEIGHTED-SUM SOP OF EE AND SE

Maximizing SE and EE are both important for HetNets. Thus, the weighted sum strategy is considered a proper choice to model the above tradeoff in an SOP, where the EE and SE are weighted summed. A dedicated balance between EE and SE can be achieved based on the system requirements by adapting the weights. For example, giving more weights to SE is important during the peak hours to serve more users, while giving more priority to EE is preferable at the off-peak time to reduce the consumed energy [10]. The MOP (13) can be modeled as an SOP as follow

$$\begin{aligned} & \max_{p_i^{[S]}} w_s EE_{SC_i} + (1 - w_s) SE_{SC_i} \\ & s.t. C_1^{[S]}, C_2^{[S]}, \end{aligned} \quad (15)$$

where $0 < w_s < 1$ is a balancing parameter between EE and SE within the SCs. Problem (15) is a non-linear fractional problem, of which optimal global solution is not guaranteed. An equivalent linear hypograph form [8], [32] can be obtained by assuming

$$\frac{w_s R_{SC_i}}{p_i^{[S]} + p_{C_i}^{[S]}} \geq \xi_{EE_i}^{[S]}. \quad (16)$$

Consequently, the hypograph equivalent form of (15) can be expressed as

$$\begin{aligned} & \max_{p_i^{[S]}} \xi_{EE_i}^{[S]} + (1 - w_s) SE_{SC_i} \\ & s.t. C_1^{[S]}, C_2^{[S]} \\ & C_3^{[S]} : w_s R_{SC_i} - \xi_{EE_i}^{[S]} (p_i^{[S]} + p_{C_i}^{[S]}) > 0 \\ & C_4^{[S]} : \xi_{EE_i}^{[S]} > 0. \end{aligned} \quad (17)$$

Similarly, the SOP of the MC problem (14) in its equivalent linear hypograph form can be formulated as

$$\begin{aligned} & \max_{p^{[M]}} \xi_{EE}^{[M]} + (1 - w_m) SE_{MC} \\ & s.t. C_1^{[M]}, C_2^{[M]} \\ & C_3^{[M]} : w_m R_{MC} - \xi_{EE}^{[M]} (p^{[M]} + p_C^{[M]}) > 0 \\ & C_4^{[M]} : \xi_{EE}^{[M]} > 0, \end{aligned} \quad (18)$$

where $0 < w_m < 1$ is the balancing parameter between EE and SE within the MC. Although problems in (17) and (18) are linear, they are functions of inseparable variables, $p_i^{[S]}$, $\forall i \in \mathcal{SC}$, and $p^{[M]}$. In other words, the transmit power from one of the BSs affects the other cells negatively through either cross-tier or co-tier interference. Thus, problems in (17) and (18) are non-convex with respect to $p_i^{[S]}$ and $p^{[M]}$ which is

computationally costly to be optimally solved. To solve (17) and (18), we propose to perform a non-cooperative game among BSs through which each BS can optimize its power individually while considering the power of other BS's as constant.

B. EXISTENCE OF NASH EQUILIBRIUM (NE)

Since each BS jointly affects others through co/cross-tier interference, the PC problem can be modeled as a game $\mathcal{G}\{\mathcal{N}, \mathcal{S}, \mathcal{U}\}$, where

- \mathcal{N} is the set of all BSs (i.e., MBS and N_{SC} SBSs) that represents the game players.
- $\mathcal{S} = \mathcal{P}_1^{[S]} \times \dots \times \mathcal{P}_{N_{SC}}^{[S]} \times \mathcal{P}^{[M]}$ is the space of the transmit power, where $\mathcal{P}_i^{[S]}$ and $\mathcal{P}^{[M]}$ are the available action space for the SBS $_i$ and MBS, respectively.
- $\mathcal{U} = \{\mathcal{U}^{[M]}, \mathcal{U}_i^{[S]} | i \in \mathcal{SC}\}$ is the set of utility of each deployed BS, where

$$\mathcal{U}_i^{[S]}(\mathbf{P}) = w_s EE_{SC_i} + (1 - w_s) SE_{SC_i}, \quad (19a)$$

$$\mathcal{U}^{[M]}(\mathbf{P}) = w_m EE_{MC} + (1 - w_m) SE_{MC}, \quad (19b)$$

where $\mathbf{P} = [p_1^{[S]}, \dots, p_{N_{SC}}^{[S]}, p^{[M]}]$ is the concatenated power vector that contains the power of all BSs. In the non-cooperative game, nash equilibrium (NE) is the point that gives a stable outcome of a game, such that all the players with conflict interests are satisfied, and no player wants to deviate. In other words, NE satisfies $\mathcal{U}_i^{[S]}(\mathbf{P}^*) > \mathcal{U}_i^{[S]}(\mathbf{P})$, $\forall i \in \mathcal{SC}$, and $\mathcal{U}^{[M]}(\mathbf{P}^*) > \mathcal{U}^{[M]}(\mathbf{P})$, $\forall \mathbf{P} \neq \mathbf{P}^*$. Since $\mathcal{U}_i^{[S]}(\mathbf{P})$ and $\mathcal{U}^{[M]}(\mathbf{P})$ have been proved in [12] to be concave functions of $p_i^{[S]}$ and $p^{[M]}$, respectively, the game \mathcal{G} has a unique NE point.

C. POWER ALLOCATION TO HetNets BASED ON THE NC GAME

In this step, the power allocated to SBSs and MBS is controlled based on the proposed NC game based technique. In the proposed NC game, the competing players, SBSs and MBS, choose their actions towards maximizing the SE and EE simultaneously and independently. Consequently, each BS (i.e., SBSs or MBS) has the opportunity to maximize its SOP of joint EE and SE by considering the power transmitted from the other BSs as a constant. We can reach the game equilibrium by intersecting the solutions of the SOP problems (17), for $\forall i \in \mathcal{SC}$, and (18).

