

Received February 22, 2019, accepted March 5, 2019, date of publication March 18, 2019, date of current version April 11, 2019. *Digital Object Identifier* 10.1109/ACCESS.2019.2905576

# Maximum Likelihood Recursive Identification for the Multivariate Equation-Error Autoregressive Moving Average Systems Using the Data Filtering

## LIJUAN LIU<sup>®1</sup>, FENG DING<sup>®1,2,3,4</sup>, LING XU<sup>1</sup>, JIAN PAN<sup>3</sup>, AHMED ALSAEDI<sup>4</sup>, AND TASAWAR HAYAT<sup>4</sup>

<sup>1</sup>Key Laboratory of Advanced Process Control for Light Industry, Ministry of Education, School of Internet of Things Engineering, Jiangnan University, Wuxi 214122, China

<sup>2</sup>College of Automation and Electronic Engineering, Qingdao University of Science and Technology, Qingdao 266061, China <sup>3</sup>School of Electrical and Electronic Engineering, Hubei University of Technology, Wuhan 430068, China

Corresponding author: Feng Ding (fding@jiangnan.edu.cn)

This work was supported in part by the National Natural Science Foundation of China under Grant 61873111 and Grant 61803183, in part by the 111 Project under Grant B12018, in part by the Graduate Education Innovation Program of Jiangsu Province under Grant KYCX17\_1458, and in part by the Natural Science Foundation of Jiangsu Province under Grant BK20180591.

**ABSTRACT** The maximum likelihood principle has wide applications in system identification. This paper studies the maximum likelihood identification problems of the multivariate equation-error systems with colored noise. The system is broken down into several subsystems based on the number of the outputs. The key is to transform the subsystem into a controlled autoregressive moving average model and a noise model. Based on the maximum likelihood principle and the data filtering technique, a filtering-based maximum likelihood recursive generalized extended least squares algorithm is presented for estimating the parameters of these two models. For comparison, a maximum likelihood recursive generalized extended least squares algorithm is presented. Finally, the simulation example results confirm the effectiveness of the two algorithms.

**INDEX TERMS** Parameter estimation, maximum likelihood, data filtering, multivariate system.

### I. INTRODUCTION

For the actual control systems, system modeling and model identification are the basis of all control problems. Parameter estimation methods can be applied to many areas [1]–[4]. Parameter estimation is the eternal theme of the identification field and has wide applications in one-dimensional and multidimensional signal processing and filtering. The research on model uncertainty has become a hot topic. Recently, a family of robust filtering approaches under model uncertainty have been introduced [5] and the robust estimation problem of a signal given noisy observations is dealt with [6]. Multivariable systems are widely used in practical control processes because of their com-

plex structures and uncertain interference. Consequently, the parameter estimation problem of different systems has attracted a lot of attention [7]–[10]. Some examples of applications include fermentation processes [11] and distillation columns [12]. Wu [13] presented a model predictive control based proportional-integral-derivative controller for the multivariable process in the distillation column. Li *et al.* [14] provided an adaptive control method to deal with the tracking control problems for the nonlinear MIMO time-varying delay systems.

The maximum likelihood principle is widely used in the field of system identification because of good statistical properties [15]. In recent years, many maximum likelihood identification algorithms have been proposed [16], [17]. For instance, Wang *et al.* [18] focused on the robust Chinese remainder theorem problem for real numbers and derived

2169-3536 © 2019 IEEE. Translations and content mining are permitted for academic research only. Personal use is also permitted, but republication/redistribution requires IEEE permission. See http://www.ieee.org/publications\_standards/publications/rights/index.html for more information.

<sup>&</sup>lt;sup>4</sup>Department of Mathematics, King Abdulaziz University, Jeddah 21589, Saudi Arabia

The associate editor coordinating the review of this manuscript and approving it for publication was Kan Liu.

a maximum likelihood estimation based robust remainder theorem algorithm. Chen and Ding [19] considered the parameter estimation problems of the multiple-input singleoutput systems and delivered a filtering based maximum likelihood recursive least squares algorithm by using the data filtering. Söderström and Soverini [20] dealt with the identification problems of the errors-in-variables models and developed a frequency domain maximum likelihood estimator. Chang and Chang [21] extended the maximum likelihood principle to the multiple-input multiple-output systems and presented a novel maximum likelihood detection algorithm which does not need QR decomposition and matrix inversion.

The data filtering technique is an effective way to solve the parameter estimation problem of systems with colored noise [22]–[24]. The main idea is to transform the system into a filtered system and a filtered noise system. The data filtering method in system identification only changes the structure of the system and does not change the relationship between the inputs and outputs [25]-[27]. In this field, Li and Liu [28] extended the filtering technique to the bilinear systems with colored noises and proposed a maximum likelihood least squares based iterative algorithm . Ding et al. [29] derived a filtering and auxiliary model based recursive least squares algorithm for the dual-rate state space systems with timedelay based on the data filtering and the auxiliary identification idea. In order to improve the estimation accuracy, Pan et al. [30] presented a filtering based multi-innovation extended stochastic gradient algorithm for multivariate moving average systems by using the filtering technique and the multi-innovation identification theory.

For the multivariate equation-error systems with autoregressive noise, Liu *et al.* [31] proposed the maximum likelihood recursive extended least squares algorithm to obtain the parameter estimates . On the basis of the work in [31], this paper considers the parameter estimation problem for the multivariate equation-error systems with autoregressive moving average noise. The main contributions of this paper are as follows.

- A filtering based maximum likelihood recursive generalized extended least squares algorithm is derived for the multivariate equation-error system with autoregressive moving average noise by using the data filtering technique and the maximum likelihood principle.
- The system is broken down into several subsystems. The identification model of the subsystem is obtained by defining the parameter vectors and the information vectors of the subsystem.
- The simulation example proved the effectiveness of the filtering based maximum likelihood recursive generalized extended least squares algorithm.

Briefly, the rest of this paper is organized as follows. The multivariate equation-error autoregressive moving average system is broken down into several subsystems and the identification model of the subsystem is given in Section II. In Sections III and IV, a maximum likelihood recursive generalized extended least squares algorithm and a filtering based maximum likelihood recursive generalized extended least squares algorithm are presented for the multivariate equationerror autoregressive moving average systems. In addition, Section V offers the numerical simulation. Some concluding remarks are given in Section VI.

#### **II. SYSTEM DESCRIPTION**

Let us introduce some notation. The symbol  $I_m$  is an  $m \times m$  identity matrix;  $1_n$  is an *n*-dimensional column vector whose elements are 1; the superscript T denotes the matrix transpose; *z* represents unit forward shift operator:  $z\mathbf{x}(r) = \mathbf{x}(r+1)$  and  $z^{-1}\mathbf{x}(r) = \mathbf{x}(r-1)$ .



**FIGURE 1.** The multivariate equation-error autoregressive moving average system.

The multivariate equation-error autoregressive moving average system in Figure 1 can be expressed as follows,

$$\boldsymbol{H}(\boldsymbol{z})\boldsymbol{y}(\boldsymbol{r}) = \boldsymbol{\Phi}_{\boldsymbol{s}}(\boldsymbol{r})\boldsymbol{\theta} + \frac{K(\boldsymbol{z})}{E(\boldsymbol{z})}\boldsymbol{v}(\boldsymbol{r}), \tag{1}$$

where  $\boldsymbol{\Phi}_{s}(r) \in \mathbb{R}^{m \times n}$  is the measured information matrix which contains the input-output data,  $\boldsymbol{y}(r) := [y_{1}(r), y_{2}(r), \cdots, y_{m}(r)]^{\mathsf{T}} \in \mathbb{R}^{m}$  is the output vector of the system,  $\boldsymbol{v}(r) := [v_{1}(r), v_{2}(r), \cdots, v_{m}(r)]^{\mathsf{T}} \in \mathbb{R}^{m}$  is a white noise vector with zero mean and variance  $\sigma_{j}^{2}$  for  $v_{j}(r), j = 1, 2, \cdots, m, \boldsymbol{\theta} \in \mathbb{R}^{n}$  is the parameter vector to be identified, and  $\boldsymbol{H}(z), \boldsymbol{E}(z)$  and  $\boldsymbol{K}(z)$  are polynomials (matrix) in the unit backward shift operator  $z^{-1}$  ( $z^{-1}\boldsymbol{y}(r) = \boldsymbol{y}(r-1)$ ):

$$H(z) := I_m + H_1 z^{-1} + H_2 z^{-2} + \dots + H_{n_h} z^{-n_h},$$
  

$$E(z) := 1 + e_1 z^{-1} + e_2 z^{-2} + \dots + e_{n_e} z^{-n_e}, \quad e_i \in \mathbb{R},$$
  

$$K(z) := 1 + k_1 z^{-1} + k_2 z^{-2} + \dots + k_{n_k} z^{-n_k}, \quad k_i \in \mathbb{R}.$$

