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Dual-Band Eight-Antenna Array Design for MIMO Applications in 5G Mobile Terminals

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ABSTRACT This paper proposes a dual-band eight-antenna array for multiple-input and multiple-output (MIMO) applications in 5G mobile terminals. The designed MIMO antenna array comprises eight L-shaped slot antennas based on stepped impedance resonators (SIRs). The required dual-resonance can be obtained by adjusting the impedance ratio of the SIR, and good impedance matching can be ensured for each antenna element by tuning the position of the microstrip feed line. The experimental results show that a measured return loss of higher than 10 dB and a measured inter-element isolation of greater than 11.2 dB have been obtained for each antenna element with a simulated total efficiency of larger than 51% across the long term evolution (LTE) band 42 (3400–3600 MHz) and LTE band 46 (5150–5925 MHz). In addition, the measured envelope correlation coefficient (ECC) is lower than 0.1 between arbitrary two antenna elements, and the proposed MIMO antenna array realizes a simulated channel capacity of higher than 36.9 bps/Hz within both operation bands. Furthermore, the MIMO antenna array can maintain acceptable radiation and MIMO performance in the presence of specific anthropomorphic mannequin (SAM) head and human hands.

INDEX TERMS Channel capacity, dual-band, multiple input and multiple output (MIMO), slot antenna, stepped impedance resonator (SIR).

I. INTRODUCTION

Due to the tremendous prosperity of modern communication systems, people have been demanding more and more rapid information transmission rate. Consequently, it is essential to possess a much greater channel capacity for modern wireless communication systems, which promotes the development of 5G communications significantly. The world is to usher 5G communications in the near future. It is well known that the multiple input and multiple output (MIMO) systems relying on multi-antennas as one of the crucial technologies for 5G operation can improve the channel capacity and enhance the communication reliability. The 5G communications needs to support higher data transmission rate, shorter time delay, and

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larger channel capacity compared to the 4G communications. The 2×2 and 4×4 MIMO antenna arrays resultantly have gradually failed to meet the rigorous requirement. Therefore, many researches have been conducted on the 8×8 MIMO antenna array designs, which are targeted to be applied in 5G mobile terminals [1], [2].

The long term evolution (LTE) band 42 (3400–3600 MHz) has been licensed as one of the sub-6GHz bands. Thus, many MIMO antenna arrays at LTE band 42 have been designed. In [3], an 8-port MIMO antenna array as well as a 16-port one have been studied, while a 12×12 MIMO antenna array has been proposed in [4]. However, the MIMO antenna array operating merely at the LTE band 42 cannot fully satisfy the demand of the forthcoming 5G multiband communications. The LTE band 46 (5150–5925 MHz), though still unlicensed, is a potential and competitive spectrum to be utilized for 5G

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TABLE 1. Optimal physical dimensions of each antenna element.

Parameters	L ₁ (mm)	L ₂ (mm)	L ₃ (mm)	L ₄ (mm)	L ₅ (mm)	W ₁ (mm)	W ₂ (mm)	W ₃ (mm)	R ₁ (mm)	R ₂ (mm)
Ant 1/Ant 4/Ant 5/Ant 8	9.6	5.2	10.0	5.5	15.0	1.5	2.7	1.8	0.6	0.4
Ant 2/Ant 3/Ant 6/ Ant 7	10.0	6.3	10.0	4.3	15.0	1.5	2.7	1.8	0.5	0.4

MIMO applications [5]. The multimode resonator, multi-stub loaded resonator, and stepped impedance resonator (SIR) are promising candidates to develop the MIMO antenna array in mobile terminals. Due to the limited space, it is challenging to arrange a large number of antenna elements in a modern commonly-sized smartphones whilst maintain good isolation and radiation efficiency.

Various techniques have been investigated in the literature to realize the multiband MIMO antenna array of 5G mobile terminals [6]–[12]. The majority of these designs are based on multimode resonators or combination of different structures. An 8-antenna MIMO array has been analyzed to operate at the LTE band 42 and LTE band 46 [6]. Two distinct radiators, namely slot antenna and monopole antenna, are responsible for different bands. A return loss of higher than 6 dB and an envelope correlation coefficient (ECC) of smaller than 0.15 have been obtained. Furthermore, it achieves a high channel capacity. However, the vertically-oriented antenna elements increase the thickness and fabrication complexity. In [7], a triple band MIMO antenna array comprising three individual radiators, including an inverted π -shaped antenna, a longer L-shaped slot antenna, and a shorter L-shaped slot antenna, has been proposed. Each band is obtained by a different number of corresponding radiators. A return loss of higher than 6 dB and an ECC of less than 0.15 have been achieved.

