

Received May 6, 2019, accepted June 2, 2019, date of publication June 10, 2019, date of current version June 26, 2019. Digital Object Identifier 10.1109/ACCESS.2019.2921801

# Stochastic Finite-Time Output Feedback **Stabilization of Feedforward Nonlinear Systems**

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This work was supported in part by the National Natural Science Foundation of China under Grant 61703194 and Grant 61773193, and in part by the Natural Science Foundation of Shandong Province under Grant ZR2019YQ28.

**ABSTRACT** The finite-time stabilization for stochastic high-order nonlinear systems is considered in this paper using output feedback. The power orders in the system nonlinearities are relaxed to be rational number greater than one. A new constructive output-feedback, finite-time controller is designed based on a finitetime observer together with the homogeneous domination. The stochastic finite-time stability analysis is given rigorously for the closed-loop stochastic system.

**INDEX TERMS** Stochastic, feedforward systems, finite-time, output feedback control.

## I. INTRODUCTION

Stabilization of feedforward systems is of practical importance [1], [2]. There are many deterministic results on feedforward systems, see [3]-[12] and the references therein. In recent years, some stabilization results have been achieved for stochastic nonlinear feedforward systems

$$d\eta_{i} = (\eta_{i+1}^{p_{i}} + f_{i}(\eta_{i+2}, \cdots, \eta_{n}, v))dt + g_{i}^{\top}(\eta_{i+2}, \cdots, \eta_{n}, v)d\omega, \quad i = 1, \cdots, n-2, d\eta_{n-1} = (\eta_{n}^{p_{n-1}} + f_{n-1}(v))dt + g_{n-1}^{\top}(v)d\omega, \quad \eta(0) = \eta_{0}, d\eta_{n} = v^{p_{n}}dt, v = \eta_{1},$$
(1)

where  $n \ge 2$ , and  $\eta = (\eta_1, \dots, \eta_n)^\top \in \mathbb{R}^n$ ,  $v \in \mathbb{R}$ ,  $y = \eta_1$  and  $\eta_0 \in R$  are the system state, input, output and initial value.  $\omega$  is an *m*-dimensional standard Wiener process defined on the complete probability space  $(\Omega, \mathcal{F}, P)$ . For  $i = 1, \dots, n$ ,  $p_i \in R_{odd}^{\geq 1} \triangleq \{ \frac{p}{q} \in R^+ : p \text{ and } q \text{ are odd integers, } p \geq q \},$ system (1) is called as high-order system if one  $p_i > 1, f_i$ :  $R^{n-i-1} \times R \to R$  and  $g_i : R^{n-i-1} \times R \to R^m$  are continuously differential with  $f_i(0, 0) = 0$ ,  $g_i(0, 0) = 0$ .

For instance, [13] studied the stochastic stabilization of nonlinear systems in feedforward form with noisy outputs. [14] constructed a state feedback stabilized controller for stochastic high-order nonlinear feedforward systems. Reference [15] considered the global stabilization of feedforward systems with time-delay. However, all of these results only consider the asymptotic stabilization.

The associate editor coordinating the review of this manuscript and approving it for publication was Feiqi Deng.

Recently, more attention have been paid on the finitetime stabilization of deterministic feedforward systems [9], stochastic lower-triangular systems [17], [18], switched stochastic nonlinear systems [19], [20]. For stochastic nonlinear feedforward systems, the finite-time stabilization is considered in [21] but where only a finite-time state feedback stabilized controller is designed.

In this paper, we will design a finite-time output feedback controller for stochastic high-order nonlinear feedforward systems (1). Based on stochastic Lyapunov theorem on finite-time stability in probability established in [22], by using the homogeneous domination method, the adding one power integrator and sign function method, constructing a  $C^2$  Lyapunov function and verifying the existence and uniqueness of solution, a continuous output feedback controller is designed to guarantee the closed-loop system finite-time stable in probability. The effectiveness of control method is showed by a simulation example.

#### **II. PRELIMINARIES**

The following notations are to be used in this paper.

*Notations:*  $\mathbb{R}^n$  denotes the *n*-dimensional Euclidean space,  $R^+$  stands for the set of all nonnegative real numbers. For a vector or matrix X,  $X^{\top}$  denotes its transpose, Tr{X} denotes its trace when X is square. ||X|| denotes Euclidean norm  $(\sum_{i=1}^{n} x_i^2)^{\frac{1}{2}}$  and Frobenius norm  $(\operatorname{Tr}\{X^{\top}X\})^{\frac{1}{2}}$  for vector X and matrix X, respectively. A function  $f : \mathbb{R}^n \to \mathbb{R}$  is  $\mathcal{C}^i$  if it is *i*-times differential.  $\mathcal{K}$  denotes the set of all functions:  $R^+ \rightarrow R^+$  $R^+$  that are continuous, strictly increasing and vanishing at zero,  $\mathcal{K}_{\infty}$  denotes the set of all functions that are of class  $\mathcal{K}$  and unbounded. sgn(x) is the sign function defined as: sgn(x) = 1 if x > 0, sgn(x) = 0 if x = 0, and sgn(x) = -1 if x < 0.

For system

$$dx = f(x)dt + g^{\top}(x)d\omega, \quad x(0) = x_0 \in \mathbb{R}^n, \qquad (2)$$

where f(x), g(x) are continuous in x, f(0) = 0, g(0) = 0,  $x_0$  is the initial value, we have the following three lemmas.

Lemma 1 [23]: Suppose continuous functions f(x) and g(x) satisfy  $||f(x)||^2 + ||g(x)||^2 \le K(1+x^2)$  for a constant K > 0, then for any  $x_0 \in \mathbb{R}^n$ , system (2) has a solution.

*Lemma 2 [17]:* If one nonnegative, radially unbounded and  $C^2$  function V(x) satisfies  $\mathcal{L}V(x)|_{(2)} \leq 0$  for all  $x \in \mathbb{R}^n$ , then system (2) has a solution for any  $x_0 \in \mathbb{R}^n$ .

*Lemma 3 [22]:* Suppose system (2) admits a unique solution, if there is a  $C^2$  function  $V : \mathbb{R}^n \to \mathbb{R}^+$ ,  $\mathcal{K}_{\infty}$  functions  $\gamma_1$  and  $\gamma_2$ , real numbers  $\lambda > 0$  and  $0 < \gamma < 1$  such that for all t > 0 and  $x \in \mathbb{R}^n$ ,  $\gamma_1(||x||) \leq V(x) \leq \gamma_2(||x||)$ ,  $\mathcal{L}V(x) \leq -\lambda V^{\gamma}(x)$ , then the trivial solution of system (2) is finite-time stable in probability.

*Lemma 4 [25]:* For  $x, y \in R$ , if  $p \ge 1 \in R$ , then

$$|x+y|^{p} \le 2^{p-1}|x^{p}+y^{p}|, \quad (|x|+|y|)^{\frac{1}{p}} \le |x|^{\frac{1}{p}}+|y|^{\frac{1}{p}}.$$
(3)

If 
$$p \in R_{odd}^{\geq 1}$$
, then  
 $|x^p - y^p| \le p(1 + 2^{p-3})|x - y|((x - y)^{p-1} + y^{p-1}),$  (4)

$$|x - y|^{p} \le 2^{p-1} |x^{p} - y^{p}|.$$
(5)

*Lemma 5:* Let real numbers  $p \ge 1$  and  $q \ge 1$  satisfy  $\frac{1}{p} + \frac{1}{q} = 1$ , then for any  $x, y \in R$  and any given positive number  $\gamma > 0, xy \le \gamma |x|^p + \frac{1}{q} (p\gamma)^{-\frac{q}{p}} |y|^q$ .