Let  $w(r) := \frac{K(z)}{E(z)}v(r) \in \mathbb{R}^m$  be the intermediate vector. Let  $\boldsymbol{\Phi}_s(r) := [\boldsymbol{\phi}_1(r), \boldsymbol{\phi}_2(r), \cdots, \boldsymbol{\phi}_m(r)]^{\mathsf{T}} \in \mathbb{R}^{m \times n}$ , where  $\boldsymbol{\phi}_j(r) \in \mathbb{R}^n$ . Let  $\boldsymbol{H}_i := [\boldsymbol{h}_{i1}^{\mathsf{T}}, \boldsymbol{h}_{i2}^{\mathsf{T}}, \cdots, \boldsymbol{h}_{im}^{\mathsf{T}}]^{\mathsf{T}} \in \mathbb{R}^{m \times m}$ , where  $\boldsymbol{h}_{ij} \in \mathbb{R}^{1 \times m}$  is the *j*th row values of  $\boldsymbol{H}_i$ . From (1), we have

$$\begin{bmatrix} y_{1}(r) \\ y_{2}(r) \\ \vdots \\ y_{m}(r) \end{bmatrix} + \begin{bmatrix} \mathbf{h}_{11} \\ \mathbf{h}_{12} \\ \vdots \\ \mathbf{h}_{1m} \end{bmatrix} z^{-1} \mathbf{y}(r) + \begin{bmatrix} \mathbf{h}_{21} \\ \mathbf{h}_{22} \\ \vdots \\ \mathbf{h}_{2m} \end{bmatrix} z^{-2} \mathbf{y}(r) + \cdots$$
$$+ \begin{bmatrix} \mathbf{h}_{nh^{1}} \\ \mathbf{h}_{nh^{2}} \\ \vdots \\ \mathbf{h}_{nh^{m}} \end{bmatrix} z^{-n_{h}} \mathbf{y}(r) = \begin{bmatrix} \boldsymbol{\phi}_{1}^{\mathrm{T}}(r) \\ \boldsymbol{\phi}_{2}^{\mathrm{T}}(r) \\ \vdots \\ \boldsymbol{\phi}_{m}^{\mathrm{T}}(r) \end{bmatrix} \boldsymbol{\theta} + \frac{K(z)}{E(z)} \begin{bmatrix} v_{1}(r) \\ v_{2}(r) \\ \vdots \\ v_{m}(r) \end{bmatrix}.$$

Afterwards, System (1) can be broken down into m subsystems. Subsystem j can be expressed as

$$\boldsymbol{H}_{j}(\boldsymbol{z})\boldsymbol{y}(r) = \boldsymbol{\phi}_{j}^{\mathrm{T}}(r)\boldsymbol{\theta} + \frac{K(\boldsymbol{z})}{E(\boldsymbol{z})}v_{j}(r), \quad j = 1, 2, \cdots, m, \qquad (2)$$

where  $\boldsymbol{H}_{j}(z) := \boldsymbol{h}_{0j} + \boldsymbol{h}_{1j}z^{-1} + \cdots + \boldsymbol{h}_{n_{h}j}z^{-n_{h}} \in \mathbb{R}^{1 \times m}$  is the *j*th row of  $\boldsymbol{H}(z)$ .

Then we have

 $\mathbf{T}(\mathbf{x})$ 

$$v_j(r) = \frac{E(z)}{K(z)} [\boldsymbol{H}_j(z) \boldsymbol{y}(r) - \boldsymbol{\phi}_j^{\mathrm{T}}(r) \boldsymbol{\theta}], \quad j = 1, 2, \cdots, m.$$

Define the subsystem parameter vectors  $\boldsymbol{h}_j \in \mathbb{R}^{mn_h}$ ,  $\boldsymbol{e} \in \mathbb{R}^{n_e}$ and  $\boldsymbol{k} \in \mathbb{R}^{n_k}$  as

$$\boldsymbol{h}_{j} := \begin{bmatrix} \boldsymbol{h}_{1j} \\ \boldsymbol{h}_{2j} \\ \vdots \\ \boldsymbol{h}_{nhj} \end{bmatrix}, \quad \boldsymbol{e} := \begin{bmatrix} e_{1} \\ e_{2} \\ \vdots \\ e_{ne} \end{bmatrix}, \quad \boldsymbol{k} := \begin{bmatrix} k_{1} \\ k_{2} \\ \vdots \\ k_{nk} \end{bmatrix}.$$

Define the subsystem parameter vectors and the information vectors as

$$\begin{split} \boldsymbol{\eta}_j &:= [\boldsymbol{\theta}^{\mathrm{T}}, \boldsymbol{h}_j^{\mathrm{T}}]^{\mathrm{T}} \in \mathbb{R}^{n+mn_h},\\ \boldsymbol{\vartheta}_j &:= [\boldsymbol{\eta}_j^{\mathrm{T}}, \boldsymbol{e}^{\mathrm{T}}, \boldsymbol{k}^{\mathrm{T}}]^{\mathrm{T}} \in \mathbb{R}^{n_0}, \quad n_0 := n + mn_h + n_e + n_k,\\ \boldsymbol{\psi}_j(r) &:= [\boldsymbol{\phi}_j^{\mathrm{T}}(r), -\boldsymbol{y}^{\mathrm{T}}(r-1), -\boldsymbol{y}^{\mathrm{T}}(r-2), \cdots, \\ &- \boldsymbol{y}^{\mathrm{T}}(r-n_h)]^{\mathrm{T}} \in \mathbb{R}^{n+mn_h},\\ \boldsymbol{\varphi}_j(r) &:= [\boldsymbol{\psi}_j^{\mathrm{T}}(r), -w_j(r-1), -w_j(r-2), \cdots, -w_j(r-n_e)] \end{split}$$

$$v_i(r-1), v_i(r-2), \cdots, v_i(r-n_k)]^{\mathrm{T}} \in \mathbb{R}^{n_0}.$$

From (2), we can get

$$y_j(r) = \boldsymbol{\varphi}_j^{\mathrm{T}}(r)\boldsymbol{\vartheta}_j + v_j(r), \qquad (3)$$

$$w_j(r) = y_j(r) - \boldsymbol{\psi}_j^{\mathrm{T}}(r)\boldsymbol{\eta}_j, \quad j = 1, 2, \cdots, m.$$
(4)

*Remark 1:* The system is broken down into m subsystems based on the number of the system outputs. The identification model of Subsystem j is shown in Equations (3) and (4).

*Remark 2:* The objective of this paper is to present maximum likelihood recursive identification algorithms to estimate the unknown parameter vectors  $\boldsymbol{\theta}$ ,  $\boldsymbol{e}$  and  $\boldsymbol{k}$  and the unknown parameter matrix  $\boldsymbol{H}_i$ .

## III. THE MAXIMUM LIKELIHOOD RECURSIVE GENERALIZED EXTENDED LEAST SQUARES ALGORITHM

According to the maximum likelihood principle, the maximum likelihood criterion function is obtained as follows [32],

$$J_j(\boldsymbol{\vartheta}_j)\Big|_{\boldsymbol{\vartheta}_{jML}} = \frac{1}{2} \sum_{r=1}^N v_j^2(r)\Big|_{\boldsymbol{\vartheta}_{jML}} = \min, \qquad (5)$$

subject to

$$v_j(r) = \frac{E(z)}{K(z)} [\boldsymbol{H}_j(z) \boldsymbol{y}(r) - \boldsymbol{\phi}_j^{\mathrm{T}}(r) \boldsymbol{\theta}]$$

The research [31] derived the maximum likelihood recursive extended least squares algorithm for the multivariate equation-error systems with autoregressive noise. Referring to the method in [31], minimizing the cost function  $J_j(\vartheta_j)$ , we can obtain the maximum likelihood recursive generalized extended least squares (ML-RGELS) algorithm:

$$\hat{\boldsymbol{\vartheta}}_{j}(r) = \hat{\boldsymbol{\vartheta}}_{j}(r-1) + \boldsymbol{L}_{j}(r)[\boldsymbol{y}_{j}(r) - \hat{\boldsymbol{\varphi}}_{j}^{\mathrm{T}}(r)\hat{\boldsymbol{\vartheta}}_{j}(r-1)], \quad (6)$$

$$\boldsymbol{L}_{j}(r) = \boldsymbol{P}_{j}(r-1)\hat{\boldsymbol{\varphi}}_{jf}(r)[1 + \hat{\boldsymbol{\varphi}}_{jf}^{\mathrm{T}}(r)\boldsymbol{P}_{j}(r-1)\hat{\boldsymbol{\varphi}}_{jf}(r)]^{-1}, \quad (7)$$

$$\boldsymbol{P}_{j}(r) = [\boldsymbol{I} - \boldsymbol{L}_{j}(r)\hat{\boldsymbol{\varphi}}_{jf}^{^{\mathrm{T}}}(r)]\boldsymbol{P}_{j}(r-1), \qquad (8)$$

$$\hat{\boldsymbol{\varphi}}_{jf}(r) = [\hat{\boldsymbol{\phi}}_{jf}^{^{\mathrm{T}}}(r), -\hat{\boldsymbol{y}}_{f}^{^{\mathrm{T}}}(r-1), -\hat{\boldsymbol{y}}_{f}^{^{\mathrm{T}}}(r-2), \cdots, \\ -\hat{\boldsymbol{y}}_{f}^{^{\mathrm{T}}}(r-n_{h}), -\hat{w}_{jf}(r-1), \\ -\hat{w}_{jf}(r-2), \cdots, -\hat{w}_{jf}(r-n_{e}), \\ \hat{v}_{jf}(r-1), \hat{v}_{jf}(r-2), \cdots, \hat{v}_{jf}(r-n_{k})]^{^{\mathrm{T}}}, \qquad (9)$$

$$\boldsymbol{\psi}_{j}(r) = [\boldsymbol{\phi}_{j}^{^{\mathrm{T}}}(r), -\boldsymbol{y}^{^{\mathrm{T}}}(r-1), -\boldsymbol{y}^{^{\mathrm{T}}}(r-2), \cdots, \\ -\boldsymbol{v}_{j}^{^{\mathrm{T}}}(r-n_{k})]^{^{\mathrm{T}}}. \qquad (10)$$

$$\hat{\boldsymbol{\varphi}}_{j}(r) = [\boldsymbol{\psi}_{j}^{\mathsf{T}}(r), -\hat{w}_{j}(r-1), -\hat{w}_{j}(r-2), \cdots, \\ -\hat{w}_{j}(r-n_{e}), \hat{v}_{j}(r-1),$$
(10)

$$\hat{v}_j(r-2), \cdots, \hat{v}_j(r-n_k)]^{\mathrm{T}},$$
 (11)

