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Comprehensive Analysis of Source and Drain Recess Depth Variations on Silicon Nanosheet FETs for Sub 5-nm Node SoC Application

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ABSTRACT Excess source and drain (S/D) recess depth (T_{SD}) variations were analyzed comprehensively as one of the most critical factors to DC/AC performances of sub 5-nm node Si-Nanosheet (NS) FETs for system-on-chip (SoC) applications. Variations of off-, on-state currents (I_{off} , I_{on}) in three-stacked NS channels and parasitic bottom transistor (tr_{pbt}) , gate capacitance (C_{gg}) , intrinsic switching delay time (τ_d) , and static power dissipation (P_{static}) are investigated quantitatively according to the T_{SD} variations. More S/D dopants diffuse into the tr_{pbt} with the deeper T_{SD} , so the I_{off} and I_{on} increase due to raised current flowing through the tr_{pbt} . Especially, the I_{off} of PFETs remarkably increases above the certain T_{SD} ($T_{SD,critical}$) compared to NFETs. Furthermore, the I_{on} contribution of each channels having the $T_{SD,critical}$ is the largest at the top NS channel and the tr_{pbt} has the ignorable I_{on} contribution. Among the NS channels, the top (bottom) NS channel has the largest (smallest) I_{on} contribution due to its larger (smaller) carrier density and velocity for both P-/NFETs. The C_{gg} also increases with the deeper T_{SD} by increasing parasitic capacitance, but fortunately, the τ_d decreases simultaneously due to the larger increasing rate of the I_{on} than that of the C_{gg} for all SoC applications. However, the P_{static} enormously increases with the deeper T_{SD} , and low power application is the most sensitive to the T_{SD} variations among the SoC applications. Comprehensive analysis of the inevitable trpbt effects on DC/AC performances is one of the most critical indicators whether Si-NSFETs could be adopted to the sub 5-nm node CMOS technology.

INDEX TERMS Nanosheet FET, parasitic bottom transistor, source/drain recess depth, sub 5-nm node, sub-sheet leakage, TCAD simulation, U-shaped source/drain.

I. INTRODUCTION

For several decades, conventional bulk planar Si-MOSFETs had been successfully scaled down, and several novel strategies were adopted from 90- to 32-nm node to improve the CMOS performance [1]–[4]. However, aggressive scaledown of planar Si-MOSFETs caused the loss of channel controllability, and short channel effects (SCEs) significantly degraded the DC/AC performances. Since 2012, as a solution, 22-nm node Si-FinFETs were adopted to industry for over-

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coming the SCEs and enabling further scale-down [5]. The Si-FinFETs technologies have been continuously improved down to 7-nm node by decreasing fin pitch and contact poly pitch and increasing aspect ratio of fins [6]–[9]. However, these kinds of fin structures need extremely high fin aspect ratio and are limited by fin pitch [10]. As a result, vertically stacked Si-Gate-All-Around (GAA) nanosheet FETs (NSFETs) were proposed as a promising candidate to replace Si-FinFETs due to those superior electrostatics below 7-nm node [11]–[14]. But both Si-FinFETs and Si-NSFETs have inevitable parasitic channels below the intrinsic channels that critically affect the leakage current, which is the killing factor

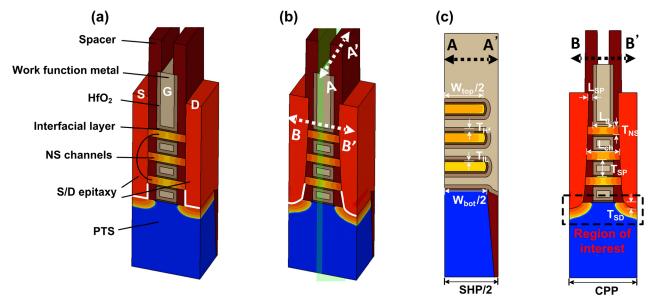


FIGURE 1. (a) Structure of the ideal S/D NSFETs and (b) the U-shaped S/D NSFETs (white lines highlight the difference between the ideal and U-shaped S/D). (c) 2-D cross-section view (A-A' and B-B') of (b).

of scaled transistors [15], [16]. Especially, the Si-NSFETs are deeply concerned because of wider parasitic channels than the Si-FinFETs.