*Lemma 6 [26]:* For any real numbers  $p = \frac{b_1}{b_0} \in R_{odd}^{\geq 1}$  with  $b_1 \geq b_0 \geq 1$  and  $a \geq 1$ , there hold

$$|x^{p} - y^{p}| \le 2^{1 - \frac{1}{b_{0}}} |\operatorname{sgn}(x)|x|^{b_{1}} - \operatorname{sgn}(y)|y|^{b_{1}}|^{\frac{1}{b_{0}}}, \quad (6)$$

$$\operatorname{sgn}(x)|x|^{\frac{1}{a}} - \operatorname{sgn}(y)|y|^{\frac{1}{a}} \le 2^{1-\frac{1}{a}}|x-y|^{\frac{1}{a}}.$$
 (7)

*Lemma* 7 [27]: If  $r \ge 1$  is any given rational number, then there are two constants  $c_0(r) > 0$  and  $1 \ge d(r) > 0$  both dependent on r such that for any  $x, y \in R$ ,

$$|\operatorname{sgn}(x)|x|^{a} - \operatorname{sgn}(y)|y|^{a}| \le c_{0}|x-y|^{d}(|x-y|^{a-d}+|y|^{a-d}).$$
(8)

## III. DESIGN OF FINITE-TIME OUTPUT FEEDBACK CONTROLLER

#### A. PROBLEM FORMULATION

To obtain the stochastic finite-time output feedback stability of system (1), in this paper, the nonlinear functions are assumed to satisfy the following conditions:

Assumption 1: There exist real numbers  $M_1 > 0$  and  $M_2 > 0$ , and a rational number  $\tau \in (-\frac{1}{2(1+\sum_{l=2}^{n} p_1 \cdots p_{l-1})}, 0)$ 

such that for  $i = 1, \cdots, n - 2$ ,

$$\begin{aligned} |f_{i}(\eta_{i+2},\cdots,\eta_{n},\nu)| &\leq M_{1} \bigg( \sum_{j=i+2}^{n} |\eta_{j}|^{\frac{r_{i}+\tau}{r_{j}}} + |\nu|^{\frac{r_{i}+\tau}{r_{n+1}}} \bigg), \\ ||g_{i}(\eta_{i+2},\cdots,\eta_{n},\nu)|| &\leq M_{2} \bigg( \sum_{j=i+2}^{n} |\eta_{j}|^{\frac{2r_{i}+\tau}{2r_{j}}} + |\nu|^{\frac{2r_{i}+\tau}{2r_{n+1}}} \bigg), \\ |f_{n-1}(\nu)| &\leq M_{1} |\nu|^{\frac{r_{n-1}+\tau}{r_{n+1}}}, \quad ||g_{n-1}(\nu)|| \leq M_{2} |\nu|^{\frac{2r_{n-1}+\tau}{2r_{n+1}}}, \quad (9) \end{aligned}$$

where

$$r_1 = \frac{1}{2}, \quad r_{i+1} = \frac{r_i + \tau}{p_i}, \ i = 1, \cdots, n.$$
 (10)

*Remark 1:* Assumption 1 holds for system (1) with  $n \ge 2$ . It's worth noting that, when n = 2, system (1) reduces to

$$d\eta_{1} = (\eta_{2}^{p_{1}} + f_{1}(v))dt + g_{1}^{\top}(v)d\omega, d\eta_{2} = v^{p_{2}}dt, y = \eta_{1}.$$
 (11)

For this case, system (11) just needs to satisfy the condition  $M_1|v|^{\frac{r_{n-1}+\tau}{r_{n+1}}}$ ,  $||g_{n-1}(v)|| \leq M_2|v|^{\frac{2r_{n-1}+\tau}{2r_{n+1}}}$  with n = 2 in Assumption 1.

*Remark 2:* For the finite-time stabilization of stochastic nonlinear systems by output feedback, several latest results such as [29]–[31] have been achieved. However, all of them consider the systems in feedback form. To our knowledge, there is no stochastic output feedback finite-time stabilization result on feedforward systems, so Assumption 1 is the weakest until now. To ensure that Assumption 1 holds, one must find a constant  $\tau \in (-\frac{1}{2(1+\sum_{l=2}^{n}p_{1}\cdots p_{l-1})}, 0)$  such that  $f_{i}$  and  $g_{i}$  satisfy (9). For system  $d\eta_{1} = \eta_{2}^{\frac{5}{3}}dt + \eta_{3}^{a}d\omega$ ,  $d\eta_{2} = \eta_{3}^{3}dt$ ,  $d\eta_{3} = vdt$ , where *a* takes any real constant inside  $(5, \frac{43}{6})$ , by choosing  $\tau = \frac{3a-15}{15-16a}$ , Assumption 1 holds due to  $g_{1} \leq |\eta_{3}|^{a} = |\eta_{3}|^{\frac{2r_{1}+\tau}{2r_{3}}}$ .

There are a variety of nonlinearities satisfying Assumption 1. For example, if there is  $\tau$  such that  $\tau + r_i = a_{i+2}r_{i+2} + \cdots + a_nr_n$ , then  $f_i = \eta_{i+2}^{a_{i+2}} \cdots \eta_n^{a_n}$  satisfies Assumption 1. It's worthy to note that some nonlinearities such as  $\sin \eta$ ,  $\ln(1 + |\eta|)$  can be bounded by  $|\eta|$ .

## B. OUTPUT FEEDBACK CONTROLLER OF SYSTEM (1)

The design of output feedback controller of system (1) consists of three parts. In Part I, the output feedback controller of the nominal system for system (1) based on a observer is designed. Part II determines the observer gains  $\ell_1, \dots, \ell_{n-1}$ . In Part III, a dynamic gain  $\Gamma$  is introduced into system (1) by the transformation, the output feedback controller of system (1) with the dynamic gain is obtained.

Part I. Output feedback controller of nominal system

In this subsection, we design a homogeneous output feedback controller for the nominal system:

$$dx_{i} = x_{i+1}^{p_{i}} dt, \quad i = 1, \cdots, n-1, dx_{n} = u^{p_{n}} dt, y = x_{1}.$$
 (12)

Denote  $\operatorname{sgn}(x)|x|^a \triangleq [x]^a$  for any  $a \in R^+$ ,  $x \in R$ . Step 1: Choose  $V_1(x_1) = m_1 \int_0^{x_1} [s]^{\frac{4-r_2p_1}{r_1}} ds$  with  $m_1 > 0$ being a constant, and define  $z_1 = [x_1]^{\frac{1}{r_1}}$ . Obviously,  $V_1(x_1)$  is  $C^2$  since  $4 - r_2p_1 \ge 1$ . The infinitesimal generator<sup>1</sup> of  $V_1$  along (12) satisfies  $\mathcal{L}V_1(x_1)|_{(12)} = m_1[z_1]^{4-r_2p_1}(x_2^{p_1} - \alpha_1^{p_1}) + m_1[z_1]^{4-r_2p_1}\alpha_1^{p_1}$ . Choosing  $\alpha_1 = -\beta_1^{r_2}[z_1]^{r_2}$  with  $\beta_1 = (\frac{a_{1,1}}{m_1})^{\frac{1}{r_2p_1}}$  and  $a_{1,1}$  being a positive constant to be designed, one has  $\mathcal{L}V_1(x_1)|_{(12)} \le -a_{1,1}z_1^4 + m_1[z_1]^{4-r_2p_1}(x_2^{p_1} - \alpha_1^{p_1})$ . Step k ( $k = 2, \dots, n$ ): We demonstrate this step by

Proposition 1 whose proof is placed in the Appendix. *Proposition 1:* Suppose that there is a  $C^2$ , positive definite and radially unbounded function  $V_{k-1}(\bar{x}_{k-1})$ , and virtual controllers  $\alpha_0, \dots, \alpha_{k-1}$  defined by  $\alpha_0 = 0, \alpha_{j-1} = -\beta_{j-1}^{r_j}[z_{j-1}]^{r_j}, z_{j-1} = [x_{j-1}]^{\frac{1}{r_{j-1}}} - [\alpha_{j-2}]^{\frac{1}{r_{j-1}}}, j = 1, \dots, k$ , such that

$$\begin{aligned} \mathcal{L}V_{k-1}(\bar{x}_{k-1})|_{(12)} \\ &\leq -\sum_{j=1}^{k-3} \left( a_{j,j} - \tilde{a}_{j+1,j} - \bar{a}_{j+1,j} - \sum_{l=j+2}^{k-1} \bar{a}_{l,j} \right) z_j^4 \\ &- (a_{k-2,k-2} - \tilde{a}_{k-1,k-2} - \bar{a}_{k-1,k-2}) z_{k-2}^4 \\ &- a_{k-1,k-1} z_{k-1}^4 + m_{k-1} [z_{k-1}]^{4-r_k p_{k-1}} (x_k^{p_{k-1}} - \alpha_{k-1}^{p_{k-1}}), \end{aligned}$$

