

Received January 5, 2019, accepted February 2, 2019, date of publication February 12, 2019, date of current version March 7, 2019.

Digital Object Identifier 10.1109/ACCESS.2019.2899035

Fast Minimization of Fixed Polarity Reed-Muller Expressions

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This work was supported in part by the National Natural Science Foundation of China under Grant 61772053, Grant 60973106, Grant 61232009, and Grant 81571142, in part by the China Postdoctoral Science Foundation under Grant 2018M641154, in part by the Scientific Science and Technology Research Projects of Universities in Hebei under Grant BJ2018012, in part by the Project of Hebei Natural Science Foundation under Grant G2018204093, in part by the Baoding Science and Technology Support Plan under Grant 15ZG050, and in part by the Hebei Science and Technology Project under Grant 15227535 and Grant 15227414.

ABSTRACT Logic minimization has recently attracted significant attention because in many applications it is important to have a compact representation as possible. In this paper, we propose a fast minimization algorithm (FMA) of fixed polarity Reed–Muller expressions (FPRMs). The main idea behind the FMA is to search the minimum FPRM with the fewest products by using the proposed binary differential evolution algorithm (BDE). The BDE can efficiently maintain population diversity and achieve a better tradeoff between the exploration and exploitation capabilities by use of proposed binary random mutation operator and improved selection operator. The experimental results on 24 MCNC benchmark circuits demonstrate that the FMA outperforms the genetic algorithm-based and simulated annealing genetic algorithm-based FPRMs minimization algorithms in terms of accuracy of solutions and solving efficiency. To the best of our knowledge, we are the first to use differential evolution algorithm to minimize FPRMs. The FMA can be extended to derive a minimum mixed polarity Reed–Muller expression.

INDEX TERMS Logic minimization, fixed polarity Reed-Muller expressions, differential evolution algorithm, mutation operator, selection operator.

I. INTRODUCTION

Logic functions can be expressed either in AND/OR/NOT based Boolean logic or in AND/XOR based Reed-Muller (RM) logic. For some circuits, such as the arithmetic circuits, parity check circuits and communication circuits, RM representation are more efficient than their Boolean representation in power, area, speed, and testability [1]–[3]. Notably, functions that do not produce efficient solutions in the Boolean representation can often be realized efficiently in the RM representation [4]–[6]. Therefore, it is necessary to establish synthesis schemes for RM logic circuits, particularly as look-up-table based Field Programmable Gate Array (FPGA) technology has become increasingly available and the relative cost of EXOR gates has become a non-critical factor restricting the use of RM design approaches [7], [8]. Fixed Polarity RM Expressions (FPRMs) are one of the canonical

RM expressions. FPRMs have attracted wide attention due to their remarkable superiority in designing easily testable circuits, detecting symmetric variable of switching functions, designing multi-level circuit and Boolean matching [9], [10]. Furthermore, for many practical functions, FPRMs require fewer products than sum-of-products expressions.

Logic minimization has become a fundamental research topic. For an n -variable logic functions, it has 2^n distinct FPRMs corresponding to 2^n different polarities. A minimum FPRM is one with the fewest products. Therefore, to search the best polarity corresponding to minimum FPRM from polarity optimization space is a typical NP-hard problem. Genetic Algorithm (GA) and its variants have been widely used in the polarity optimization of FPRMs due to the simplicity, robustness and inherent parallelism of GA. (e.g., [11]–[15]).

However, the existing polarity optimization approaches, which are mainly based on GA or its variants, have a low convergence speed or are easily trapped into the local optimal

The associate editor coordinating the review of this manuscript and approving it for publication was Walter Didimo.

solution due to the inherent defects of GA. Firstly, the selection operation makes the variance and entropy of population evolve toward the reduction, which would result in the decrease of population diversity. Secondly, there exists a contradiction between the dispersion of population and convergence, which leads to the slow convergence speed. Lastly, the selection mechanism based on fitness tends to the pure random selection when there is not much difference among the fitness of individuals, which would make the GA fall into the local optimal solution.

In this paper, we propose a Fast Minimization Algorithm (FMA) of FPRMs. Compared to existing minimization algorithms of FPRMs, our main contributions are as follows.

