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# Effects of Some Operation Parameters on the Performance of a Reservoir Computing System Based on a Delay Feedback Semiconductor Laser With Information Injection by Current Modulation

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**ABSTRACT** Via a Santa Fe time series prediction task, the performance of a reservoir computing (RC) system based on a delay feedback semiconductor laser (SL) under current modulation is experimentally investigated and evaluated in detail. Experimental results show that some key operation parameters including the modulation index, the feedback ratio and the SL bias current, seriously affect the prediction performance of this RC system. By optimizing parameter settings, the degradation of the RC performance with the increase of the SL bias current can be efficiently alleviated, and the RC system can still realize a good performance with a NMSE smaller than 0.1 even for the SL biased at a high level of three times threshold current  $I_{th}$ . This research is an improvement to a previously experimental report in which such RC system can achieve a good performance only under a SL bias level near  $I_{th}$ .

**INDEX TERMS** Reservoir computing, semiconductor laser, delay feedback, current modulation, Santa Fe time series, prediction performance.

## I. INTRODUCTION

Nowadays, novel computational paradigms have attracted widespread attention due to the increasing demand for massive information processing [1], [2]. Among these novel methods emerged in the past decades, recurrent neural network (RNN) is one of the most successfully neuro-inspired approaches. RNN is composed of a large number of artificial neurons with recursive connections, and thereby can efficiently perform time-dependent tasks such as speech recognition and time-series prediction [3]. However, properly training RNN is difficult due to the widely known problems of gradient vanishing and exploding [4], which partly limits its large scale deployment in practical application. This drawback can be overcome by reservoir computing (RC), which is originally known as echo state network and liquid

state machine and is introduced at the beginning of 21st century [5], [6]. RC can significantly simplify the training procedure by adopting a fixed connection weights in a reservoir and only training the readout weights and has been successfully applied in some fields [7], [8]. For early-stage RC systems, the reservoir is made up by a network of a larger number of interconnected nonlinear nodes, which results in the hardware implementation of RC very challenging. In 2011, Appeltant *et al.* innovatively implemented a RC based on a simple nonlinear delay system [9], where the complex network is replaced by a single nonlinear element with delay feedback and the large number of nonlinear nodes are replaced by the virtual nodes distributed in a feedback loop. Compared with the earlier RC systems, this RC based on a single nonlinear node has a great advantage in hardware implementation. Since then, nonlinear delay RC systems receive more attention. In 2012, two optoelectronic delay RC experiments were carried out [10], [11], where the

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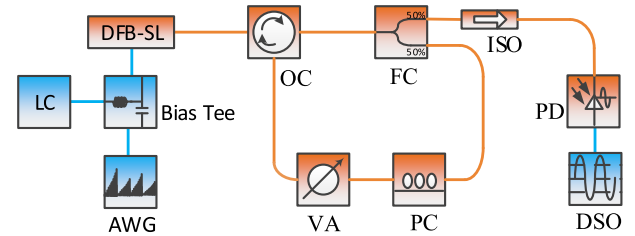
reservoir is constituted by a Mach-Zehnder modulator with optoelectronic feedback. In the same year, a RC system based on a semiconductor optical amplifier with optical feedback was implemented, and all-optical information processing can be realized [12]. Shortly afterwards, another all-optical RC system was proposed and experimentally investigated by Brunner *et al.* [13], in which the reservoir is made up of a semiconductor laser (SL) with optical feedback. This SL-based RC system possesses some unique virtues such as low energy consumption, high information processing rate, true parallelism and relatively easy hardware implementation, and it has been further explored in a few years [14]–[24].

For a SL-based RC system, the information can be injected into the reservoir via optical injection or current modulation. During the past few years, SL-based RC systems under optical information injection have been extensively studied [15]–[23], but few attentions are paid on SL-based RC systems under electrical information injection [13], [14], [24]. Comparatively speaking, a SL-based RC system with electrical information injection is relatively simpler than that with optical information injection in hardware implementation since the components related to optical injection are not required. Via a classification task, this SL-based RC system with electrical information injection was firstly proposed and preliminarily tested [13], and the experimental results showed that a good performance can be obtained only when the SL is biased at a small current range near its threshold but the performance will be degraded for the SL biased at a high current level. After considering that a wide operation range of the SL bias current is necessary towards practical applications for SL-based RC under current modulation, very recently, our group has numerically demonstrated that this RC system can achieve a good performance within a wide current range by optimizing some parameters [24].