$$\hat{\phi}_{jf}(r) = \phi_j(r) + \sum_{i=1}^{n} \hat{e}_i(r-1)\phi_j(r-i) - \sum_{i=1}^{n_k} \hat{k}_i(r-1)\hat{\phi}_{jf}(r-i),$$
(12)

$$\hat{\mathbf{y}}_{f}(r) = \mathbf{y}(r) + \sum_{i=1}^{n_{e}} \hat{e}_{i}(r)\mathbf{y}(r-i) - \sum_{i=1}^{n_{k}} \hat{k}_{i}(r)\hat{\mathbf{y}}_{f}(r-i),$$
(13)

$$\hat{w}_{jf}(r) = y_j(r) + \sum_{i=1}^{n_n} \hat{h}_{ij}(r) \mathbf{y}(r-i) - \boldsymbol{\phi}_j^{\mathsf{T}}(r) \hat{\boldsymbol{\theta}}(r) - \sum_{i=1}^{n_k} \hat{k}_i(r) \hat{w}_{jf}(r-i),$$
(14)

$$\hat{v}_{jf}(r) = v_j(r) - \sum_{i=1}^{n_k} \hat{k}_i(r) \hat{v}_{jf}(r-i),$$
(15)

$$\hat{\boldsymbol{\vartheta}}_{j}(r) = [\hat{\boldsymbol{\eta}}_{j}^{\mathrm{T}}(r), \hat{\boldsymbol{e}}_{j}^{\mathrm{T}}(r), \hat{\boldsymbol{k}}_{j}^{\mathrm{T}}(r)]^{\mathrm{T}},$$
(16)

$$\hat{w}_j(r) = y_j(r) - \boldsymbol{\psi}_j^{\mathrm{T}}(r)\hat{\boldsymbol{\eta}}_j(r), \qquad (17)$$

$$\hat{\nu}_j(r) = y_j(r) - \hat{\boldsymbol{\varphi}}_j^{\mathrm{T}}(r)\hat{\boldsymbol{\vartheta}}_j(r), \qquad (18)$$

$$\hat{\boldsymbol{\theta}}(r) = \frac{\boldsymbol{\theta}_1(r) + \boldsymbol{\theta}_2(r) + \dots + \boldsymbol{\theta}_m(r)}{m},$$
(19)

$$\hat{e}(r) = \frac{\hat{e}_1(r) + \hat{e}_2(r) + \dots + \hat{e}_m(r)}{m},$$
 (20)

$$\hat{k}(r) = \frac{\hat{k}_1(r) + \hat{k}_2(r) + \dots + \hat{k}_m(r)}{m},$$
(21)

$$\hat{\boldsymbol{\vartheta}}(r) = [\hat{\boldsymbol{\theta}}^{\mathrm{T}}(r), \hat{\boldsymbol{h}}_{1}^{\mathrm{T}}(r), \hat{\boldsymbol{h}}_{2}^{\mathrm{T}}(r), \cdots, \hat{\boldsymbol{h}}_{m}^{\mathrm{T}}(r), \hat{\boldsymbol{e}}^{\mathrm{T}}(r), \hat{\boldsymbol{k}}^{\mathrm{T}}(r)]^{\mathrm{T}}.$$
(22)

The flowchart of the ML-RGELS algorithm for computing  $\hat{\vartheta}(r)$  is shown in Figure 2.





## IV. THE FILTERING BASED MAXIMUM LIKELIHOOD RECURSIVE GENERALIZED EXTENDED LEAST SQUARES ALGORITHM

The data filtering technology is an effective method to solve the parameter estimation problems of the colored noise systems. A filtering based maximum likelihood recursive generalized extended least squares algorithm is derived in this section. Through the data filtering, the subsystem can be transformed into a controlled autoregressive moving average model and a noise model.

Define

$$\boldsymbol{\phi}_{1j}(r) := E(z)\boldsymbol{\phi}_j^{\mathrm{T}}(r) \in \mathbb{R}^n,$$
  
$$\boldsymbol{y}_1(r) := E(z)\boldsymbol{y}(r) \in \mathbb{R}^m.$$

Multiplying both sides of Equation (2) by E(z) yields

$$\boldsymbol{H}_{j}(z)\boldsymbol{E}(z)\boldsymbol{y}(r) = \boldsymbol{E}(z)\boldsymbol{\phi}_{j}^{\mathrm{T}}(r)\boldsymbol{\theta} + \boldsymbol{K}(z)\boldsymbol{v}_{j}(r).$$
(23)

Then Equation (23) can be written as

$$\boldsymbol{H}_{j}(z)\boldsymbol{y}_{1}(r) = \boldsymbol{\phi}_{1j}(r)\boldsymbol{\theta} + K(z)v_{j}(r).$$
(24)

This is the controlled autoregressive moving average model. Define the parameter vector  $\boldsymbol{\vartheta}_{1j}$  and the information vector  $\boldsymbol{\varphi}_{1j}(r)$ :

$$\boldsymbol{\vartheta}_{1j} := [\boldsymbol{\theta}^{\mathsf{T}}, \boldsymbol{h}_{1j}, \boldsymbol{h}_{2j}, \cdots, \boldsymbol{h}_{n_h j}, k_1, k_2, \cdots, k_{n_k}]^{\mathsf{T}} \in \mathbb{R}^{n_1},$$
  

$$n_1 := n + mn_h + n_k,$$
  

$$\boldsymbol{\varphi}_{1j}(r) := [\boldsymbol{\phi}_{1j}^{\mathsf{T}}(r), -\boldsymbol{y}_1^{\mathsf{T}}(r-1), -\boldsymbol{y}_1^{\mathsf{T}}(r-2), \cdots,$$
  

$$-\boldsymbol{y}_1^{\mathsf{T}}(r-n_h), v_j(r-1),$$
  

$$v_j(r-2), \cdots, v_j(r-n_k)]^{\mathsf{T}} \in \mathbb{R}^{n_1}.$$

Then we can get the identification model of this model:

$$y_{1j}(r) = \boldsymbol{\varphi}_{1j}^{\mathrm{T}}(r)\boldsymbol{\vartheta}_{1j} + v_j(r).$$
<sup>(25)</sup>

Define the intermediate variable

$$w_j(r) := \frac{K(z)}{E(z)} v_j(r).$$
(26)

Define the information vectors  $\boldsymbol{\varphi}_{ej}(r)$  and  $\boldsymbol{\varphi}_{kj}(r)$ :

$$\varphi_{ej}(r) := [-w_j(r-1), -w_j(r-2), \cdots, \\ -w_j(r-n_e)]^{\mathsf{T}} \in \mathbb{R}^{n_e},$$

$$\varphi_{kj}(r) := [v_j(r-1), v_j(r-2), \cdots, v_j(r-n_k)]^{\mathsf{T}} \in \mathbb{R}^{n_k}.$$

Then Equation (26) can be written as

$$w_j(r) = [1 - E(z)]w_j(r) + K(z)v_j(r)$$
  
=  $\boldsymbol{\varphi}_{ej}^{\mathrm{T}}(r)\boldsymbol{e} + \boldsymbol{\varphi}_{kj}^{\mathrm{T}}(r)\boldsymbol{k} + v_j(r).$  (27)

Define and minimize the criterion function

$$J_1(\boldsymbol{\vartheta}_{1j}) := \frac{1}{2} \sum_{r=1}^N v_j^2(r) \Big|_{\boldsymbol{\vartheta}_{1jML}} = \min$$

subject to

$$v_j(r) = \frac{1}{K(z)} [\boldsymbol{H}_j(z) \boldsymbol{y}_1(r) - \boldsymbol{\phi}_{1j}(r) \boldsymbol{\theta}].$$
(28)

Let  $\hat{h}_j(r)$ ,  $\hat{e}(r)$ ,  $\hat{k}(r)$  and  $\hat{\vartheta}_{1j}(r)$  be the estimates of  $h_j$ , e, k and  $\vartheta_{1j}$  at time r, respectively,

$$\hat{\boldsymbol{h}}_{j}(r) := [\hat{\boldsymbol{h}}_{1j}(r), \hat{\boldsymbol{h}}_{2j}(r), \cdots, \hat{\boldsymbol{h}}_{n_{h}j}(r)]^{\mathsf{T}}, \hat{\boldsymbol{e}}(r) := [\hat{e}_{1}(r), \hat{e}_{2}(r), \cdots, \hat{e}_{n_{e}}(r)]^{\mathsf{T}}, \hat{\boldsymbol{k}}(r) := [\hat{k}_{1}(r), \hat{k}_{2}(r), \cdots, \hat{k}_{n_{k}}(r)]^{\mathsf{T}}, \hat{\boldsymbol{\vartheta}}_{1i}(r) := [\hat{\boldsymbol{\theta}}^{\mathsf{T}}(r), \hat{\boldsymbol{h}}^{\mathsf{T}}_{i}(r), \hat{\boldsymbol{k}}^{\mathsf{T}}(r)]^{\mathsf{T}}.$$

Then the estimates of E(z) and F(z) at time *r* can be written as

$$\hat{E}(r,z) := 1 + \hat{e}_1(r)z^{-1} + \hat{e}_2(r)z^{-2} + \dots + \hat{e}_{n_e}(r)z^{-n_e},$$
  
$$\hat{K}(r,z) := 1 + \hat{k}_1(r)z^{-1} + \hat{k}_2(r)z^{-2} + \dots + \hat{k}_{n_k}(r)z^{-n_k}.$$

Furthermore, the estimates  $\hat{\phi}_{1j}(r)$  and  $\hat{y}_1(r)$  of  $\phi_{1j}(r)$  and  $y_1(r)$  can be obtained by

$$\hat{\phi}_{1j}(r) = \hat{E}(r-1, z)\phi_j^{\mathsf{T}}(r) 
= \phi_j^{\mathsf{T}}(r) + \hat{e}_1(r-1)\phi_j^{\mathsf{T}}(r-1) + \cdots 
+ \hat{e}_{n_e}(r-1)\phi_j^{\mathsf{T}}(r-n_e), 
\hat{y}_1(r) = \hat{E}(r-1, z)y(r) 
= y(r) + \hat{e}_1(r-1)y(r-1) + \cdots 
+ \hat{e}_{n_e}(r-1)y(r-n_e).$$