1) We propose a Binary Differential Evolution (BDE) algorithm to solve the discrete binary-encoded combination optimization problem, which can efficiently maintain population diversity and achieve a better tradeoff between the exploration and exploitation capabilities by use of proposed binary random mutation operator and improved selection operator.

2) We propose a FPRMs minimization algorithm, called FMA, which uses the BDE to search the minimum FPRM with the fewest products. To the best of our knowledge, we are the first to use differential evolution algorithm to minimize FPRMs. It can be extended to derive a minimum mixed polarity Reed-Muller expression.

3) We compare FMA with the GA based and Simulated Annealing GA (SAGA) based FPRMs minimization algorithms on MCNC benchmark circuits. Experimental results demonstrate the effectiveness and superiority of FMA.

The remainder of this paper is structured as follows. In Section II, we introduce a few preliminaries that are relevant to our study. The fast minimization algorithm of FPRMs is described in detail in Section III. Section IV presents the experimental results. Our conclusions are presented in Section V.

II. PRELIMINARIES

A. FPRM

Any n -variable Boolean function may be represented canonically in a sum-of-products form as

$$f(x_{n-1}, x_{n-2}, \dots, x_0) = \sum_{i=0}^{2^n-1} a_i m_i \quad (1)$$

where Σ is an OR operator and m_i are the minterms. a_i are the coefficient of minterms, and $a_i = 1$ or 0 represents the presence or absence of minterms, respectively.

By applying Shannon theorem, the Boolean function can be expressed as follows:

$$\begin{aligned} f &= \overline{x_{n-1}}f(0, x_{n-2}, \dots, x_0) + x_{n-1}f(1, x_{n-2}, \dots, x_0) \\ &= (1 \oplus x_{n-1})f(0, x_{n-2}, \dots, @_0) \oplus x_{n-1}f(1, x_{n-2}, \dots, x_0) \\ &= f(0, x_{n-2}, \dots, x_0) \oplus x_{n-1}[f(0, x_{n-2}, \dots, x_0) \\ &\quad \oplus f(1, x_{n-2}, \dots, x_0)] \end{aligned} \quad (2)$$

Therefore, the XOR/AND expansion corresponding to variant x_{n-1} of Boolean function is as follows:

$$f = f_0(x_{n-2}, \dots, x_0) \oplus x_{n-1}f_1(x_{n-2}, \dots, x_0) \quad (3)$$

where $f_1(x_{n-2}, \dots, x_0) = f(0, x_{n-2}, \dots, x_0) \oplus f(1, x_{n-2}, \dots, x_0)$, and $f_0(x_{n-2}, \dots, x_0) = f(0, x_{n-2}, \dots, x_0)$.

By applying Shannon theorem to each variant in turn, the Boolean function can be expressed by a FPRM as follows:

$$f^p(x_{n-1}, x_{n-2}, \dots, x_0) = \bigoplus_{i=0}^{2^n-1} b_i \pi_i \quad (4)$$

where $\bigoplus \Sigma$ denotes the modulo-2 addition, and $\pi_i = x_{n-1}x_{n-2} \dots x_0$ represents the products of FPRM. $b_i \in \{0, 1\}$ represents whether or not π_i appears in the function, and $p = (p_{n-1}p_{n-2} \dots p_0)$ is the polarity. In a FPRM, each variable appears either in true or in complement form, but not both. The polarity of FPRM can be represented by replacing each variable with 0 or 1 depending on whether the variable is used in true or complement, respectively. When a variable is used in true (complement), it can be replaced with 0 (1).

B. DIFFERENTIAL EVOLUTION

Differential Evolution (DE) algorithm, which was proposed by Storn and Price [22], is a simple yet powerful global optimization algorithm to deal with continuous optimization problems. It has become a new research hotspot in evolutionary computation and has been successfully applied in scientific and engineering fields. Similar to GA, DE is also a population based algorithm, which is stochastic in nature to find global solution in feasible individual space. Moreover, only a few control parameters are required in comparison with other competing heuristic optimization methods [17]. The DE mainly includes population initialization and three evolutionary operators (i.e., mutation, crossover and selection) that are used to update the population.

