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Analysis of Fractional Order Sliding Mode Control in a D-STATCOM Integrated Power Distribution System

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ABSTRACT At present, the disturbances like the voltage fluctuations, resulting from the grid's complexities and unbalanced load conditions, create severe power quality concerns like total harmonic distortion (THD) and voltage unbalance factor (VUF) of the grid voltage. Though the custom power devices such as distribution-static compensators (D-STATCOMs) improve these power quality concerns, however, the accompanying controller plays the substantial role. Therefore, this paper proposes a fractional-order sliding mode control (FOSMC) for a D-STATCOM to compensate the low power distribution system by injecting/absorbing a specific extent of the reactive power under disturbances. FOSMC is a non-linear robust control in which the sliding surface is designed by using the Riemann-Liouville (*RL*) function and the chattering phenomenon is minimized by using the exponential reaching law. The stability of FOSMC is evidenced by employing the Lyapunov stability criteria. Moreover, the performance of the proposed FOSMC is further accessed while doing its parametric variations. The complete system is demonstrated with a model of 400V, 180kVA radial distributor along with D-STATCOM under two test scenarios in MATLAB/Simulink environment. The results of the proposed controller are compared with the fixed frequency sliding mode control (FFSMC) and conventional proportional-integral (PI) control. The results validate the superiority of the proposed controller in terms of rapid tracking, fast convergence, and overall damping with very low THD and VUF.

INDEX TERMS Power quality, custom power devices, distribution static compensator, fractional order sliding mode control, total harmonic distortion, voltage unbalance factor.

NOMENCLATURE

Abbreviations:

| | |
|-------|----------------------------------------|
| AMF | Asymmetric Membership Function |
| CFNN | Compensatory Fuzzy Neural Network |
| CHBMC | Cascaded H-bridge Multilevel Converter |
| CSC | Current Source Converter |
| DC | Direct Current |

| | |
|-----------|---------------------------------------------|
| DCVSA | Discrete Continuous Vortex Search Algorithm |
| D-STATCOM | Distribution Static Compensator |
| DG | Distribution Generation |
| DPC | Direct Power Control |
| FCS | Finite Control Set |
| FLC | Fuzzy Logic Control |
| GOA | Grasshopper's Optimization Algorithm |
| HV | High Voltage |
| IP | Integral Proportional |
| LADRC | Linear Active Disturbance Rejection Control |

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| | |
|----------|----------------------------------------------------------------|
| LV | Low voltage |
| MC | Matrix Converter |
| MG | Microgrid |
| MPC | Model Predictive Control |
| NFC | Neuro Fuzzy Control |
| PID | Proportional Integral Derivative |
| PLL | Phase Lock Loop |
| PR | Proportional Resonant |
| PSO | Particle Swarm Optimization |
| PV | Photovoltaic |
| RL | Riemann Liouville |
| SMC | Sliding Mode Control |
| SDBC-MMC | Single Delta Bridge Cell-Modular Multi-level cascade converter |
| SRLMMN | Sign Regressor Least Mean Mixed Norm |
| SVC | Static VAR Compensator |
| SPWM | Sinusoidal Pulse Width Modulation |
| THD | Total Harmonic Distortion |
| TSA | Tree Seed Algorithm |
| VSC | Voltage Source Converter |
| VUF | Voltage Unbalance Factor |

I. INTRODUCTION

In the power distribution system, power quality implies to keep the AC bus sinusoidal voltage waveform at the measured voltage and frequency [1]. The waveforms of voltage and current must be purely sinusoidal and free from any interruptions [2]. However, the grid transients and unbalanced loads may distort the waveform and cause voltage instability [3]. These distortions may disseminate all over the power distribution system [4]. Power quality concerns incorporate grid transients (voltage sag/swell), voltage and current distortion, voltage spike, voltage unbalance, noise, harmonics, and flickers. These can initiate abnormal operation of the equipment or assuredly trip the protection devices [5]. To cope with these concerns, fixed and switch reactors/capacitors banks and static VAR compensators (SVC) are established and executed [6]. However, these kinds of standard facilities have a few drawbacks such as large size, additional losses, inadequate bandwidth, and slow response time [7]. The two-level converters can also be employed to compensate the low power distribution system. However, the limited power rating and switching losses of the two-level converters resulting from high-frequency operation in the high power and high voltage applications are considered constraints [8]. In recent times, D-STATCOM has gained enough popularity with the development of semiconductor devices due to its high-power density, small size, minimal losses, fast response, and broad compensation range [9]. D-STATCOM is a power quality conditioner, and capable of supplying and absorbing reactive power, and is integrated into the grid via a VSC as well [10]. D-STATCOM is linked in parallel with the load and can operate uninterruptedly, attaining a power factor near the unity and preventing operational concerns [11]. For reactive power compensation, the D-STATCOM is one of the fastest

and most reliable device. By monitoring the magnitude of voltage, the exchange of reactive power between the power distribution system and D-STATCOM can be controlled to alleviate the power quality concerns [12]. Furthermore, an efficient control scheme is also required to rapidly achieve the necessary compensation in a low power distribution system [13].

In recent years, many scholars have accomplished a lot of research work regarding D-STATCOM integration in the power distribution system using linear control strategies. The work in [14], has analyzed the execution of D-STATCOM to reduce the undesirable effect generated by the DG linked to a grid. The voltage profile is improved via the DPC approach employing PI control during steady-state and voltage sag conditions. In [15], the authors have proposed a D-STATCOM to compensate the unbalanced conditions resulting in dc voltage oscillation via PR control. A PID control is employed for SDBC-MMC to compensate the single-phase load in a power distribution system for power factor correction [16]. The authors in [17], have presented a cascaded multi-level inverter-based D-STATCOM during grid transients and load alterations via PI control for reactive power compensation. In [18], the authors have proposed a three-phase four-wire VSC based D-STATCOM using PI control. The D-STATCOM model is utilized under an unbalanced load condition to stabilize the voltage. In [19], the authors have proposed a D-STATCOM for the distribution system to mitigate voltage unbalanced effect created by U-V or V-V winding of transformer employing PI control. In [20], the authors have proposed a D-STATCOM for a renewable energy-based distribution system. The PI control is applied to alleviate the voltage sag/swell problem. The authors in [21], have proposed a CSC-based D-STATCOM which utilizes the minimum DC voltage for the reactive power compensation to eliminate the harmonics and reduces the switching losses using the PI control. In [22], the authors have proposed a combined PV-STATCOM via IP and PID control under an unbalanced non-linear load to reduce the harmonics, zero-sequence component, and the THD. The dynamic response of MG is enhanced by using the D-STATCOM with an improved PID control based on the application of GOA [23]. In [24], the authors have proposed a D-STATCOM based on CSC to enhance the power quality and consistency during step alter in load utilizing PI control. The CHBMC based D-STATCOM using PI control is proposed in [25] to enrich the DC voltage and reactive power control. The control parameters of the PI control are optimized by using the PSO. The authors in [26], have proposed a D-STATCOM to enhance the power quality problem resulted from an unbalanced load condition using the PI control. In [27], the authors have proposed a PI control for parallel operation among D-STATCOMs for the reactive power-sharing, and to compensate the harmonics resulting from the non-linear load. The authors in [28], have proposed a PI control by utilizing the PLL algorithm for a D-STATCOM to enhance the power factor and dc voltage regulation. In [29], the authors have proposed a multi-level CHB D-STATCOM

via PI control to enhance the power factor by eliminating the harmonics and the switching losses. In [30], the authors have investigated the execution of D-STATCOM for reactive power compensation required by doubly-fed wind farm linked to the grid. The power quality concerns alleviated by D-STATCOM in [31] employing PR control and comb filter during unbalanced and non-linear scenarios. Though, the linear control strategies have major limitations such as tuning of control constraints, inadequate transient response [32], and deficient reference tracking in the situation of challenging a significant error [33].