Define the filtered information vector

$$\hat{\boldsymbol{\varphi}}_{1jf}(r) := -\frac{\partial v_j(r)}{\partial \boldsymbol{\vartheta}_{1j}} \bigg|_{\boldsymbol{\vartheta}_{1j}(r-1)} \\ = -\bigg[\frac{\partial v_j(r)}{\partial \boldsymbol{\theta}}, \frac{\partial v_j(r)}{\partial \boldsymbol{h}_{1j}}, \frac{\partial v_j(r)}{\partial \boldsymbol{h}_{2j}}, \cdots, \frac{\partial v_j(r)}{\partial \boldsymbol{h}_{nhj}}, \\ \frac{\partial v_j(r)}{\partial k_1}, \frac{\partial v_j(r)}{\partial k_2}, \cdots, \frac{\partial v_j(r)}{\partial k_{n_k}}\bigg]_{\boldsymbol{\vartheta}_{1j}(r-1)}^{\mathrm{T}}.$$
(29)

From (28), the filtering values are calculated as follows:

$$\begin{aligned} \frac{\partial v_j(r)}{\partial \boldsymbol{\theta}} \bigg|_{\hat{\boldsymbol{\vartheta}}_j(r-1)} &= -[\hat{K}(r-1,z)]^{-1} \hat{\boldsymbol{\phi}}_{1j}(r) = -\hat{\boldsymbol{\phi}}_{jf}(r), \\ \frac{\partial v_j(r)}{\partial \boldsymbol{h}_{ij}} \bigg|_{\hat{\boldsymbol{\vartheta}}_j(r-1)} &= [\hat{K}(r-1,z)]^{-1} \hat{\boldsymbol{y}}_1(r-i) = z^{-i} \hat{\boldsymbol{y}}_f(r), \\ \frac{\partial v_j(r)}{\partial k_i} \bigg|_{\hat{\boldsymbol{\vartheta}}_j(r-1)} &= -[\hat{K}(r-1,z)]^{-1} \hat{v}_j(r-i) = -z^{-i} \hat{v}_{jf}(r), \end{aligned}$$

where  $\hat{\boldsymbol{\phi}}_{if}(r)$ ,  $\hat{\boldsymbol{y}}_{f}(r)$  and  $\hat{v}_{if}(r)$  can be computed by

$$\begin{aligned} \hat{\phi}_{jf}(r) &:= [\hat{K}(r-1,z)]^{-1} \hat{\phi}_{1j}(r) \\ &= \hat{\phi}_{1j}(r) - \sum_{i=1}^{n_k} \hat{k}_i(r-1) \hat{\phi}_{jf}(r-i), \\ \hat{y}_f(r) &:= [\hat{K}(r-1,z)]^{-1} \hat{y}_1(r) \\ &= \hat{y}_1(r) - \sum_{i=1}^{n_k} \hat{k}_i(r-1) \hat{y}_f(r-i), \\ \hat{v}_{jf}(r) &:= [\hat{K}(r-1,z)]^{-1} \hat{v}_j(r) \\ &= \hat{v}_j(r) - \sum_{i=1}^{n_k} \hat{k}_i(r-1) \hat{v}_{jf}(r-i). \end{aligned}$$

Then the filtered information vector  $\hat{\varphi}_{jf}(r)$  can be rewritten as

$$\hat{\boldsymbol{\varphi}}_{jf}(r) = [\hat{\boldsymbol{\phi}}_{jf}^{^{\mathrm{T}}}(r), -\hat{\boldsymbol{y}}_{f}^{^{\mathrm{T}}}(r-1), -\hat{\boldsymbol{y}}_{f}^{^{\mathrm{T}}}(r-2), \cdots, -\hat{\boldsymbol{y}}_{f}^{^{\mathrm{T}}}(r-n_{h}), \\ \hat{v}_{jf}(r-1), \hat{v}_{jf}(r-2), \cdots, \hat{v}_{jf}(r-n_{k})]^{^{\mathrm{T}}}.$$

From (25), we have  $v_j(r) = y_{1j}(r) - \boldsymbol{\varphi}_{1j}^{\mathsf{T}}(r)\boldsymbol{\vartheta}_{1j}$ . Replacing  $y_{1j}(r)$ ,  $\boldsymbol{\varphi}_{1j}(r)$  and  $\boldsymbol{\vartheta}_{1j}$  with  $\hat{y}_{1j}(r)$ ,  $\hat{\boldsymbol{\varphi}}_{1j}(r)$  and  $\boldsymbol{\vartheta}_{1j}(r)$ , the estimate  $\hat{v}_j(r)$  of  $v_j(r)$  can be calculated as follows:

$$\hat{v}_{j}(r) = \hat{y}_{1j}(r) - \hat{\varphi}_{1j}^{\mathsf{T}}(r)\hat{\vartheta}_{1j}(r),$$
  
$$\hat{\varphi}_{1j}(r) = [\hat{\varphi}_{1j}^{\mathsf{T}}(r), -\hat{y}_{1}^{\mathsf{T}}(r-1), -\hat{y}_{1}^{\mathsf{T}}(r-2), \cdots, -\hat{y}_{1}^{\mathsf{T}}(r-n_{h}),$$
  
$$\hat{v}_{j}(r-1), \hat{v}_{j}(r-2), \cdots, \hat{v}_{j}(r-n_{k})]^{\mathsf{T}}.$$

Minimizing the cost function  $J_1(\vartheta_{1j})$ , then the filtering based maximum likelihood recursive generalized extended least squares algorithm for estimating  $\vartheta_{1j}$  can be summarized as follows:

$$\hat{\boldsymbol{\vartheta}}_{1j}(r) = \hat{\boldsymbol{\vartheta}}_{1j}(r-1) + \boldsymbol{L}_{1j}(r)[\hat{y}_{1j}(r) - \hat{\boldsymbol{\varphi}}_{1j}^{\mathsf{T}}(r)\hat{\boldsymbol{\vartheta}}_{1j}(r-1)],$$
(30)

$$\boldsymbol{L}_{1j}(r) = \frac{\boldsymbol{P}_{1j}(r-1)\hat{\boldsymbol{\varphi}}_{1jf}(r)}{1+\hat{\boldsymbol{\varphi}}_{1jf}^{\mathsf{T}}(r)\boldsymbol{P}_{1j}(r-1)\hat{\boldsymbol{\varphi}}_{1jf}(r)},$$
(31)

$$\boldsymbol{P}_{1j}(r) = [\boldsymbol{I} - \boldsymbol{L}_{1j}(r)\hat{\boldsymbol{\varphi}}_{1jf}^{\mathsf{T}}(r)]\boldsymbol{P}_{1j}(r-1), \qquad (32)$$

$$\hat{\boldsymbol{\theta}}(r) = \frac{1}{m} [\hat{\boldsymbol{\theta}}_1(r) + \hat{\boldsymbol{\theta}}_2(r) + \dots + \hat{\boldsymbol{\theta}}_m(r)], \qquad (33)$$

$$\hat{k}(r) = \frac{1}{m} [\hat{k}_1(r) + \hat{k}_2(r) + \dots + \hat{k}_m(r)],$$
 (34)

$$\hat{\boldsymbol{\varphi}}_{1jf}(r) = [\hat{\boldsymbol{\phi}}_{jf}^{T}(r), -\hat{\boldsymbol{y}}_{f}^{T}(r-1), -\hat{\boldsymbol{y}}_{f}^{T}(r-2), \cdots, \\ -\hat{\boldsymbol{y}}_{f}^{T}(r-n_{h}), \hat{v}_{jf}(r-1), \\ \hat{v}_{jf}(r-2), \cdots, \hat{v}_{jf}(r-n_{k})]^{T},$$
(35)

$$\hat{\boldsymbol{\varphi}}_{1j}(r) = [\hat{\boldsymbol{\varphi}}_{1j}^{\mathsf{T}}(r), -\hat{\boldsymbol{y}}_{1}^{\mathsf{T}}(r-1), -\hat{\boldsymbol{y}}_{1}^{\mathsf{T}}(r-2), \cdots, \\ -\hat{\boldsymbol{y}}_{1}^{\mathsf{T}}(r-n_{h}), \hat{v}_{j}(r-1), \\ \hat{v}_{j}(r-2), \cdots, \hat{v}_{j}(r-n_{k})],$$
(36)

$$\hat{\phi}_{jf}(r) = \hat{\phi}_{1j}(r) - \sum_{i=1}^{n_k} \hat{k}_i(r-1)\hat{\phi}_{jf}(r-i), \qquad (37)$$

$$\hat{\mathbf{y}}_{f}(r) = \hat{\mathbf{y}}_{1}(r) - \sum_{i=1}^{n_{k}} \hat{k}_{i}(r-1)\hat{\mathbf{y}}_{f}(r-i), \qquad (38)$$

$$\hat{v}_{jf}(r) = \hat{v}_j(r) - \sum_{i=1}^{n_k} \hat{k}_i(r-1)\hat{v}_{jf}(r-i),$$
(39)

$$\hat{\boldsymbol{\phi}}_{1j}(r) = \boldsymbol{\phi}_j^{\mathrm{T}}(r) + \sum_{i=1}^{n_e} \hat{e}_i(r-1)\boldsymbol{\phi}_j^{\mathrm{T}}(r-i), \qquad (40)$$

$$\hat{\mathbf{y}}_{1}(r) = \mathbf{y}(r) + \sum_{i=1}^{n_{e}} \hat{e}_{i}(r-1)\mathbf{y}(r-i),$$
(41)

$$\hat{v}_{j}(r) = \hat{y}_{1j}(r) - \hat{\varphi}_{1j}^{\mathsf{T}}(r)\hat{\vartheta}_{1j}(r), \qquad (42)$$

$$\hat{\boldsymbol{\vartheta}}_{1j}(r) = [\hat{\boldsymbol{\eta}}_j(r), \hat{\boldsymbol{k}}^{\mathrm{T}}(r)]^{\mathrm{T}}.$$
(43)