Meanwhile, in ultra-scaled transistor, impacts of process variations on DC/AC performances are inevitable inherent problems. Among these variations, source/drain (S/D) process variations should be carefully controlled because those are directly related to source-to-drain leakage current as well as drive current. So far, several studies about the impacts of the S/D process variations on the Si-FinFETs including S/D epitaxy shape, depth [17]–[19], S/D length [20], [21], and S/D doping concentration [22] were addressed, but the studies on the vertically stacked Si-NSFETs were rarely reported.

In this paper, we quantitatively investigated the impacts of S/D process variations on the DC/AC performances of sub 5-nm node Si-NSFETs. We analyzed off- and on-state currents (I_{off} , I_{on}), gate capacitance (C_{gg}), and switching delay time (τ_d) according to S/D epitaxy shapes and S/D depth variations in the following sections. In Section II, we explained the device design and simulation methodology of the sub 5-nm node NSFETs. In Section III, we investigated the sensitivity of the I_{off} , I_{on} , C_{gg} , τ_d , and static power dissipation (P_{static}) to the S/D process variations. Finally, we gave a conclusion in Section IV.

II. DEVICE DESIGN AND SIMULATION METHODOLOGY

We simulated the sub 5-nm node three-stacked NSFETs with advanced physics models using Sentaurus TCAD simulator [23]. Drift-diffusion model was considered self-consistently with Poisson and carrier continuity equations for carrier transport. Density-gradient model was included to consider quantum confinement effects of the channel region [24], [25].

Slotboom bandgap narrowing model was also included for doping-induced bandgap narrowing in overall device regions [26]. Lombardi mobility model was considered to calculate the mobility degradation induced by transverse field at the interfaces [27]. Inversion and accumulation layer and thin-layer mobility models were included to consider Coulomb, phonon, and surface roughness scatterings [28], [29]. Low-field ballistic mobility and high-field saturation models were also included [30]. Furthermore, we considered recombination using Shockley-Read-Hall, Auger, and Hurkx band-to-band tunneling models [31]–[33]. Finally, a deformation potential model was considered for strain dependency of effective mass, effective density-of-states, carrier mobility, and band structure [34].

Fig. 1 shows the structure of the n-type NSFETs having ideal and U-shaped S/D epitaxies, and its cross-section view. All the NSFETs were formed by fully considering doping, diffusion, and strain effect using Sentaurus process simulator [23]. In reality, S/D recess profile cannot be a perfect vertical shape, so we adopted the U-shaped S/D recess profile, which was taken from [11]. In the U-shaped S/D, inner spacer length (L_{sp}) could be the shortest (longest) at the top (bottom) side, while gate length (L_g) was the same regardless of NS channels positions. Channel length (L_{ch}) was defined as distance from the source to the drain epitaxies in each of the NS channels. In addition, Si_{0.5}Ge_{0.5} and Si_{0.98}C_{0.02} epitaxies were formed as the S/D to induce compressive and tensile stress to the NS channels for p/n-type FETs (P-/NFETs), respectively.

Highly doped $Si_{0.5}Ge_{0.5}$ [12] ($Si_{0.98}C_{0.02}$) epitaxies with boron (phosphorus) of 5×10^{20} (1×10^{20}) cm⁻³ were formed as the S/D of the PFETs (NFETs), and silicon NS channels and substrate were undoped (1×10^{15} cm⁻³). Punchthrough stopper (PTS) was doped with phosphorus (boron) of 2×10^{18} cm⁻³ in the PFETs (NFETs). Detailed geometry parameters for the sub 5-nm node NSFETs were summarized

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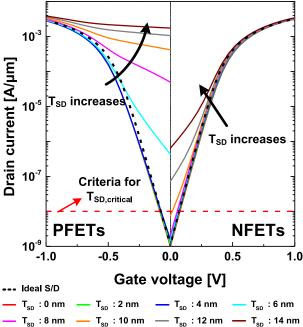