In this paper, an 8-antenna MIMO array is proposed which operates at the LTE band 42 and LTE band 46 for future 5G mobile terminals. The MIMO antenna array is composed of eight slot antenna elements based on SIRs excited by shorted microstrip feed lines printed on the top side of the substrate. The slot antenna elements are realized in the ground plane printed on the bottom side of the substrate. The proposed MIMO antenna array has been simulated and validated by the measurements of a fabricated prototype. Good agreement between the measurement and simulation results has been observed, showing that the MIMO antenna array achieves good return loss, isolation, ECC, total efficiency, and channel capacity. The effects of the display panel on the performance of the MIMO antenna array have also been studied. Besides, the ECC and total efficiency of the MIMO antenna array also have been simulated and analyzed using CST Studio Suite 2016, in the presence of one hand, two hands and specific anthropomorphic mannequin (SAM) head.

II. DUAL-BAND EIGHT-ANTENNA MIMO ARRAY DESIGN

The geometrical configuration of the proposed dual-band 8-antenna MIMO array is given in Figure 1. The whole system is mounted on a FR4 ($\varepsilon_r = 4.4$, $\tan \delta = 0.025$) substrate with

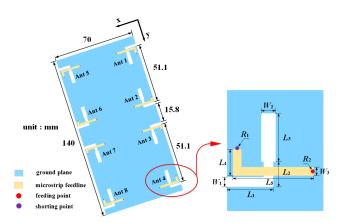


FIGURE 1. Geometrical configuration of the proposed 8-port dual-band MIMO antenna array.

a size of $140 \, \text{mm} \times 70 \, \text{mm} \times 1 \, \text{mm}$. Each slot antenna element based on a SIR is excited by a $50 \, \Omega$ microstrip line etched on the top side of the substrate. For each microstrip feed line, its one end is connected with a SMA connector and the other end is shorted to the ground to produce an inductance, and thus good impedance matching can be achieved.

According to [13], the multiband antenna can be attained by adjusting the impedance ratio and electrical length ratio of the SIR. With the purpose to obtain dual operation bands, the above two parameters need to be carefully selected. To simplify the calculation process and subsequent design, the electrical length ratio is assumed to be 1.0, the fundamental frequency is 3500 MHz, and the first harmonic frequency is 5500 MHz. Based on the above assumptions, the calculated impedance ratio is approximately 1.6. Hence, the slot lines with characteristic impedances of 40 Ω and 64 Ω , are used to accomplish the design. The lengths and positions of the microstrip feed lines can be adjusted, so that the designed MIMO antenna array can well operate in the required two frequency bands. Finally, the optimal physical dimensions for each antenna element are listed in Table 1.

III. SIMULATED RESULTS AND ANALYSIS

For the designed 8-port MIMO antenna array, the simulated *S*-parameters results and MIMO performance are presented in this section. Furthermore, the influences of the designed antenna array in the presence of user's hands and SAM head are also investigated and analyzed. Moreover, the effects of the display panel are also studied.

A. S-PARAMETERS AND EQUIVALENT SURFACE CURRENT DISTRIBUTIONS

The simulated S-parameters are shown in Figure 2. Note that only the simulated results of Ant 1 to Ant 4 are depicted due



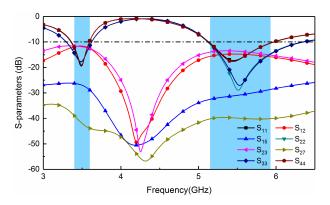


FIGURE 2. Simulated S-parameters of the MIMO antenna array.

to the symmetry of the designed MIMO antenna array. It can be obviously seen that the return loss of the designed 8-port MIMO antenna array are greater than 10 dB within the two frequency bands. Since the antenna elements along the same long side are close to each other, the mutual coupling, i.e. S_{12} and S_{23} , are relatively larger. However, the isolations are still greater than 11.8 dB which is acceptable. In comparison, the distances between the antenna elements on the same narrow side are further, and the mutual coupling, i.e. S_{15} , is therefore less than -25 dB across the two frequency bands. The mutual coupling between nonadjacent antenna elements are certainly rather weak.