For the identification model in (27), minimizing the cost function

$$\boldsymbol{J}_1(\boldsymbol{e}) := \sum_{r=1}^N [w_j(r) - \boldsymbol{\varphi}_{ej}^{\mathrm{T}}(r)\boldsymbol{e} - \boldsymbol{\varphi}_{kj}^{\mathrm{T}}(r)\boldsymbol{k}]^2,$$

we can obtain the following recursive least squares relation to estimate the parameter vector *e*:

$$\hat{\boldsymbol{e}}_{j}(r) = \hat{\boldsymbol{e}}_{j}(r-1) + \boldsymbol{L}_{ej}(r)[\hat{w}_{j}(r) - \boldsymbol{\varphi}_{ej}^{\mathrm{T}}(r)\hat{\boldsymbol{e}}(r-1) - \hat{\boldsymbol{\varphi}}_{kj}^{\mathrm{T}}(r)\hat{\boldsymbol{k}}(r)], \boldsymbol{L}_{ej}(r) = \boldsymbol{P}_{ej}(r-1)\boldsymbol{\varphi}_{ej}(r)[1 + \boldsymbol{\varphi}_{ej}^{\mathrm{T}}(r)\boldsymbol{P}_{ej}(r-1)\boldsymbol{\varphi}_{ej}(r)]^{-1}, \boldsymbol{P}_{ej}(r) = [\boldsymbol{I} - \boldsymbol{L}_{ej}(r)\boldsymbol{\varphi}_{ej}^{\mathrm{T}}(r)]\boldsymbol{P}_{ej}(r-1), \quad \boldsymbol{P}_{ej}(0) = p_{0}\boldsymbol{I}_{n_{e}}.$$

$$\hat{\varphi}_{ej}(r) := [-\hat{w}_j(r-1), -\hat{w}_j(r-2), \cdots, -\hat{w}_j(r-n_e)]^{\mathrm{T}} \in \mathbb{R}^{n_e}.$$

From (4), we have  $w_j(r) = y_j(r) - \psi_j^{\mathsf{T}}(r)\eta_j$ , replacing  $\eta_j$  with its estimate  $\hat{\eta}_j(r)$ , the estimate  $\hat{w}_j(r)$  of  $w_j(r)$  can be calculated as follows:

$$\hat{w}_j(r) = y_j(r) - \boldsymbol{\psi}_j^{\mathrm{T}}(r)\hat{\boldsymbol{\eta}}_j(r).$$
(44)

Therefore, the recursive generalized extended least squares algorithm for estimating the parameter vector e is summarized as follows:

$$\hat{\boldsymbol{e}}_{j}(r) = \hat{\boldsymbol{e}}_{j}(r-1) + \boldsymbol{L}_{ej}(r)$$

$$\times [\hat{w}_{j}(r) - \hat{\boldsymbol{\varphi}}_{ej}^{\mathsf{T}}(r)\hat{\boldsymbol{e}}(r-1) - \hat{\boldsymbol{\varphi}}_{kj}^{\mathsf{T}}(r)\hat{\boldsymbol{k}}(r)], \qquad (45)$$

$$\boldsymbol{L}_{ej}(r) = \boldsymbol{P}_{ej}(r-1)\hat{\boldsymbol{\varphi}}_{ej}(r)[1+\hat{\boldsymbol{\varphi}}_{ej}^{\mathsf{T}}(r)\boldsymbol{P}_{ej}(r-1)\hat{\boldsymbol{\varphi}}_{ej}(r)]^{-1},$$
(46)

$$\boldsymbol{P}_{ej}(r) = [\boldsymbol{I} - \boldsymbol{L}_{ej}(r)\hat{\boldsymbol{\varphi}}_{ej}^{\mathrm{T}}(r)]\boldsymbol{P}_{ej}(r-1), \qquad (47)$$

$$\hat{\boldsymbol{e}}(r) = \frac{1}{m} [\hat{\boldsymbol{e}}_1(r) + \hat{\boldsymbol{e}}_2(r) + \dots + \hat{\boldsymbol{e}}_m(r)], \qquad (48)$$

$$\hat{\boldsymbol{\varphi}}_{ej}(r) = [-\hat{w}_j(r-1), -\hat{w}_j(r-2), \cdots, -\hat{w}_j(r-n_e)]^{\mathrm{T}},$$
(49)

$$\hat{\varphi}_{kj}(r) = [\hat{v}_j(r-1), \hat{v}_j(r-2), \cdots, \hat{v}_j(r-n_k)]^T,$$
(50)

$$\hat{w}_j(r) = y_j(r) - \boldsymbol{\psi}_j^{\mathrm{T}}(r)\hat{\boldsymbol{\eta}}_j(r), \qquad (51)$$

$$\boldsymbol{\psi}_{j}(r) = [\boldsymbol{\phi}_{j}^{\mathrm{T}}(r), -\boldsymbol{y}^{\mathrm{T}}(r-1), -\boldsymbol{y}^{\mathrm{T}}(r-2), \cdots, \\ -\boldsymbol{y}^{\mathrm{T}}(r-n_{h})]^{\mathrm{T}}, \qquad (52)$$
$$\hat{\boldsymbol{\vartheta}}(r) = [\hat{\boldsymbol{\theta}}^{\mathrm{T}}(r), \hat{\boldsymbol{h}}_{1}^{\mathrm{T}}(r), \hat{\boldsymbol{h}}_{2}^{\mathrm{T}}(r), \cdots, \hat{\boldsymbol{h}}_{m}^{\mathrm{T}}(r), \hat{\boldsymbol{e}}^{\mathrm{T}}(r), \hat{\boldsymbol{k}}^{\mathrm{T}}(r)]^{\mathrm{T}}.$$

Equations (30)–(43) and (45)–(53) make up the filtering based maximum likelihood recursive generalized extended least squares (F-ML-RGELS) algorithm. The computation process of the F-ML-RGELS algorithm is summarized as follows.

- 1) To initialize. Let r = 1, and give the initial values  $\hat{\boldsymbol{\vartheta}}_{1j}(0) = 1_{n_1}/p_0$ ,  $\hat{\boldsymbol{e}}_j(0) = 1_{n_e}/p_0$ ,  $\hat{\boldsymbol{\phi}}_{jf}(r) = 1_n/p_0$ ,  $\hat{\boldsymbol{y}}_f(r-i) = 1_m/p_0$ ,  $\hat{v}_{jf}(r-i) = 1/p_0$ ,  $\hat{v}_j(r-i) = 1/p_0$ ,  $\hat{w}_j(r-i) = 1/p_0$ ,  $\boldsymbol{P}_{1j}(0) = p_0 \boldsymbol{I}_{n_1}$ ,  $\boldsymbol{P}_{ej}(0) = p_0 \boldsymbol{I}_{n_e}$ ,  $i = 1, 2, \cdots, \max[n_h, n_e, n_k]$ ,  $p_0 = 10^6$ ,  $j = 1, 2, \cdots, m$ . Set the data length *L*.
- 2) Gather the observation data y(r) and  $\Phi_s(r)$ , compute the filtered information vector  $\hat{\phi}_{1j}(r)$  and the filtered output vector  $\hat{y}_1(r)$  by (40) and (41), and construct the information vectors  $\hat{\varphi}_{1j}(r)$  and  $\psi_j(r)$  by (36) and (52).
- 3) Calculate the filtered vector  $\hat{\phi}_{jf}(r)$  using (37) and construct the filtered information vector  $\hat{\phi}_{1if}(r)$  using(35).
- 4) Calculate the gain vector  $L_{1j}(r)$  and the covariance matrix  $P_{1j}(r)$  using (31) and (32).
- 5) Calculate the parameter estimate  $\hat{\boldsymbol{\vartheta}}_{1j}(r)$  using (30),  $(j = 1, 2, \dots, m)$ .
- 6) Read  $\hat{\eta}_j(r)$  from  $\hat{\vartheta}_j(r)$  in (43) and calculate  $\hat{w}_j(r)$  and  $\hat{v}_j(r)$  using (51) and (42).

- 7) Compute the information vectors  $\hat{\varphi}_{ej}(r)$  and  $\hat{\varphi}_{kj}(r)$  by (49) and (50).
- 8) Compute the gain vector  $L_{ej}(r)$  by (46) and the covariance matrix  $P_{ej}(r)$  by (47).
- 9) Update the parameter estimate  $\hat{e}_i(r)$  by (45).
- 10) Compute the filtered output vector  $\hat{y}_f(r)$  and  $\hat{v}_{jf}(r)$  by (38) and (39).
- 11) Compute the parameter estimates  $\hat{\theta}(r)$ ,  $\hat{k}(r)$  and  $\hat{e}(r)$  by (33), (34) and (48).
- 12) Update the parameter estimate  $\hat{\vartheta}(r)$  by (53).
- 13) Increase *r* by 1 and go to Step 2.

*Remark 1:* The system is broken down into several subsystems based on the number of the system outputs. In this way, the maximum likelihood principle can be used more easily in parameter identification.

*Remark 2:* The data filtering method in system identification only changes the structure of the system and does not change the relationship between the inputs and outputs. The subsystem to be identified is transformed into a controlled autoregressive moving average model and a noise model.