To find an expression for the power of the SBS $_i$, $p_i^{[S]}$, we need first to solve its equivalent unconstrained Lagrangian equation, $\mathcal{L}_i^{[S]}$, in (20), as shown at the bottom of the next page. Equation (20) is solved by considering $p^{[M]}$, $p_j^{[S]}$, $\forall j \in \mathcal{SC} \setminus i$, as a constant (i.e., constant co-tier and cross-tier interference). In that case, (20) is convex w.r.t. $p_i^{[S]}$ [7], [33], where the parameters $\lambda_i^{[S]}$, $\mu_{i,n}^{[S]}$, $\beta_i^{[S]}$ are the Lagrangian multipliers (LMs) related to the SOP of the SBS $_i$. By assuming that SU $_{i,N_{SU}}$ is the user with the worst channel condition within the SC $_i$, we can find an expression for $p_i^{[S]}$ by substituting $\kappa_{i,n}^{[S]}$ with its value in (3), and then tacking the first derivative

of $\mathcal{L}_i^{[S]}$ w.r.t. the $p_{i,N_{SU}}^{[S]}$ for $n = N_{SU}$. Thus, we can obtain equation (21).

$$\frac{\partial \mathcal{L}_i^{[S]}}{\partial p_{i,N_{SU}}^{[S]}} = \frac{(1 - w_s + \beta_i^{[S]} w_s) |h_{i,N_{SU}}^{[S]}|^2}{(|h_{i,N_{SU}}^{[S]}|^2 p_i^{[S]} + \sum_{\substack{j=1 \\ j \neq i}}^{N_{SC}} |f_{j,i,N_{SU}}^{[S]}|^2 p_j^{[S]} + |g_{i,N_{SU}}^{[S]}|^2 p^{[M]} + \sigma^2) \ln 2 - \beta_i^{[S]} \xi_{EE_i}^{[S]} - \lambda_i^{[S]} + \mu_{i,N_{SU}}^{[S]} |h_{i,N_{SU}}^{[S]}|^2}. \quad (21)$$

By setting the $\frac{\partial \mathcal{L}_i^{[S]}}{\partial p_{i,N_{SU}}^{[S]}} = 0$, the optimal value for $p_i^{[S]}$, $\forall i \in \mathcal{SC}$ can be given as

$$p_i^{[S]} = \frac{(1 - w_s + \beta_i^{[S]} w_s)}{(\beta_i^{[S]} \xi_{EE_i}^{[S]} + \lambda_i^{[S]} - \mu_{i,N_{SU}}^{[S]} |h_{i,N_{SU}}^{[S]}|^2) \ln 2} - \sum_{\substack{j=1 \\ j \neq i}}^{N_{SC}} \frac{|f_{j,i,N_{SU}}^{[S]}|^2}{|h_{i,N_{SU}}^{[S]}|^2} p_j^{[S]} - \frac{|g_{i,N_{SU}}^{[S]}|^2}{|h_{i,N_{SU}}^{[S]}|^2} p^{[M]} - \frac{\sigma^2}{|h_{i,N_{SU}}^{[S]}|^2}. \quad (22)$$

Similarly, to find an expression for the power of the MBS, $p^{[M]}$, we need to solve its equivalent unconstrained Lagrangian equation in (23), as shown at the bottom of the next page, $\mathcal{L}^{[M]}$, by considering $p_i^{[S]}$, $\forall i$, as a constant (i.e., constant cross-tier interference). Thus, (23) is convex w.r.t. $p^{[M]}$, where the parameters $\lambda^{[M]}$, $\mu_n^{[M]}$, $\beta^{[M]}$ are the LMs related to the SOP of the MBS. By assuming that MU $_{N_{MU}}$ is the user with the worst channel condition within the MC, we can find an expression for $p^{[M]}$ by replace $\kappa_n^{[M]}$ with its value in (4), and then tacking the first derivative of $\mathcal{L}^{[M]}$ w.r.t. the $p_{N_{MU}}^{[S]}$ for $n = N_{MU}$. Thus, we can obtain equation (24).

$$\frac{\partial \mathcal{L}^{[M]}}{\partial p_{N_{MU}}^{[M]}} = \frac{(1 - w_m + \beta^{[M]} w_m) |h_{N_{MU}}^{[M]}|^2}{(|h_{N_{MU}}^{[M]}|^2 p^{[M]} + \sum_{j=1}^{N_{SC}} |g_{j,N_{MU}}^{[M]}|^2 p_j^{[S]} + \sigma^2) \ln 2 - \beta^{[M]} \xi_{EE}^{[M]} - \lambda^{[M]} + \mu_{N_{MU}}^{[M]} |h_{N_{MU}}^{[M]}|^2}. \quad (24)$$

By setting the $\frac{\partial \mathcal{L}^{[M]}}{\partial p_{N_{MU}}^{[M]}} = 0$, the optimal value for $p^{[M]}$ can be expressed by

$$p^{[M]} = \frac{(1 - w_m + \beta^{[M]} w_m)}{(\beta^{[M]} \xi_{EE}^{[M]} + \lambda^{[M]} - \mu_{N_{MU}}^{[M]} |h_{N_{MU}}^{[M]}|^2) \ln 2} - \sum_{j=1}^{N_{SC}} \frac{|g_{j,N_{MU}}^{[M]}|^2}{|h_{N_{MU}}^{[M]}|^2} p_j^{[S]} - \frac{\sigma^2}{|h_{N_{MU}}^{[M]}|^2}. \quad (25)$$

The set of equations consisting of (22), $\forall i \in \mathcal{SC}$, and (25) are deterministic since they are a linear function of each other. Thus, by substitutionally solving this set of equations, an equilibrium solution can be obtained. Moreover, in (22) and (25), we consider the channel and the co/cross-channel of the user with the worst channel condition in each cell since if we ensure the minimum rate for the worst channel condition user, we can provide higher rate than the minimum rate for other users within the cell. Besides, the proposed NC game dramatically decreases the signaling needed to be shared among cells since we need to share only the channel and the co/cross-channel of the worst user in each cell with the CCU.

On the other hand, the equilibrium level is affected by the values given to the LMs. One of the simplest and fastest ways to find optimum values for these multipliers is to utilize one of the metaheuristic algorithms. In this work, we adopt the differential evolution (DE) algorithm proposed in [34] as a fast convergence algorithm to find near-optimum values for the LMs. Algorithm 1 presents the proposed non-cooperative game for joint EE and SE and the details regarding the updating based DE for the LMs, where \mathcal{T}_{max} is the maximum number of iterations.