In the context of non-linear control strategies, the authors in [34] have investigated the performance of D-STATCOM via dead-beat repetitive control to diminish the current prediction error for improved reactive power compensation. However, the control has few drawbacks such as parameter variance and external disturbances [35]. The authors in [36] have proposed a D-STATCOM to compensate the unbalanced voltage condition via hysteresis current controller. Though, in the digital platform, the modeling and implementation of hysteresis control are complex and suffer from the variable switching frequency [37]. A PI control based on fuzzy logic is executed for the D-STATCOM to maintain the voltage stability in the power distribution system [38]. Even though, FLC has its limitations due to the fact that it cannot handle situations outside its proposed fuzzy rules, and it is difficult to prove the stability of FLC [39]. The authors in [40], have proposed an NFC for a D-STATCOM during load variations. In [41], the authors have proposed a D-STATCOM via CFNN-AMF controller for the dc voltage regulation and power quality enrichment. However, the dynamic response of the controller is moderate. In [42], a shunt connected D-STATCOM is investigated in the power distribution system to compensate the reactive power required for load balancing. They proposed an SRLMMN control for fast convergence and to reduce steady-state error. The authors in [43], have proposed a cluster balancing control to alleviate the effect of voltage sag from a single-phase, two-phase, and three-phase power distribution system. Despite that, while acknowledging the learning algorithm, the controller needed additional computational time [44]. In [45], the authors have proposed an enhanced LADRC for D-STATCOM to competently control the DC voltage and the reference reactive power during voltage sag and load variations. The authors in [46], have proposed a capacitor-less D-STATCOM based on MC by employing FCS-MPC for reactive power compensation under non-linear load conditions. Though, the switching losses reduce the system reliability due to high fidelity [47]. Thus, an adaptive MPC-based capacitor less D-STATCOM is proposed in [48] to diminish the switching frequency and THD. However, to reach the desired accomplishment the MPC necessitates precise control parameters. Moreover, the cost function calculation requires more computation burden [49]. In [50], the authors have proposed a digital current control for D-STATCOM during unbalanced and non-linear load conditions. The authors in [51], have investigated the execution

of D-STATCOM in a T-type configuration by utilizing a simple carrier-based modulation control method to minimize the voltage fluctuation and harmonics at dc link. The optimum location and sizing of D-STATCOM are determined in [52] by employing DCVSA to reduce the operational cost. In [53], the authors have proposed a novel TSA-based D-STATCOM to improve the execution and consistency during normal and critical load conditions.

Sliding mode control is a well-known non-linear control that offers robustness, fast dynamic response, stability, enhanced regulation properties, and highly compatible with power converters [54]. An SMC is employed for a D-STATCOM evaluation and design [55] to compensate the power distribution system under voltage sag. Nevertheless, the control eliminates the chattering problem and stabilizes the output voltage to some extent. Despite this, the response time is very slow which may cause stability issues. The authors in [56], have proposed a three-level four-leg D-STATCOM for reactive power compensation using second-order SMC during grid faults and unbalanced load circumstances. Though, mathematical modeling and execution are quite complex. Moreover, during unbalanced load conditions, the THD of the output is high as well. The authors in [57], have proposed and investigate the performance of a D-STATCOM based on wind farm DFIG via fractional-order SMC. Nevertheless, the voltage unbalance factor (VUF) is not taken into account under unbalanced load conditions. In [58], a photovoltaic (PV) based D-STATCOM is investigated under harmonics and load variations while utilizing SMC. Though, the robustness and tracking error of this control scheme is nearly degraded. An integral SMC is employed for a D-STATCOM to compensate the load alterations [59]. Although, the control provides a good tracking response under unbalanced load conditions. Nevertheless, the output voltage waveform indicates the chattering problem and stability issue as well. In [60], the authors have proposed an SMC for a cascaded STATCOM to compensate the voltage unbalance and voltage sag problem. Although, the response time of control to reach a steady-state is very slow. In [61], the authors have investigated the performance of D-STATCOM by using the input-output feedback linearization scheme. They have proposed SMC to improve the three-phase voltage at LV AC bus and dc side voltage of D-STATCOM. Though, the robustness of the controller is lacking. In [62], the authors have proposed an SMC for D-STATCOM under DC link voltage variations to enrich the system stability. The authors in [63], have proposed an SMC for the D-STATCOM to eliminate the harmonics created by nonlinear load. Nevertheless, the size and cost of the system is increase due to large values of output filter selected to eliminate the harmonics. The authors in [64] have proposed an enhanced D-STATCOM model under voltage sag/swell using SMC. The authors claim that the system efficiency increases, although its performance is inadequate in terms of THD under voltage sag/swell.

After a comprehensive review of the literature survey, following conclusions could be drawn.

- The key emphasis of the D-STATCOM based power distribution system is to improve the power quality in the low power distribution systems.
- As a result of the escalating number of power electronic devices, grid transients, and sensitive loads, the future trend highly demands improved power quality in the low power distribution systems by employing the D-STATCOM. Consequently, more advanced controller is required to tackle these problems. The application of a non-linear robust controllers is found deficient in the published literature.
- The FOSMC is not implemented in the literature studies for power quality problems such as grid transients (voltage sag/swell) and unbalanced load conditions.
- Besides, the performance of the FOSMC under its parametric variations on a D-STATCOM is found missing.
- The stability analysis of the applied controllers, and the phenomenon of chattering elimination is deficient as well.
- The essential measure VUF for evaluating a controller under unbalanced load conditions has been seen to be lacking in published literature for this particular problem.

In this paper, the authors have proposed the FOSMC for D-STATCOM to compensate the low power distribution system under disturbances. A model of radial distributor along with D-STATCOM is examined under grid transients (voltage sag/swell) and unbalanced load conditions. The D-STATCOM injects/absorbs the specific amount of reactive power at the LV AC bus depending upon the disturbances. Besides, the performance of the FOSMC under its parametric variations on a D-STATCOM is discussed as well. In the aspect of robustness, the FOSMC is superior by virtue of its additional design parameters comprising of adjustable non-integer differentiator and integrator [65]. This improves the dynamic response of the system, establishes more degree of freedom, ensures the fast convergence, and gives an adequate method to eliminate the chattering phenomenon [66]. The conventional SMC has certain challenges such as chattering phenomenon and inadequate convergence which might result in unpredicted high-frequency oscillations and diminish the system execution [67]. FOSMC strategy, even though confirms to be competent of alleviating the chattering intensity in the sliding surface and quickening the reaching speed as well [68]. In this paper, the Riemann-Liouville (*RL*) function is utilized to design the sliding surface. Furthermore, the stability of the FOSMC is evidenced by employing Lyapunov stability criteria. To summarize, the authors' key contributions include:

- The D-STATCOM integrated power system is proposed to improve the power quality under disturbances.
- Two test scenarios have been established and examined. The grid transients (voltage sag/swell) are considered at the LV AC bus in the first scenario. Likewise, in second scenario an unbalanced load conditions are considered

at the LV AC bus. D-STATCOM sustains the voltage at LV AC bus by injecting/absorbing a certain extent of reactive power in both cases.

- FOSMC is proposed to drive the D-STATCOM under the disturbances.
- Besides, performance of the FOSMC under its parametric variations on a D-STATCOM is discussed as well.
- The sliding surface of the proposed FOSMC is designed by using the Riemann Liouville (*RL*) function. Additionally, the chattering phenomenon is minimized by using the exponential reaching law.
- The stability of the FOSMC is evidenced by employing Lyapunov stability criteria.
- The THD and VUF of the output voltage have been considered to evaluate the controller's performance.
- Furthermore, the experimental results of the proposed control are compared with FFSMC and PI control as well.
- The performance of proposed FOSMC outperformed the literature study reference [57] in terms of controller robustness and limited scope of study.

Rest of the paper is arranged in the following approach. Section II describes the model structure, design, and control of D-STATCOM. Section III derives the mathematical modelling of FOSMC for D-STATCOM. Section IV proves the stability of proposed FOSMC. Section V made the results and discussions. Section VI concludes the paper.

II. MODEL STRUCTURE, DESIGN, AND CONTROL OF D-STATCOM

Fig. 1 shows the simplified model of D-STATCOM configuration [69] which includes a Y-Δ and Δ-Δ distribution transformer, dynamic load (linear and unbalanced), main grid, and a D-STATCOM.

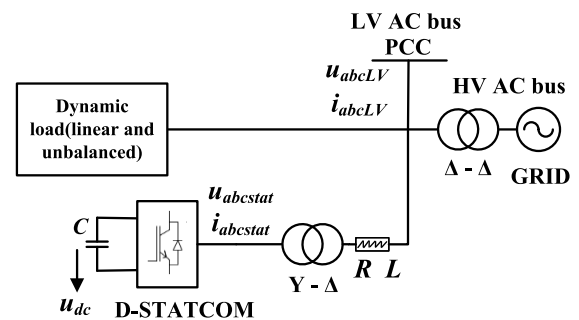


FIGURE 1. Simplified model of D-STATCOM configuration.