*Remark 3:* Compared with the ML-RGELS algorithm, the introduction of the data filtering technique improves the parameter estimation accuracy of the F-ML-RGELS algorithm.

## V. EXAMPLE

Consider the following multivariate equation-error autoregressive moving average model:

$$H(z)\mathbf{y}(r) = \boldsymbol{\Phi}_{s}(r)\boldsymbol{\theta} + \frac{K(z)}{E(z)}\mathbf{v}(r),$$
  

$$\boldsymbol{\theta} = [\theta_{1}, \theta_{2}]^{\mathrm{T}} = [-0.85, 0.61]^{\mathrm{T}},$$
  

$$E(z) = 1 + e_{1}z^{-1} = 1 + 0.01z^{-1},$$
  

$$K(z) = 1 + k_{1}z^{-1} = 1 - 0.26z^{-1},$$
  

$$H(z) = \mathbf{I} + \mathbf{H}_{1}z^{-1}$$
  

$$= \begin{bmatrix} 1 & 0 \\ 0 & 1 \end{bmatrix} + \begin{bmatrix} h_{11} & h_{12} \\ h_{21} & h_{22} \end{bmatrix} z^{-1}$$
  

$$= \begin{bmatrix} 1 & 0 \\ 0 & 1 \end{bmatrix} + \begin{bmatrix} 0.23 & 0.50 \\ -0.70 & 0.53 \end{bmatrix} z^{-1},$$
  

$$\boldsymbol{\vartheta}_{1} = [\theta_{1}, \theta_{2}, h_{11}, h_{12}, e_{1}, k_{1}]^{\mathrm{T}}$$
  

$$= [-0.85, 0.61, 0.23, 0.50, 0.01, -0.26]^{\mathrm{T}},$$
  

$$\boldsymbol{\vartheta}_{2} = [\theta_{1}, \theta_{2}, h_{21}, h_{22}, e_{1}, k_{1}]^{\mathrm{T}}$$
  

$$= [-0.85, 0.61, -0.70, 0.53, 0.01, -0.26]^{\mathrm{T}},$$
  

$$\boldsymbol{\vartheta} = [\theta_{1}, \theta_{2}, h_{11}, h_{12}, h_{21}, h_{22}, e_{1}, k_{1}]^{\mathrm{T}}$$
  

$$= [-0.85, 0.61, 0.23, 0.50, -0.70, 0.53, 0.01, -0.26]^{\mathrm{T}},$$

In simulation, the data length L = 3000.  $\Phi_s(r)$  is a  $2 \times 2$  matrix sequence which contains the input-output data,  $\mathbf{y}(r) \in \mathbb{R}^2$  is the output vector of this model, and  $\mathbf{v}(r) \in \mathbb{R}^2$  as the white noise vector with zero mean and the noise variances  $\sigma_1^2 = \sigma_2^2 = \sigma^2 = 0.40^2$ . Applying the ML-RGELS algorithm and the F-ML-RGELS algorithm to estimate the parameters of this model, the estimation errors



**FIGURE 3.** The F-ML-RGELS and ML-RGELS estimation errors versus r ( $\sigma^2 = 0.40^2$ ).

 $\delta := \|\hat{\boldsymbol{\vartheta}}(r) - \boldsymbol{\vartheta}\| / \|\boldsymbol{\vartheta}\|$  versus *r* are shown in Figure 3. The F-ML-RGELS parameter estimates are plotted in Figures 5.

Applying the F-ML-RGELS algorithm to estimate the parameters of this model with different variances  $\sigma^2 = 0.40^2$  and  $\sigma^2 = 1.00^2$ , the parameter estimates and their errors are shown in Table 1.

To show the effectiveness of the estimated model obtained by the F-ML-RGELS algorithm, we choose the parameter estimates of the seventh row in Table 1 as the final estimated model, that is

$$\begin{bmatrix} 1+0.22024z^{-1} & 0.49714z^{-1} \\ -0.70579z^{-1} & 1+0.51896z^{-1} \end{bmatrix} \mathbf{y}(r) \\ = \mathbf{\Phi}_s(r) \begin{bmatrix} -0.85481 \\ 0.61899 \end{bmatrix} + \frac{1-0.26985z^{-1}}{1-0.00524z^{-1}} \mathbf{v}(r).$$

Then, the model predicted outputs can be represented as

$$\hat{y}_{1}(r) = y_{1}(r) - \hat{y}_{1f}(r) - \hat{h}_{11}z^{-1}\hat{y}_{1f}(r) - \hat{h}_{12}z^{-1}\hat{y}_{2f}(r) + \hat{\Phi}_{1f}\theta_{1} + \hat{\Phi}_{2f}\theta_{2} = y_{1}(r) - \hat{y}_{1f}(r) - 0.22024\hat{y}_{1f}(r-1) - 0.49714\hat{y}_{2f}(r-1) - 0.85481\hat{\Phi}_{1f} + 0.61899\hat{\Phi}_{2f}, \hat{y}_{2}(r) = y_{2}(r) - \hat{y}_{2f}(r) - \hat{h}_{21}z^{-1}\hat{y}_{2f}(r) - \hat{h}_{22}z^{-1}\hat{y}_{2f}(r) + \hat{\Phi}_{3f}\theta_{1} + \hat{\Phi}_{4f}\theta_{2} = y_{2}(r) - \hat{y}_{2f}(r) + 0.70579\hat{y}_{1f}(r-1) - 0.51896\hat{y}_{2f}(r-1) - 0.85481\hat{\Phi}_{3f} + 0.61899\hat{\Phi}_{4f}, \hat{y}_{1f}(r) = \hat{E}(z)y_{1}(r) + [1 - \hat{K}(z)]y_{1f}(r) = y_{1}(r) - 0.00524y_{1}(r-1) + 0.26985\hat{y}_{1f}(r-1) ,$$

**TABLE 1.** The F-ML-RGELS estimates and errors ( $\sigma^2 = 0.40^2$ ,  $\sigma^2 = 1.00^2$ ).



**FIGURE 4.** The F-ML-RGELS estimation errors versus *r* with different  $\sigma^2$ .

$$\begin{split} \hat{y}_{2f}(r) &= \hat{E}(z)y_2(r) + [1 - \hat{K}(z)]y_{2f}(r) \\ &= y_2(r) - 0.00524y_2(r-1) + 0.26985\hat{y}_{2f}(r-1), \\ \hat{\Phi}_{1f}(r) &= \hat{E}(z)\Phi_{s1}(r) + [1 - \hat{K}(z)]\hat{\Phi}_{1f}(r) \\ &= \Phi_{s1}(r) - 0.00524\Phi_{s1}(r-1) + 0.26985\hat{\Phi}_{1f}(r-1), \\ \hat{\Phi}_{2f}(r) &= \hat{E}(z)\Phi_{s2}(r) + [1 - \hat{K}(z)]\hat{\Phi}_{2f}(r) \\ &= \Phi_{s2}(r) - 0.00524\Phi_{s2}(r-1) + 0.26985\hat{\Phi}_{2f}(r-1), \\ \hat{\Phi}_{3f}(r) &= \hat{E}(z)\Phi_{s3}(r) + [1 - \hat{K}(z)]\hat{\Phi}_{3f}(r) \\ &= \Phi_{s3}(r) - 0.00524\Phi_{s3}(r-1) + 0.26985\hat{\Phi}_{3f}(r-1), \\ \hat{\Phi}_{4f}(r) &= \hat{E}(z)\Phi_{s4}(r) + [1 - \hat{K}(z)]\hat{\Phi}_{4f}(r) \\ &= \Phi_{s4}(r) - 0.00524\Phi_{s4}(r-1) + 0.26985\hat{\Phi}_{4f}(r-1). \end{split}$$

The root mean square error is used to describe the error between the true outputs and the predicted outputs, which is defined as

$$Error 1 = \left[\frac{1}{200} \sum_{r=3001}^{3200} [\hat{y}_1(r) - y_1(r)]^2\right]^{1/2} = 0.36164,$$
  
$$Error 2 = \left[\frac{1}{200} \sum_{r=3001}^{3200} [\hat{y}_2(r) - y_2(r)]^2\right]^{1/2} = 0.41900.$$

The outputs  $y_1(r)$  and  $y_2(r)$ , the predicted outputs  $\hat{y}_1(r)$ and  $\hat{y}_2(r)$ , and the prediction errors  $\hat{y}_1(r) - y_1(r)$  and  $\hat{y}_2(r) - y_2(r)$  of the estimated model versus *r* are shown in Figure 6.

From Table 1 and Figures 3–6, we can draw the following conclusions.

$\sigma^2$	r	$\theta_1$	$\theta_2$	$h_{11}$	$h_{12}$	$h_{21}$	$h_{22}$	$e_1$	$k_1$	$\delta$ (%)
$0.40^2$	100	-0.85854	0.59889	0.22024	0.50297	-0.66589	0.48864	0.01748	-0.36808	8.16610
	200	-0.85068	0.61407	0.21900	0.50950	-0.66597	0.52706	-0.01884	-0.32568	5.40846
	500	-0.83328	0.61413	0.18083	0.49930	-0.69214	0.53772	-0.03551	-0.29131	5.13112
	1000	-0.85197	0.61342	0.20620	0.49541	-0.70810	0.52730	0.00783	-0.29902	3.13939
	2000	-0.85804	0.62433	0.21766	0.49521	-0.70815	0.52600	-0.00140	-0.27563	2.00807
	3000	-0.85487	0.61899	0.22024	0.49714	-0.70579	0.51896	-0.00524	-0.26985	1.76011
$1.00^{2}$	100	-0.71768	0.35596	0.58101	0.50332	-0.61302	0.58499	0.04045	-0.32425	31.43129
	200	-0.73484	0.41986	0.46386	0.51492	-0.63071	0.58371	-0.08238	-0.36237	24.20665
	500	-0.73629	0.48025	0.25654	0.47016	-0.67163	0.57796	-0.16672	-0.39682	19.42952
	1000	-0.79842	0.52193	0.16576	0.47628	-0.70224	0.54589	-0.10725	-0.42308	15.78351
	2000	-0.83022	0.58493	0.12299	0.47718	-0.70724	0.53387	-0.08452	-0.39438	13.38358
	3000	-0.83520	0.58882	0.11410	0.48316	-0.70735	0.52012	-0.07125	-0.37942	12.58205
True values		-0.85000	0.61000	0.23000	0.50000	-0.70000	0.53000	0.01000	-0.26000	



**FIGURE 5.** The F-ML-RGELS estimates versus r ( $\sigma^2 = 0.40^2$ ).



**FIGURE 6.** The system outputs, the predicted outputs and the prediction errors versus *r* for the F-ML-RGELS ( $\sigma^2 = 0.40^2$ ).