D. POWER ALLOCATION TO NOMA USERS

In this step, the power allocated to each NOMA macro and small user, $p_n^{[M]}$ and $p_{i,n}^{[S]}$, is optimized while considering both the QoS constraint and SIC condition. By optimizing the power allocated to the SBSs and MBS using the NC-EE-SE algorithm, the cross-tier and co-tier interference can be treated as constant when estimating the allocated power to NOMA users. Thus, the values of $p_n^{[M]}$ and $p_{i,n}^{[S]}$ can be independently estimated. To satisfy a required QoS, the power allocated to NOMA users should guarantee that the users' SINRs do not fall below a threshold value. The SINR threshold at the $SU_{i,n}$ can be

expressed as:

$$\begin{aligned} \gamma_{i,n}^{[S]} &\geq \theta_{i,n}^{[S]} \\ \frac{p_{i,n}^{[S]} \kappa_{i,n}^{[S]}}{\sum_{l=1}^{n-1} p_{i,l}^{[S]} \kappa_{i,n}^{[S]} + 1} &\geq \theta_{i,n}^{[S]}, \end{aligned} \quad (26)$$

where $\theta_{i,n}^{[Sb]} = 2^{r_{i,n}^{[Sb]}} - 1$ is the SINR threshold value at $SU_{i,n}$. Using simple manipulations, Eq. (26) can be reformulated as

$$p_{i,n}^{[S]} \geq \frac{\theta_{i,n}^{[S]}}{\kappa_{i,n}^{[S]}} \left(\sum_{l=1}^{n-1} p_{i,l}^{[S]} \kappa_{i,n}^{[S]} + 1 \right). \quad (27)$$

Furthermore, in the existence of interference or noise, to decode a signal of a NOMA user from the superimposed higher SIC ordering users, a minimum power difference is required [5] such that

$$\left(\alpha_{i,n}^{[S]} - \sum_{l=1}^{n-1} \alpha_{i,l}^{[S]} \right) p_{i,n}^{[S]} \kappa_{i,n-1}^{[S]} \geq \delta_{diff}^{[S]}, \quad (28)$$

where $\delta_{diff}^{[S]}$ denotes the minimum signal power to noise difference among SUs. Equation (28) shows the mandatory condition to achieve successful SIC, i.e., the decoded and the remaining undetectable signals should be accurately distinguished. Also, by simple manipulations, Eq. (28) can reformulated as

$$p_{i,n}^{[S]} \geq \frac{\delta_{diff}^{[S]}}{\kappa_{i,n-1}^{[S]}} + \sum_{l=1}^{n-1} p_{i,l}^{[S]}, \quad (29)$$

By changing the inequality condition in (27) and (29) to equality, the minimum power allocated to the $SU_{i,n}$ can be calculated as

$$p_{i,n}^{[S]} = \frac{\theta_{i,n}^{[S]} (\kappa_{i,n}^{[S]} \delta_{diff}^{[S]} - \kappa_{i,n-1}^{[S]})}{\kappa_{i,n}^{[S]} \kappa_{i,n-1}^{[S]} (\theta_{i,n}^{[S]} - 1)}. \quad (30)$$

$$\begin{aligned} \mathcal{L}_i^{[S]} = (1 - w_s + \beta_i^{[S]} w_s) \sum_{n=1}^{N_{SU}} \log_2 \left(\frac{\sum_{l=1}^n p_{i,l}^{[S]} \kappa_{i,n}^{[S]} + 1}{\sum_{l=1}^{n-1} p_{i,l}^{[S]} \kappa_{i,n}^{[S]} + 1} \right) - \beta_i^{[S]} \xi_{EE_i}^{[S]} (p_i^{[S]} + p_{C_i}^{[S]}) + \lambda_i^{[S]} \left(p_{max}^{[S]} - \sum_{n=1}^{N_{SU}} p_{i,n}^{[S]} \right) \\ + \sum_{n=1}^{N_{SU}} \mu_{i,n}^{[S]} \left(\sum_{l=1}^n p_{i,l}^{[S]} \kappa_{i,n}^{[S]} + 1 - 2^{r_{i,n}^{[S]}} \left\langle \sum_{l=1}^{n-1} p_{i,l}^{[S]} \kappa_{i,n}^{[S]} + 1 \right\rangle \right), \end{aligned} \quad (20)$$

$$\begin{aligned} \mathcal{L}^{[M]} = (1 - w_m + \beta^{[M]} w_m) \sum_{n=1}^{N_{MU}} \log_2 \left(\frac{\sum_{l=1}^n p_l^{[M]} \kappa_n^{[M]} + 1}{\sum_{l=1}^{n-1} p_l^{[M]} \kappa_n^{[M]} + 1} \right) - \beta^{[M]} \xi_{EE}^{[M]} (p^{[M]} + p_C^{[M]}) + \lambda^{[M]} \left(p_{max}^{[M]} - \sum_{n=1}^{N_{MU}} p_n^{[M]} \right) \\ + \sum_{n=1}^{N_{MU}} \mu_n^{[M]} \left(\sum_{l=1}^n p_l^{[M]} \kappa_n^{[M]} + 1 - 2^{r_{i,n}^{[M]}} \left\langle \sum_{l=1}^{n-1} p_l^{[M]} \kappa_n^{[M]} + 1 \right\rangle \right), \end{aligned} \quad (23)$$

TABLE 2. The per iteration complexity order of the investigated algorithms.

Algorithm	Complexity
Exhaustive Search	$\mathcal{O}(\binom{P_{max}-P_{min}}{N_{SC}}/\zeta)$, ζ is power diff. step
PC with pricing [12]	$\mathcal{O}(N_{SC}^3)$
SCA [11]	$\mathcal{O}(N_{SC}N_{SU} + N_{MU})^2$
CCCP [10]	$\mathcal{O}(\log(1/\varepsilon))$, $\varepsilon = 10^{-3}$ is the error tolerance
Dinkelbach with NC [18]	$\mathcal{O}(N_{SU}^2 + N_{MU}^2)$
LDD [17]	$\mathcal{O}(N_{SC}N_{SU})$
NC-EE [8]	$\mathcal{O}(N_{SC})$
NC-SE [9]	$\mathcal{O}(N_{SC})$
Proposed NC-EE-SE	$\mathcal{O}(N_{SC})$

Thus, Eq.(30) represents the minimum power allocated to each NOMA SU to satisfy both QoS constraint and SIC condition, simultaneously. Moreover, the number of clustered NOMA SUs per SC that can access the same time/frequency resource, N_{SU} , is constrained by the condition $C_1^{[S]}$. Similarly, the minimum power allocation to the $p_n^{[M]}$ can be calculated as

$$p_n^{[M]} = \frac{\theta_n^{[M]}(\kappa_n^{[M]}\delta_{diff}^{[M]} - \kappa_{n-1}^{[M]})}{\kappa_n^{[M]}\kappa_{n-1}^{[M]}(\theta_n^{[M]} - 1)}, \quad (31)$$

where $\delta_{diff}^{[M]}$ is the minimum signal power to noise difference among MUs and $\theta_n^{[M]}$ is SINR threshold at MU_n . Equation (30) represents the minimum power should allocated to each NOMA MU_n to satisfy both QoS constraint and SIC condition. Also, the number of clustered NOMA MUs that can access the same time/frequency resource, N_{MU} , is constrained by the condition $C_1^{[M]}$.