To vividly illustrate the two resonant modes of the SIR-based slot antenna element, the simulated equivalent surface current distributions of Ant 1, respectively, at 3500 MHz and 5500 MHz are displayed in Figure 3. At 3500 MHz, it can be easily seen that the slot antenna operates at the fundamental quarter-wavelength resonant mode. In addition, two current nulls along the slot can be found when the antenna operates at 5500 MHz, i.e. the first harmonic resonant mode.

B. MIMO PERFORMANCE

The ECC defines the correlation level between the radiation patterns of any two antenna elements. It is the most important indicator to judge the performance such as diversity gain of a MIMO systems [14], [15]. The smaller the ECC is, the higher the diversity gain can be ensured. It is generally required that the ECC must be less than 0.5 to provide reasonable diversity gain [16].

The ECC between two antenna elements can be accurately calculated from their farfield radiation patterns using (1)

$$\rho_{ij} = \frac{\left| \iint_{4\pi} \overline{F}_i(\theta, \phi) \cdot \overline{F}_j^*(\theta, \phi) d\Omega \right|^2}{\iint_{4\pi} \left| \overline{F}_i(\theta, \phi) \right|^2 d\Omega \cdot \iint_{4\pi} \left| \overline{F}_j(\theta, \phi) \right|^2 d\Omega}$$
(1)

where ρ_{ij} denotes the ECC between element Ant i and Ant j, $\overline{F}_i(\theta, \phi)$ denotes the farfield radiation patterns of Ant i, and * denotes the Hermitian transpose of a matrix. Here, note that $\overline{F}_i(\theta, \phi)$ and $\overline{F}_j(\theta, \phi)$ are complex vectors. The ECC also

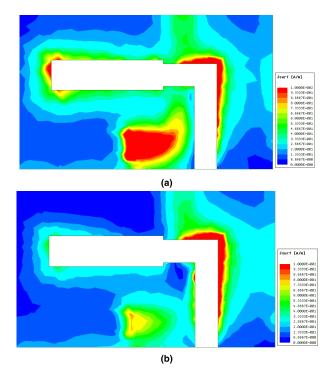


FIGURE 3. Simulated equivalent surface current distributions of Ant 1. (a) 3500 MHz. (b) 5500 MHz.

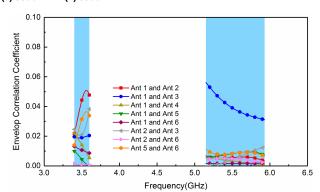


FIGURE 4. Simulated ECCs of the MIMO antenna array.

can be estimated from the S-parameters using (2)

$$\rho_{ij} = \frac{\left|S_{ii}^* S_{ij} + S_{ji}^* S_{jj}\right|^2}{\left(1 - \left(\left|S_{ii}\right|^2 + \left|S_{ji}\right|^2\right)\right) \left(1 - \left(\left|S_{jj}\right|^2 + \left|S_{ij}\right|^2\right)\right)} \tag{2}$$

where S_{ii} is the reflection coefficient of element Ant i, and S_{ij} ($i \neq j$) is the transmission coefficient between element Ant i and element Ant j.

The simulated ECCs calculated from the farfield patterns are presented in Figure 4. Within both frequency bands, the ECCs between any two neighbor antennas are smaller than 0.1, while those between two nonadjacent antennas are even less than 0.02. It is sufficient to prove that any two antenna elements of the proposed MIMO antenna array have a mall correlation level within the two frequency bands.

The channel capacity means the maximum achievable data transmission rate [17], [18]. For a MIMO antenna array,



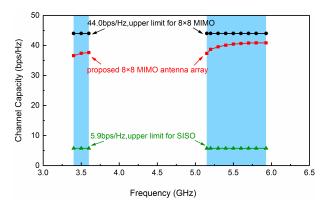


FIGURE 5. Simulated channel capacity when the SNR is 20 dB.