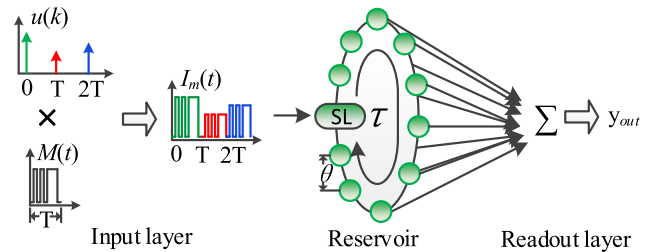
In this work, we experimentally investigate a RC system based on an optical delay feedback SL under current modulation in detail. Via a Santa Fe time series prediction task, we evaluate the influences of some key parameters such as the modulation index, the feedback ratio, and the SL bias current on the RC performances, and the dependence of the RC performance on the system rest state is examined. By adopting optimized operation parameters, the RC system can realize a good prediction performance even for the SL biased at a high current of three times its threshold.

## II. EXPERIMENTAL SETUP

Figure 1 shows our experimental setup. A distributed feedback semiconductor laser (DFB-SL) is taken as the nonlinear node of the reservoir, whose bias current and temperature is controlled by a laser controller (LC, ILX-Lightwave, LDC-3724C) with an accuracy of 0.01 mA and 0.01°C, respectively. The modulation signal is generated by an arbitrary waveform generator (AWG, Tektronix, AWG70001A, 1.5KSa/s-50GSa/s) with a 50 Ω output impedance. A Bias Tee (Picosecond, 5541A, 80 KHz-26 GHz) is used to combine the bias current and the modulation signal. The feedback loop is composed of an optical circulator (OC), a fiber coupler (FC), a polarization controller (PC), and a variable attenuator (VA). The PC keeps the polarization direction of the feedback light parallel with the emitted light, and the VA controls the feedback ratio. In the sampling section, an optical isolator (ISO) is utilized to suppress the unwanted reflection. The optical signal is converted into an electrical signal by a photo-detector (PD, New Forus, 1544-B, 12 GHz bandwidth). The electrical signal is recorded by a digital storage oscilloscope (DSO, Agilent, DSO-X 91604A, 40 GSa/s, 16 GHz bandwidth).



**FIGURE 1.** Experimental setup of a RC system based on an optical feedback SL under current modulation. DFB-SL: distributed feedback semiconductor laser; OC: optical circular; FC: fiber coupler; ISO: isolator; PD: photo-detector; PC: polarization controller; VA: variable attenuator; LC: laser controller; AWG: arbitrary waveform generator; DSO: digital storage oscilloscope.



**FIGURE 2.** Schematic diagram of the operating principle for a RC system based on an optical feedback SL under current modulation.

loop is composed of an optical circulator (OC), a fiber coupler (FC), a polarization controller (PC), and a variable attenuator (VA). The PC keeps the polarization direction of the feedback light parallel with the emitted light, and the VA controls the feedback ratio. In the sampling section, an optical isolator (ISO) is utilized to suppress the unwanted reflection. The optical signal is converted into an electrical signal by a photo-detector (PD, New Forus, 1544-B, 12 GHz bandwidth). The electrical signal is recorded by a digital storage oscilloscope (DSO, Agilent, DSO-X 91604A, 40 GSa/s, 16 GHz bandwidth).