The main grid is linked at the LV AC bus via Δ-Δ distribution transformer. The 1500V D-STATCOM [70] is linked in shunt with the main grid via Y-Δ distribution transformer at LV AC bus and connected to a capacitor by its dc bus. The magnitude of the voltage generated by a D-STATCOM is not the same as grid voltage, although have the same phase. The compensation process of the D-STATCOM is altered on the

divergence among voltage at LV AC bus and D-STATCOM voltage. A D-STATCOM injects the reactive power while considered $u_{abcstat}$ is higher than u_{abcLV} . Likewise, a D-STATCOM absorbs the reactive power while considered $u_{abcstat}$ is lower than u_{abcLV} . To design the three-phase D-STATCOM, the value of C_{dc} can be obtained by using Eq. (1) [70]

$$C_{dc} = \frac{3u_s \Delta I_L T}{u_{cmax}^2 - u_{dc}^2} \quad (1)$$

where, C_{dc} , u_{dc} , u_{cmax} , ΔI_L , u_s , and T denotes the dc capacitor, capacitor voltage, pre-set threshold of capacitor voltage, 5% of load current (current ripple ratio), D-STATCOM peak phase voltage, and time period of one cycle, respectively. The high harmonics that are multiples of switching frequency are removed by using the three-phase filter. In the absence of a filter, the noise will add in line with injected voltages. To overcome this problem, an LC filter is utilized at the secondary side of the distribution transformer. The value of the filter inductor and capacitor is chosen appropriately by using Eq. (2) [87] and Eq. (3) [88].

$$L = \frac{\sqrt{3}m u_{dc}}{12f_s a \Delta I_L} \quad (2)$$

$$f_s = \frac{1}{2\pi \sqrt{LC}} \quad (3)$$

where, terms m , f_s , a indicates modulation index, switching frequency, and overloading factor, respectively. Fig. 2 shows the proposed D-STATCOM control scheme. The three-phase voltages and currents (u_{abcLV} , i_{abcLV} , $u_{abcstat}$, $i_{abcstat}$) are signified as the voltages and currents at LV AC bus and D-STATCOM output. While the transformer and line parameters are characterized by R and L . The direct quadrature (dq) approach is applied to both voltages and currents to enhanced reference tracking for FOSMC loops and respective SPWM gate signal generation. The transformation angle θ is sensed at the LV AC bus by applying a PLL. Fig. 3 represents the flowchart of the D-STATCOM compensation procedure.

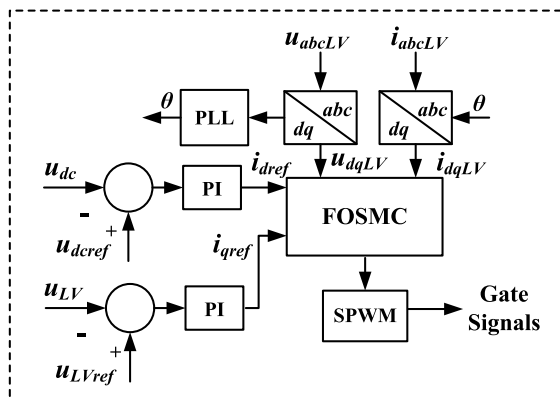


FIGURE 2. Proposed D-STATCOM control scheme.

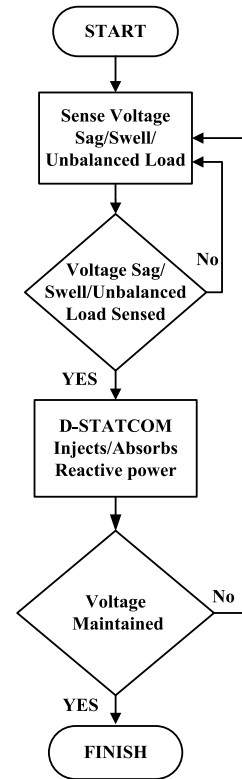


FIGURE 3. Flowchart of D-STATCOM compensation procedure.

III. MATHEMATICAL MODEL OF D-STATCOM USING FRACTIONAL ORDER SLIDING MDOE CONTROL

The D-STATCOM dynamics in abc -stationary reference frame is modelled as [72]

$$\frac{di_{abcLV}}{dx} = \frac{-Ri_{abcLV}}{L} + \frac{u_{abcstat}}{L} - \frac{u_{abcLV}}{L} \quad (4)$$

In synchronous reference frame (dq structure) along with addition of D-STATCOM modulating signal ($u_{stat} = u_{dc}/2m$), and $\omega = d\theta/dt$, Eq. (4) can be modified as

$$\frac{di_{dLV}}{dt} = \frac{-Ri_{dLV}}{L} + \omega i_{qLV} + \frac{u_{dc}m}{2L} - \frac{u_{dLV}}{L} \quad (5)$$

$$\frac{di_{qLV}}{dt} = \frac{-Ri_{qLV}}{L} - \omega i_{dLV} + \frac{u_{dc}m}{2L} - \frac{u_{qLV}}{L} \quad (6)$$

With the addition of model uncertainty terms (ρ_d, ρ_q) [65], Eq. (5) and Eq. (6) can be modified as

$$\frac{di_{dLV}}{dt} = f_d(u_{dLV}, u_{qLV}, i_{dLV}, i_{qLV}) + xm_d + \rho_d \quad (7)$$

$$\frac{di_{qLV}}{dt} = f_q(u_{dLV}, u_{qLV}, i_{dLV}, i_{qLV}) + xm_q + \rho_q \quad (8)$$

where,

$$x = \frac{u_{dc}}{2L} \quad (9)$$

The most significant cause of model uncertainties includes variation in system parameters such as mechanical stresses,

expansion, self-thermal alterations, load demand, a failure rate of transmission lines, generation output, line outage and generator outage [83] can significantly disrupt the system stability and reduces the system performance [84]. The influence of uncertain parameters is usually not substantial for unstressed systems. When the stability margin declines, though, system behavior becomes considerably additional sensitive to parameter perturbations [85]. In this aspect, the robust FOSMC is designed against external disturbance and parameter uncertainties. Besides, the utilization of fractional-order differentiator and integrator offers an extra degree of freedom and enhances the system convergent properties [86].

The non-linear functions terms $f_d(\cdot)$ and $f_q(\cdot)$ are described as

$$f_d = \frac{di_{dLV}}{dt} = \frac{-Ri_{dLV}}{L} + \omega i_{qLV} - \frac{u_{dLV}}{L} \quad (10)$$

$$f_q = \frac{di_{qLV}}{dt} = \frac{-Ri_{qLV}}{L} - \omega i_{dLV} - \frac{u_{qLV}}{L} \quad (11)$$

The proposed FOSMC non-integer sliding surface for d and q control loops are described as [69]

$$S_d = e_d + \lambda D^{\alpha-1}(\text{sig}(e_d)^\gamma) \quad (12)$$

$$S_q = e_q + \lambda D^{\alpha-1}(\text{sig}(e_q)^\gamma) \quad (13)$$

where,

$$e_d = i_{dref} - i_{dLV} \quad (14)$$

$$e_q = i_{qref} - i_{qLV} \quad (15)$$

The e_d and e_q terms are related to signify current reference tracking error, $D^{\alpha-1}$ signifies the fractional integral term of $(\alpha - 1)$ order, and α, γ, λ signifies the positive parameters along design choices of $(\alpha < 1$ and $\gamma < 1)$.

The well-known $\text{sig}(\cdot)^\gamma$ function is described as [65]

$$\text{sig}(x)^\gamma = |x|^\gamma \text{sgn}(x) \quad (16)$$

where, $\text{sgn}(\cdot)$ signifies the sign function described as

$$\text{sgn}(x) = \begin{cases} \frac{x}{|x|}, & \text{if } x \neq 0 \\ 0, & \text{if } x = 0 \end{cases} \quad (17)$$

Now taking the derivative of Eq. (12) and Eq. (13) and substituting the Eq. (14) and Eq. (15) into Eq. (12) and Eq. (13), the proposed sliding surface equation with the inclusion of RL function [73] is written as

$$S_d^\bullet = \dot{i}_{dref} - \dot{i}_{dLV} + \lambda_{RL} D^\alpha(\text{sig}(e_d)^\gamma) \quad (18)$$

$$S_q^\bullet = \dot{i}_{qref} - \dot{i}_{qLV} + \lambda_{RL} D^\alpha(\text{sig}(e_q)^\gamma) \quad (19)$$

By substituting Eq. (10) and Eq. (11) into Eq. (18) and Eq. (19), the modified equation can be written as

$$S_d^\bullet = f_d(\cdot) + x m_d + \rho_d + \lambda_{RL} D^\alpha(\text{sig}(e_d)^\gamma) \quad (20)$$

$$S_q^\bullet = f_q(\cdot) + x m_q + \rho_q + \lambda_{RL} D^\alpha(\text{sig}(e_q)^\gamma) \quad (21)$$

Based on Eq. (20) and Eq. (21), the proposed control law ensures the current reference tracking error convergence and

generate the modulating signals m_d and m_q for SPWM can be written as

$$m_d = \frac{-[f_d(\cdot) + \lambda_{RL} D^\alpha(\text{sig}(e_d)^\gamma) + k_d \text{sgn}(s_d)]}{x} \quad (22)$$

$$m_q = \frac{-[f_q(\cdot) + \lambda_{RL} D^\alpha(\text{sig}(e_q)^\gamma) + k_q \text{sgn}(s_q)]}{x} \quad (23)$$

where, terms k_d and k_q signifies the FOSMC sliding gains. Fig. 4 shows the control diagram of FOSMC for D-STATCOM in q -axis. Likewise, same methodology is adopted for d -axis. Fig. 5 shows the flowchart of proposed FOSMC strategy.