FIGURE 2. Transfer characteristic of the ideal (black dashed line) and the U-shaped S/D NSFETs (color lines) according to the T_{SD} for both P-/NFETs. The I_{off} is fixed to 1 nA/ μ m in the ideal S/D NSFETs and in the U-shaped S/D having the T_{SD} of 0. Drain current is normalized to the SHP.

in Table 1. Dielectric constants were 3.9, 5.0, and 22.0 in IL, spacer, and HfO₂ for overall regions of the NSFETs, respectively. Each contact resistance (R_{cont}) of the source and drain defined for contact width was fixed as an optimistic value of 50 $\Omega \cdot \mu m$ for both P-/NFETs [36], and operating voltage ($|V_{dd}|$) was 0.7 V. All drain currents (I_{ds}) in this work were normalized to sheet pitch (SHP), and threshold voltage (V_{th}) was extracted using constant current method at $W_{ch}/L_g \times 10^{-7} \text{A}/\mu\text{m}$, where $W_{ch} = 3 \times (W_{top} + W_{bot} + W_{top} +$ $2 \times T_{NS}$). Finally, the NSFETs structures of this paper were built by process simulation mimicking the real process flow of [11], while electrical simulations were conducted using fully-calibrated physical model parameters of 10-nm node Si-FinFETs including saturation velocity and low-field ballistic coefficient [8], [35]. Here, I-V data of [11] could not be used for calibrating the physical model parameters, because those were presented as arbitrary units. Moreover, to estimate the I_{on}/I_{off} characteristics of the sub 5-nm node NSFET properly, calibrating the mobility and velocity model from matured and well-known Si-FinFETs technology was reasonable approach. For calibrating our TCAD deck to the hardware data properly, a higher S/D doping concentration (N_{SD}) of the PFETs than the NFETs was adopted for subthreshold swing (SS) and drain-induced barrier lowering (DIBL), which are relatively poorer for the PFETs than for the NFETs.

III. RESULTS AND DISCUSSION

A. OFF-STATE ($|V_{GS}| = 0$, $|V_{ds}| = |V_{dd}|$) ANALYSIS

Fig. 2 shows transfer characteristics of the NSFETs according to excess S/D depth (T_{SD}) for both P-/NFETs and several DC characteristics are summarized in Table 2. Here, the V_{th} of

TABLE 1. Geometry parameters for Sub 5-nm node NSFETs.

Parameters	Values [nm]			
Contact poly pitch	CPP	40		
Sheet pitch	SHP	56		
Gate length	L_{g}	12		
Channel length (top/middle/bottom)	L_{ch}	Ideal: $22 / 22 / 22$ U-shaped: $18.5 / 20 / 22$ (when $T_{SD}=0$)		
NS, Spacing thickness	T_{NS} , T_{SP}	5, 10		
NS width (top, bottom)	W_{top} , W_{bot}	40, 44		
Inner spacer length	L_{sp}	Ideal S/D : 5 U-shaped S/D : 3 - 5		
Interfacial layer / HfO ₂ thickness	T_{IL},T_{Hf}	1, 2		
Excess S/D depth	T_{SD}	0 - 14		

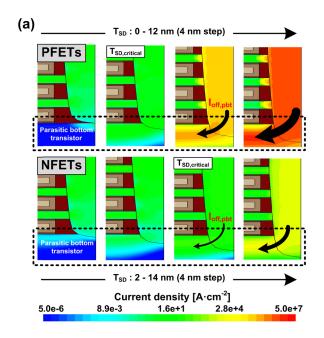
the PFETs at the T_{SD} of 10-14 nm was not shown because they do not approach to the $W_{ch}/L_g \times 10^{-7}$ A/ μ m even in very large positive gate voltage ($|V_{gs}|$). The T_{SD} splits from 0 to 14 nm, and drain voltage (| V_{ds} |) is fixed to the | V_{dd} |. The I_{off} and I_{on} are defined as the I_{ds} when the $|V_{gs}|$ is 0 and the $|V_{dd}|$, respectively. The ideal S/D NSFETs show the smallest SS comparing to the U-shaped S/D NSFETs due to the longest L_{ch} , which mitigated the SCEs. In U-shaped S/D NSFETs, both I_{off} and I_{on} increase with the deeper T_{SD} , and especially, the I_{off} remarkably increases at the certain points of the T_{SD} . To quantify these points, we defined critical excess S/D depth $(T_{SD,critical})$ as the maximum T_{SD} where the I_{off} is smaller than 10 times of I_{off} at the T_{SD} of 0. The definition of the $T_{SD,critical}$ is based on a I_{off} criteria for system-on chip (SoC) applications, wherer the devices having the I_{off} of 10 nA/um are no longer available for standard performance applications [37]. The $T_{SD,critical}$ is 4 nm for the PFETs and 10 nm for the NFETs, and it means that DC performance variations to the T_{SD} are more sensitive in the PFETs than the NFETs.