1) POPULATION INITIALIZATION

In DE, each individual is an n -dimensional vector that represents a candidate solution to the problem, which is randomly created in the search domain as follows:

$$x_{ij} = x_j^l + r \times (x_j^u - x_j^l) \quad (5)$$

where $i \in \{1, 2, \dots, NP\}$, $j \in \{1, 2, \dots, n\}$, x_{ij} is the j -th component of the i -th individual x_i , x_j^l and x_j^u are the lower and upper bounds of x_j , respectively. NP is the population size, and r is a uniformly distributed random number in $[0, 1]$. In initialization, all individuals are randomly generated with the uniform probability distribution.

2) MUTATION OPERATOR

The basic idea of mutation is that the difference vector between two individuals is taken, and a scaled version of the difference vector is added to a third individual to create a new candidate solution. Various other mutation schemes are listed in Table 1 [18], where x_{best} is the optimal individual in current

TABLE 1. Mutation schemes of DE.

Mutation schemes	Mathematical equation
Best/1/exp	$v_i = x_{best} + F \times (x_{r1} - x_{r2})$
Rand/1/exp	$v_i = x_{r1} + F \times (x_{r2} - x_{r3})$
Rand-to-Best/1/exp	$v_i = x_i + F \times (x_{r1} - x_{r2})$
Best/2/exp	$v_i = x_{best} + F \times (x_{r1} + x_{r2} - x_{r3} - x_{r4})$
Rand/2/exp	$v_i = x_{r1} + F \times (x_{r2} + x_{r3} - x_{r4} - x_{r5})$

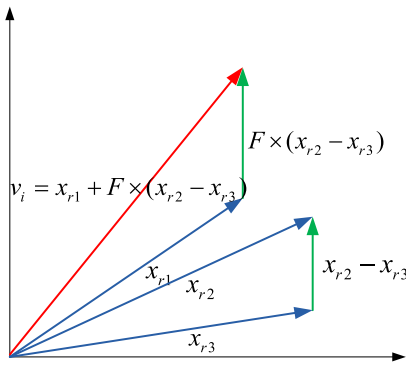


FIGURE 1. "DE/rand/1" mutation operation.

population. Equation (6) is one of most popular DE mutation schemes, which can be described by Fig.1.

$$v_i = x_{r1} + F \times (x_{r2} - x_{r3}) \quad (6)$$

where v_i is a mutation vector, and F is the scaling factor that controls the amplification of the differential vector. x_{r1} , x_{r2} and x_{r3} are candidate solutions, and $r1$, $r2$ and $r3$ are randomly selected that satisfy $r1, r2, r3 \in \{1, 2, \dots, NP\}$ and $r1 \neq r2 \neq r3 \neq i$.

3) CROSSOVER OPERATOR

The trial vector u_i is generated by crossing the target vector x_i with its mutation vector v_i . The widely used binomial crossover is defined as follows:

$$u_{ij} = \begin{cases} v_{ij}, & rand < CR \text{ or } j = r \\ x_{ij}, & otherwise \end{cases} \quad (7)$$

where j is the index of the dimensionality n , and $rand$ is a stochastic number taken from the uniform distribution $[0,1]$. CR is the constant crossover rate in $[0,1]$, and r is the random number in $[1,n]$.

4) SELECTION OPERATOR

The fitness value $f(x_i)$ of the target vector x_i is compared with the fitness value $f(u_i)$ of trial vector u_i to create the new population in the next generation. The winner will survive for the next generation. Taking the minimization problem as an example, the selection process is depicted as:

$$x_i^{next} = \begin{cases} u_i, & f(u_i) \geq f(x_i) \\ x_i, & otherwise \end{cases} \quad (8)$$

III. FAST MINIMIZATION OF FPRMS

In this section, we propose a binary version of DE according to the polarity characteristic of FPRMs, called BDE, to find the optimal solution in binary optimization space. Moreover, based on the BDE, we propose a FPRMs minimization algorithm, called FMA, which uses the BDE to search the best polarity corresponding to minimum FPRM with the fewest products. The overview of the FMA is depicted in Fig.2.