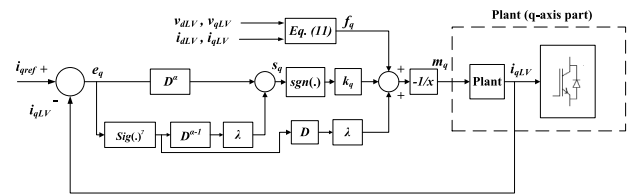


FIGURE 4. Control diagram of FOSMC for D-STATCOM in q -axis.

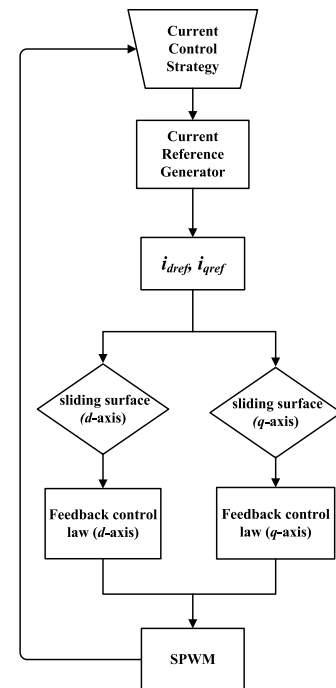


FIGURE 5. Flowchart of FOSMC strategy.

IV. STABILITY ANALYSIS

Considering the definition of Lyapunov function for the stability of proposed FOSMC is described as [74]

$$V(t) = \frac{1}{2}(S_d^2 + S_q^2) \quad (24)$$

By taking the time derivative of Eq. (24) one obtains

$$V^\bullet(t) = S_d S_d^\bullet + S_q S_q^\bullet \quad (25)$$

Based on Eq. (7) and Eq. (8), and Eq. (25) the time derivative Lyapunov function signified as

$$V^\bullet(t) = S_d(f_d(\cdot) + xm_d + \rho_d + \lambda_{RL}D^\alpha(\text{sig}(e_d)^\gamma) + S_q(f_q(\cdot) + xm_q + \rho_q + \lambda_{RL}D^\alpha(\text{sig}(e_q)^\gamma)) \quad (26)$$

By substituting m_d and m_q from Eq. (22) and Eq. (23) into Eq. (26), the modified equation is written as

$$V^\bullet(t) = S_d(\rho_d - k_d \text{sgn}(S_d)) + S_q(\rho_q - k_q \text{sgn}(S_q)) \quad (27)$$

Considering $\text{sgn}(S_d) = |S_d|/S_d$ and $\text{sgn}(S_q) = |S_q|/S_q$

$$V^\bullet(t) = (S_d \rho_d - k_d |S_d|) + (S_q \rho_q - k_q |S_q|) \quad (28)$$

Now selecting $(k_d = |\rho_d| + \xi_d)$ and $(k_q = |\rho_q| + \xi_q)$, where ξ_d and ξ_q signify as positive parameters. Eq. (28) one finds

$$V^\bullet(t) \leq -\xi_d |S_d| - \xi_q |S_q| \leq -\min(\xi_d, \xi_q)(|S_d| + |S_q|) = -\min(\xi_d, \xi_q) \|S\|_1 \quad (29)$$

The minimum value of ξ_d and ξ_q represents minimum (ξ_d, ξ_q) . Eq. (29) verifies the stability condition and finite time convergence of proposed FOSMC sliding surface S_d and S_q . Consequently, $V^\bullet(t)$ is negative, and the proposed FOSMC system is asymptotic stable [69].

Remarks:

Afterwards the reaching condition is guaranteed, the Lemma (Lyapunov stability theorem) can be conferred to examine the system stability during the sliding phase.

Lemma 1: The system is asymptotic stable approaching 0 when the positive definite function $V(t) > 0$ exists and satisfy $V^\bullet(t) < 0$, and $V(t)$ is negative definite [81].

Lemma 2: The system is unstable approaching 0 when the positive definite function $V(t) > 0$ exists and satisfy $V^\bullet(t) > 0$, and $V(t)$ is positive definite [82].

V. SYSTEM BASED ANALYSIS AND RESULT DISCUSSIONS

In this section, a detailed model of D-STATCOM, based on Fig. 1 [69] has been developed and accomplished in MATLAB Simulink environment to validate the performance of the proposed FOSMC under two test scenarios. Besides, the performance of the FOSMC under its parametric variations on a D-STATCOM is discussed as well. The implementation of the proposed FOSMC is compared with FFSMC and PI control. In the first test scenario, the main grid signified by a power generator resulting in grid transients (voltage sag/swell) at the LV AC bus. In the second test scenario, the voltage and current are unbalanced, resulting from unbalanced load conditions at the LV AC bus. D-STATCOM injects/absorbs the required reactive power to keep the constant voltage at the LV AC bus in both scenarios. The parameters of the proposed power distribution system are presented in TABLE 1 [69]. The control parameters of the proposed FOSMC and FFSMC are presented in TABLE 2 [69] and TABLE 3 [75], respectively. The parameters of PI control are chosen according to the optimal setting to achieve fast output stabilization and are presented in TABLE 4 [76].

TABLE 1. Parameters of proposed electrical power distribution system.

| Symbol | Parameter | Value |
|----------|----------------------------------|-----------|
| u_{LV} | Nominal Voltage at LV AC bus | 312V |
| L_{ac} | Dynamic Load | 50kVA |
| ω | Angular Frequency | 3.77rad/s |
| L | Coupling Inductance of D-STATCOM | 2.89mH |
| R | Coupling Resistance D-STATCOM | 0.1mH |
| f_s | Switching Frequency of D-STATCOM | 10kHz |
| Q | Nominal Power of D-STATCOM | +50kVAR |
| u_{dc} | DC Link Voltage of D-STATCOM | 1500V |

TABLE 2. Parameters of fractional order SMC.

| Symbol | Parameter | Value |
|------------|----------------------------------|-----------------|
| γ | Fractional Order SMC Parameter | 0.9 |
| λ | Fractional Order SMC Parameter | 1500 |
| α | Fractional Order SMC Coefficient | 0.5 |
| k_d, k_q | Fractional Order Sliding Gains | 5×10^6 |

TABLE 3. Parameters of fixed frequency SMC.

| Symbol | Parameter | Value |
|--------------|-------------------------------|-------|
| ϵ_d | Fixed Frequency SMC Parameter | 200 |
| k_d | Fixed Frequency SMC Gain | 100 |
| ϵ_q | Fixed Frequency SMC Parameter | 200 |
| k_q | Fixed Frequency SMC Gain | 50 |

TABLE 4. Parameters of PI control.

| Symbol | Parameter | Value |
|----------|----------------------------------|-------|
| k_{pi} | Proportional Gain (Current Loop) | 1 |
| k_{it} | Integral Gain (Current Loop) | 3.3 |
| k_{pv} | Proportional Gain (Voltage Loop) | 5 |
| k_{iv} | Integral Gain (Voltage Loop) | 10 |

A. STEADY STATE PERFORMANCE AND FOSMC PARAMETERS SELECTION

The proposed FOSMC accomplishment is based on its control parameter variations [77]. Besides, the resemblance between proposed FOSMC, FFSMC, and PI control strategies required a steady-state test to evaluate the execution. The foremost concerns of the execution assessment are the chattering phenomenon and amplitude of the error signal [78]. The chattering phenomenon ruins the system stability and is accounted as a low order noise in the power distribution system [73]. Due to the instability of error signals among error boundaries, the chattering problem is created [77]. The conventional SMC does not have any control on chattering phenomena however the FOSMC is competent to diminish the error signal and convergence within the fraction of chattering time interval via choosing the suitable value of fractional coefficient α [65]. The fractional-order can be selected by examining the complete realization of control outcomes with varied control parameters. The increase of fractional coefficient α reduces the effect of the $(D^{\alpha-1}(\text{sig}(e_d)^\gamma))$ integral term on the FOSMC sliding surface, and simultaneously increases the signal tracking error [65], and reduces the convergence time [69]. Similarly, due to the presence of model

uncertainties and external disturbances, the fractional order sliding gains k_d and k_q are chosen to satisfy the reaching condition [79].