- The F-ML-RGELS and ML-RGELS estimation errors are becoming small as the data length *r* increases. This shows that the two algorithms are effective see Figure 3.
- The estimation errors are becoming smaller as the noise variance decreases under the same data length in the F-ML-RGELS algorithm see Table 1 and Figure 4.

- The predicted outputs of the F-ML-RGELS algorithm are very close to the true outputs see Figures 6.
- The F-ML-RGELS algorithm has smaller parameter estimation errors than the ML-RGELS algorithm see Figure 3.

#### **VI. CONCLUSIONS**

This paper considers the parameter identification problems for the multivariate equation-error autoregressive moving average systems. An F-ML-RGELS algorithm is proposed for the multivariate equation-error systems by using the data filtering technique and the maximum likelihood principle. In addition, an ML-RGELS algorithm is presented as a comparison. The numerical example shows that the F-ML-RGELS algorithm is effective and has smaller parameter estimation errors than the ML-RGELS algorithm. The proposed algorithm can be extended to other multivariate systems [33], [34] and other fields [35]–[38]. The proposed algorithms in this paper can combine other identification methods [39]–[47], statistical strategies [48]–[55] and other methods [56]–[62] to study parameter identification of different systems and can be applied to other fields [63]–[71].

#### REFERENCES

- Y. Cao, P. Li, and Y. Zhang, "Parallel processing algorithm for railway signal fault diagnosis data based on cloud computing," *Future Gener. Comput. Syst.*, vol. 88, pp. 279–283, Nov. 2018.
- [2] Y. Cao, L. C. Ma, S. Xiao, X. Zhang, and W. Xu, "Standard analysis for transfer delay in CTCS-3," *Chin. J. Electron.*, vol. 26, no. 5, pp. 1057–1063, 2017.
- [3] Y. Cao, Y. Wen, X. Meng, and W. Xu, "Performance evaluation with improved receiver design for asynchronous coordinated multipoint transmissions," *Chin. J. Electron.*, vol. 25, no. 2, pp. 372–378, 2016.
- [4] Y. Z. Zhang, Y. Cao, Y. H. Wen, L. Liang, and F. Zou, "Optimization of information interaction protocols in cooperative vehicle-infrastructure systems," *Chin. J. Electron.*, vol. 27, no. 2, pp. 439–444, 2018.
- [5] M. Zorzi, "Convergence analysis of a family of robust Kalman filters based on the contraction principle," *SIAM J. Control Optim.*, vol. 55, no. 5, pp. 3116–3131, 2017.
- [6] M. Zorzi, "On the robustness of the Bayes and Wiener estimators under model uncertainty," *Automatica*, vol. 83, no. 9, pp. 133–140, 2017.
- [7] L. Xu, "A proportional differential control method for a time-delay system using the Taylor expansion approximation," *Appl. Math. Comput.*, vol. 236, pp. 391–399, Jun. 2014.
- [8] L. Xu, "Application of the Newton iteration algorithm to the parameter estimation for dynamical systems," *J. Comput. Appl. Math.*, vol. 288, pp. 33–43, Nov. 2015.
- [9] L. Xu, L. Chen, and W. L. Xiong, "Parameter estimation and controller design for dynamic systems from the step responses based on the Newton iteration," *Nonlinear Dyn.*, vol. 79, no. 3, pp. 2155–2163, 2015.
- [10] L. Xu, "The damping iterative parameter identification method for dynamical systems based on the sine signal measurement," *Signal Process.*, vol. 120, pp. 660–667, Mar. 2016.
- [11] C. B. Youssef, V. Guillou, and A. Olmos-Dichara, "Modelling and adaptive control strategy in a lactic fermentation process," *Control Eng. Pract.*, vol. 8, no. 11, pp. 1297–1307, 2000.
- [12] S. Y. Park and C. H. Han, "A nonlinear soft sensor based on multivariate smoothing procedure for quality estimation in distillation columns," *Comput. Chem. Eng.*, vol. 24, nos. 2–7, pp. 871–877, 2000.
- [13] S. Wu, "Multivariable PID control using improved state space model predictive control optimization," *Ind. Eng. Chem. Res.*, vol. 54, no. 20, pp. 5505–5513, 2015.

- [14] D. P. Li, D. J. Li, Y. J. Liu, S. C. Tong, and C. L. P. Chen, "Approximationbased adaptive neural tracking control of nonlinear MIMO unknown timevarying delay systems with full state constraints," *IEEE Trans. Cybern.*, vol. 47, no. 10, pp. 3100–3109, Oct. 2017.
- [15] X. Q. Yang, B. Huang, and H. J. Gao, "A direct maximum likelihood optimization approach to identification of LPV time-delay systems," *J. Franklin Inst.*, vol. 353, no. 8, pp. 1862–1881, 2016.
- [16] J. T. Jiao, K. Venkat, Y. J. Han, and T. Weissman, "Maximum likelihood estimation of functionals of discrete distributions," *IEEE Trans. Inf. The*ory, vol. 63, no. 10, pp. 6774–6798, Oct. 2017.
- [17] M. Imani and U. M. Braga-Neto, "Maximum-likelihood adaptive filter for partially observed Boolean dynamical systems," *IEEE Trans. Signal Process.*, vol. 65, no. 2, pp. 359–371, Jan. 2017.
- [18] W. Wang, X. Li, W. Wang, and X.-G. Xia, "Maximum likelihood estimation based robust Chinese remainder theorem for real numbers and its fast algorithm," *IEEE Trans. Signal Process.*, vol. 63, no. 13, pp. 3317–3331, Jul. 2015.
- [19] F. Y. Chen and F. Ding, "The filtering based maximum likelihood recursive least squares estimation for multiple-input single-output systems," *Appl. Math. Model.*, vol. 40, no. 3, pp. 2106–2118, 2016.
- [20] T. Söderström and U. Soverini, "Errors-in-variables identification using maximum likelihood estimation in the frequency domain," *Automatica*, vol. 79, pp. 131–143, May 2017.
- [21] M.-X. Chang and W.-Y. Chang, "Maximum-likelihood detection for MIMO systems based on differential metrics," *IEEE Trans. Signal Pro*cess., vol. 65, no. 14, pp. 3718–3732, Jul. 2017.
- [22] J. Pan, X. H. Yang, H. F. Cai, and B. X. Mu, "Image noise smoothing using a modified Kalman filter," *Neurocomputing*, vol. 173, no. 3, pp. 1625–1629, 2016.
- [23] D. Sigalov, T. Michaeli, and Y. Oshman, "LMMSE filtering in feedback systems with white random modes: Application to tracking in clutter," *IEEE Trans. Autom. Control*, vol. 59, no. 9, pp. 2549–2554, Sep. 2014.
- [24] C. Wang and T. Tang, "Several gradient-based iterative estimation algorithms for a class of nonlinear systems using the filtering technique," *Nonlinear Dyn.*, vol. 77, no. 3, pp. 769–780, 2014.
- [25] M. V. Kulikova and J. V. Tsyganova, "Constructing numerically stable Kalman filter-based algorithms for gradient-based adaptive filtering," *Int. J. Adapt. Control Signal Process.*, vol. 29, no. 11, pp. 1411–1426, 2015.
- [26] J. Dong, M. S. Liao, L. Zhang, and J. Y. Gong, "A unified approach of multitemporal SAR data filtering through adaptive estimation of complex covariance matrix," *IEEE Trans. Geosci. Remote Sens.*, vol. 56, no. 9, pp. 5320–5333, Sep. 2018.
- [27] S. Y. Zhao, Y. S. Shmaliy, C. K. Ahn, and F. Liu, "Adaptive-horizon iterative UFIR filtering algorithm with applications," *IEEE Trans. Ind. Electron.*, vol. 65, no. 8, pp. 6393–6402, Aug. 2018.
- [28] M. H. Li and X. M. Liu, "The least squares based iterative algorithms for parameter estimation of a bilinear system with autoregressive noise using the data filtering technique," *Signal Process.*, vol. 147, pp. 23–34, Jun. 2018.
- [29] F. Ding, X. Liu, and Y. Gu, "An auxiliary model based least squares algorithm for a dual-rate state space system with time-delay using the data filtering," *J. Franklin Inst.*, vol. 353, no. 2, pp. 398–408, 2016.
- [30] J. Pan, X. Jiang, X. K. Wan, and W. Ding, "A filtering based multiinnovation extended stochastic gradient algorithm for multivariable control systems," *Int. J. Control, Autom. Syst.*, vol. 15, no. 3, pp. 1189–1197, 2017.
- [31] L. J. Liu, F. Ding, and Q. M. Zhu, "Recursive identification for multivariate autoregressive equation-error systems with autoregressive noise," *Int. J. Syst. Sci.*, vol. 49, no. 13, pp. 2763–2775, 2018.
- [32] W. Wang, F. Ding, and J. Y. Dai, "Maximum likelihood least squares identification for systems with autoregressive moving average noise," *Appl. Math. Model.*, vol. 36, no. 5, pp. 1842–1853, May 2012.
- [33] Z. W. Ge, F. Ding, L. Xu, A. Alsaedi, and T. Hayat, "Gradient-based iterative identification method for multivariate equation-error autoregressive moving average systems using the decomposition technique," *J. Franklin Inst.*, vol. 356, no. 3, pp. 1658–1676, 2019.
- [34] Q. Y. Liu and F. Ding, "Auxiliary model-based recursive generalized least squares algorithm for multivariate output-error autoregressive systems using the data filtering," *Circuits Syst. Signal Process.*, vol. 38, no. 2, pp. 590–610, 2019.