In this experiment, data pre-processing (masking in input layer) and post-processing (training and testing in output layer) are performed offline. Figure 2 is a schematic diagram to clearly illustrate the operating principle of the RC. As shown in the diagram, before original data  $u(k)$  is fed into the reservoir, it is sampled with a period  $T$  (we set  $T = \tau$ , where  $\tau$  is the delay time in feedback loop) and multiplied by a mask  $M(t)$ . The mask defines the coupling weights between input data and virtual nodes, it is a piecewise constant function with a period  $T$  and keeps a constant over an interval  $\theta$ . To maximize the diversity of virtual node states, the mask is usually constituted by random sequences such as binary mask [25], [26], multi-level mask [27], and chaotic mask [28]. The mask used in this experiment is a binary mask, where the mask values are randomly extracted from  $\{0, 1\}$  with equal probability. The masked data  $I_m(t)$ , *i.e.* the signal output by AWG, can be regarded as a continuous time signal which keeps a constant over  $\theta$  [9], and it is fed into the reservoir serially to modulate the SL current. Under current modulation and optical feedback, the SL presents nonlinear

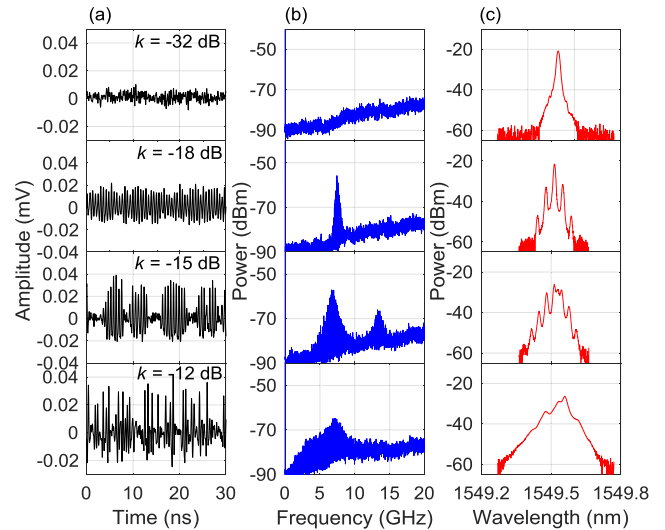
transient responses. The SL responses at different time are sampled and recorded for post-processing. In post-processing procedure, we define  $N$  virtual nodes distributed in a period  $\tau$  with an interval  $\theta$ . Then,  $N$  states of the virtual nodes can be collected within  $\tau$ , and the state matrix  $X^{M \times N}$  of the reservoir is obtained after  $M$  successive  $\tau$ . It should be noted that, in this experiment, the sampling rate of DSO is 40 GHz with a sampling step length of 0.025 ns. As a result, when the interval  $\theta$  is larger than 0.025 ns, several sample points will be recorded in each  $\theta$ . Here we define the virtual node states as the average of the sampling points around the middle of the corresponding time window over duration of  $\theta/2$  [12].

The RC output is obtained by a linear combination of the virtual node states, i.e.  $y_{out} = W_{out}X$ , where  $W_{out}$  is the output weight and need to be optimized through training. During training process, the mean square error between the reservoir output  $y_{out}$  and the expected values  $y_d$  is minimized, and then  $W_{out} = (X^T X)^{-1} X^T y_d$  can be obtained via a standard regression algorithm.

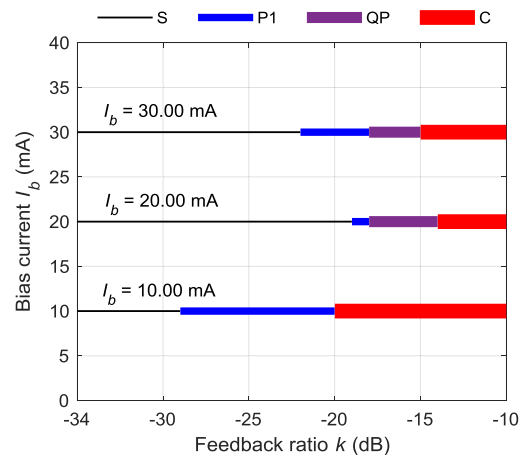
### III. RESULTS AND DISCUSSION

For successful information processing, a reservoir should fulfill some important properties including consistent responses, high-dimensional state space and fading memory [14], [29], which are related to the system dynamics. Previous studies have shown that an SL can exhibit rich nonlinear behaviors under external perturbation [30]–[33], and in this experiment the perturbations refer to feedback light and input signal. First, we focus on the variation of the system rest state (the system state without input data) with the feedback ratio  $k$ , where  $k$  is defined as a ratio of the feedback power and the output power of the DFB-SL at free-running.