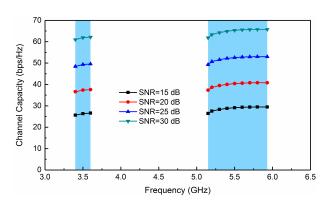


FIGURE 6. Simulated channel capacities with varying SNRs.

the larger the number of antenna elements is, normally the higher the channel capacity is. In general, the channel capacity of a MIMO antenna array can be calculated using (3) [19]

$$C = \log_2 \left[\det \left(I_N + \frac{SNR}{N} H H^* \right) \right]$$
 (3)

where C is the channel capacity, SNR is the signal-to-noise ratio, N is the number of the antenna elements, I_N denotes a $N \times N$ identity matrix, and H is the matrix of the channel capacity.

The simulated channel capacity of the 8-antenna MIMO array is depicted in Figure 5. The communication channel is assumed to be an independently and identically distributed channel with the Rayleigh fading environment [6]. Given the SNR is 20 dB, by averaging 1,00,000 random Rayleigh fading channel realizations within the two frequency bands [7], the channel capacity of an ideal single input and single output (SISO) system is 5.9 bps/Hz, and that of an ideal 8×8 MIMO system should be 44 bps/Hz. The channel capacity of the proposed 8-antenna MIMO array is more than 36.9 bps/Hz across both operation bands, i.e. only 7.1 bps/Hz less than the theoretical upper bound. Within the LTE band 46, the channel capacity is higher than 38.5 bps/Hz with a maximum of 40.9 bps/Hz. Moreover, the channel capacities with varying SNRs are shown in Figure 6. While the SNR is increased by 5 dB, the channel capacity is enhanced by 11 bps/Hz approximately.

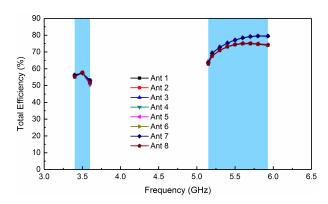


FIGURE 7. Simulated total efficiency of each antenna element.

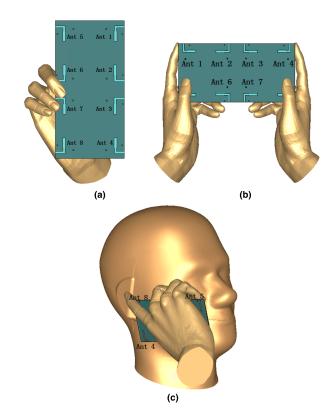


FIGURE 8. Three typical utility scenarios for mobile terminals. (a) Single hand. (b) Two hands. (c) SAM head and one hand.

C. TOTAL EFFICIENCY

The total efficiency of each antenna element of the proposed 8-antenna MIMO array is simulated and plotted in Figure 7. In the simulations, the matched 50 Ω loads are connected to other ports except the excited element. The total efficiencies are about from 51% to 59% and from 62% to 80% across the 3500-MHz band and 5500-MHz band, respectively.

D. EFFECTS OF SAM HEAD AND USER'S HANDS

Since the proposed 8-antenna MIMO array is to be utilized in 5G mobile terminals, the impacts of human head and

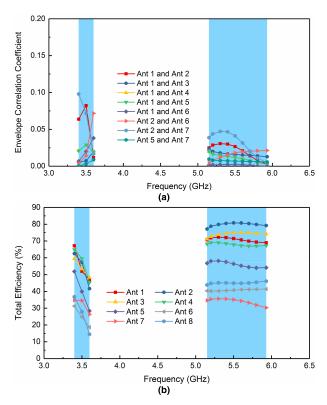
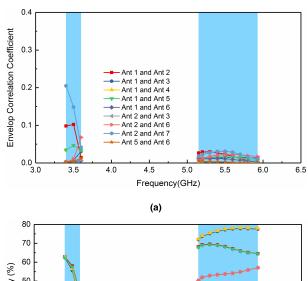


FIGURE 9. Simulated results of the proposed MIMO antenna array with single hand. (a) ECC. (b) Total efficiency.

hands on the antenna array need to be investigated. Therefore, the performance of the 8-antenna MIMO array in the presence of single hand, two hands, and SAM head have been analyzed and discussed, as illustrated in Figure 8.