E. COMPLEXITY ANALYSIS

The computational complexity of the proposed NC-EE-SE technique is bounded by the complexity of computing $p_i^{[S]}$ and $p^{[M]}$ from (22) and (25), respectively. Since Eqs. (22) and (25) contain only summation operators, the overall computational complexity of the proposed NC-EE-SE per iteration is upper bounded by $\mathcal{O}(N_{SC})$. The bottleneck complexity orders of different algorithms are listed in Table 2. It is obvious from Table 2 that the proposed algorithm has the same complexity as NC-SE and NC-EE, and much lower complexity than the other compared techniques.

V. NUMERICAL RESULTS AND DISCUSSION

A. SIMULATION PARAMETERS

The assumed simulation parameters are listed in Table 3. Two non-overlapped SCs³ are uniformly adopted under the coverage of MC. Each BS (i.e., MBS or SBSs) serves only two NOMA users per time/frequency resources; one is called a cell-center user (i.e., $n = 1$), while the other is called a

³In this paper, we choose $N_{SC} = 2$ as in [9]. However, the proposed algorithm is general for the case of $N_{SC} > 2$.

Algorithm 1 Proposed NC-EE-SE

- **Initialize:** LMs vector $\forall i \in \mathcal{SC}$, $\mathbf{e}_{M_i} = [\lambda_i^{[S]0}, \beta_i^{[S]0}, \mu_{i,1}^{[S]0}, \lambda_i^{[M]0}, \beta_i^{[M]0}, \mu_1^{[M]0}]$.
- **Set:** $w_s, w_m, \xi_{EE_i}^{[S]}, \xi_{EE}^{[M]}, \mathcal{T}_{max}$
- **while** $t \leq \mathcal{T}_{max}$ **do**

1) Generating a trial LMs vector by applying the DE algorithm:

- 1.1 Produce three random LMs vectors $\mathbf{e}_{M_1}, \mathbf{e}_{M_2}$, and \mathbf{e}_{M_3} .
- 1.2 Estimate the mutant vector \mathbf{v}^t :

$$\mathbf{v}^t = \mathbf{e}_{M_i}^t + f(\mathbf{e}_{M_1} - \mathbf{e}_{M_2}).$$

1.3 Estimate the trial vector \mathbf{u}^t :

$$\mathbf{u}^t = \mathbf{e}_{M_3} + \mathbf{e}^t \cdot (\mathbf{v}^t - \mathbf{e}_{M_3}).$$

- 2) Substitutionally solving equations (22), $\forall i \in \mathcal{SC}$, and (25) to calculate $p^{[M]t}$ and $p_i^{[S]t}$, $\forall i \in \mathcal{SC}$.
- 3) Updating the LMs vector $\mathbf{e}_{M_i}^{t+1}$ as:

$$\mathbf{e}_{M_i}^{t+1} = \begin{cases} \mathbf{u}^t & \mathcal{L}^{[S]}(p_i^{[S]t}) > \mathcal{L}_i^{[S]}(p_i^{[S]t-1}) \\ & \mathcal{L}^{[M]}(p^{[M]t}) > \mathcal{L}^{[M]}(p^{[M]t-1}) \\ \mathbf{e}_{M_i}^t & \text{Otherwise} \end{cases}$$

4) $t \leftarrow t + 1$

if $|\mathcal{L}^{[S]}(p_i^{[S]t}) - \mathcal{L}_i^{[S]}(p_i^{[S]t-1})| \leq \epsilon$ **then**

if $|\mathcal{L}^{[M]}(p^{[M]t}) - \mathcal{L}^{[M]}(p^{[M]t-1})| \leq \epsilon$ **then**

return $p_i^{[S]t+1}$ and $p^{[M]t+1}$.

break.

end if

end if

5) $t \leftarrow t + 1$

end while

where

- f is a system-defined scaling factor.
- \mathbf{e}^t is a binary random vector.

TABLE 3. Simulation parameters.

Parameter	Value	Parameter	Value
$p_{max}^{[M]}$	46 dBm	$\delta_{diff}^{[S]} = \delta_{diff}^{[M]}$	10 dBm
$p_{max}^{[S]}$	26 dBm	$d_{i,1}^{[S]}$	5 m to 25 m
Noise Power density	-174 dBm/Hz	$d_{i,2}^{[S]}$	25 m to 50 m
MC radius	500 m	$d_1^{[M]}$	50 m to 200 m
SC radius	50 m	$d_2^{[M]}$	200 m to 300 m
N_{SC}	2	$p_c^{[M]}$	27 dBm
N_{MU}, N_{SU}	2, 2	$p_{E_i}^{[S]}$	20 dBm
$\{r_{i,1}^{[S]}, r_{i,2}^{[S]}\}$	$\{0.5, 1\}$	$\{r_1^{[M]}, r_2^{[M]}\}$	$\{1, 2\}$

cell-edge user (i.e., $n = 2$). $SU_{i,1}$ and MU_1 are randomly distributed over an area of ranges $d_{i,1}^{[S]}$ and $d_1^{[M]}$ far from the center of their BSs, respectively, while $SU_{i,2}$ and MU_2

are randomly spread over an area of ranges $d_{i,2}^{[S]}$ and $d_2^{[M]}$, respectively. Moreover, the channel coefficients are randomly produced by the multiplication of the free space path loss and the Rayleigh fading with zero mean and unit variance as in [35].

Simulation results compare the performance in terms of the achieved EE and SE among techniques; 1) NC-SE [9], where only SE is taken into account, 2) NC-EE [8], where only EE is taken into account, and 3) the proposed NC-EE-SE scheme #1. In these techniques, the same power is allocated to all SBSs based on the user with the worst channel condition among all SBSs. NC-SE [9] and NC-EE [8] techniques can be considered as the optimum performance of non-cooperative game based SE or EE for the case of the same allocated power to all SBSs. In addition, the comparison includes the results of 4) the proposed NC-EE-SE scheme #2, where the SBSs are allocated with different power by solving (22), $\forall i \in \mathcal{S}_C$, to manage the co-tier interference as long as the cross-tier interference. Also, we assume that $w_m = w_s = w$, and $\xi_{EE}^{[M]} = \xi_{EE_i}^{[S]} = \xi$.