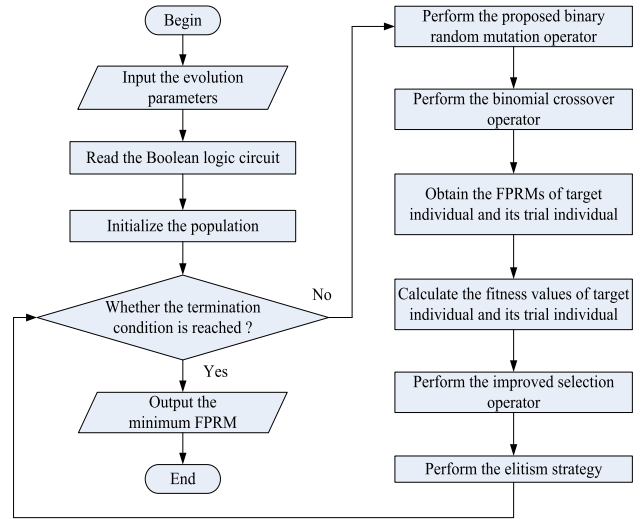


FIGURE 2. Overview of FMA.

A. BINARY DIFFERENTIAL EVOLUTION ALGORITHM

Since the polarity of FPRM can be represented by replacing each variable with 0 or 1 depending on whether the variable is used in true or complement, the polarity optimization of FPRMs is binary-encoded combinational optimization problem. However, the standard DE and most of its improved variants operate in the continuous space, which are not suitable for solving polarity optimization problems of FPRMs. Moreover, the DE has not yet been used as optimizer for RM circuits until now.

In this section, we propose a binary version of DE, called BDE, which enables the DE to operate in binary spaces. The structure of the BDE is similar to the standard DE so that the advantages of DE, such as easy implementation and parameter tuning, are inherited. Moreover, to enhance the global convergence ability, we introduce the elitism strategy to the BDE, namely the worst individual in current population is replaced by the elitism individual. In addition, we propose a binary random mutation operator and an improved selection operator to increase the population diversity and improve the global searching ability.

1) POPULATION INITIALIZATION

In the initialization process, the initial individuals are randomly generated as follows:

$$x_{ij} = \begin{cases} 1, & r < 0.5 \\ 0, & r \geq 0.5 \end{cases} \quad (9)$$

After a random number r between 0 and 1 is generated, the j -th element of i -th individual x_{ij} is set to 1 (namely, polarity 1) if r is less than 1/2; otherwise, x_{ij} is set to 0 (namely, polarity 0).

2) FITNESS FUNCTION

In BDE, the fitness function is used to evaluate the quality of individuals. Since the higher the fitness, the better the individual, the fitness functions is defined as follows:

$$fitness(x_i) = 1.0/products(x_i) \quad (10)$$

where $fitness(x_i)$ represents the fitness value of individual x_i , and $products(x_i)$ represents the number of products corresponding to individual x_i .

3) MUTATION OPERATOR

The use of logical mutation operator proposed in [19] has been widely reported in many studies. The logical mutation operator replaces the subtraction, multiplication and addition operator with XOR, AND and OR operators. It is represented as follows:

$$v_i = x_{r1} \otimes F \otimes (x_{r2} \oplus x_{r3}) \quad (11)$$

In the logical mutation operator, the subtraction of two random selected vectors was represented by XOR operator, the multiplication factor F was replaced by the AND operator, and the OR operator was applied to replace the operation for the addition of two randomly selected vectors. However, the OR operator has a higher probability of production of bit 1 in the evolution process, which could reduce the search diversity of the optimal solution.

To avoid the premature convergence and increase the population diversity, we propose a binary random mutation operator, which is represented as follows:

$$v_{i,j} = \begin{cases} x_{i,j} + (-1)^{x_{i,j}} \cdot |x_{r2,j} - x_{r3,j}|, & rand \geq 0.5 \\ x_{best,j} + (-1)^{x_{best,j}} \cdot |x_{r2,j} - x_{r3,j}|, & rand < 0.5 \end{cases} \quad (12)$$

where $v_{i,j}$ represents the j -th element of mutation vector v_i . $r2, r3 \in \{1, 2, \dots, NP\}$ and $r2 \neq r3 \neq i$. x_{best} is the optimal individual in current population, and $rand$ is a random number in $[0,1]$.

Since the polarity of FPRMs is taken as 0 or 1, the absolute value of difference vector is only 0 or 1 and the mutated variable is still 0-1 variable. Therefore, the binary random mutation operator satisfies the closure. Moreover, whether or not one dimension variable can be mutated depends on the difference vector. Specifically, the $x_{i,j}$ or $x_{best,j}$ can be mutated (from 0 to 1 or from 1 to 0) when $|x_{r2,j} - x_{r3,j}| = 1$, which could increase the population diversity, improve the global searching ability and prevent the algorithm trapping into the local optimal solution. In addition, the probability of $x_{r2,j}$ is equal to $x_{r3,j}$ is increasing along with the population evolution. The $x_{i,j}$ or $x_{best,j}$ remain the same when $|x_{r2,j} - x_{r3,j}| = 0$, which could accelerate the convergence speed.