B. PERFORMANCE EVALUATION UNDER VOLTAGE SAG/SWELL OF MAIN GRID

In this test scenario, the proposed D-STATCOM model employing FOSMC is evaluated under voltage sag/swell in an entire simulation of 0.2s. Initially, the nominal voltage of 312V is maintained at LV AC bus under normal conditions. The voltage sag initiates at 0.05s results in a voltage decline of 10% at LV ac bus. Similarly, voltage swell initiates at 0.1sec results in a voltage rise of 10% at LV AC bus. The system returns to normal condition in the simulation time of 0.15s. Fig. 6(a) shows the three-phase voltage waveform at LV AC bus during voltage sag/swell. Fig. 6(b) shows the RMS voltage waveform at LV AC bus during voltage sag/swell. Fig. 6(c) shows the load active and reactive power waveform under voltage sag/swell.

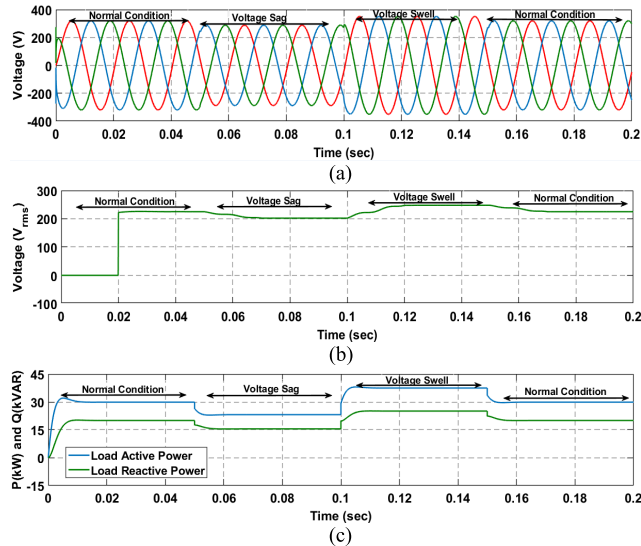


FIGURE 6. (a) Three-phase voltage at LV AC bus under voltage sag/swell of Main Grid (b) RMS Voltage at LV AC bus under voltage sag/swell of Main Grid (c) Load Active and Reactive power under voltage sag/swell of main grid.

Besides the operation of the proposed D-STATCOM, the voltage at the LV AC bus is made constant under voltage sag/swell. The D-STATCOM maintains the voltage at the LV AC bus while injects/absorbs the reactive power required. Fig. 7(a) and Fig. 7(b) shows the three-phase and RMS voltage waveform at LV AC bus during voltage sag/swell with D-STATCOM while employing PI control. Fig. 7(c) and Fig. 7(d) shows the three-phase and RMS voltage waveform at LV AC bus during voltage sag/swell with D-STATCOM while employing FFSMC. Fig. 7(e) and Fig. 7(f) shows the three-phase and RMS voltage waveform at LV AC bus during voltage sag/swell with D-STATCOM while employing FOSMC. Nevertheless, the execution of the proposed FOSMC is extremely exceptional regarding rapid reference

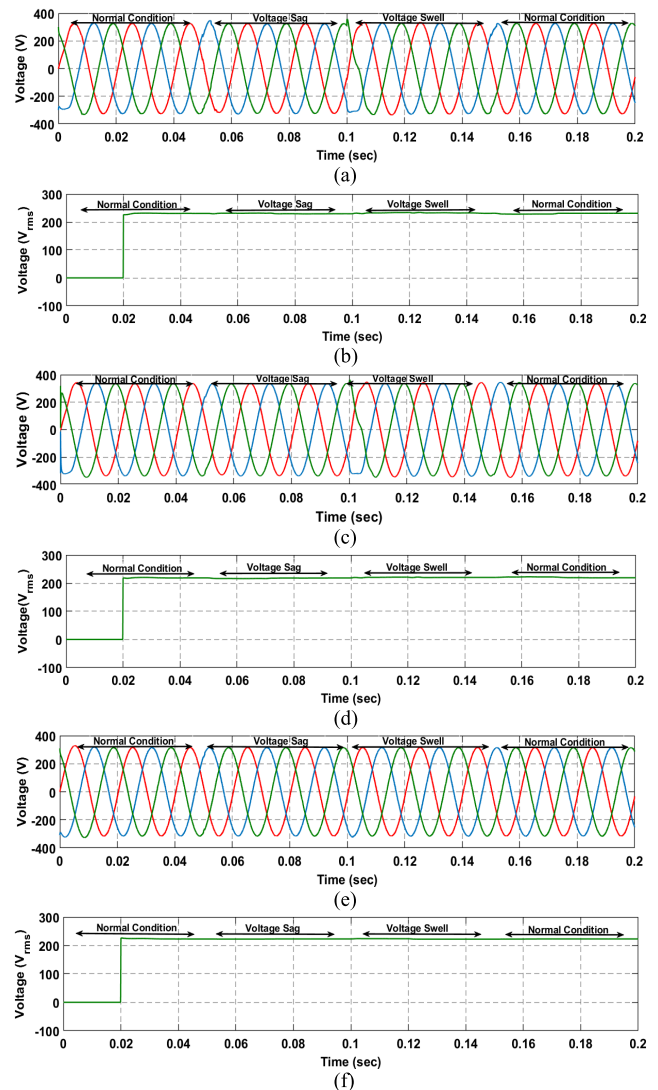


FIGURE 7. (a) Three-phase Voltage at LV AC bus under voltage sag/swell of main Grid with D-STATCOM while employing PI control (b) RMS Voltage at LV AC bus under voltage sag/swell of main Grid with D-STATCOM while employing PI control (c) Three-phase Voltage at LV AC bus under voltage sag/swell of main Grid with D-STATCOM while employing FFSMC (d) RMS Voltage at LV AC bus under voltage sag/swell of main Grid with D-STATCOM while employing FFSMC (e) Three-phase Voltage at LV AC bus under voltage sag/swell of main Grid with D-STATCOM while employing FOSMC (f) RMS Voltage at LV AC bus under voltage sag/swell of main Grid with D-STATCOM while employing FOSMC.

tracking and fast convergence in contrast to FFSMC and PI control. The assessment of voltage THD at LV AC bus during voltage sag/swell is listed in TABLE 5. It is evident that the voltage THD of proposed FOSMC during voltage sag/swell results in 0.52% in contrast to FFSMC and PI control and are well under the IEEE standards which verifies the competency and exceptional performance of proposed control.

Fig. 8(a) shows the load active and reactive power under voltage sag/swell with D-STATCOM at LV AC bus while employing PI control. Similarly, Fig. 8(b) shows the load active and reactive power under voltage sag/swell with D-STATCOM at LV AC bus while employing FFSMC.

TABLE 5. Assessment of voltage THD at LV AC bus during voltage Sag/Swell.

| Controller Strategy | Load Type | THD (%) |
|---------------------|---------------|---------|
| PI Control | Balanced load | 2.17 |
| FFSMC | Balanced load | 0.84 |
| FOSMC | Balanced load | 0.52 |

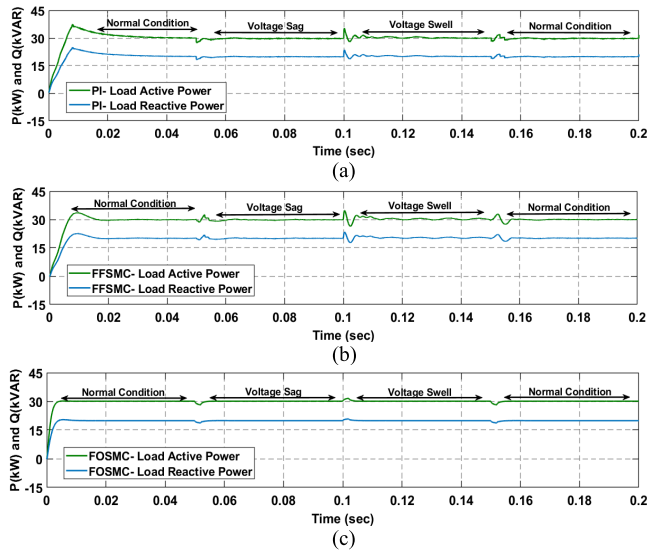


FIGURE 8. (a) Load Active and Reactive power under voltage sag/swell with D-STATCOM while employing PI control (b) Load active and reactive power under voltage sag/swell with D-STATCOM while employing FFSMC (c) Load active and reactive power under voltage sag/swell with D-STATCOM while employing FOSMC.

Fig. 8(c) shows the load active and reactive power under voltage sag/swell with D-STATCOM at LV AC bus while employing FOSMC. The execution of the proposed FOSMC can be seen exceptional in terms of fast convergence, upright damping, and instant tracking, in distinction to PI control and FFSMC.