B. SIMULATION RESULTS

The performance of the proposed NC-EE-SE algorithm is compared with the state of the state-of-the-art algorithms in terms of total EE ($EE_{MC} + \sum_{i=1}^{N_{SC}} EE_{SC_i}$), and total SE, ($SE_{MC} + \sum_{i=1}^{N_{SC}} SE_{SC_i}$), at different values of signal-to-noise ratio (SNR) in Figs. 2(a) and 2(b), respectively. Fig.2 shows that the proposed NC-SE-EE provides higher SE and EE than the conventional OMA and NOMA. Also, Fig.2 shows that by tuning the parameters ξ , the proposed NC-SE-EE, for both Sch#1 and Sch#2, can give comparable performance to the D.C. programming [7], [10] algorithm and very close to the NC-SE [9] and the exhaustive search, while much improving the EE over the compared schemes. In other words, the proposed algorithm is able to find a better tradeoff point that improves EE and SE than conventional approaches.

For different values of the tradeoff balancing parameter w , the performance of the proposed NC-EE-SE is investigated in terms of the total EE and the total SE versus the SNR in Figs. 3(a) and 3(b), respectively. In general, increasing w from 0 to 1 means that the NC game will allocate the power so as to improve EE and sacrifice the SE. However, by choosing appropriate values for w and ξ , we can get a much higher EE without significantly losing in the SE. Also, it appears from Fig. 3 that by controlling the power of each SBS separately using the proposed scheme #2, we can improve the EE over scheme #1 with almost the same SE. It is also worth noting that at $w = 1$ (i.e., only EE is taken into account), the proposed scheme #2 can improve the EE over the NC-EE [8] without sacrificing in the SE.

The results for $\xi = 0, w = 0$, where SE is only taken into account is shown in Fig. 4. It is obvious that the proposed scheme #2 can improve the EE over NC-SE [9] with the same SE level. The results in Figs. 3 and 4 confirm that by adequately allocating the power at each BS below the maximum

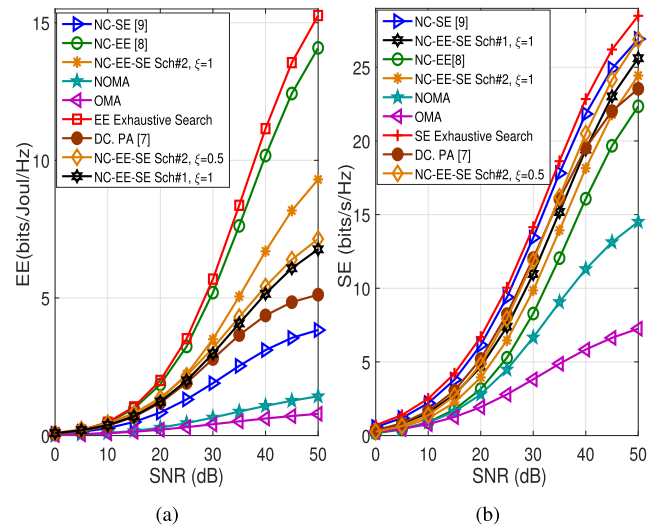


FIGURE 2. The performance of the proposed NC-EE-SE versus the state-of-the-art techniques in terms of (a) total EE, and (b) total SE versus SNR at $w = 0.8$.

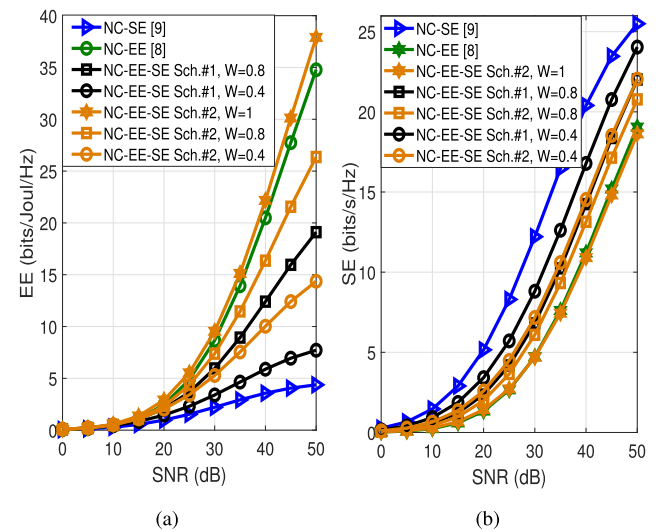


FIGURE 3. The performance of the proposed NC-EE-SE versus NC-EE [8] and NC-SE [9] in terms of (a) total EE, and (b) total SE versus SNR for different values of w , and $\xi = 3$.

power, we can find some points where the decrease in the signal power is compensated by the reduction in the level of interference to sustain the users' QoS while preserving the emitted energy.

The effect of increasing w on the EE and SE of the proposed algorithm for different ξ values is shown in Figs. 5(a) and 5(a), respectively. Algorithms NC-EE and NC-SE do not depend on the w . Increasing w will objective the game more towards the EE and far from the SE. In other words, by increasing w , the NC game will decrease the power in order to improve the EE. Also, it is obvious from Fig. 5 that increasing ξ will direct the game more in the direction of the EE rather than SE.

At different values of SNR and threshold rates, $r_{th}^{[S]}$, the outage probabilities of the $SU_{i,n}$ for the proposed

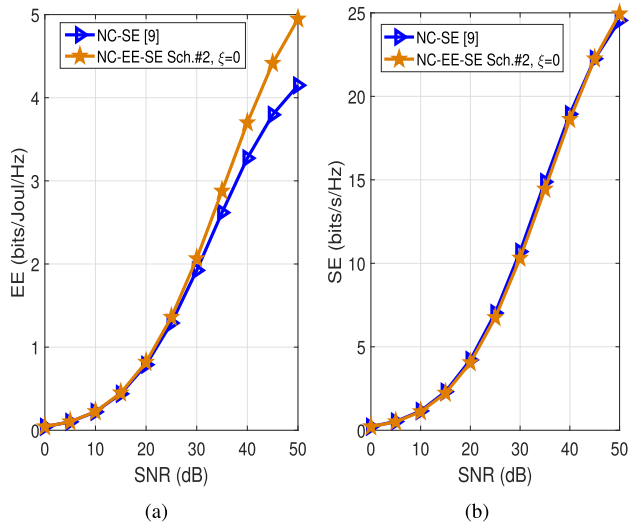


FIGURE 4. The performance of the proposed NC-EE-SE versus NC-SE [9] in terms of (a) total EE, and (b) total SE versus SNR for $w = 0$, and $\xi = 0$.