During the experiment, the DFB-SL is stabilized at 20°C, and its threshold current  $I_{th}$  is about 8.50 mA at free-running. For a fixed bias current  $I_b$ , the SL can be entered into different dynamical states under optical feedback with different feedback ratio  $k$ . Figure 3 presents the temporal waveforms, power spectra and optical spectra of the SL under  $I_b = 20$  mA and different  $k$ . For  $k = -32$  dB (Row 1), the output power is nearly a constant with small ripples due to the noise, the optical spectrum is a typical single mode shape, and the power spectrum is similar to the noise floor. As a result, the SL can be determined to operate at a stable (S) state. For  $k = -18$  dB (Row 2), the temporal waveform behaves a periodic oscillation, whose fundamental frequency is close to the relaxation oscillation frequency and is about 7.3 GHz obtained from the power spectrum. The optical spectrum presents multiple peaks with equal interval. All these features indicate that the SL operates at a period-one (P1) state. For  $k = -15$  dB (Row 3), some new frequencies emerge around the fundamental frequency (6.5 GHz) and its harmonic frequency (13.0 GHz), which is the feature of a quasi-periodic (QP) state. As for  $k = -12$  dB (Row 4), the time series fluctuates dramatically, the optical spectrum is broadened obviously, and the corresponding power spectrum continuously covers a



**FIGURE 3.** Temporal waveforms (Column a), power spectra (Column b), and optical spectra (Column c) of a SL under  $I_b = 20$  mA and  $k = -32$  dB (Row 1),  $k = -18$  dB (Row 2),  $k = -15$  dB (Row 3), and  $k = -12$  dB (Row 4), respectively.



**FIGURE 4.** Variations of the system rest state with the feedback ratio  $k$  under three different SL bias current  $I_b$  of 10 mA, 20 mA and 30 mA. S: stable state; P1: period-one state; QP: quasi-periodic state; C: chaotic state.

very broad frequency range. Under this circumstance, the SL can be judged as a chaotic (C) state.

We choose three different cases of SL bias current  $I_b$  to present the continuous variation of the system rest state with  $k$ , and the results are shown in Fig. 4. For  $I_b = 10.00$  mA, the SL is sensitive to optical feedback, it operates at a stable state only when  $k$  is smaller than  $-29$  dB. In the range of  $-29$  dB  $\leq k \leq -20$  dB, the SL exhibits a period-one (P1) state. When  $k$  is further increased, the SL exhibits a chaotic state. In the case of  $I_b = 20.00$  mA, the SL tends to be less sensitive to feedback and operates at a stable state until  $k$  is increased to  $-18$  dB. When  $k$  is further increased, the SL sequentially presents P1, quasi-periodic (QP) state, and chaotic state. Similarly, as for the case of  $I_b = 30.00$  mA, the SL operates at a stable state for  $k \leq -22$  dB, and with

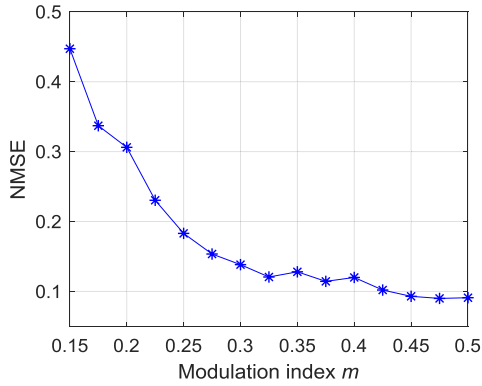


FIGURE 5. NMSEs as a function of  $m$  for  $I_b = 20.00$  mA,  $\theta = 0.2$  ns and  $k = -28$  dB.

further increase of  $k$  the SL sequentially exhibits P1, QP and chaotic state. From this test, one can see that in the three cases of  $I_b$ , the system rest state undergoes a change from stable state to chaotic state with the increase of  $k$ , which will be helpful for assessing the relation between the RC performance and the system rest state in the following researches.