Fig. 9(a) shows the injected/absorbed reactive power by D-STATCOM under voltage sag/swell while employing PI control. Fig. 9(b) shows the injected/absorbed reactive

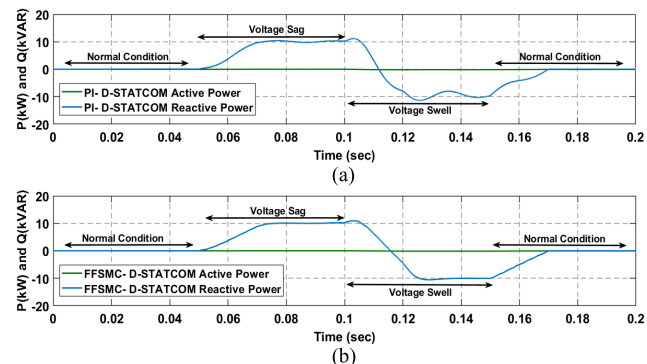


FIGURE 9. (a) Injected/absorbed Reactive power by D-STATCOM under voltage sag/swell while employing PI control (b) Injected/absorbed Reactive power by D-STATCOM under voltage sag/swell while employing FFSMC.

power by D-STATCOM under voltage sag/swell while employing FFSMC. Fig. 10 shows the variation of α , k_d and k_q , and its influence on rise time, convergence time, tracking error, and time to reach the steady-state under voltage sag/swell. The optimum value of α can be retrieved by varying it from 0 to 0.9. Initially, the value of the fractional coefficient α is chosen from [65]. From Fig. 10(a), it can be observed that the increase of $\alpha = 0.8$ increases the tracking error and chattering. Likewise, the decrease of $\alpha = 0.2$ increases the overshoot, rise time, and time to reach the steady-state, can be seen in Fig. 10(c). In the context of k_d and k_q , the initial values are selected from [69]. The decrease in the values of k_d and k_q results in the increase of tracking error and the time to reach the steady-state and it can be seen in Fig. 10(d). Meanwhile, the increase of k_d and k_q results in the decrease of tracking error and the time to reach the steady-state. Therefore, the value of FOSMC parameters must be selected in such a way that a compromise is achieved among the rise time, overshoot, convergence time, tracking error, and chattering free smooth output. The reference reactive power value is admirably tracked by the controlled trajectories when α is selected to be 0.5, and k_d, k_q are selected to be 5×10^6 , evidenced in Fig. 10(b). The peak overshoot and convergence

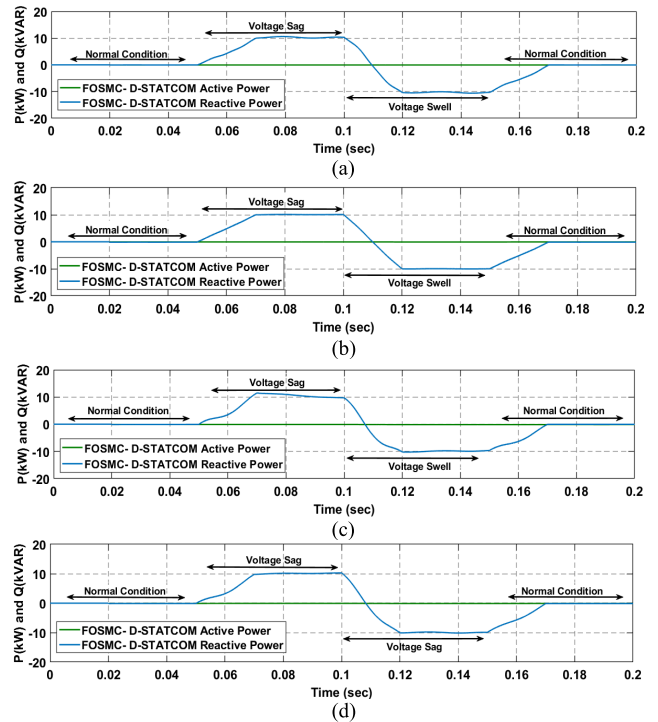


FIGURE 10. (a) Injected/absorbed reactive power by D-STATCOM under voltage sag/swell while employing FOSMC with $\alpha = 0.2$ and $k_d, k_q = 5 \times 10^6$ (b) Injected/absorbed reactive power by D-STATCOM under voltage sag/swell while employing FOSMC with $\alpha = 0.5$ and $k_d, k_q = 5 \times 10^6$ (c) Injected/absorbed reactive power by D-STATCOM under voltage sag/swell while employing FOSMC with $\alpha = 0.8$ and $k_d, k_q = 5 \times 10^6$ (d) Injected/absorbed reactive power by D-STATCOM under voltage sag/swell while employing FOSMC with $\alpha = 0.5$ and $k_d, k_q = 3 \times 10^4$.

time are enhanced because of non-integer order instantaneous error as per the major characteristic of FOSMC. It can be noted that the proposed FOSMC illustrates the exceptional execution over the FFSMC and PI control in the reactive power tracking, due to the fact that the proposed FOSMC is competent of precisely tracking the reference value with the lowest chattering.

Fig. 11(a) shows the dc voltage of D-STATCOM under voltage sag/swell while employing PI control. Fig. 11(b) shows the dc voltage of D-STATCOM under voltage sag/swell while employing FFSMC. Fig. 11(c) shows the dc voltage of D-STATCOM under voltage sag/swell while employing the proposed FOSMC. It can be noted that the execution of the proposed FOSMC is superior to FFSMC and PI control regarding fast convergence and fast-tracking.

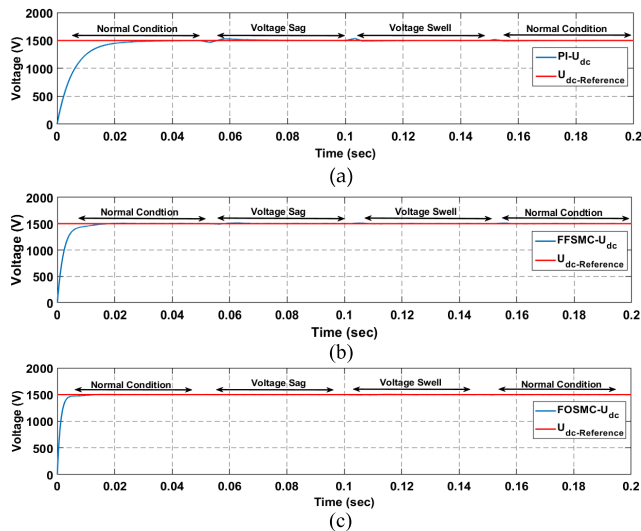


FIGURE 11. (a) dc voltage of D-STATCOM under voltage sag/swell while employing PI Control (b) dc voltage of D-STATCOM under voltage sag/swell while employing FFSMC (c) dc voltage of D-STATCOM under voltage sag/swell while employing FOSMC.

C. PERFORMANCE EVALUATION UNDER UNBALANCED LOAD CONDITIONS

In this test scenario, the proposed D-STATCOM model employing FOSMC is evaluated under unbalanced load conditions in an entire simulation of 0.2s. Initially, the balanced load of $P = 30$ kW and $Q = 20$ kVAR is applied to the system. At 0.07sec, a balanced load is disconnected from the system and an unbalanced load of ($P = 31$ kW and $Q = 18$ kVAR for phase a, $P = 29$ kW and $Q = 17$ kVAR for phase b, $P = 32$ kW and $Q = 22$ kVAR for phase c) is imposed on the system, results in distorted current waveform at LV AC bus, and subsequently, the voltage also got distorted. At 0.15s, the unbalanced load is switch off and a normal load is again imposed on the system. Fig. 12(a) and Fig. 12(b) shows the three-phase and RMS current at the LV AC bus under unbalanced load conditions. Fig. 12(c) and Fig. 12(d) shows the three-phase and RMS distorted voltage at the LV AC bus resulting from an unbalanced current.

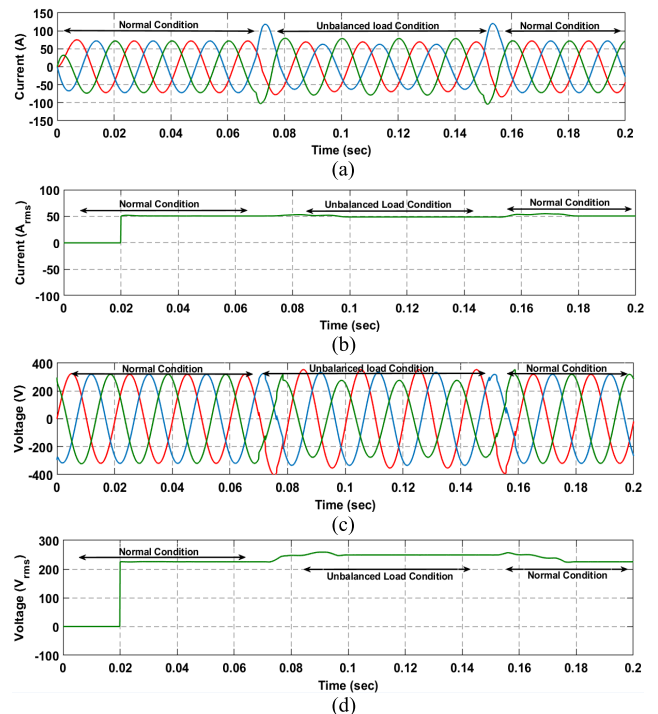


FIGURE 12. (a) Three-phase Current at LV AC bus under unbalanced load condition (b) RMS Current at LV AC bus under unbalanced load condition (c) Three-phase Voltage at LV AC bus under unbalanced load condition (d) RMS Voltage at LV AC bus under unbalanced load condition.