At the single hand condition, i.e. the screen is positioned vertically, the involved simulated utility scenario is shown in Figure 8(a). As shown in Figure 9(a), the ECCs (ρ_{12} and ρ_{27}) increase within the LTE band 42, but still keep less than 0.1. Although the value is larger than that without the hand, it is rather small and acceptable. Yet, the total efficiencies of Ant 6 and Ant 8 close to the hand are reduced to 18%, which are nearly unusable at the LTE band 42. Hence, just as shown in Figure 9(b), the 8-antenna MIMO array is equivalent to a 6-antenna MIMO array applicable for 5G communications as well.

Furthermore, as shown in Figure 8(b), the screen is placed horizontally. At the two hands condition, Figure 10(a) shows that the ECCs (ρ_{12} and ρ_{27}) increase within the 3500-MHz band, and they are kept less than 0.2 and usable for mobile communication systems. By observing Figure 10(b) closely, the minimum total efficiencies of Ant 5 and Ant 8 close to the human hands are reduced to 13% at the 3500-MHz band, while other antennas can still work well with acceptable total efficiencies greater than 26%. Hence, at the reading condition, the designed 8 × 8 MIMO antenna array is equivalent to a 6 × 6 antenna array.



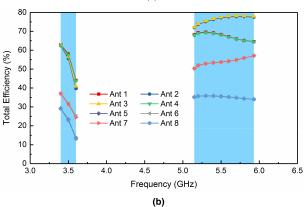


FIGURE 10. Simulated results of the proposed MIMO antenna array with two hands. (a) ECC. (b) Total efficiency.

Figure 8(c) presents the involved simulated utility scenario in the presence of the SAM head and single hand. Figure 11 plots the simulated ECC and total efficiency. It can be found that the ECCs increase within the two frequency bands, but maintain less than 0.3. Though larger than those of the above two utility scenarios, the ECC remains lower than 0.5 and acceptable. As presented in Figure 11(b), Ant i (i = 1, 2, 3, 4) can work well with the total efficiency of higher than 20% approximately within the two frequency bands. However, the total efficiencies of Ant i (i = 5, 6, 7, 8) close to the SAM head and hand diminish to 14%, which are almost unusable at the LTE band 42. Thus, the designed 8×8 MIMO antenna array is equivalent to a 4×4 MIMO antenna array, which can still be applied for 5G communications.

E. EFFECTS OF DISPLAY PANEL

Figure 12 shows the simulated S-parameters of the 8-antenna MIMO array with and without a common display screen. As the structure of the designed antenna array is symmetric, only the simulated results of Ant 1 and Ant 2 are given. As shown in Figure 12, the display panel is mounted on a grass ($\varepsilon_r = 5.5$) substrate with a size of 124 mm \times 62 mm \times 1 mm. It can be observed that there is a minor frequency shifting due to the existence of the display screen. However, we can still obtain a return loss of greater than 10 dB across the desired two frequency bands.



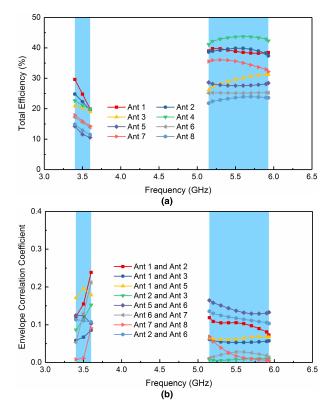


FIGURE 11. Simulated results of the proposed MIMO antenna array with SAM head and single head. (a) ECC. (b) Total efficiency.

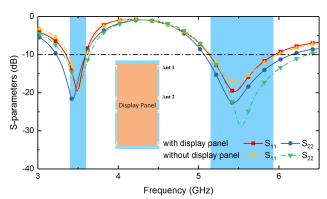


FIGURE 12. Simulated return loss of Ant1 and Ant2 with and without the display panel.

IV. EXPERIMENTAL RESULTS AND DISCUSSION

The manufactured 8-antenna MIMO array is given in Figure 13. The microstrip feed lines mounted on the top side of the substrate are shorted to the ground plane through vias. The *S*-parameters and farfield radiation patterns of each antenna element are measured in a microwave anechoic chamber.