Next, we inject data into the reservoir to evaluate the performance of the RC system. Here, we select a benchmark task of Santa Fe time series prediction. The goal of this task is to predict the trajectory of a chaotic time series one step ahead the input data, where the chaotic time series are experimentally obtained from a far-infrared laser operating at chaotic state [34]. In this work, we use 3000 points for training and the subsequent 1000 points for testing. The prediction performance is quantified by the normalized mean square error (NMSE) between the reservoir output  $y_{out}$  and the expected values  $y_d$ . The NMSE can be expressed as:

$$NMSE = \frac{1}{L} \sum_{n=1}^L (y_{out}(n) - y_d(n))^2 / var(y_d) \quad (1)$$

where  $L$  is the total number of tested data, and  $var$  denotes the variance. In this research, the obtained NMSE is a mean value over four runs.  $NMSE = 0$  indicates a perfect prediction while  $NMSE = 1$  indicates a completely failed prediction. In general, for implementing a Santa Fe time series prediction task in hardware, a RC system with  $NMSE \leq 0.1$  is considered to be acceptable [16]. It is worth mentioning that very recently a RC system with a NMSE about 0.016 has been experimentally realized [35].

Since the data are electrically injected into the reservoir via directly modulating the SL current, the modulation index  $m$  as a key parameter may affect the prediction performance, where  $m$  is defined as  $I_m/I_b$  and  $I_m$  is the modulation current amplitude. In order to investigate the influence of  $m$  on NMSEs, we set the virtual node interval  $\theta$  to 0.2 ns and choose a stable rest state as implemented in [13]. Under the case of  $\theta = 0.2$  ns and a fixed delay time  $\tau$  of 38.6 ns, there are 193 virtual nodes. The variation of NMSE with  $m$  for  $I_b = 20.00$  mA and  $k = -28$  dB is shown in Fig. 5. As shown

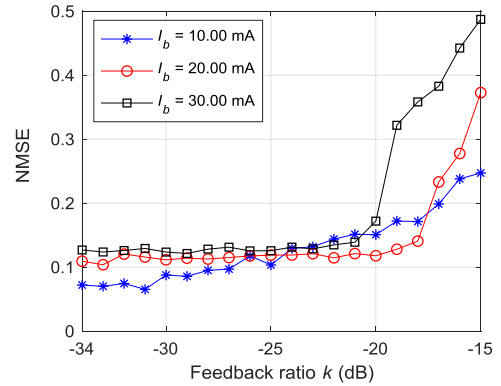
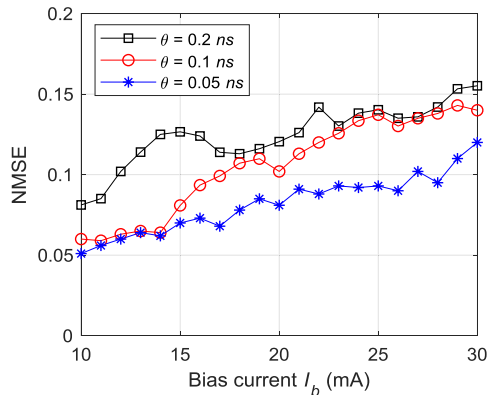


FIGURE 6. NMSEs as a function of  $k$  under three different cases of  $I_b$  for  $\theta = 0.2$  ns and  $m = 0.33$ .

in the diagram, for a small  $m$  of 0.15, the system yields a significantly bad performance of  $NMSE = 0.45$ . The NMSE decreases rapidly when  $m$  increases from 0.15 to 0.3 but decreases slowly when  $m$  is larger than 0.3. Obviously, under this circumstance, a relatively large  $m$  is helpful to improve the prediction performance. Taking into account the influence of  $m$  on NMSE and combining the maximum output amplitude of our used AWG, we set  $m = 0.33$  in the following tests. Under this case, we can evaluate the RC performance within a range of  $I_b \leq 30.00$  mA by adjusting the output amplitude of the AWG and without adding an electric amplifier.