$$(13)$$

then by defining  $z_k = [x_k]^{\frac{1}{r_k}} - [\alpha_{k-1}]^{\frac{1}{r_k}}$ , there is a virtual controller  $\alpha_k = -\beta_k^{r_{k+1}}[z_k]^{r_{k+1}}$  such that the *k*th Lyapunov function

$$V_k(\bar{x}_k) = V_{k-1}(\bar{x}_{k-1}) + m_k W_k(\bar{x}_k),$$
  

$$W_k(\bar{x}_k) = \int_{\alpha_{k-1}}^{x_k} \left[ [s]^{\frac{1}{r_k}} - [\alpha_{k-1}]^{\frac{1}{r_k}} \right]^{4 - r_{k+1}p_k} ds \quad (14)$$

satisfies

$$\mathcal{L}V_{k}(\bar{x}_{k})|_{(12)} \leq -\sum_{j=1}^{k-2} \left( a_{j,j} - \tilde{a}_{j+1,j} - \bar{a}_{j+1,j} - \sum_{l=j+2}^{k} \bar{a}_{l,j} \right) z_{j}^{4} - (a_{k-1,k-1} - \tilde{a}_{k,k-1} - \bar{a}_{k,k-1}) z_{k-1}^{4} - a_{k,k} z_{k}^{4} + m_{k} [z_{k}]^{4 - r_{k+1}p_{k}} (x_{k+1}^{p_{k}} - \alpha_{k}^{p_{k}}),$$
(15)

where  $m_{k-1}$ ,  $m_k$ ,  $\tilde{a}_{j+1,j}$ ,  $\bar{a}_{j+1,j}$ ,  $\bar{a}_{l,j}$ ,  $\tilde{a}_{k,k-1}$ ,  $\bar{a}_{k,k-1}$ ,  $j = 1, \dots, k - 2, l = j + 2, \dots, k$ , are positive constants,  $a_{1,1}, \dots, a_{k-1,k-1}, a_{k,k}$  are positive constants to be designed,  $\beta_1, \dots, \beta_k$  are positive constants dependent on  $a_{1,1}, \dots, a_{k-1,k-1}, a_{k,k}$ .

At step *n*, choose  $V_n(x) = \sum_{k=1}^n m_k W_k(\bar{x}_k)$  with  $m_1, \dots, m_n$  being positive constants and design

$$u^{*} = -\beta_{n}^{r_{n+1}}[z_{n}]^{r_{n+1}},$$
  

$$z_{j} = [x_{j}]^{\frac{1}{r_{j}}} - [\alpha_{j-1}]^{\frac{1}{r_{j}}},$$
  

$$\alpha_{0} = 0, \quad \alpha_{j-1} = -\beta_{j-1}^{r_{j}}[z_{j-1}]^{r_{j}}, \quad j = 1, \cdots, n, \quad (16)$$

<sup>1</sup>For  $C^2$  function V(x), the infinitesimal generator of V(x) along system (2) is defined as  $\mathcal{L}V(x) = (\frac{\partial V(x)}{\partial x})^{\top} f(x) + \frac{1}{2} \operatorname{Tr}\{g(x) \frac{\partial^2 V(x)}{\partial x^2} g^{\top}(x)\}$ , where  $\frac{1}{2} \operatorname{Tr}\{g(x) \frac{\partial^2 V(x)}{\partial x^2} g^{\top}(x)\}$  is called the Hessian term of  $\mathcal{L}$ . then there are positive constants  $\beta_1, \dots, \beta_n$  such that

$$\mathcal{L}V_{n}(x)|_{(12)} \leq -\sum_{j=1}^{n-2} \left( a_{j,j} - \tilde{a}_{j+1,j} - \bar{a}_{j+1,j} - \sum_{l=j+2}^{n} \bar{a}_{l,j} \right) z_{j}^{4} - (a_{n-1,n-1} - \tilde{a}_{n,n-1} - \bar{a}_{n,n-1}) z_{n-1}^{4} - a_{n,n} z_{n}^{4} + m_{n} [z_{n}]^{4 - r_{n+1}p_{n}} (u^{p_{n}} - u^{*p_{n}}), \quad (17)$$

where  $f_n = 0$ ,  $g_n = 0$ ,  $\tilde{a}_{j+1,j}$ ,  $\bar{a}_{j+1,j}$ ,  $\bar{a}_{l,j}$ ,  $\tilde{a}_{n,n-1}$ ,  $\bar{a}_{n,n-1}$ ,  $j = 1, \dots, n-2, l = j+2, \dots, n$ , are positive constants,  $a_{1,1}, \dots, a_{n,n}$  are positive constants to be designed.

Introduce a homogeneous reduced-order observer:

$$d\hat{\xi}_k = -\ell_{k-1}\hat{x}_k^{p_{k-1}}dt, \quad \hat{x}_k = [\hat{\xi}_k + \ell_{k-1}\hat{x}_{k-1}]^{\frac{r_k}{r_{k-1}}}, \quad (18)$$

where  $k = 2, \dots, n$ . Using the certainty equivalence principle together with (16), one obtains the implementable output feedback controller for system (12):

$$u = \hat{\alpha}_n = -\beta_n^{r_{n+1}} \left[ [\hat{x}_n]^{\frac{1}{r_n}} + \sum_{i=1}^{n-1} \left( \prod_{j=i}^{n-1} \beta_j \right) [\hat{x}_i]^{\frac{1}{r_i}} \right]^{r_{n+1}}.$$
 (19)

Part II. Determination of observer gains  $\ell_1, \dots, \ell_{n-1}$ 

For  $k = 2, \dots, n$ , define  $e_k = [x_k^{p_{k-1}} - \hat{x}_k^{p_{k-1}}]^{\frac{1}{r_k p_{k-1}}}$  and choose

$$V = V_n + \sum_{k=2}^n \int_{[\hat{\xi}_k + \ell_{k-1} x_{k-1}]}^{[x_k] \frac{4 - r_k p_{k-1}}{r_k}} \frac{4 - r_k p_{k-1}}{r_{k-1}} \left( [s]^{\frac{r_{k-1}}{4 - r_k p_{k-1}}} - (\hat{\xi}_k + \ell_{k-1} x_{k-1}) \right) ds \triangleq \sum_{k=2}^n U_k + V_n.$$
(20)

Obviously,  $\frac{4-r_k p_{k-1}}{r_k} \ge \frac{4-r_k p_{k-1}}{r_{k-1}} \ge 1$  imply that  $\sum_{k=2}^n U_k$  is  $C^1$ . Using (12), (18), (20) yields

 $\mathcal{L}U_k$ 

$$=\frac{4-r_{k}p_{k-1}}{r_{k}}\Big([x_{k}]^{\frac{r_{k-1}}{r_{k}}}-(\hat{\xi}_{k}+\ell_{k-1}x_{k-1})\Big)|x_{k}|^{\frac{4-r_{k}p_{k-1}-r_{k}}{r_{k}}} \cdot x_{k+1}^{p_{k}}-\ell_{k-1}[e_{k}]^{r_{k}p_{k-1}}\Big([x_{k}]^{\frac{4-r_{k}p_{k-1}}{r_{k}}}-[\hat{x}_{k}]^{\frac{4-r_{k}p_{k-1}}{r_{k}}}\Big) -\ell_{k-1}[e_{k}]^{r_{k}p_{k-1}}\Big([\hat{x}_{k}]^{\frac{4-r_{k}p_{k-1}}{r_{k}}}-[\hat{\xi}_{k}+\ell_{k-1}x_{k-1}]^{\frac{4-r_{k}p_{k-1}}{r_{k-1}}}\Big),$$
(21)

where  $x_{n+1} = u$ .

We can easily verify the following Proposition 2 directly by Lemmas 3,6, we omit the proof here.

Proposition 2: For  $k = 2, \dots, n$ , there holds  $-\ell_{k-1}[e_k]^{r_k p_{k-1}}([x_k]^{\frac{4-r_k p_{k-1}}{r_k}} - [\hat{x}_k]^{\frac{4-r_k p_{k-1}}{r_k}}) \le -\ell_{k-1}$  $2^{1-\frac{4-r_k p_{k-1}}{r_k p_{k-1}}} e_k^4.$ 

To further estimate (21), we give the following Proposition 3-Proposition 5 whose proofs are in the Appendix.