- [35] H. Xu, F. Ding, and E. F. Yang, "Modeling a nonlinear process using the exponential autoregressive time series model," *Nonlinear Dyn.*, vol. 95, no. 3, pp. 2079–2092, Feb. 2019.
- [36] Y. Gu, Y. X. Chou, J. C. Liu, and Y. Ji, "Moving horizon estimation for multirate system with time-varying time delays," *J. Franklin Inst.*, vol. 356, no. 4, pp. 2325–2345, 2019.
- [37] Y. Gu, J. C. Liu, X. L. Li, Y. X. Chou, and Y. Ji, "State space model identification of multirate processes with time-delay using the expectation maximization," *J. Franklin Inst.*, vol. 356, no. 3, pp. 1623–1639, 2019.
- [38] L. Xu, F. Ding, and Q. M. Zhu, "Hierarchical newton and least squares iterative estimation algorithm for dynamic systems by transfer functions based on the impulse responses," *Int. J. Syst. Sci.*, vol. 50, no. 1, pp. 141–151, 2019.
- [39] F. Ding, H. Chen, L. Xu, J. Dai, Q. Li, and T. Hayat, "A hierarchical least squares identification algorithm for Hammerstein nonlinear systems using the key term separation," *J. Franklin Inst.*, vol. 355, no. 8, pp. 3737–3752, 2018.
- [40] Y. Cao, H. Lu, and T. Wen, "A safety computer system based on multisensor data processing," *Sensors*, vol. 19, no. 4, p. 818, 2019. doi: 10.3390/s19040818.
- [41] Y. Cao, Y. Zhang, T. Wen, and P. Li, "Research on dynamic nonlinear input prediction of fault diagnosis based on fractional differential operator equation in high-speed train control system," *Chaos*, vol. 29, no. 1, 2019, Art. no. 013130. doi: 10.1063/1.5085397.
- [42] L. J. Wan and F. Ding, "Decomposition-based gradient estimation algorithms for multivariate equation-error autoregressive systems using the multi-innovation theory," *Circuits Syst. Signal Process.*, 38, 2019. doi: 10.1007/s00034-018-1014-2.
- [43] J. Pan, H. Ma, X. Jiang, W. Ding, and F. Ding, "Adaptive gradient-based iterative algorithm for multivariable controlled autoregressive moving average systems using the data filtering technique," *Complexity*, vol. 2018, Jul. 2018, Art. no. 9598307. doi: 10.1155/2018/9598307.
- [44] F. Ding, L. Xu, F. E. Alsaadi, and T. Hayat, "Iterative parameter identification for pseudo-linear systems with ARMA noise using the filtering technique," *IET Control Theory Appl.*, vol. 12, no. 7, pp. 892–899, 2018.
- [45] X. Zhang, L. Xu, F. Ding, and T. Hayat, "Combined state and parameter estimation for a bilinear state space system with moving average noise," *J. Franklin Inst.*, vol. 355, no. 6, pp. 3079–3103, 2018.
- [46] X. Zhang, F. Ding, A. Alsaadi, and T. Hayat, "Recursive parameter identification of the dynamical models for bilinear state space systems," *Nonlinear Dyn.*, vol. 89, no. 4, pp. 2415–2429, 2017.
- [47] X. Zhang, F. Ding, L. Xu, and E. F. Yang, "State filtering-based least squares parameter estimation for bilinear systems using the hierarchical identification principle," *IET Control Theory Appl.*, vol. 12, no. 12, pp. 1704–1713, 2018.
- [48] C. C. Yin and J. S. Zhao, "Nonexponential asymptotics for the solutions of renewal equations, with applications," *J. Appl. Probab.*, vol. 43, no. 3, pp. 815–824, 2006.
- [49] H. Gao and C. Yin, "The perturbed Sparre Andersen model with a threshold dividend strategy," J. Comput. Appl. Math., vol. 220, nos. 1–2, pp. 394–408, 2008.
- [50] C. Yin and K. C. Yuen, "Optimality of the threshold dividend strategy for the compound Poisson model," *Statist. Probab. Lett.*, vol. 81, no. 12, pp. 1841–1846, 2011.
- [51] C. Yin, Y. Shen, and Y. Wen, "Exit problems for jump processes with applications to dividend problems," *J. Comput. Appl. Math.*, vol. 245, pp. 30–52, Jun. 2013.
- [52] C. Yin and Y. Wen, "Optimal dividend problem with a terminal value for spectrally positive Lévy processes," *Insurance Math. Econ.*, vol. 53, no. 3, pp. 769–773, 2013.
- [53] C. Yin and C. Wang, "The perturbed compound Poisson risk process with investment and debit interest," *Methodol. Comput. Appl. Probab.*, vol. 12, no. 3, pp. 391–413, 2010.
- [54] C. Yin, Y. Wen, and Y. Zhao, "On the optimal dividend problem for a spectrally positive Lévy process," *Astin Bull.*, vol. 44, no. 3, pp. 635–651, 2014.
- [55] Y. Wen and C. Yin, "Solution of Hamilton-Jacobi-Bellman equation in optimal reinsurance strategy under dynamic VaR constraint," *J. Function Spaces*, 2019, Art. no. 6750892. doi: 10.1155/2019/6750892.
- [56] L. Xu and F. Ding, "Parameter estimation algorithms for dynamical response signals based on the multi-innovation theory and the hierarchical principle," *IET Signal Process.*, vol. 11, no. 2, pp. 228–237, 2017.

- [57] L. Xu, F. Ding, Y. Gu, A. Alsaedi, and T. Hayat, "A multi-innovation state and parameter estimation algorithm for a state space system with d-step state-delay," *Signal Process.*, vol. 140, pp. 97–103, Nov. 2017.
- [58] L. Xu and F. Ding, "Parameter estimation for control systems based on impulse responses," *Int. J. Control, Autom., Syst.*, vol. 15, no. 6, pp. 2471–2479, Dec. 2017.
- [59] L. Xu, "The parameter estimation algorithms based on the dynamical response measurement data," *Adv. Mech. Eng.*, vol. 9, no. 11, pp. 1–12, Nov. 2017. doi: 10.1177/1687814017730003.
- [60] L. Xu and F. Ding, "Iterative parameter estimation for signal models based on measured data," *Circuits Syst. Signal Process.*, vol. 37, no. 7, pp. 3046–3069, Jul. 2018.
- [61] L. Xu, W. L. Xiong, A. Alsaedi, and T. Hayat, "Hierarchical parameter estimation for the frequency response based on the dynamical window data," *Int. J. Control Autom. Syst.*, vol. 16, no. 4, pp. 1756–1764, Aug. 2018.
- [62] X. Y. Li, H. X. Li, and B. Y. Wu, "Piecewise reproducing kernel method for linear impulsive delay differential equations with piecewise constant arguments," *Appl. Math. Comput.*, vol. 349, pp. 304–313, May 2019.
- [63] X.-S. Zhan, L.-L. Cheng, J. Wu, Q.-S. Yang, and T. Han, "Optimal modified performance of MIMO networked control systems with multi-parameter constraints," *ISA Trans.*, vol. 84, no. 1, pp. 111–117, 2019.
- [64] P. Gong, W. Wang, F. Li, and H. Cheung, "Sparsity-aware transmit beamspace design for FDA-MIMO radar," *Signal Process.*, vol. 144, pp. 99–103, Mar. 2018.
- [65] Z. H. Rao et al., "Research on a handwritten character recognition algorithm based on an extended nonlinear kernel residual network," KSII Trans. Internet Inf. Syst., vol. 12, no. 1, pp. 413–435, 2018.

- [66] N. Zhao *et al.*, "Contract design for relay incentive mechanism under dual asymmetric information in cooperative networks," *Wireless Netw.*, vol. 24, no. 8, pp. 3029–3044, 2018.
- [67] J. Pan, W. Li, and H. Zhang, "Control algorithms of magnetic suspension systems based on the improved double exponential reaching law of sliding mode control," *Int. J. Control Autom. Syst.*, vol. 16, no. 6, pp. 2878–2887, 2018.
- [68] F. Yang, P. Zhang, and X. X. Li, "The truncation method for the Cauchy problem of the inhomogeneous Helmholtz equation," *Applicable Anal.*, vol. 98, no. 5, pp. 991–1004, Apr. 2019.
- [69] W. Zhang, L. Xue, and X. Jiang, "Global stabilization for a class of stochastic nonlinear systems with SISS-like conditions and time delay," *Int. J. Robust Nonlinear Control*, vol. 28, no. 13, pp. 3909–3926, 2018.
- [70] D. Zhao, Y. Wang, Y. Li, and S. X. Ding, "H<sub>∞</sub> fault estimation for two-dimensional linear discrete time-varying systems based on Krein space method," *IEEE Trans. Syst., Man, Cybern. Syst.*, vol. 48, no. 12, pp. 2070–2079, Dec. 2018.
- [71] Y. Wang, Y. Si, B. Huang, and Z. Lou, "Survey on the theoretical research and engineering applications of multivariate statistics process monitoring algorithms: 2008–2017," *Can. J. Chem. Eng.*, vol. 96, no. 10, pp. 2073–2085, 2018.

Authors' photographs and biographies not available at the time of publication.

. . .