In the above test, the prediction performance is analyzed under a stable rest state with  $I_b = 20.00$  mA and  $k = -28$  dB. Further, we investigate the dependence of the RC prediction performance on the system rest state. For convenience of comparison with Fig. 4, we still choose three different cases of  $I_b$  (10.00 mA, 20.00 mA, and 30.00 mA) and continually change  $k$  from  $-34$  dB to  $-15$  dB, and the results obtained under  $\theta = 0.2$  ns and  $m = 0.33$  are presented in Fig. 6. For  $I_b = 10.00$  mA, NMSEs fluctuate around 0.07 when  $k < -30$  dB. Once  $k$  is larger than  $-30$  dB (the system rest state is P1 as shown in Fig. 3), NMSEs gradually increase. In the cases of  $I_b = 20.00$  mA and 30.00 mA, by comparing Fig. 4 and Fig. 6, NMSEs are maintained at a relatively low level when the system operates at a stable rest state, but NMSEs increase rapidly with  $k$  once the system operates at oscillated rest state. From this test, one can find that the prediction performance of the RC strongly depends on the system rest state. For the three different cases of bias current, relatively low prediction errors can be achieved under the system operating at stable rest state owing to that the system can generate consistently transient response. However, once the system presents oscillated rest state, its transient response will become more complicated under the stimulation of input data, which results in the degradation of prediction performance.

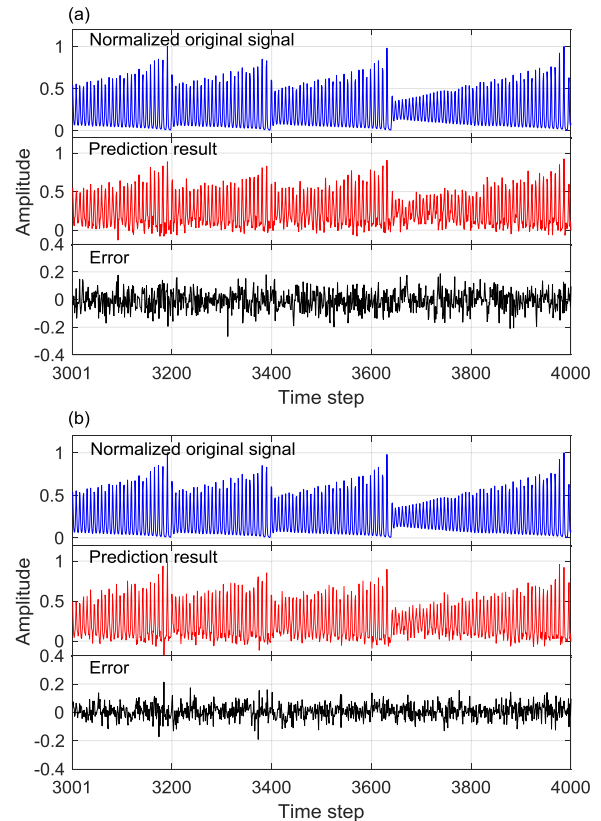
According to the results shown in Fig. 6, for  $k = -31$  dB, NMSE obtained under three different  $I_b$  of 10.00 mA, 20.00 mA, and 30.00 mA is about 0.07, 0.11 and 0.13, respectively. It seems that a lower  $I_b$  will obtain a smaller NMSE. In order to identify the influence of  $I_b$  on NMSE,



**FIGURE 7.** NMSEs as a function of  $I_b$  under three different cases of  $\theta$  for  $m = 0.33$  and  $k = -31$  dB.