Proposition 3: For  $k = 2, \dots, n-1$ , there holds  $\frac{4-r_k p_{k-1}}{r_k} ([x_k]^{\frac{r_{k-1}}{r_k}} - (\hat{\xi}_k + \ell_{k-1} x_{k-1}))|x_k|^{\frac{4-r_k p_{k-1} - r_k}{r_k}} x_{k+1}^{p_k}$   $\leq b_1 \sum_{j=k-1}^{k+1} z_j^4 + c_{1,k} e_k^4 + h_{1,k} (\ell_{k-1}) e_{k-1}^4, \text{ where } b_1 \text{ and } c_{1,k}$ are positive constants,  $h_{1,k}$  is a positive constant related to  $\ell_{k-1}$ . Proposition 4: For  $k = 3, \dots, n$ , there holds  $-\ell_{k-1}[e_k]^{r_k p_{k-1}}([\hat{x}_k]^{\frac{4-r_k p_{k-1}}{r_k}} - [\hat{\xi}_k + \ell_{k-1} x_{k-1}]^{\frac{4-r_k p_{k-1}}{r_{k-1}}}) \leq b_2 \sum_{j=k-1}^k z_j^4 + c_{2,k} e_k^4 + h_{2,k} (\ell_{k-1}) e_{k-1}^4$ , where  $b_2$  and  $c_{2,k}$  are positive constants,  $h_{2,k}$  is a positive constant related to  $\ell_{k-1}$ .

Proposition 5: There holds  $\frac{4-r_np_{n-1}}{r_n}([x_n]^{\frac{r_{n-1}}{r_n}} - (\hat{\xi}_n + \ell_{n-1}x_{n-1}))|x_n|^{\frac{4-r_np_{n-1}-r_n}{r_n}}u^{p_n} \le b_3 \sum_{j=1}^n z_j^4 + c_3 \sum_{k=2}^n e_k^4 + h_3(\ell_{n-1})e_{n-1}^4$ , where  $b_3$  and  $c_3$  are positive constants,  $h_3$  is a positive constant related to  $\ell_{n-1}$ .

We estimate  $m_n[z_n]^{4-r_{n+1}p_n}(u^{p_n} - u^{*p_n})$  in (17) by the following proposition whose proof is in the Appendix.

Proposition 6:  $m_n[z_n]^{4-r_{n+1}p_n}(u^{p_n}-u^{*p_n}) \le b_4 \sum_{k=1}^n z_k^4 + c_4 \sum_{k=2}^n e_k^4.$ 

From (17), (20), (21) and Propositions 2-6, it follows that

$$\begin{split} \mathcal{L}V|_{(12),(18),(19)} \\ &\leq -\sum_{j=1}^{n-2} \left( a_{j,j} - \tilde{a}_{j+1,j} - \bar{a}_{j+1,j} - \sum_{l=j+2}^{n} \bar{a}_{l,j} - 2b_1 \\ &- 2b_2 - b_3 - b_4 \right) z_j^4 - (a_{n-1,n-1} - \tilde{a}_{n,n-1} - \bar{a}_{n,n-1} \\ &- 2b_1 - 2b_2 - b_3 - b_4 ) z_{n-1}^4 - (a_{n,n} - 2b_1 - 2b_2 - b_3 \\ &- b_4 ) z_n^4 - \left( 2^{1 - \frac{4 - r_{2P_1}}{r_{2P_1}}} \ell_1 - c_{1,2} - h_{1,3}(\ell_2) - h_{2,3}(\ell_2) \right) \\ &- c_3 - c_4 \right) e_2^4 - \sum_{k=3}^{n-2} \left( 2^{1 - \frac{4 - r_k p_{k-1}}{r_k p_{k-1}}} \ell_{k-1} - c_{1,k} - c_{2,k} \right) \\ &- h_{1,k+1}(\ell_k) - h_{2,k+1}(\ell_k) - c_3 - c_4 \right) e_k^4 \\ &- \left( 2^{1 - \frac{4 - r_{n-1} p_{n-2}}{r_{n-1} p_{n-2}}} \ell_{n-2} - c_{1,n-1} - c_{2,n-1} - h_{2,n}(\ell_{n-1}) \right) \\ &- c_3 - h_3(\ell_{n-1}) - c_4 \right) e_{n-1}^4 - \left( 2^{1 - \frac{4 - r_{n} p_{n-1}}{r_n p_{n-1}}} \ell_{n-1} \right) \\ &- c_{2,n} - c_3 - c_4 \right) e_n^4. \end{split}$$

By designing constants  $a_{1,1}, \dots, a_{n,n}, \ell_1, \dots, \ell_{n-1}$  as

$$a_{j,j} \geq 1 + \tilde{a}_{j+1,j} + \bar{a}_{j+1,j} + \sum_{\substack{l=j+2\\l=j+2}}^{n} \bar{a}_{l,j} + 2b_1 + 2b_2 + b_3 + b_4, \quad j = 1, \cdots, n-1, \\ a_{n,n} \geq 1 + 2b_1 + 2b_2 + b_3 + b_4, \\ \ell_{n-1} \geq 2^{\frac{4-r_{n}p_{n-1}}{r_{n}-1}-1} (1 + c_{2,n} + c_3 + c_4), \\ \ell_{n-2} \geq 2^{\frac{4-r_{n}p_{n-2}}{r_{n-1}p_{n-2}}-1} (1 + c_{1,n-1} + c_{2,n-1} + c_4 + c_3 + h_3(\ell_{n-1}) + h_{2,n}(\ell_{n-1})), \\ \ell_{k-1} \geq 2^{\frac{4-r_{k}p_{k-1}}{r_{k}p_{k-1}}-1} (1 + c_{1,k} + c_{2,k} + h_{1,k+1}(\ell_k) + h_{2,k+1}(\ell_k) + c_3 + c_4), \quad k = n-2, \cdots, 3, \\ \ell_1 \geq 2^{\frac{4-r_{2}p_1}{r_{2}p_1}-1} (1 + c_{1,2} + h_{1,3}(\ell_2) + h_{2,3}(\ell_2) + c_3 + c_4),$$
(22)

one obtains

$$\mathcal{L}V|_{(12),(18),(19)} \le -\sum_{k=1}^{n} z_k^4 - \sum_{k=2}^{n} e_k^4.$$
 (23)

Denote  $\mathcal{X} = (x_1, \dots, x_n, \hat{\xi}_2, \dots, \hat{\xi}_n)^\top$ , the closed-loop system (12), (18), (19) can be rewritten as

$$d\mathcal{X} = \Phi(\mathcal{X})dt = (x_2^{p_1}, \cdots, u^{p_n}, -\ell_1 \hat{x}_2^{p_1}, \cdots, -\ell_{n-1} \hat{x}_n^{p_{n-1}})^\top dt.$$
(24)

Introducing the dilation weight<sup>2</sup>  $\Delta = (r_1, \dots, r_n, r_1, \dots, r_{n-1})$  for  $\mathcal{X}$ , we have the following proposition whose proof is in the Appendix.

Proposition 7: There exist positive constants  $\pi_1, \pi_2$  and  $\pi_3$  such that  $\pi_1 \|\mathcal{X}\|_{\Delta}^{4-\tau} \leq V(\mathcal{X}) \leq \pi_2 \|\mathcal{X}\|_{\Delta}^{4-\tau}, \mathcal{L}V|_{(19),(24)} \leq -\pi_3 \|\mathcal{X}\|_{\Delta}^4.$ 

*Part III. Output feedback controller of system (1)* Introduce the transformation

$$x_i = \Gamma^{-d_i} \eta_i, \quad u = \Gamma^{-d_{n+1}} v, \ i = 1, \cdots, n,$$
 (25)

then system (1) can be converted into

$$dx_{i} = (\Gamma x_{i+1}^{p_{i}} + \tilde{f}_{i}(x_{i+2}, \cdots, x_{n}, u))dt + \tilde{g}_{i}^{\top}(x_{i+2}, \cdots, x_{n}, u)d\omega, \quad i = 1, \cdots, n-2, dx_{n-1} = (\Gamma x_{n}^{p_{n-1}} + \tilde{f}_{n-1}(u))dt + \tilde{g}_{n-1}^{\top}(u)d\omega, dx_{n} = \Gamma u^{p_{n}}dt, \quad x(0) = x_{0}(\eta_{0}), y = x_{1},$$
(26)

where  $d_1 = 0$ ,  $d_i = \frac{d_{i-1}+1}{p_{i-1}}$ ,  $i = 2, \dots, n, 0 < \Gamma \le 1$  is a constant to be determined,  $\tilde{f}_i = \Gamma^{-d_i} f_i$  and  $\tilde{g}_i = \Gamma^{-d_i} g_i$ ,  $i = 1, \dots, n-1$ .