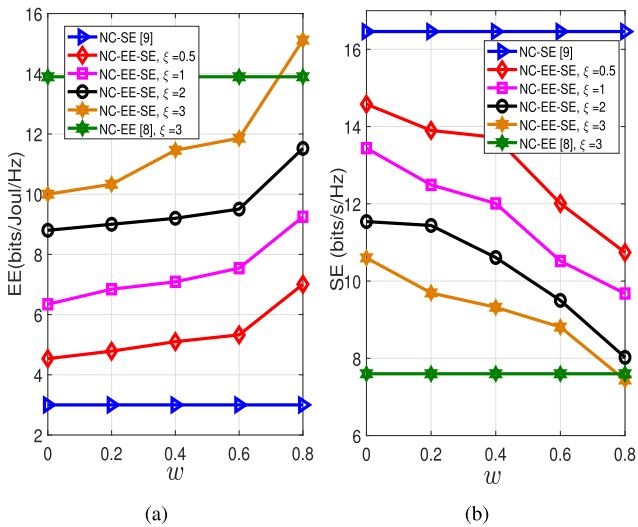


FIGURE 5. The performance of the proposed NC-EE-SE scheme #2 in terms of total EE versus w for different values of ξ .

NC-EE-SE is plotted in Fig.6(a) and Fig.6(b) against baseline approaches. The outage probability in Fig. 6 is plotted according to equations (28) and (29) in [13]. From Fig. 6, it is clear that the outage performance of the proposed NC-EE-SE is better than NC-EE and comparable to NC-SE and the exhaustive search technique. Also, the outage performance can be improved as ξ and w goes to zero. In other words, the proposed NC-SE-EE can give acceptable outage performance as long as the value of SE threshold $r_{th}^{[S]}$ is feasible and compatible with the value of the EE threshold, ξ .

Fig.7 shows the convergence behavior of the NC-EE-SE. It can be observed that adopting the DE algorithm to obtain optimum values for the LMs forces the allocated power to each BS to reach its stable status after a limited number of iterations. Moreover, the convergence is guaranteed, even for $N_{SC} > 2$, as long as the maximum power and QoS threshold values are feasible. Accordingly, the proposed NC-EE-SE is

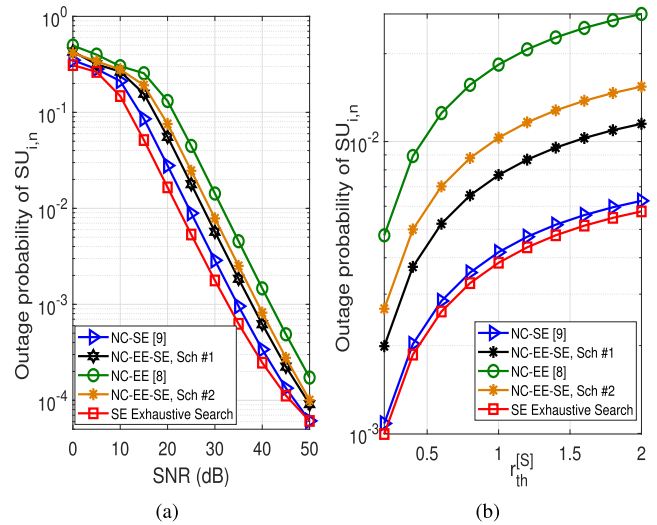


FIGURE 6. The performance of the proposed NC-EE-SE algorithm in terms of outage probability of the $SU_{i,n}$ at various values of a) SNR, and b) threshold rate $r_{th}^{[S]}$.

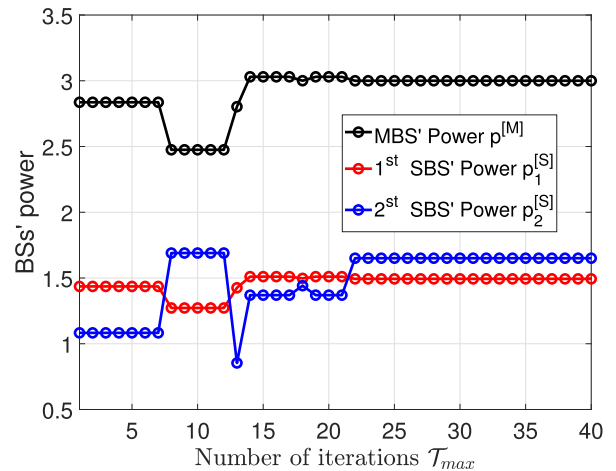


FIGURE 7. Convergence behavior of the proposed NC-EE-SE scheme #2 game for $T_{max} = 40$.

cost-efficient in terms of convergence time in addition to the hardware complexity.

VI. CONCLUSION

In this work, the EE-SE tradeoff in NOMA HetNets has been studied in terms of interference management and PC. The EE-SE tradeoff has been modeled as a non-convex MOP. The MOP has been relaxed into a convex SOP by adopting the weighted sum strategy and the hypograph transformation. Then, a non-cooperative game-based technique, NC-EE-SE, has been proposed to allocate the power to the SBSs and MBS in a competitive manner to jointly maximize their EE and SE based on the system requirements. Then, a closed-form formula has been proposed to control the power allocated to NOMA users taking into account both QoS and SIC condition. From the discussed results, properly choosing the balancing parameters and the EE threshold value can improve the tradeoff between EE and SE. MIMO can provide extra

degree of freedom that can be useful for our NOMA-HetNets systems in terms of accommodating more users or mitigating part of the interference, which will be considered in our future work.