we continually change  $I_b$  from 10.00 mA to 30.00 mA with  $m = 0.33$  and  $k = -31$  dB. Meanwhile, considering that the virtual node interval  $\theta$  also affects the RC performance, we choose three different cases of  $\theta$  (0.2 ns, 0.1 ns, and 0.05 ns) to analyze the variation of NMSE with  $I_b$ . As shown in Fig. 7, the variations of NMSEs under three different  $\theta$  present similar trends, *i.e.* NMSEs rise with fluctuations with the increase of  $I_b$  from 10.00 mA to 30.00 mA. For  $\theta = 0.2$  ns, NMSEs increase from 0.08 to 0.16, and a good performance is realized only in the vicinity of the SL threshold current. In the case of  $\theta = 0.1$  ns, NMSEs increase from 0.06 to 0.14, and a good performance can be obtained when  $I_b \leq 17.00$  mA. As for  $\theta = 0.05$  ns, NMSEs increase from 0.05 to 0.12, and  $I_b$  for yielding a good performance is enlarged to 27.00 mA (about three times threshold current). It can be concluded that under this experimental condition, a relatively low bias current is helpful to achieve a good prediction performance, which is due to better consistent response for SL biased at lower current [14]. Additionally, the performance degradation with the increase of  $I_b$  can be efficiently alleviated by adopting a short interval  $\theta$ .

Finally, to intuitively illustrate the influence of the virtual node interval  $\theta$  on the prediction performance, Fig. 8 displays corresponding results under  $I_b = 10.00$  mA, where  $\theta$  is set as 0.2 ns in Fig. 8(a) and 0.05 ns in Fig. 8(b), respectively. Obviously, the prediction errors for  $\theta = 0.05$  ns are smaller than that for  $\theta = 0.2$  ns. Additionally, we also test the RC performance under a shorter  $\theta$  of 0.025 ns, but the prediction fails. The reasons for these experimental results may be explained as follows. A relatively short  $\theta$  may increase the connectivity of the adjacent virtual nodes [9], [18] and results in a large  $N$  ( $= \tau/\theta$ ), which is helpful to increase the state space dimension of the RC system. As a result, the prediction errors for  $\theta = 0.05$  ns are smaller than that for  $\theta = 0.2$  ns. However, as pointed out in Ref. [9], for a too short  $\theta$ , too much averaging takes place and the RC system does not respond to the external input. Therefore, there exists an optimal value of  $\theta$ , which is related to the relaxation oscillation frequency  $f_{ro}$  of the SL. It has been demonstrated that the SL-based



**FIGURE 8.** Experimental results of a Santa Fe time series prediction task under  $I_b = 10.00$  mA, where (a)  $\theta = 0.2$  ns and (b)  $\theta = 0.05$  ns.

RC performs well when  $\theta = 0.2T_{ro}$  ( $T_{ro} = 1/f_{ro}$ ) [13], [24]. Since  $f_{ro}$  is about 3.3 GHz ( $T_{ro} = 0.303$  ns) for  $I_b = 10$  mA, the optimized value of  $\theta$  is about 0.06 ns. Under this case, the obtained result under  $\theta = 0.05$  ns (shown in Fig. 8(b)) is close to the optimized result.

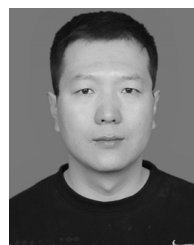
#### IV. CONCLUSION

In summary, we experimentally investigate the influences of some key operation parameters on the performance of a RC system based on an optical feedback SL with information injection by current modulation. Via a Santa Fe time series prediction task, the influences of the modulation index, the feedback ratio and the SL bias current on the RC performance are assessed in detail. The relation between the system rest state and the RC performance is specified, and the optimal parameter settings are confirmed for realizing a good prediction performance. Experimental results show that under a modulation index of  $m = 0.33$ , a relatively low SL bias current is helpful to achieve a good RC performance, and a minimum NMSE of 0.05 is obtained in the vicinity of SL threshold current with a data processing rate of 25.9MSa/s. Moreover, though the prediction performance decreases with the increase of the SL bias current, the performance degradation can be efficiently alleviated by adopting a short virtual node interval of  $\theta = 0.05$  ns. Under this case, the RC system can yield a good performance even under a high SL bias

current of three times its threshold. We believe that this research would be helpful to deeply understand the SL-based RC system under electrical information injection and then further exploit its potential applications.

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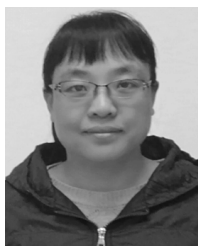
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