Figure 14 compares the simulated and experimental return loss. Only those of Ant 1 and Ant 2 are given on account of the symmetric structure of the 8-antenna MIMO array. There exists subtle differences between the measured and simulated results because of the fabrication errors, SMA welding issues and environment effects. However, the measured results still have a good agreement with the simulated results. Compared

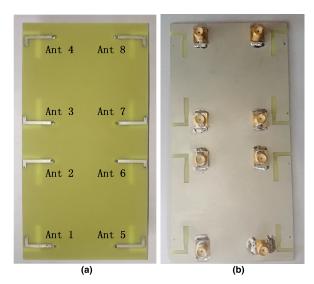


FIGURE 13. Prototype of the manufactured dual-band 8-antenna MIMO array. (a) Top view. (b) Bottom view.

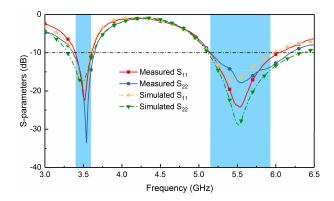


FIGURE 14. Simulated and experimental return loss of Ant1 and Ant2.

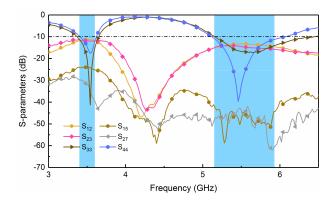


FIGURE 15. Experimental S-parameters of the manufactured MIMO antenna array.

with the simulated results, the performance of Ant 1 becomes slightly better within the LTE band 46. The measured 10-dB impedance bands are 220 MHz ranging from 3390 MHz to 3610 MHz and 920 MHz ranging from 5120 MHz to 6040 MHz.

The other experimental representative *S*-parameters of the manufactured 8-antenna MIMO array are given in Figure 15. Because of the structural symmetry, only the measured return

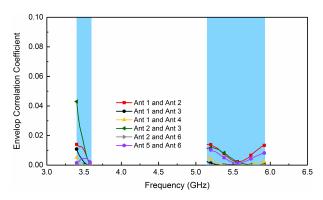


FIGURE 16. Experimental ECC results of the MIMO antenna array.

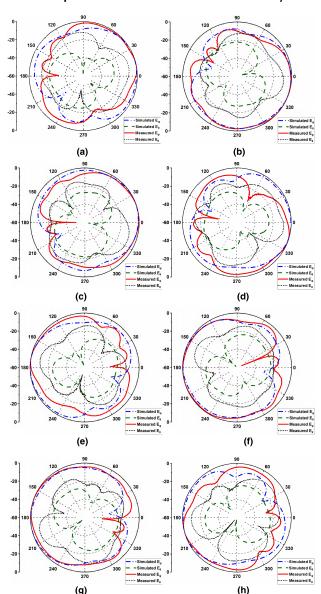


FIGURE 17. Normalized simulated and experimental farfield patterns at 3500 MHz in the x-y plane. (a) Ant1. (b) Ant 2. (c) Ant 3. (d) Ant 4. (e) Ant 5. (f) Ant 6. (g) Ant 7. (h) Ant 8.

loss of Ant i (i = 3, 4) are given. It shows that Ant 3 has a 10-dB impedance bandwidths of 7.4% (3380–3640 MHz) and

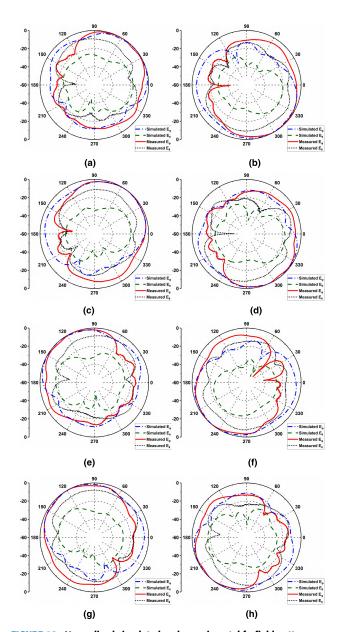


FIGURE 18. Normalized simulated and experimental farfield patterns 5500 MHz in the x-y plane. (a) Ant1. (b) Ant 2. (c) Ant 3. (d) Ant 4. (e) Ant 5. (f) Ant 6. (g) Ant 7. (h) Ant 8.