To analyze which channels mainly contribute to the I_{off} , the I_{off} density of the U-shaped S/D NSFETs according to the T_{SD} is shown in Fig 3a. For both P-/NFETs, the I_{off} density of the NS channels does not change significantly regardless the T_{SD} . Meanwhile, the I_{off} of parasitic bottom channel $(I_{off,pbt})$ in parasitic bottom transistor (tr_{pbt}) increases as the T_{SD} increases. The $I_{off,pbt}$ increases with deepen T_{SD} , because more S/D dopants inevitably diffuse into the PTS region and reduce the V_{th} of the tr_{pbt} . Especially, the $I_{off,pbt}$ density of the PFETs severely increases above the $T_{SD,critical}$ because the S/D dopant concentration of the tr_{pbt} for the PFETs begins to increase remarkably above the $T_{SD,CRITICAL}$. On the other hand, in the NFETs, the S/D dopant concentration linearly increases according to the T_{SD} increase and its amount is smaller than the PFETs' (Fig. 3b). As a result, the PFETs are more sensitive than the NFETs to the T_{SD} variations and show larger $I_{off,pbt}$ density and SCEs.

More details on the PFETS, a critical factor of high I_{off} sensitivity to the T_{SD} is channel stress (S_{zz}) differences between the NS channels and the tr_{pbt} . Fig. 4 shows transfer characteristics of the PFETs having the N_{SD} of 1×10^{20} cm⁻³ according to the T_{SD} (note that the annealing condition has

	PFETs			NFETs				
T _{SD} [nm]	SS [mV/dec]	V _{th} [V]	$I_{off}[nA/\mu m]$	$I_{on}[mA/\mu m]$	SS [mV/dec]	$V_{th}\left[V ight]$	$I_{\rm off} [nA/\mu m]$	I _{on} [mA/μm]
Ideal	79.8	-0.272	1.00	1.13	73.9	0.244	1.00	1.41
0	86.2	-0.294	1.00	0.99	77.8	0.264	1.00	1.32
2	87.0	-0.293	1.10	0.98	78.0	0.260	1.16	1.34
4	90.4	-0.292	1.54	0.98	78.3	0.256	1.34	1.37
6	240.9	-0.161	425	1.05	78.7	0.252	1.60	1.39
8	400.1	0.571	4.92×10^4	1.26	80.7	0.248	2.17	1.41
10	1.37×10^{3}		4.17×10^5	1.63	105.5	0.241	7.77	1.44
12	8.83×10^{3}	>> 0.571	1.06×10^6	2.11	203.0	0.220	73.4	1.48
14	1.67×10^4		1.74×10^6	2.59	275.9	0.127	646	1.53

TABLE 2. SS, I_{off}, I_{on}, and V_{th} of the sub 5-nm node NSFETs according to the T_{SD}.