4) IMPROVED SELECTION OPERATOR

The traditional greed selection operator chooses the winner with higher fitness value from target vector x_i and trial vector u_i for the next generation by comparing their fitness values. However, compared to other individuals of population, the loser may have higher fitness value. Therefore, the loser is ignored will omit better information and reduce the convergence speech. To preserve the better individual for next generation, the improved selection operation is as follows:

$$W = U \cup X \quad (13)$$

where U and X represent the populations consisting of individuals u_i and $x_i (i \in \{1, 2, \dots, NP\})$, respectively. W contains all the individuals of U and X , and the NP better individuals are selected as child population from W .

B. ALGORITHM DESCRIPTION

Based on the above description, the FMA is illustrated in Algorithm 1, in which “ G_{max} ” represents the maximum number of iteration, “ NP ” represents the population size, and “ D ” represents the dimensionality of decision variable. As shown in Algorithm 1, the FPRMs of target individual and its trial individual are obtained by using the fixed polarity conversion algorithm [11] according to their binary codes. Then, their fitness values are calculated according to their FPRMs and fitness function.

Algorithm 1 FMA

Input: The evolutionary parameters

Output: The minimum FPRM

1: Read the Boolean logic circuit;

2: $g \leftarrow 0$;

3: Initialize the population;

4: **while** ($g < G_{max}$)

5: **for** $i = 1$ to NP **do**

6: Perform proposed binary random mutation operator;

7: **for** $j = 1$ to D **do**

8: Perform the binomial crossover operator;

9: **end for**

10: Obtain FPRMs of target and trial individuals;

11: Calculate the fitness values of target and trial individuals;

12: Perform proposed improved selection operator;

13: **end for**

14: Perform the elitism strategy;

15: $g \leftarrow g + 1$;

16: **End**

17: Output the minimum FPRM

IV. EXPERIMENTAL RESULTS

The FMA has been implemented in C language, and the programs were compiled by the GNU C compiler. The results were obtained by using a PC with Intel Core i7

TABLE 2. Parameters and evolution operators settings for GAFMA, SAGAFMA, and FMA.

Parameter	GAFMA	SAGAFMA	FMA
The population size	100	100	100
The crossover probability	0.80	0.80	0.70
The mutation probability	0.01	0.01	-
The selection operator	Roulette wheel selection	Roulette wheel selection	Proposed improved greed selection
The crossover operator	One-point crossover	One-point crossover	Binomial crossover
The mutation operator	Basic bit mutation	Basic bit mutation	Proposed binary random mutation
Termination criterion	There is no improvement on the optimal solution in population over 10 iterations		

TABLE 3. Comparison of GAFMA, SAGAFMA, and FMA on the accuracy of solutions.

Name	Input	GAFMA			SAGAFMA			FMA		
		min	avg	std	min	avg	std	min	avg	std
b3	3	2	2	0	2	2	0	2	2	0
xor3	3	3	3	0	3	3	0	3	3	0
bw	5	8	8	0	8	8	0	8	8	0
xor5	5	5	5	0	5	5	0	5	5	0
m1	6	3	3	0	3	3	0	3	3	0
Z5xp1	7	3	3.20	0.60	3	3	0	3	3	0
lin	7	58	59.30	2.10	49	49	0	49	49	0
ex5	8	12	12.40	1.20	8	8	0	8	8	0
m3	8	5	5.90	1.37	5	5	0	5	5	0
rd84	8	57	60.10	3.30	54	54	0	54	54	0
1020	10	72	75.20	6.40	72	72	0	72	72	0
ex1010	10	56	59.60	7.20	56	56	0	56	56	0
br1	12	26	28.40	2.62	23	23	0	23	23	0
14_4color	14	25	32.00	3.69	14	14	0	14	14	0
table3	14	92	104.80	9.43	75	75	0	75	75	0
dk48	15	6	10.00	1.79	6	6.40	0.80	6	6	0
alcom	16	78	86.60	4.74	71	71	0	71	71	0
table5	17	647	664.70	13.07	618	621.20	1.60	618	618	0
src1	18	33	37.00	2.00	12	13.20	1.83	12	12.20	0.60
in2	19	325	355.10	10.72	277	283.30	4.12	277	277	0
mark1	20	420	431.40	17.41	356	368.80	6.40	356	356	0
mux	21	27	28.00	2.00	23	23	0	23	23	0
duke2	22	302	303.60	3.20	261	277.10	10.54	261	265.60	9.20
cordic	23	3419	3531.40	59.44	2148	2172.40	15.01	2148	2149.60	3.20