23.6% (5140–6440 MHz), and those of Ant 4 are 6.3% (3400 –3620 MHz) and 17.4% (5120–6080 MHz). Compared with the simulations, the experimental isolation (S_{12} and S_{23}) are relatively larger but better than 11.2 dB. In consistent with the simulations, the measured isolation between neighboring antennas on the narrow side, i.e. S_{15} , is relatively smaller that is greater than 23 dB within the 3500-MHz band and 5500-MHz band.

Using (2), the ECCs between arbitrary antenna elements are calculated, as depicted in Figure 16. The ECC is less than 0.1 within both frequency bands, validating that the proposed MIMO antenna array has good diversity performance.

The simulated and measured normalized farfield radiation patterns in the x-y plane are presented in Figures 17 and 18,



TABLE 2. Performance comparison between the proposed MIMO antenna array and reported ones.

Ref.	Bandwidth (MHz)	Dimensions (mm ³)	Isolation (dB)	ECC	No. of Ants.	Total Efficiency (%)
Proposed	3400-3600	$140 \times 70 \times 1$	≥11.2	<0.08	8	51—59 (lower band)
	5150-5925	140 / 70 / 1		_0.00		62—80 (upper band)
[3]	3400—3800	$150\times75\times0.8$	NG	≤0.5	8 and 16	30–52 (8 × 8) 30–52 (16 × 16)
[4]	3400—3600	$150\times75\times0.8$	≥12.5	≤0.2	12	55-70
[6]	3400—3600	$150 \times 75 \times 0.8$	≥12	≤0.1 (lower band) ≤0.04 (upper band)	8	50–56 (lower band)
	5150-5925	130 1173 110.0				53–65 (upper band)
[7]	3400—3800	$150 \times 80 \times 0.8$	≥12	≤0.15 (lower band)	8 (lower band)	41-82 (lower band)
	5150-5925	130 ~ 60 ~ 6.6		≤0.1 (upper band)	6 (upper band)	47–79 (upper band)
[9]	3400-3800	$150 \times 80 \times 0.8$	≥11	≤0.15 (lower band)	10	42–65 (lower band)
	5150-5925	150 60 ^ 0.6		≤0.05 (upper band)		62–82 (upper band)

respectively, at 3500 MHz and 5500 MHz. Apart from the antenna element under test, all others are connected to 50 Ω matched loads. It can be observed that the measured results achieve a good agreement with the simulated ones. Besides, it can be concluded that the measured E_{φ} (co-polarization) is 9 dB larger than E_{θ} (cross polarization) in the main E-plane. Therefore, the antenna elements can achieve a relatively pure horizontal polarization. In detail, the Ant i (i = 1, 2, 3, 4) mainly radiates towards the +x direction at 3500 MHz, while the maximum radiation direction of the others is generally in the -x direction. Similar radiation patterns can also be found at 5500 MHz. Therefore, the proposed MIMO antenna array exhibits a good omnidirectional property, demonstrating the feasibility for 5G mobile terminal applications.

Table II summarizes the performance comparison between the proposed MIMO antenna array and the previous reported ones. When compared with those in [3] and [4], the proposed 8-antenna MIMO array is able to cover both LTE band 42 and LTE band 46. In addition, the proposed MIMO antenna array has the lowest ECC. It also owns the best impedance matching with a return loss of larger than 10 dB across the entire desired frequency bands. Furthermore, the proposed MIMO antenna array has higher total efficiencies. Therefore, the proposed MIMO antenna array generally outperforms the reported ones.

V. CONCLUSION

This paper investigates thoroughly a dual-band eight-antenna array based on slot SIRs for future 5G MIMO applications. It realizes a return loss of higher than 10 dB, an inter-element isolation of greater than 11.2 dB, an ECC of lower than 0.1, and a total efficiency of more than 51% within the LTE band 42 and LTE band 46 for each antenna element. The measured results well agree with the simulations. The channel capacity is expected to be as high as 36.9–37.9 bps/Hz and 38.5–40.9 bps/Hz within the two frequency bands, respectively. A low ECC and good MIMO performance also can be maintained in the presence of the SAM head and

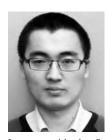
human hands. Hence, the proposed 8×8 MIMO antenna array can be a promising candidate for 5G multiband MIMO applications in mobile terminals.

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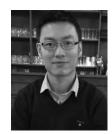
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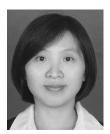


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