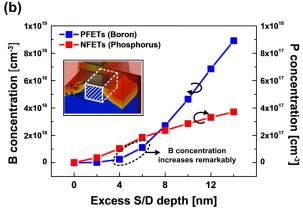


FIGURE 3. (a) The I_{Off} density profile in the U-shaped S/D NSFETs according to the T_{SD} for both P-/NFETs. The tr_{pbt} is specified as black dashed box. (b) Boron and phosphorus concentrations in the tr_{pbt} for the P-/NFETs, respectively. The concentrations is averaged over the volume from the interface between the interfacial layer and the PTS region to depth $T_{SD}+5$ nm (see the white cuboid in the inset).

been also revised accordingly to reproduce the experimental SS [8].). For accurate characteristic projection of the PFETs having the N_{SD} of 1×10^{20} cm⁻³, S/D dopant profile along

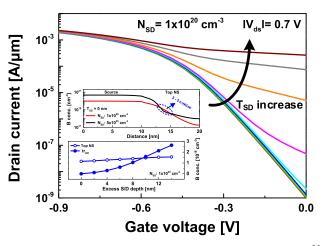


FIGURE 4. Transfer characteristics of the PFETs having the N_{SD} of 1 \times 10²⁰ cm⁻³ according to the T_{SD} (the legends are the same as the Fig. 2.). Inset above shows boron concentration profile along the source epitaxy to the top NS channel. Inset below shows averaged boron concentrations in the top NS channel and the tr_{pbt} over the volume according to the T_{SD} .

the NS channel is adjusted to 3-5 nm/dec (inset above in the Fig. 4) [38]–[42]. Note that the PFETs having lower N_{SD} still suffer from severe upsurge of the I_{off} as the T_{SD} increases ($T_{SD,critical}$ is 6 nm). Typically, the compressive S_{zz} of the PFETs retards boron diffusion into the silicon channels [43], and compressive S_{zz} in the tr_{pbt} is much smaller than NS channels. As a result, more S/D dopants can easily diffuse into the tr_{pbt} than the NS channels (inset below in the Fig. 4), and it significantly degrades the SCEs, regardless the N_{SD} of the PFETs.

B. ON-STATE ($|V_{GS}| = |V_{ds}| = |V_{dd}|$) ANALYSIS

Fig. 5 shows the I_{on} density of the U-shaped S/D NSFETs according to the T_{SD} in each NS channel and tr_{pbt} , where S/D stress effects are fully considered as [44]. Unlike the off-state operation, on-state operation shows different dependency on the T_{SD} with device types. First in the PFETs, as the T_{SD} increases, the I_{on} density in the tr_{pbt} ($I_{on,pbt}$) remarkably increases, whereas the I_{on} densities in the NS channels decrease. On the other hand, in the NFETs, the $I_{on,pbt}$ density slightly increases, but the I_{on} densities in the NS channels

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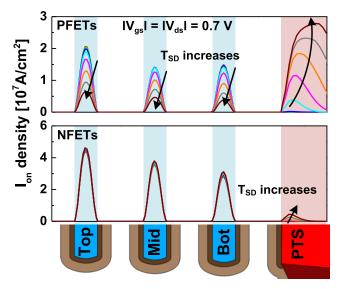


FIGURE 5. The I_{on} densities of each NS channel and the tr_{pbt} in the U-shaped S/D NSFETs according to the T_{SD} . The I_{on} densities were extracted at the center of channel length and width. The legends are the same as the Fig. 2.

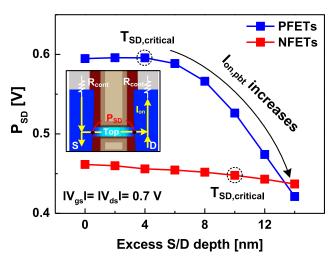


FIGURE 6. The P_{SD} according to the T_{SD} . The P_{SD} is extracted at the center of channel thickness and width of the top NS channel. Inset represents the definition of the P_{SD} and the S/D contact resistance.

rarely vary. In common, increasing S/D dopants diffusion into the PTS region with the deeper T_{SD} induces the $I_{on,pbt}$ density increases for both P-/NFETs. The different dependencies of the I_{on} densities on the T_{SD} in the n-/p-channel NS channels are mainly analyzed with potential differences between the ends of the NS channels (P_{SD}). The P_{SD} is defined as the valence (conduction) band energy difference between source and drain epitaxies of the PFETs (NFETs) and the P_{SD} is extracted at the center of channel thickness and width of the NS channel (inset of Fig. 6). We investigated the P_{SD} of the top NS only because the other two NS channels also have the same P_{SD} tendencies as the top NS channel (Fig. 6). Increasing $I_{on,pbt}$ with the deeper T_{SD} causes the larger potential reduction by the R_{cont} , so the P_{SD} also reduces. Especially,