In view of (18), the observer of system (26) is constructed as:

$$d\hat{\xi}_{k} = -\Gamma \ell_{k-1} \hat{x}_{k}^{p_{k-1}} dt, \hat{x}_{k} = [\hat{\xi}_{k} + \ell_{k-1} \hat{x}_{k-1}]^{\frac{r_{k}}{r_{k-1}}}, \quad k = 2, \cdots, n, \quad (27)$$

where  $\ell_1, \dots, \ell_{n-1}$  are chosen in (22). Then, (19), (26) and (27) can be rewritten in the compact form:

$$d\mathcal{X} = (\Gamma \Phi(\mathcal{X}) + F(\mathcal{X}))dt + G^{\top}(\mathcal{X})d\omega, \qquad (28)$$

where  $\Phi(\mathcal{X})$  is defined in (24),  $F = (\tilde{f}_1, \dots, 0, 0, \dots, 0)^{\top}$ ,  $G = (\tilde{g}_1, \dots, 0, 0, \dots, 0)$ . Noting that  $\Phi$  in (28) has the same structure as (24), by adopting the same Lyapunov function  $V(\mathcal{X})$  as in the preceding subsection, one can conclude from Proposition 7 that

$$\mathcal{L}V|_{(19),(28)} \leq -\Gamma\pi_3 \|\mathcal{X}\|_{\Delta}^4 + \frac{\partial V}{\partial \mathcal{X}}F + \frac{1}{2}\mathrm{Tr}\left\{G\frac{\partial^2 V}{\partial \mathcal{X}^2}G^{\top}\right\}.$$
(29)

We now estimate the last two terms on the right-hand side of (29) by the following proposition whose proof is in the Appendix.

 $<sup>^{2}</sup>$ For a more precise definition of dilation and homogeneity, please refer to [28].

*Proposition 8:* There exists a positive constant  $\gamma_0$  such that

$$\frac{\partial V(\mathcal{X})}{\partial \mathcal{X}} F(\mathcal{X}) + \frac{1}{2} \operatorname{Tr} \left\{ G(\mathcal{X}) \frac{\partial^2 V(\mathcal{X})}{\partial \mathcal{X}^2} G^{\top}(\mathcal{X}) \right\} \leq \gamma_0 \Gamma^{1+\nu} \|\mathcal{X}\|_{\Delta}^4.$$
(30)

Substituting Proposition 8 into (29), and choosing  $\Gamma = \min\{(\frac{\pi_3}{2\gamma_0})^{\frac{1}{\nu}}, 1\}$ , one gets

$$\mathcal{L}V(\mathcal{X})|_{(19),(28)} \le -\Gamma(\pi_3 - \gamma_0 \Gamma^{\nu}) \|\mathcal{X}\|_{\Delta}^4 \le -\frac{\pi_3 \Gamma}{2} \|\mathcal{X}\|_{\Delta}^4.$$
(31)

Up to now, by (19), (25), (27), one obtains output feedback controller of system (1):

$$v = -\Gamma^{d_{n+1}} \beta_n^{r_{n+1}} \left[ [\hat{x}_n]^{\frac{1}{r_n}} + \sum_{i=1}^{n-1} \left( \prod_{j=i}^{n-1} \beta_j \right) [\hat{x}_i]^{\frac{1}{r_i}} \right]^{r_{n+1}}, \quad (32)$$

where  $\hat{x}_1 = \eta_1$  and  $\hat{x}_2, \dots, \hat{x}_n$  are observed by (27).

## **IV. FINITE-TIME STABILITY**

We state the main result in this paper.

Theorem 1: When Assumption 1 holds, the equilibrium  $\eta = 0$  of the closed-loop system (1) with (27), (32) is finite-time stable in probability.

*Proof:* With (31), Lemma 2, the closed-loop stochastic system (19) and (28) has a solution for any initial value  $\mathcal{X}_0 \in \mathbb{R}^n$ . Since  $\Gamma \Phi(\mathcal{X}) + F(\mathcal{X})$  and  $G^{\top}(\mathcal{X})$  are  $C^1$  on  $\mathbb{R}^n/\{0\}$ , the closed-loop stochastic system obviously has a unique solution in forward time for any  $\mathcal{X}_0 \in \mathbb{R}^n/\{0\}$ . When  $\mathcal{X}_0 = 0$ , the closed-loop stochastic system has a unique zero solution, that is, when  $\mathcal{X}_0 = 0$ ,  $P(||\mathcal{X}(t; \mathcal{X}_0)|| \ge r) = 0$  for all r > 0 and  $t \ge 0$ , which can be verified by the following counter-evidence:

Suppose that there are  $r_0 > 0$  and  $t_0 \ge 0$  such that

$$P(\|\mathcal{X}(t_0; \mathcal{X}_0)\| \ge r_0) > \varepsilon_0, \tag{33}$$

where  $1 \ge \varepsilon_0 \ge 0$ . By Definition 2.1 and Theorem 2.2 in [32, Sec. 4.2], one concludes from (31) that the trivial solution of stochastic system (19) and (28) is stable in probability, that is, for every pair of  $\varepsilon \in (0, 1)$  and r > 0, there exists a  $\delta = \delta(\varepsilon, r) > 0$  such that  $P(||\mathcal{X}(t; \mathcal{X}_0)|| < r$  for all  $t \ge 0) \ge 1 - \varepsilon$  whenever  $||\mathcal{X}_0|| \le \delta$ . This conclusion implies that for above  $r_0$  and  $\varepsilon_0$ ,  $P(||\mathcal{X}(t; \mathcal{X}_0)|| < r_0$  for all  $t \ge 0) \ge 1 - \varepsilon_0$ . This contradicts (33).

According to the lemmas of weighted homogeneity in [28], there is a constant  $\pi_0 > 0$  such that

$$\mathcal{L}V(\mathcal{X})|_{(19),(28)} \le -\frac{\pi_3}{2} \Gamma \|\mathcal{X}\|_{\Delta}^4 \le -\pi_0 V(\mathcal{X})^{\frac{4}{4-\tau}}.$$
 (34)

Since the closed-loop system (19), (28) has a unique solution, then it follows from  $0 < \frac{4}{4-\tau} < 1$ , Lemma 3 and Proposition 7 that the origin of the closed-loop stochastic system is finite-time stable in probability. Since (25) is an equivalent transformation, one concludes that origin of closedloop stochastic system (1), (27), (32) is finite-time stable in probability.

## **V. A SIMULATION EXAMPLE**

Without loss of generality, we consider a simple example:

$$d\eta_{1} = \left(\eta_{2}^{\frac{5}{3}} + 0.2 \ln \left(1 + |v|^{\frac{25}{13}}\right)\right) dt + 0.05v^{2} d\omega, d\eta_{2} = v dt, y = \eta_{1}.$$
 (35)

By choosing  $\tau = -\frac{1}{27} \in (-\frac{1}{2(1+p_1)}, 0) = (-\frac{3}{16}, 0)$  and (10), one has  $r_1 = \frac{1}{2}, r_2 = \frac{5}{18}, r_3 = \frac{13}{54}$ . Assumption 1 holds with  $f_1 = 0.2 \ln(1 + |v|^{\frac{25}{13}}) \le |v|^{\frac{25}{13}} = |v|^{\frac{r_1+\tau}{r_3}}, g_1 = 0.05v^2 \le 0.05|v|^{\frac{2r_1+\tau}{2r_3}}$ .

Following the design procedure as in Section III, one constructs the finite-time output feedback controller:

$$v = -\Gamma^{\frac{8}{5}} \beta_2^{\frac{13}{54}} \left[ [\hat{x}_2]^{\frac{18}{5}} + \beta_1 [\eta_1]^2 \right]^{\frac{13}{54}},$$
  
$$d\hat{\xi}_2 = -\Gamma \ell_1 \hat{x}_2^{\frac{5}{3}} dt, \quad \hat{x}_2 = [\hat{\xi}_2 + \ell_1 \eta_1]^{\frac{5}{9}}, \qquad (36)$$

where  $\beta_1 > 0$ ,  $\beta_2 > 0$ ,  $\ell_1 > 0$ ,  $0 < \Gamma \le 1$  are design parameters.