Besides the operation of the proposed D-STATCOM, the voltage at the LV AC bus is made constant under unbalanced load conditions. The D-STATCOM maintains the voltage at the LV AC bus while injecting the reactive power required. Fig. 13(a) and Fig. 13(b) shows the three-phase and RMS voltage at LV AC bus under unbalanced load conditions with D-STATCOM while employing PI control. Fig. 13(c) and Fig. 13(d) shows the three-phase and RMS voltage at LV AC bus under unbalanced load conditions with D-STATCOM while employing FFSMC. Fig. 13(e) and Fig. 13(f) shows the three-phase and RMS voltage at LV AC bus under unbalanced load conditions with D-STATCOM while employing FOSMC. However, the execution of the proposed FOSMC is extremely exceptional regarding rapid reference tracking and fast convergence in contrast to FFSMC and PI control. The assessment of voltage THD and VUF at LV AC bus during unbalanced load conditions is reported in TABLE 6. It can be observed that the voltage THD and VUF of proposed FOSMC during unbalanced load conditions result in 0.97% and 0.0014% in contrast to FFSMC and PI control and are well under the IEEE standards which verifies the competency and exceptional performance of FOSMC.

Fig. 14(a) shows the load active and reactive power under unbalanced load conditions with D-STATCOM at LV AC bus while employing PI control. Similarly, Fig. 14(b) shows the load active and reactive power under unbalanced load conditions with D-STATCOM at LV AC bus while employing FFSMC. Fig. 14(c) shows the load active and reactive power

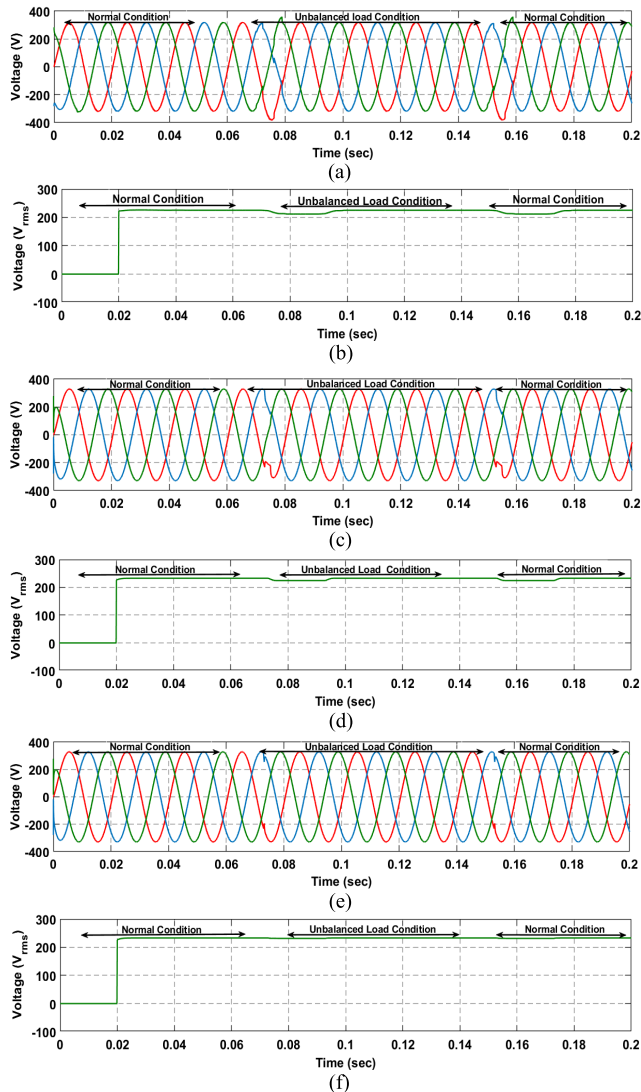


FIGURE 13. (a) Three-phase Voltage at LV AC bus under unbalanced load condition with D-STATCOM while employing PI control (b) RMS Voltage at LV AC bus under unbalanced load condition with D-STATCOM while employing PI control (c) Three-phase Voltage at LV AC bus under unbalanced load condition with D-STATCOM while employing FFSMC (d) RMS Voltage at LV AC bus under unbalanced load condition with D-STATCOM while employing FFSMC (e) Three-phase Voltage at LV AC bus under unbalanced load condition with D-STATCOM while employing FOSMC (f) RMS Voltage at LV AC bus under unbalanced load condition with D-STATCOM while employing FOSMC.

TABLE 6. Assessment of voltage THD at LV AC bus during unbalanced load conditions.

| Controller Strategy | THD (%) | VUF (%) |
|---------------------|---------|---------|
| PI Control | 3.63 | 0.71 |
| FFSMC | 1.96 | 0.02 |
| FOSMC | 0.97 | 0.0014 |

under unbalanced load conditions with D-STATCOM at LV AC bus while employing FOSMC. The execution of the proposed FOSMC can be seen exceptional in terms of fast convergence, upright damping, and instant tracking, in distinction to PI control and FFSMC.

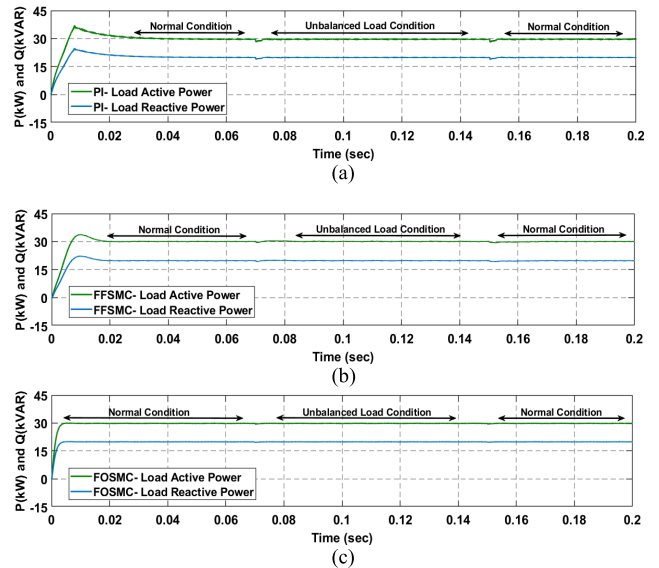


FIGURE 14. (a) Load Active and Reactive power under unbalanced load conditions with D-STATCOM while employing PI control (b) Load Active and Reactive power under unbalanced load conditions with D-STATCOM while employing FFSMC (c) Load Active and Reactive power under unbalanced load conditions with D-STATCOM while employing FOSMC.

Fig. 15(a) shows the reactive power injected by D-STATCOM under unbalanced load conditions while employing PI control. Fig. 15(b) shows the reactive power injected by D-STATCOM under unbalanced load conditions while employing FFSMC. Fig. 16 shows the variation of α , k_d and k_q , and its influence on rise time, convergence time, tracking error, and time to reach the steady-state under unbalanced load conditions. The optimum value of α can be retrieved by varying it from 0 to 0.9. Initially, the value of the fractional coefficient α is chosen from [65]. From Fig. 16(a), it can be observed that the increase of $\alpha = 0.8$ increases the tracking error and chattering. Likewise, the decrease of $\alpha = 0.2$ increases the overshoot, rise time, and time to reach the steady-state, can be seen in Fig. 16(c). In the context of k_d and k_q , the initial values are selected from [69]. The decrease in the values of k_d and k_q results in the increase

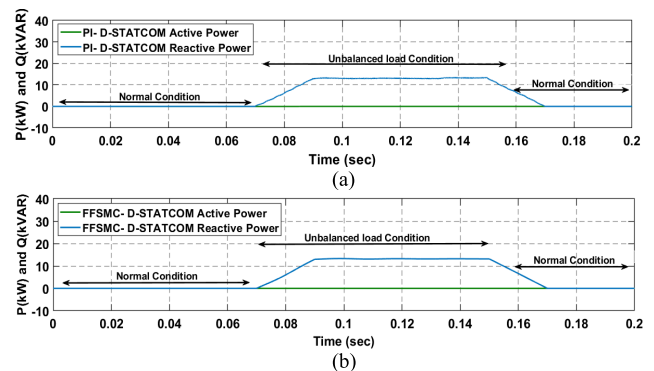


FIGURE 15. (a) Injected Reactive power by D-STATCOM under voltage sag/swell while employing PI control (b) Injected Reactive power by D-STATCOM under voltage sag/swell while employing FFSMC.