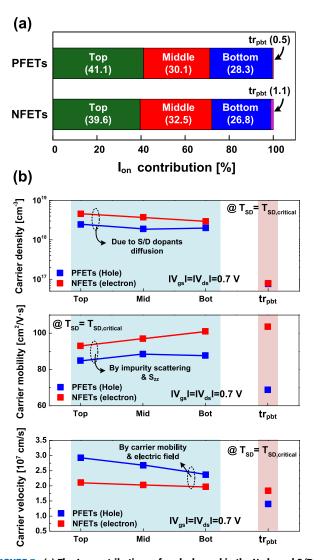


FIGURE 7. (a) The I_{OD} contributions of each channel in the U-shaped S/D NSFETs having the $T_{SD} = T_{SD,critical}$. The I_{OD} contributions of each channel was evaluated by integrating the I_{OD} densities over the channel cross-section area at the center of the channels (see Fig. 1c). (b) Averaged carrier densities, mobilities and velocities of each channel in the U-shaped S/D NSFETs having the $T_{SD} = T_{SD,critical}$ over the channel volumes surrounded by interfacial layers.

the $I_{on,pbt}$ remarkably increases above the $T_{SD,critical}$ in the PFETs but increases a little in the NFETs. As a result, the P_{SD} significantly decreases for the PFETs and lowers the I_{on} densities of the NS channels, but the P_{SD} does not vary much for the NFETs.

To reveal the origin of the I_{on} , we analyzed the I_{on} contributions of each channel using the U-shaped S/D NSFETs with T_{SD} of the $T_{SD,critical}$ (Fig. 7a). For comparison, the I_{on} contributions of the ideal S/D NSFETs was also investigated in [45]. The I_{on} contributions of each channel was evaluated by integrating I_{on} density over the channel cross-section area at the center of channels (see Fig. 1c). The top NS channel has the largest I_{on} contribution for both P-/NFETs and those are 41.1 % and 39.6 %, respectively. Interestingly, the tr_{pbt} has a negligible I_{on} contribution (0.5 and 1.1 %



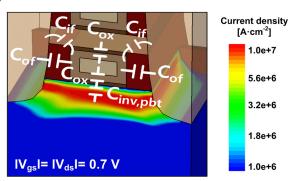
TABLE 3. Relative comparison of electrical parameters of each NS channels and dominant factors.

Parameters	PFETs	NFETs			
	$Top > Mid \approx Bot$	Top > Mid > Bot			
Carrier density	S/D dopant diffusion & S_{zz}				
Carrier mobility	$Top \le Mid \approx Bot$	Top < Mid < Bot			
	impurity scattering & Szz				
Carrier velocity	Top > Mid > Bot	$Top \approx Mid \approx Bot$			
	carrier mobility & electric field				
I _{on} contribution	Top > Mid > Bot				
	carrier density & velocity				

for the P-/NFETs, respectively), contrary to the case of the I_{off} (Fig. 3). In addition, the bottom NS channel shows the smallest I_{on} contribution for both P-/NFETs (28.3, 26.8 %, respectively).