REFERENCES

- [1] V. Jungnickel, K. Manolakis, W. Zirwas, B. Panzner, V. Braun, M. Lossow, M. Sternad, R. Apelfrojd, and T. Svensson, "The role of small cells, coordinated multipoint, and massive MIMO in 5G," *IEEE Commun. Mag.*, vol. 52, no. 5, pp. 44–51, May 2014.
- [2] Y. Saito, Y. Kishiyama, A. Benjebbour, T. Nakamura, A. Li, and K. Higuchi, "Non-orthogonal multiple access (NOMA) for cellular future radio access," in *Proc. IEEE 77th Veh. Technol. Conf. (VTC Spring)*, Jun. 2013, pp. 1–5.
- [3] W. Shin, M. Vaezi, B. Lee, D. J. Love, J. Lee, and H. V. Poor, "Coordinated beamforming for multi-cell MIMO-NOMA," *IEEE Commun. Lett.*, vol. 21, no. 1, pp. 84–87, Jan. 2017.
- [4] J. Zhao, Y. Liu, K. K. Chai, A. Nallanathan, Y. Chen, and Z. Han, "Spectrum allocation and power control for non-orthogonal multiple access in HetNets," *IEEE Trans. Wireless Commun.*, vol. 16, no. 9, pp. 5825–5837, Sep. 2017.
- [5] D. Ni, L. Hao, Q. T. Tran, and X. Qian, "Power allocation for downlink NOMA heterogeneous networks," *IEEE Access*, vol. 6, pp. 26742–26752, 2018.
- [6] Y. Gu, W. Saad, M. Bennis, M. Debbah, and Z. Han, "Matching theory for future wireless networks: Fundamentals and applications," *IEEE Commun. Mag.*, vol. 53, no. 5, pp. 52–59, May 2015.
- [7] Z. Song, Q. Ni, and X. Sun, "Distributed power allocation for nonorthogonal multiple access heterogeneous networks," *IEEE Commun. Lett.*, vol. 22, no. 3, pp. 622–625, Mar. 2018.
- [8] Y. Zhang, Y. Zhu, W. Xia, F. Shen, X. Zuo, F. Yan, and L. Shen, "Game-based power control for downlink non-orthogonal multiple access in HetNets," in *Proc. IEEE Global Commun. Conf. (GLOBECOM)*, Dec. 2018, pp. 206–212.
- [9] A. Nasser, O. Muta, M. ElSabrouty, and H. Gacanin, "Interference mitigation and power allocation scheme for downlink MIMO-NOMA HetNet," *IEEE Trans. Veh. Technol.*, vol. 68, no. 7, pp. 6805–6816, Jul. 2019.
- [10] S. Zhang, N. Zhang, G. Kang, and Z. Liu, "Energy and spectrum efficient power allocation with NOMA in downlink HetNets," *Phys. Commun.*, vol. 31, pp. 121–132, Dec. 2018.
- [11] B. Huang and A. Guo, "Spectral and energy efficient resource allocation for massive MIMO HetNets with wireless backhaul," *IEEE Wireless Commun. Lett.*, vol. 8, no. 2, pp. 552–555, Apr. 2018.
- [12] C. C. Coskun and E. Ayanoglu, "Energy- and spectral-efficient resource allocation algorithm for heterogeneous networks," *IEEE Trans. Veh. Technol.*, vol. 67, no. 1, pp. 590–603, Jan. 2018.
- [13] A. Nasser, O. Muta, M. ElSabrouty, and H. Gacanin, "Compressive sensing based spectrum allocation and power control for NOMA HetNets," *IEEE Access*, vol. 7, pp. 98495–98506, 2019.
- [14] S. Rezvani, N. M. Yamchi, M. R. Javan, and E. A. Jorswieck, "Resource allocation in virtualized CoMP-NOMA HetNets: Multi-connectivity for joint transmission," *IEEE Trans. Commun.*, vol. 69, no. 6, pp. 4172–4185, Jun. 2021.
- [15] S. Aboagye, A. Ibrahim, T. M. N. Ngatched, and O. A. Dobre, "Joint user association and power control for area spectral efficiency maximization in HetNets," in *Proc. IEEE 92nd Veh. Technol. Conf. (VTC-Fall)*, Nov. 2020, pp. 1–6.
- [16] J. Tang, J. Luo, M. Liu, D. K. C. So, E. Alsusa, G. Chen, K.-K. Wong, and J. A. Chambers, "Energy efficiency optimization for NOMA with SWIPT," *IEEE J. Sel. Topics Signal Process.*, vol. 13, no. 3, pp. 452–466, Jun. 2019.
- [17] M. Ataeshojai, R. C. Elliott, W. A. Krzymień, C. Tellambura, and J. Melzer, "Energy-efficient resource allocation in single-RF load-modulated massive MIMO HetNets," *IEEE Open J. Commun. Soc.*, vol. 1, pp. 1738–1764, 2020.
- [18] S. Sobhi-Givi, M. G. Shayesteh, and H. Kalbkhani, "Energy-efficient power allocation and user selection for mmWave-NOMA transmission in M2M communications underlying cellular heterogeneous networks," *IEEE Trans. Veh. Technol.*, vol. 69, no. 9, pp. 9866–9881, Sep. 2020.
- [19] S. Ghosh and D. De, "E²M³: Energy-efficient massive MIMO-MISO 5G HetNet using stackelberg game," *J. Supercomput.*, pp. 1–35, Apr. 2021. [Online]. Available: <https://link.springer.com/article/10.1007/s11227-021-03809-1>, doi: 10.1007/s11227-021-03809-1.
- [20] H. Zhao, Z. Liu, and Y. Sun, "Energy efficiency optimization for SWIPT in K-user MIMO interference channels," *Phys. Commun.*, vol. 27, pp. 197–202, Apr. 2018.
- [21] W. Zhu, P. Xu, H. Jiang, and Y. He, "A resource allocation and cell association algorithm for energy efficiency in HetNets," *Int. J. Commun. Syst.*, vol. 32, no. 16, p. e4123, Nov. 2019.
- [22] D. Xu, X. Chen, C. Wu, S. Zhang, S. Xu, and S. Cao, "Energy-efficient subchannel and power allocation for HetNets based on convolutional neural network," in *Proc. IEEE 89th Veh. Technol. Conf. (VTC-Spring)*, Apr. 2019, pp. 1–5.
- [23] W. U. Khan, X. Li, A. Ihsan, Z. Ali, B. M. Elhalawany, and G. A. S. Sidhu, "Energy efficiency maximization for beyond 5G NOMA-enabled heterogeneous networks," *Peer Peer Netw. Appl.*, vol. 14, pp. 3250–3264, 2021.
- [24] S. Aboagye, A. Ibrahim, T. M. N. Ngatched, and O. A. Dobre, "VLC in future heterogeneous networks: Energy- and spectral-efficiency optimization," in *Proc. IEEE Int. Conf. Commun. (ICC)*, Jun. 2020, pp. 1–7.
- [25] O. Amin, E. Bedeer, M. H. Ahmed, and O. A. Dobre, "Energy efficiency-spectral efficiency tradeoff: A multiobjective optimization approach," *IEEE Trans. Veh. Technol.*, vol. 65, no. 4, pp. 1975–1981, Apr. 2015.
- [26] H. Gao, M. Wang, and T. Lv, "Energy efficiency and spectrum efficiency tradeoff in the D2D-enabled HetNet," *IEEE Trans. Veh. Technol.*, vol. 66, no. 11, pp. 10583–10587, Nov. 2017.
- [27] F. H. Panahi, F. H. Panahi, and T. Ohtsuki, "Spectrum-aware energy efficiency analysis in K-tier 5G HetNets," *Electronics*, vol. 10, no. 7, p. 839, Apr. 2021.
- [28] Z. Ding, F. Adachi, and H. V. Poor, "The application of MIMO to non-orthogonal multiple access," *IEEE Trans. Wireless Commun.*, vol. 15, no. 1, pp. 537–552, Jan. 2016.
- [29] Y. Fu, M. Zhang, L. Salaiin, C. W. Sung, and C. S. Chen, "Zero-forcing oriented power minimization for multi-cell MISO-NOMA systems: A joint user grouping, beamforming, and power control perspective," *IEEE J. Sel. Areas Commun.*, vol. 38, no. 8, pp. 1925–1940, Aug. 2020.
- [30] A. Nasser, O. Muta, H. Gacanin, and M. ElSabrouty, "Joint user pairing and power allocation with compressive sensing in NOMA systems," *IEEE Wireless Commun. Lett.*, vol. 10, no. 1, pp. 151–155, Jan. 2021.
- [31] Y. Fu, L. Salaiin, C. W. Sung, and C. S. Chen, "Subcarrier and power allocation for the downlink of multicarrier NOMA systems," *IEEE Trans. Veh. Technol.*, vol. 67, no. 12, pp. 11833–11847, Dec. 2018.
- [32] S. Wang, M. Ge, and W. Zhao, "Energy-efficient resource allocation for OFDM-based cognitive radio networks," *IEEE Trans. Commun.*, vol. 61, no. 8, pp. 3181–3191, Aug. 2013.
- [33] H. Zhang, D.-K. Zhang, W.-X. Meng, and C. Li, "User pairing algorithm with SIC in non-orthogonal multiple access system," in *Proc. IEEE Int. Conf. Commun. (ICC)*, May 2016, pp. 1–6.
- [34] S.-M. Guo, C.-C. Yang, P.-H. Hsu, and J. S. H. Tsai, "Improving differential evolution with a successful-parent-selecting framework," *IEEE Trans. Evol. Comput.*, vol. 19, no. 5, pp. 717–730, Oct. 2015.
- [35] S. Ali, E. Hossain, and D. I. Kim, "Non-orthogonal multiple access (NOMA) for downlink multiuser MIMO systems: User clustering, beamforming, and power allocation," *IEEE Access*, vol. 5, pp. 565–577, 2017.