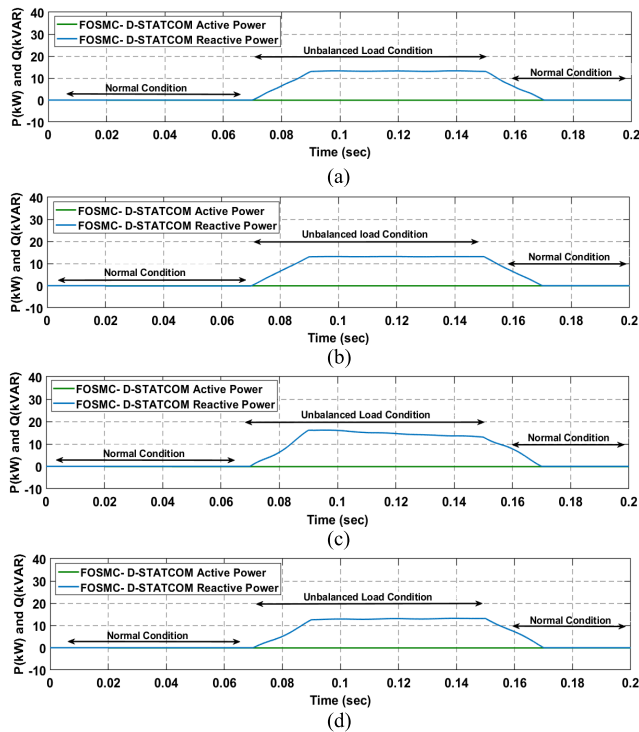


FIGURE 16. (a) Injected reactive power by D-STATCOM under unbalanced load conditions while employing FOSMC with $\alpha = 0.2$ and $k_d, k_q = 5 \times 10^6$ (b) Injected reactive power by D-STATCOM under Unbalanced load conditions while employing FOSMC $\alpha = 0.5$ and $k_d, k_q = 5 \times 10^6$ (c) Injected reactive power by D-STATCOM under unbalanced load conditions while employing FOSMC $\alpha = 0.8$ and $k_d, k_q = 5 \times 10^6$ (d) Injected reactive power by D-STATCOM under unbalanced load conditions while employing FOSMC $\alpha = 0.5$ and $k_d, k_q = 3 \times 10^4$.

of tracking error and the time to reach the steady-state and it can be seen in Fig. 16(d). Meanwhile, the increase of k_d and k_q results in the decrease of tracking error and the time to reach the steady-state. Therefore, the value of FOSMC parameters must be selected in such a way that a compromise is achieved among the rise time, overshoot, convergence time, tracking error, overall damping along with chattering free smooth output. Therefore, the reference reactive power value is admirably tracked by the controlled trajectories when α is selected to be 0.5 and k_d, k_q are selected to be 5×10^6 , evidenced in Fig. 16(b). The peak overshoot and convergence time are enhanced because of non-integer order instantaneous error as per the major characteristic of FOSMC. It can be noted that the proposed FOSMC illustrates the exceptional execution over the FFSMC and PI control in the reactive power tracking, due to the fact that the proposed FOSMC is competent of precisely tracking the reference value with the lowest chattering.

Fig. 17(a) shows the dc voltage of D-STATCOM under unbalanced load conditions while employing PI control. Fig. 17(b) shows the dc voltage of D-STATCOM under unbalanced load conditions while employing FFSMC. Fig. 17(c) shows the dc voltage of D-STATCOM under unbalanced load conditions while employing the proposed FOSMC. It can be

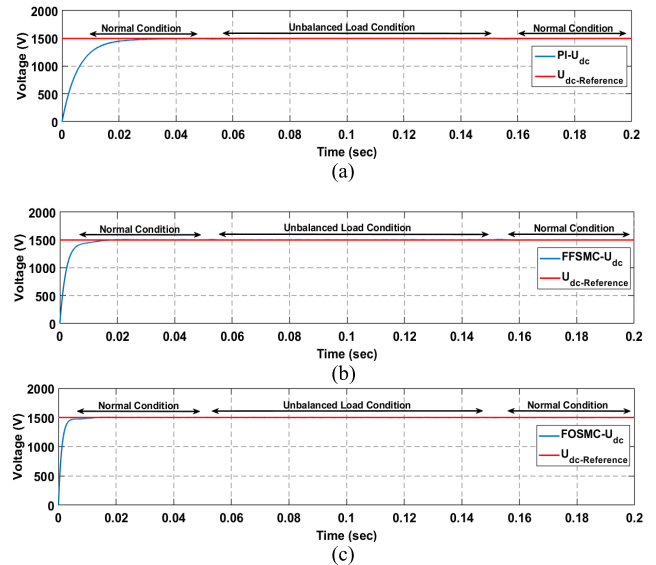


FIGURE 17. (a) dc voltage of D-STATCOM under unbalanced load conditions while employing PI Control (b) dc voltage of D-STATCOM under unbalanced load conditions while employing FFSMC (c) dc voltage of D-STATCOM under unbalanced load conditions while employing FOSMC.

noted that the execution of the proposed FOSMC is exceptional to FFSMC and PI control regarding fast convergence and fast-tracking.

Table 7 shows the assessment of different control strategies present in the study with the proposed FOSMC. Table 8 shows the assessment of the proposed FOSMC with existing SMC strategies while enhancing power quality. The assessment is made in terms of response time, accuracy, robustness,

TABLE 7. Comparison with different control strategies.

| Reference | Control Schemes | Compensation | Controller Complexity |
|-----------|--------------------|---------------------------------------|-----------------------|
| [17] | PI Control | Voltage Sag/Swell | Low |
| [36] | Hysteresis Control | Unbalanced Load | Medium |
| [38] | FLC | Voltage Sag/Swell | Low |
| [71] | MPC | Voltage Unbalanced | Medium |
| Proposed | FOSMC | Voltage Sag/Swell and Unbalanced Load | Lowest |

TABLE 8. X-tics comparison of proposed FOSMC with current SMC strategies.

| X-tics | Proposed FOSMC | [55] | [56] | [80] | [60] |
|-----------------------|----------------|--------|------|------|----------|
| Response Time | Fast | Fast | Fast | Fast | Medium |
| Accuracy | Very High | High | High | High | High |
| Robustness | Very High | Medium | High | High | Moderate |
| Chattering | Lowest | Low | Low | - | Low |
| Transient Response | Very Fast | Fast | Fast | Fast | Fast |
| Voltage THD (%) | 0.52 | - | - | 1.02 | - |
| Steady-State Time (s) | 0.02 | 0.3 | - | - | - |

chattering, THD, and steady-state time. The performance of the proposed control is excellent in all performance parameters.

VI. CONCLUSION

In this paper, the authors have proposed a FOSMC based D-STATCOM to compensate the low power distribution system under disturbances such as voltage sag/swell and unbalanced load conditions. Besides, the performance of the FOSMC under its parametric variations is discussed as well. The complete system is demonstrated with a model of 400V, 180kVA radial distributor along with D-STATCOM under two test scenarios in MATLAB/Simulink environment. In the first test scenario, the grid transients (voltage sag/swell) are considered at the LV AC bus. Likewise, in the second test scenario, the unbalanced load conditions are considered at the LV AC bus. D-STATCOM sustains the voltage at LV AC bus by injecting/absorbing a certain extent of reactive power under voltage sag/swell and unbalanced load conditions.

The results of the proposed controller are compared with fixed frequency sliding mode control (FFSMC) and conventional proportional-integral (PI) control. The results validate the superiority of the proposed controller in terms of rapid tracking, fast convergence, and overall damping with very low THD, and VUF. In the first test scenario, the voltage THD of proposed FOSMC during voltage sag/swell results in 0.52% in contrast to FFSMC and PI control which have THD of 0.84% and 2.17% respectively. In the second test scenario, the voltage THD of proposed FOSMC during unbalanced load conditions results in 0.97% in contrast to FFSMC and PI control which have THD of 1.96% and 3.63%. Likewise, the VUF under unbalanced load conditions with proposed FOSMC is 0.0014% in contrast to FFSMC and PI control which have VUF of 0.02% and 0.71%.

In terms of assessment with existing SMC schemes, the proposed FOSMC has a very high response time, very high accuracy, very high robustness, lowest chattering along with low THD and VUF. The proposed model could be realized on the hardware platform for real-time verification purposes in future applications.

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