The reason why each of the NS channels has the different I_{on} contributions can be clarified using carrier density, mobility, and velocity. Typically, current density is proportional to the product of the carrier densitive and the velocity (Fig. 7b). First of all, the carrier density is proportional to gate overdrive voltage ($|V_{gs}| - |V_{th}|$). More S/D dopant diffusion into the channels decreases the $|V_{th}|$ of each channels, resulting in higher carrier density at the same $|V_{gs}|$. The S/D dopant diffusion mainly occurs at S/D annealing steps due to its high annealing temperature. In this step, both the L_{ch} and the S_{zz} complexly affect the S/D dopant diffusion. The shorter L_{ch} , the deeper S/D dopants diffuse from the S/D to the channels, whereas compressive (tensile) S_{77} retards boron (phosphorus) diffusion into silicon channels [43], [46], [47]. Furthermore, the larger S/D volume beside top-side NS channel typically induces more compressive (tensile) S_{77} than the bottom-side channels in the PFETs (NFETs), but the S_{zz} differences among the NS channels for the NFETs is neglectible. Therefore, these two factors (the L_{ch} and the S_{zz}) result in the S/D dopant concentrations in channels having order of top> middle \approx bottom (top> midddle> bottom) NS channel in the PFETs (NFETs), and the carrier densities of each channels are also in this order. Secondly, carrier mobilities is critically affected by impurity scattering and the S_{77} . The carriers in the top-side NS channel are mostly suffered from impurity scattering due to the largest amount of S/D dopants in the channel. Then, although the S_{zz} can boost the carrier mobility, the carrier mobilities of each NS channls show almost inversely proprotional to the amounts of impurities. Finally, the carrier velocities of the PFETs are the largest (smallest) in the top (bottom) NS channel because the shorter L_{ch} makes stonger electric field along the channel direction. On the other hand, the carrier velocities of the NFETs are almost the same among the NS channels because the smaller (larger) carrier mobilities of the top (bottom) NS channel

(a)



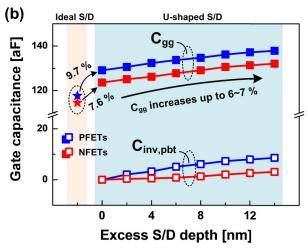


FIGURE 8. (a) Oxide and parasitic capacitance components related to the C_{gg} in the U-shaped S/D NSFETs. (b) The C_{gg} and $C_{inv,pbt}$ of the NSFETs at $|V_{gs}|=|V_{ds}|=0.7$ V according to the T_{SD} .

can be compensated (degraded) in carrier velocities by its larger (smaller) electric field. Therefore, for these reasons, I_{on} contributions of each channel are determined like the Fig. 7a, and we summaized the dominant factors determining the I_{on} contributions of each NS channel in Table 3.

C. AC PERFORMANCE ANALYSIS

In this section, the effects of the tr_{pbt} on AC operation are investigated comprehensively. In the NSFETs, the C_{gg} consists of intrinsic gate oxide capacitance (C_{ox}) and parasitic capacitance (C_{par}) consisting of inversion capacitance of the tr_{pbt} $(C_{inv,pbt})$, inner (C_{if}) and outer fringe capacitances (C_{of}) (Fig. 8a). Fig. 8b shows the C_{gg} and $C_{inv,pbt}$ of the NSFETs according to the T_{SD} . The C_{gg} was extracted at frequency of 1 MHz, and $C_{inv,pbt}$ was extracted as follow:

- 1) Integrating inversion charge density in the PTS region over the PTS volume along the V_{gs} .
- 2) Differentiating the results of 1) over the V_{gs} , then getting the series-connected capacitance, $C_{inv,pbt} + C_{ox}$.
- 3) Subtracting the C_{ox} from the results of 2), then obtaining the $C_{inv,pbt}$ itself. Here, $C_{ox} = L_g \times W_{bot} \times 3.9 \times \varepsilon_0$ /EOT, where ε_0 is the vacuum permittivity and EOT is the effective oxide thickness with T_{IL} of 1 nm and T_{Hf} of 2 nm.

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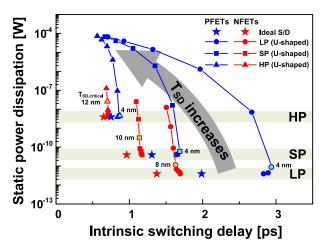


FIGURE 9. the P_{static} versus the τ_d of the sub 5-nm node NSFETs according to the T_{SD} for three SoC applications: LP, SP, and HP. Star symbols are from the ideal S/D NSFETs and the others are from the U-shaped S/D NSFETs with different T_{SD} (the symbols filled in green are from when the T_{SD} is the $T_{SD,critical}$.).