AHMED NASSER (Member, IEEE) received the B.Sc. degree (Hons.) in electronics and communications engineering from Suez Canal University, Egypt, in 2012, the M.Sc. degree in electronics and communications engineering from Egypt-Japan University for Science and Technology (E-JUST), Egypt, in 2016, and the dual Ph.D. degree from Kyushu University, Japan, and E-JUST, in 2020. He is currently an Assistant Professor with the Faculty of Engineering, Suez Canal University.

His current research interests include HetNets interference management, NOMA, mMIMO channel estimation, interference alignment, digital signal processing, and emerging technologies for 6G wireless networks. In addition, he has good experience in game theory, compressive sensing, and data-driven-based applications for wireless communications.



OSAMU MUTA (Member, IEEE) received the B.E. degree (associate) from Sasebo Institute of Technology, in 1994, the B.E. degree from Ehime University, in 1996, the M.E. degree from Kyushu Institute of Technology, in 1998, and the Ph.D. degree from Kyushu University, in 2001. In 2001, he joined the Graduate School of Information Science and Electrical Engineering, Kyushu University, as an Assistant Professor. Since 2010, he has been an Associate Professor with the Center for Japan-Egypt Cooperation in Science and Technology, Kyushu University. His current research interests include signal processing techniques for wireless communications and powerline communications, MIMO techniques, interference coordination techniques, and nonlinear distortion compensation techniques for high-power amplifiers. He is a Senior Member of the Institute of Electronics, Information and Communication Engineering (IEICE). He was a recipient of the 2005 Active Research Award in the IEICE Technical Committee of Radio Communication Systems, the Chairperson's Award for excellent paper in the IEICE Technical Committee of Communication Systems, in 2014, 2015, and 2017, and the 2020 IEICE Communication Society Best Paper Award.



HARIS GACANIN (Fellow, IEEE) received the Dipl.-Ing. degree in electrical engineering from the University of Sarajevo, in 2000, and the M.Sc. and Ph.D. degrees from Tohoku University, Japan, in 2005 and 2008, respectively. He was with Tohoku University from 2008 to 2010 first as Japan Society for Promotion of Science (JSPS) Postdoctoral Fellow and later, as an Assistant Professor. He joined Alcatel-Lucent (now Nokia) in 2010, where he worked as a Physical-Layer Expert, the Research Director, and the Department Head with Nokia Bell Labs until 2020. From 2018 to 2020, he was an Adjunct Professor with the University of Leuven (KU Leuven), Belgium. He is currently a Full (Chair) Professor with RWTH Aachen University, Germany. His professional interests are related to broad areas of digital signal processing and artificial intelligence with applications in communication systems. He has more than 200 scientific publications (journals, conferences, and patent applications) and invited/tutorial talks.

He is a Distinguished Lecturer of IEEE Vehicular Technology Society and an Associate Editor of *IEEE Communications Magazine*, while he served as an Editor for *IEICE Transactions on Communications* and *IET Communications*. He is a Senior Member of the Institute of Electronics, Information and Communication Engineering (IEICE) and acted as the General Chair and a Technical Program Committee Member of various IEEE conferences. He was a recipient of several Nokia Innovation Awards, the IEICE Communication System Study Group Best Paper Award (joint 2014, 2015, and 2017), the 2013 Alcatel-Lucent Award of Excellence, the 2012 KDDI Foundation Research Award, the 2009 KDDI Foundation Research Grant Award, the 2008 JSPS Postdoctoral Fellowships for Foreign Researchers, the 2005 Active Research Award in Radio Communications, the 2005 Vehicular Technology Conference (VTC 2005-Fall) Student Paper Award from IEEE VTS Japan Chapter, and the 2004 Institute of IEICE Society Young Researcher Award. He was awarded by Japanese Government (MEXT) Research Scholarship in 2002.



MAHA ELSABROUTY (Senior Member, IEEE) received the B.Sc. degree (Hons.) in electronics and electrical communication engineering from Cairo University, Egypt, and the M.Sc. and Ph.D. degrees in electrical engineering from the University of Ottawa. She is currently with Egypt-Japan University for Science and Technology (E-JUST). Her current research interests include massive MIMO techniques, interference management in HetNets, cognitive radio, intelligent techniques for wireless communications, and green communication systems.

...