Choose  $(\eta_1(0), \eta_2(0), \hat{\xi}_2(0))^{\top} = (0.04, -0.002, 0.02)^{\top},$  $\Gamma = 0.09, \ell_1 = 500, \beta_1 = 1, \beta_2 = 20.$  Fig.1 shows the effectiveness of control method.



FIGURE 1. The responses of closed-loop system (35)-(36).

#### **VI. A CONCLUDING REMARK**

The finite-time output feedback stabilization in probability of stochastic high-order nonlinear feedforward autonomous systems is considered. In this paper, we initially list some lemmas which will deal with the nonlinearities with powers being any rational number and bigger than one, then based on the homogeneous domination and stochastic Lyapunov finite-time stability theory, the stochastic finite-time output feedback stabilizer is designed and analysed. A remaining problem is whether the finite-time stabilization can be established for stochastic nonlinear feedforward nonautonomous systems.

#### **APPENDIX**

*Proof of Proposition 1:* Firstly,  $V_k(\bar{x}_k)$  is positive definite and radially unbounded, one can see the detailed proof in [27].

In view of  $-\frac{1}{\sum_{l=1}^{n} p_1 \cdots p_{l-1}} < \tau < 0$ , one deduces that  $\frac{1}{r_k} \ge 2$ ,  $4 - r_{k+1}p_k \ge 2$ ,  $k = 1, \cdots, n$ . This implies that  $V_k$  is  $C^2$ . Define  $z_k = [x_k]^{\frac{1}{r_k}} - [\alpha_{k-1}]^{\frac{1}{r_k}}$ , it follows from (13) that

$$\begin{aligned} \mathcal{L}V_{k-1}(\bar{x}_{k-1})|_{(12)} \\ &\leq -\sum_{j=1}^{k-3} \left( a_{j,j} - \tilde{a}_{j+1,j} - \bar{a}_{j+1,j} - \sum_{l=j+2}^{k-1} \bar{a}_{l,j} \right) z_{j}^{4} \\ &- (a_{k-2,k-2} - \tilde{a}_{k-1,k-2} - \bar{a}_{k-1,k-2}) z_{k-2}^{4} - a_{k-1,k-1} z_{k-1}^{4} \\ &+ m_{k} [z_{k}]^{4-r_{k+1}p_{k}} (x_{k+1}^{p_{k}} - \alpha_{k}^{p_{k}}) + m_{k} [z_{k}]^{4-r_{k+1}p_{k}} \alpha_{k}^{p_{k}} \\ &+ m_{k-1} [z_{k-1}]^{4-r_{k}p_{k-1}} (x_{k}^{p_{k-1}} - \alpha_{k-1}^{p_{k-1}}) + m_{k} \sum_{j=1}^{k-1} \frac{\partial W_{k}}{\partial x_{j}} x_{j+1}^{p_{j}}. \end{aligned}$$

$$(37)$$

Lemmas 5,6 yield

$$m_{k-1}[z_{k-1}]^{4-r_k p_{k-1}} (x_k^{p_{k-1}} - \alpha_{k-1}^{p_{k-1}}) \\ \leq \tilde{a}_{k,k-1} z_{k-1}^4 + \lambda_{k,1} z_k^4, \quad (38)$$

where  $\tilde{a}_{k,k-1}$  is a positive constant,  $\lambda_{k,1}$  is a positive constant dependent on  $\tilde{a}_{k,k-1}$ .

With the help of Lemmas 5,6,8 and the integral mean value theorem, one can deduce that

$$m_{k} \sum_{j=1}^{k-1} \frac{\partial W_{k}}{\partial x_{j}} \cdot x_{j+1}^{p_{j}}$$

$$\leq c \sum_{j=1}^{k-1} |z_{k}|^{3-\tau} \left( |z_{j}|^{1-r_{j}} + |z_{j-1}|^{1-r_{j}} \right) \left( |z_{j+1}|^{r_{j+1}p_{j}} + |z_{j}|^{r_{j+1}p_{j}} \right)$$

$$\leq \sum_{j=1}^{k-1} \bar{a}_{k,j} z_{j}^{4} + \lambda_{k,2} z_{k}^{4}, \qquad (39)$$

where  $\bar{a}_{k,1}, \dots, \bar{a}_{k,k-1}$  are positive constants,  $c, \lambda_{k,2}$  are appropriate positive constants,  $\lambda_{k,2}$  is dependent on  $\bar{a}_{k,1}, \dots, \bar{a}_{k,k-1}$ .

Defining  $\lambda_k = \lambda_{k,1} + \lambda_{k,2}$ , choosing  $\beta_k = (\frac{a_{k,k} + \lambda_k}{m_k})^{\frac{1}{r_{k+1}p_k}}$ with  $a_{k,k} > 0$  being a positive constant to be designed, and substituting (38), (39) into (37), one gets (15).

*Proof of Proposition 3:* By (16), (18), Lemmas 4,5,7, one deduces that

$$\frac{4 - r_{k}p_{k-1}}{r_{k}} ([x_{k}]^{\frac{r_{k-1}}{r_{k}}} - (\hat{\xi}_{k} + \ell_{k-1}x_{k-1}))x_{k+1}^{p_{k}}|x_{k}|^{\frac{4 - r_{k}p_{k-1} - r_{k}}{r_{k}}} \\
\leq \frac{4 - r_{k}p_{k-1}}{r_{k}} ([x_{k}]^{\frac{r_{k-1}}{r_{k}}} - [\hat{x}_{k}]^{\frac{r_{k-1}}{r_{k}}}| + \ell_{k-1}|x_{k-1} - \hat{x}_{k-1}|) \\
\cdot |x_{k}|^{\frac{4 - r_{k}p_{k-1} - r_{k}}{r_{k}}}|x_{k+1}|^{p_{k}} \\
\leq (c|e_{k}|^{d_{1,k}r_{k}}(|e_{k}|^{r_{k-1} - d_{1,k}r_{k}} + |z_{k}|^{r_{k-1} - d_{1,k}r_{k}} \\
+ |z_{k-1}|^{r_{k-1} - d_{1,k}r_{k}}) + \bar{h}_{1,k}(\ell_{k-1})|e_{k-1}|^{r_{k-1}}) \\
\cdot (|z_{k}|^{4 - r_{k}p_{k-1} - r_{k}} + |z_{k}|^{4 - r_{k}p_{k-1} - r_{k}} + |z_{k}|^{r_{k+1}p_{k}}) \\
\leq b_{1}\sum_{j=k-1}^{k+1} z_{j}^{4} + c_{1,k}e_{k}^{4} + h_{1,k}(\ell_{k-1})e_{k-1}^{4},$$
(40)

where  $\bar{c}_{1,k} = c_0(\frac{r_{k-1}}{r_k})$ ,  $d_{1,k} = d(\frac{r_{k-1}}{r_k})$ ,  $\bar{h}_{1,k}$  is a positive constant related to  $\ell_{k-1}$ , and throughout this paper we use c to indicate a general constant.

*Proof of Proposition 4:* From (16), (18), Lemmas 4,5,7, it follows that

$$-\ell_{k-1}[e_{k}]^{r_{k}p_{k-1}}([\hat{x}_{k}]^{\frac{4-r_{k}p_{k-1}}{r_{k}}} - [\hat{\xi}_{k} + \ell_{k-1}x_{k-1}]^{\frac{4-r_{k}p_{k-1}}{r_{k-1}}})$$

$$\leq \ell_{k-1}^{1+d_{2,k}}\bar{c}_{2,k}|e_{k}|^{r_{k}p_{k-1}}|x_{k-1} - \hat{x}_{k-1}|^{d_{2,k}}$$

$$\cdot (|\hat{x}_{k}|^{\frac{4-r_{k}p_{k-1}-d_{2,k}r_{k-1}}{r_{k}}} + |\ell_{k-1}(x_{k}^{p_{k-1}} - \hat{x}_{k}^{p_{k-1}})|^{\frac{4-r_{k}p_{k-1}}{r_{k-1}}} - d_{2,k})$$

$$\leq b_{2}\sum_{j=k-1}^{k}z_{j}^{4} + c_{2,k}e_{k}^{4} + h_{2,k}(\ell_{k-1})e_{k-1}^{4}, \qquad (41)$$

where  $\bar{c}_{2,k} = c_0(\frac{4-r_k p_{k-1}}{r_{k-1}}), d_{2,k} = d(\frac{4-r_k p_{k-1}}{r_{k-1}}), \bar{h}_{2,k}$  is a positive constant related to  $\ell_{k-1}$ .