The ideal S/D NSFETs have the smallest C_{gg} due to their longest L_{sp} and negligible $I_{on,pbt}$. The C_{gg} of the U-shaped S/D NSFETs with the T_{SD} of 0 are 9.7 % and 7.8% larger than the ideal S/D NSFETs' for the P-/NFETs, respectively. Meanwhile, the C_{gg} of the U-shaped S/D NSFETs increases up to 7 % as the T_{SD} increases. Because gate oxide thickness and width of the tr_{pbt} are not varied with the deeper T_{SD} , the C_{ox} is not varied much, then increase of the C_{par} dominantly induces the increase of the C_{gg} . Interestingly, the $C_{inv,pbt}$ is negligible when $T_{SD}=0$, however, its contribution to the increase of the C_{par} gradually increases as the T_{SD} deepens.

The $P_{static}(=I_{off}\times |V_{dd}|)$ versus the $\tau_d(=C_{gg}\times |V_{dd}|/I_{on})$ of the NSFETs is also presented for SoC applications: low power (LP), standard performance (SP), and high performance (HP) (Fig. 9). The I_{off} is fixed to 10^{-10} A/ μ m for the LP, 10^{-9} A/ μ m for the SP, and 10^{-7} A/ μ m for the HP. The ideal S/D NSFETs have the smaller τ_d than the U-shaped S/D NSFETs with the T_{SD} of 0 because of their superior I_{on} and smaller C_{gg} (Table 2 and Fig. 8b). In the U-shaped NSFETs, the τ_d decreases with the deeper T_{SD} due to the increasing rate of the I_{on} being larger than the increasing rate of the C_{gg} in both P/NFETs. However, the P_{static} significantly increases as the T_{SD} increases because tremendous amount of current flow in the tr_{pbt} at off-state operation. Fortunately, when the T_{SD} is less than the $T_{SD,critical}$, the variations of the τ_d and P_{static} according to the T_{SD} are small. In addition, the τ_d and P_{static} of the PFETs are more sensitive to the T_{SD} than those of the NFETs because the tr_{pbt} of the PFETs drives more leakage current than the NFETs'. Furthermore, the LP is more sensitive to the T_{SD} than the SP and the HP. Therefore, the T_{SD} should be controlled less than $T_{SD,critical}$ to prevent the serious variations of the τ_d and P_{static} , especially in the PFETs than the NFETs and for the LP than the other applications.

IV. CONCLUSION

The T_{SD} variations of the sub 5-nm node NSFETs were thoroughly investigated in terms of the DC/AC performances. The tr_{pbt} is a critical killing factor of the NSFETs in advanced technology node. The deeper T_{SD} induces more S/D dopant diffusion into the tr_{pbt} and lowers the V_{th} of tr_{pbt} , so the I_{off} and the I_{on} increase due to the tr_{pbt} . Especially, the I_{off} remarkably increases above the $T_{SD,critical}(4, 10 \text{ nm})$ for the P-/NFETs, respectively), and the I_{off} of the PFETs is more severely degraded than that of the NFETs because the compressive S_{zz} of the tr_{pbt} is inevitably smaller than the S_{zz} of the NS channels for the PFETs. At the $T_{SD,critical}$, the top NS channel is the largest contributor to the I_{on} for both P-/NFETs, but fortunately, the tr_{pbt} is the negligible contributor to the I_{on} (0.5, 1.1 % for the P-/NFETs, respectively). The C_{gg} also increases due to the increase of the C_{par} as the T_{SD} deepens, especially, the contribution of the $C_{inv,pbt}$ to the C_{par} also gradually increases. However, the τ_d decreases as the T_{SD} increases due to the increasing rate of the I_{on} being larger than the increasing rate of the C_{gg} in both P/NFETs for all the SoC applications. Finally, the P_{static} enormously varies compared to the τ_d . The τ_d and P_{static} are more sensitive to the T_{SD} variations in the PFETs than in the NFETs and for the LP than the other applications. Therefore, the sub 5-nm node NSFETs ensures immune to the T_{SD} variations if the I_{off} can be controlled only.

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