*Proof of Proposition 5:* By (16), (18), (19), Lemmas 4,5,7, one deduces that

$$\frac{4 - r_n p_{n-1}}{r_n} ([x_n]^{\frac{r_{n-1}}{r_n}} - (\hat{\xi}_n + \ell_{n-1} x_{n-1}))|x_n|^{\frac{4 - r_n p_{n-1} - r_n}{r_n}} u^{p_n}$$

$$\leq \bar{c}_3 |e_n|^{d_{1,n}r_n} (|z_n|^{4 - r_n p_{n-1} - r_n} + |z_{n-1}|^{4 - r_n p_{n-1} - r_n})$$

$$\cdot (|e_n|^{r_{n-1} - d_{1,n}r_n} + |z_n|^{r_{n-1} - d_{1,n}r_n} + |z_{n-1}|^{r_{n-1} - d_{1,n}r_n})$$

$$\cdot (\sum_{j=2}^n |e_j|^{r_{n+1}p_n} + \sum_{j=1}^n |z_j|^{r_{n+1}p_n}) + \bar{h}_3(\ell_{n-1})|e_{n-1}|^{r_{n-1}}$$

$$\cdot (|z_n|^{4 - r_n p_{n-1} - r_n} + |z_{n-1}|^{4 - r_n p_{n-1} - r_n})$$

$$\cdot (\sum_{j=2}^n |e_j|^{r_{n+1}p_n} + \sum_{j=1}^n |z_j|^{r_{n+1}p_n})$$

$$\leq b_3 \sum_{j=1}^n z_j^4 + c_3 \sum_{k=2}^n e_k^4 + h_3(\ell_{n-1})e_{n-1}^4,$$

where  $\bar{c}_3$  is a positive constant,  $\bar{h}_3$  is a positive constant related to  $\ell_{n-1}$ .

*Proof of Proposition 6:* With (16), (19) and Lemmas 4,5,7, one can deduce that

$$\begin{split} m_{n}[z_{n}]^{4-r_{n+1}p_{n}}(u^{p_{n}}-u^{*p_{n}}) \\ &\leq \tilde{c}_{4}|z_{n}|^{4-r_{n+1}p_{n}}\sum_{\substack{k=2\\ k=2}}^{n}|x_{k}^{p_{k-1}}-\hat{x}_{k}^{p_{k-1}}|^{d_{3,k}} \\ &\cdot(|x_{k}^{p_{k-1}}-\hat{x}_{k}^{p_{k-1}}|^{\frac{r_{n+1}p_{n}}{r_{k}p_{k-1}}-d_{3,k}}+|x_{k}|^{\frac{r_{n+1}p_{n}}{r_{k}}-d_{3,k}p_{k-1}}) \\ &\leq \bar{c}_{4}|z_{n}|^{4-r_{n+1}p_{n}}\sum_{\substack{k=2\\ k=2}}^{n}|e_{k}|^{d_{3,k}r_{k}p_{k-1}}(|e_{k}|^{r_{n+1}p_{n}-d_{3,k}r_{k}p_{k-1}} \\ &+|z_{k}|^{r_{n+1}p_{n}-d_{3,k}r_{k}p_{k-1}}+|z_{k-1}|^{r_{n+1}p_{n}-d_{3,k}r_{k}p_{k-1}}) \\ &\leq b_{4}\sum_{k=1}^{n}z_{k}^{4}+c_{4}\sum_{k=2}^{n}e_{k}^{4}, \end{split}$$

where  $\tilde{c}_4$ ,  $\bar{c}_4$  are positive constants,  $d_{3,j} = d(\frac{r_{n+1}p_n}{r_jp_{j-1}})$ . *Proof of Proposition 7:* Obviously *V* is homogeneous of degree  $4 - \tau$  and the term  $-\sum_{k=1}^n z_k^4 - \sum_{k=2}^n e_k^4 = -W$  on right-hand side of (23) is is homogeneous of degree 4, which is verified by

$$\begin{split} V(\Delta_{\varepsilon}(\mathcal{X})) &= \sum_{k=1}^{n} \int_{\varepsilon'^{k} \alpha_{k-1}}^{\varepsilon'^{k} x_{k}} [[s]^{\frac{1}{r_{k}}} - \varepsilon[\alpha_{k-1}]^{\frac{1}{r_{k}}}]^{4-r_{k+1}p_{k}} ds \\ &+ \sum_{k=2}^{n} \int_{[\varepsilon'^{k-1} \hat{\xi}_{k} + \ell_{k-1}\varepsilon'^{k-1} x_{k-1}]}^{[\varepsilon'^{k} x_{k}]^{\frac{4-r_{k}p_{k-1}}{r_{k}}}} ([\iota]^{\frac{r_{k-1}}{4-r_{k}p_{k-1}}} \\ &- \varepsilon^{r_{k-1}}(\hat{\xi}_{k} + \ell_{k-1}x_{k-1})) d\iota \\ &= \varepsilon^{4-\tau} V(\mathcal{X}), \\ &- W(\Delta_{\varepsilon}(\mathcal{X})) \\ &= -\varepsilon^{4} \sum_{k=1}^{n} (\varepsilon[x_{k}]^{\frac{1}{r_{k}}} + \sum_{i=1}^{k-1} (\prod_{j=i}^{i-1} \beta_{j})\varepsilon[x_{i}]^{\frac{1}{r_{i}}})^{4} \\ &- \varepsilon^{4} \sum_{k=2}^{n} (x_{k}^{p_{k-1}} - [\hat{\xi}_{k} + \ell_{k-1}[\hat{\xi}_{k-1} + \ell_{k-2}[\cdots \\ &+ \ell_{2}[\hat{\xi}_{2} + \ell_{1}x_{1}]^{\frac{r_{2}}{r_{1}}} \cdots ]^{\frac{r_{k-2}}{r_{k-3}}}]^{\frac{r_{k-1}}{r_{k-1}}} ]^{\frac{r_{k}p_{k-1}}{r_{k-1}}} \\ &= -\varepsilon^{4} W(\mathcal{X}). \end{split}$$

Noting that  $V(\mathcal{X})$  and  $W(\mathcal{X})$  are both positive definite and radially unbounded, by the lemmas of weighted homogeneity in [28], one can find positive constants  $\pi_1$ ,  $\pi_2$ , and  $\pi_3$  such that Proposition 7 holds.

Proof of Proposition 8: By (16), (19), (25), Lemmas 4,9 and Assumption 1,  $|\tilde{f}_i| \leq \gamma_1 \Gamma^{1+\nu} \|\mathcal{X}\|_{\Delta}^{r_i+\tau}$ ,  $\|\tilde{g}_i\| \leq \gamma_2 \Gamma^{1+\nu} \|\mathcal{X}\|_{\Delta}^{2r_i+\tau}$ , where  $0 < \nu = \min_{\substack{i=1,\cdots,n-1, \ j=i+2,\cdots,n+1}} \{\frac{d_j(r_i+\tau)}{r_j} - d_i - 1\}$ ,

 $\gamma_1$ ,  $\gamma_2$  are positive constants. Then

$$\frac{\partial V(\mathcal{X})}{\partial \mathcal{X}} F(\mathcal{X}) + \frac{1}{2} \operatorname{Tr} \left\{ \partial G(\mathcal{X}) \frac{\partial^2 V(\mathcal{X})}{\partial \mathcal{X}^2} G^{\top}(\mathcal{X}) \right\}$$

$$= \sum_{j=1}^{n-1} \frac{\partial V}{\partial x_j} \tilde{f}_j + \frac{1}{2} \sum_{i,j=1}^{n-1} \operatorname{Tr} \{ \tilde{g}_i \frac{\partial^2 V}{\partial x_i \partial x_j} \tilde{g}_j^{\top} \}$$

$$\leq \sum_{j=1}^{n-1} \left| \frac{\partial V}{\partial x_j} \right| \cdot |\tilde{f}_j| + \bar{\gamma}_0 \sum_{i,j=1}^{n-1} \left\| \frac{\partial^2 V}{\partial x_i \partial x_j} \right\| \cdot \|\tilde{g}_i\| \cdot \|\tilde{g}_j^{\top}\|$$

$$\leq \bar{\gamma}_{01} \Gamma^{1+\nu} \sum_{j=1}^{n-1} \left\| \mathcal{X} \right\|_{\Delta}^{4-\tau-r_j} \cdot \left\| \mathcal{X} \right\|_{\Delta}^{r_j+\tau}$$

$$+ \bar{\gamma}_{02} \Gamma^{1+\nu} \sum_{i,j=1}^{n-1} \left\| \mathcal{X} \right\|_{\Delta}^{4-\tau-r_i-r_j} \cdot \left\| \mathcal{X} \right\|_{\Delta}^{\frac{2r_i+\tau}{2}} \cdot \left\| \mathcal{X} \right\|_{\Delta}^{\frac{2r_j+\tau}{2}}$$

$$\leq \gamma_0 \Gamma^{1+\nu} \left\| \mathcal{X} \right\|_{\Delta}^{4,}, \qquad (42)$$

where  $\bar{\gamma}_0$ ,  $\bar{\gamma}_{01}$ ,  $\bar{\gamma}_{02}$  and  $\gamma_0$  are appropriate constants.

#### REFERENCES

- A. R. Teel, "A nonlinear small gain theorem for the analysis of control systems with saturation," *IEEE Trans. Autom. Control*, vol. 41, no. 9, pp. 1256–1270, Sep. 1996.
- [2] G. Kaliora and A. Astolfi, "Nonlinear control of feedforward systems with bounded signals," *IEEE Trans. Autom. Control*, vol. 49, no. 11, pp. 1975–1990, Nov. 2004.
- [3] M. Jankovic, R. Sepulchre, and P. V. Kokotovic, "Constructive Lyapunov stabilization of nonlinear cascade systems," *IEEE Trans. Autom. Control*, vol. 41, no. 12, pp. 1723–1735, Dec. 1996.

- [4] R. Sepulchre, M. Jankovic, and P. V. Kokotovic, "Integrator forwarding: A new recursive nonlinear robust design," *Automatica*, vol. 33, pp. 979–984, May 1997.
- [5] Y. Xudong, "Universal stabilization of feedforward nonlinear systems," *Automatica*, vol. 39, pp. 141–147, Jan. 2003.
- [6] C. Qian and J. Li, "Global output feedback stabilization of uppertriangular nonlinear systems using a homogeneous domination approach," *Int. J. Robust Nonlinear Control*, vol. 16, pp. 441–463, Jun. 2006.
- [7] M. Krstic, "Input delay compensation for forward complete and strict-feedforward nonlinear systems," *IEEE Trans. Autom. Control*, vol. 55, no. 2, pp. 287–303, Feb. 2010.
- [8] X. Zhang, Q. Liu, L. Baron, and E.-K. Boukas, "Feedback stabilization for high order feedforward nonlinear time-delay systems," *Automatica*, vol. 47, pp. 962–967, May 2011.
- [9] S. Ding, C. Qian, S. Li, and Q. Li, "Global stabilization of a class of uppertriangular systems with unbounded or uncontrollable linearizations," *Int. J. Robust Nonlinear Control*, vol. 21, pp. 271–294, Feb. 2011.
- [10] I. Karafyllis and M. Krstic, "Global stabilization of feedforward systems under perturbations in sampling schedule," *SIAM J. Control Optim.*, vol. 50, no. 3, pp. 1389–1412, 2012.
- [11] W. Zha, J. Zhai, and S. Fei, "Global output feedback control for a class of high-order feedforward nonlinear systems with input delay," *ISA Trans.*, vol. 52, no. 4, pp. 494–500, Jul. 2013.
- [12] X.-J. Xie and X.-H. Zhang, "Further results on global state feedback stabilization of nonlinear high-order feedforward systems," *ISA Trans.*, vol. 53, pp. 341–346, Mar. 2014.
- [13] S. Battilotti, "Stochastic stabilization of nonlinear systems in feedforward form with noisy outputs," *IEEE Trans. Autom. Control*, vol. 50, no. 1, pp. 100–105, Jan. 2005.
- [14] G.-D. Zhao and X.-J. Xie, "A homogeneous domination approach to state feedback of stochastic high-order feedforward nonlinear systems," *Int. J. Control*, vol. 86, no. 5, pp. 966–976, 2013.
- [15] C.-R. Zhao and X.-J. Xie, "Global stabilization of stochastic high-order feedforward nonlinear systems with time-varying delay," *Automatica*, vol. 50, pp. 203–210, Jan. 2014.
- [16] S. P. Bhat and D. S. Bernstein, "Finite-time stability of continuous autonomous systems," *SIAM J. Control Optim.*, vol. 38, no. 3, pp. 751–766, 2000.
- [17] X.-H. Zhang, K. Zhang, and X.-J. Xie, "Finite-time stabilization of stochastic high-order nonlinear systems," *Asian J. Control*, vol. 18, pp. 2244–2255, Nov. 2016.
- [18] Z.-L. Zhao and Z.-P. Jiang, "Finite-time output feedback stabilization of lower-triangular nonlinear systems," *Automatica*, vol. 96, pp. 259–269, Oct. 2018.
- [19] Z. Song and J. Zhai, "Finite-time stabilization for switched stochastic nonlinear systems," in *Proc. Chin. Automat. Congr. (CAC)*, 2017, pp. 2104–2108.
- [20] F. Wang, B. Chen, Y. Sun, and C. Lin, "Finite time control of switched stochastic nonlinear systems," *Fuzzy Sets Syst.*, vol. 365, pp. 140–152, Jun. 2018.
- [21] X.-J. Xie, X.-H. Zhang, and K. Zhang, "Finite-time state feedback stabilisation of stochastic high-order nonlinear feedforward systems," *Int. J. Control*, vol. 89, no. 7, pp. 1332–1341, 2016.
- [22] J. Yin, S. Khoo, Z. Man, and X. Yu, "Finite-time stability and instability of stochastic nonlinear systems," *Automatica*, vol. 47, pp. 2671–2677, Dec. 2011.
- [23] A. V. Skorokhod, Studies in the Theory of Random Processes. Boston, MA, USA: Addison-Wesley, 1965.
- [24] J. Yin and S. Khoo, "Continuous finite-time state feedback stabilizers for some nonlinear stochastic systems," *Int. J. Robust Nonlinear Control*, vol. 25, pp. 1581–1600, Jul. 2015.
- [25] C. Qian and W. Lin, "A continuous feedback approach to global strong stabilization of nonlinear systems," *IEEE Trans. Autom. Control*, vol. 46, no. 7, pp. 1061–1079, Jul. 2001.
- [26] Z.-Y. Sun, X.-H. Zhang, and X.-J. Xie, "Continuous global stabilisation of high-order time-delay nonlinear systems," *Int. J. Control*, vol. 86, no. 6, pp. 994–1007, 2013.
- [27] X.-H. Zhang, K. Zhang, and X.-J. Xie, "Finite-time output feedback stabilization of nonlinear high-order feedforward systems," *Int. J. Robust Nonlinear Control*, vol. 26, pp. 1794–1814, May 2016.
- [28] J. Li and C. Qian, "Global finite-time stabilization of a class of uncertain nonlinear systems using output feedback," in *Proc. 44th IEEE Conf. Decis. Control*, Dec. 2005, pp. 2652–2657.

- [29] H. Wang and Q. Zhu, "Finite-time stabilization of high-order stochastic nonlinear systems in strict-feedback form," *Automatica*, vol. 54, pp. 284–291, Apr. 2015.
- [30] Q. Lan, S. Li, S. Khoo, and P. Shi, "Global finite-time stabilisation for a class of stochastic nonlinear systems by output feedback," *Int. J. Control*, vol. 88, no. 3, pp. 494–506, 2015.
- [31] Z. Song and J. Zhai, "Global finite-time stabilization for switched stochastic nonlinear systems via output feedback," *J. Franklin Inst.*, vol. 356, pp. 1379–1395, Feb. 2019.
- [32] X. R. Mao, Stochastic Differential Equations and Their Applications. Chichester, U.K.: Horwood Publishing, 2007.
- [33] X.-H. Zhang and X.-J. Xie, "Global state feedback stabilisation of nonlinear systems with high-order and low-order nonlinearities," *Int. J. Control*, vol. 87, no. 3, pp. 642–652, 2014.



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