3.40GHz with 4G RAM under Linux. We compared the FMA with the GA based FPRMs Minimization Algorithm (GAFMA) [20] and SAGA based FPRMs Minimization Algorithm (SAGAFMA) [21] on 24 randomly selected MCNC benchmark circuits. Moreover, we ran the GAFMA, SAGAFMA, and FMA 10 times on each circuit to reduce the impact of randomness on the results. Table 2 gives the parameters and evolution operators settings for GAFMA, SAGAFMA, and FMA.

A. COMPARISON OF GAFMA, SAGAFMA, AND FMA ON THE ACCURACY OF SOLUTIONS

The comparison of GAFMA, SAGAFMA, and FMA on the accuracy of solutions are listed in Table 3. We took the optimal solution (namely, minimum number of products corresponding to best polarity), average value and standard deviation as experimental data. Column 1 shows the circuit name. Column 2 shows the number of input variables. “min”, “avg” and “std” represent the obtained optimal solution,

TABLE 4. Comparison of GAFMA, SAGAFMA, and FMA on the average number of iterations and run time.

Name	Input	GAFMA		SAGAFMA		FMA		Save1(%)		Save2(%)	
		iteration	time(s)	iteration	time(s)	iteration	time(s)	iteration	time(s)	iteration	time(s)
b3	3	1	0.20	1	0.18	1	0.03	0.00	85.00	0.00	83.33
xor3	3	1	0.19	1	0.18	1	0.05	0.00	73.68	0.00	72.22
bw	5	6	0.96	3	0.42	1	0.06	83.33	93.75	66.67	85.71
xor5	5	8	0.82	4	0.57	1	0.05	87.50	93.90	75.00	91.23
m1	6	12	1.18	4	0.61	1	0.06	91.67	94.92	75.00	90.16
Z5xp1	7	12	1.22	5	0.63	1	0.06	91.67	95.08	80.00	90.48
lin	7	14	7.34	6	3.70	1	0.14	92.86	98.09	83.33	96.22
ex5	8	15	2.85	8	1.44	2	0.37	86.67	87.02	75.00	74.31
m3	8	14	2.21	6	1.28	3	0.55	78.57	75.11	50.00	57.03
rd84	8	22	14.92	7	5.62	4	0.62	81.82	95.84	42.86	88.97
1020	10	24	23.03	9	7.76	6	4.60	75.00	80.03	33.33	40.72
ex1010	10	20	15.78	12	6.14	7	4.81	65.00	69.52	41.67	21.66
br1	12	31	17.02	16	9.37	8	3.44	74.19	79.79	50.00	63.29
14_4color	14	55	30.08	23	15.26	10	7.05	81.82	76.56	56.52	53.80
table3	14	70	37.40	27	18.70	18	11.62	74.29	68.93	33.33	37.86
dk48	15	62	36.32	26	16.43	20	9.59	67.74	73.60	23.08	41.63
alcom	16	69	67.85	31	34.41	23	17.53	66.67	74.16	25.81	49.06
table5	17	81	84.99	52	52.90	31	22.42	61.73	73.62	40.38	57.62
src1	18	73	173.55	37	139.34	29	54.63	60.27	68.52	21.62	60.79
in2	19	84	158.64	41	127.88	33	32.26	60.71	79.66	19.51	74.77
mark1	20	127	240.91	40	195.61	30	60.75	76.38	74.78	25.00	68.94
mux	21	95	226.33	34	174.26	27	68.90	71.58	69.56	20.59	60.46
duke2	22	116	262.72	45	223.15	34	75.18	70.69	71.38	24.44	66.31
cordic	23	145	318.05	67	257.18	42	95.84	71.03	69.87	37.31	62.73

average value and standard deviation respectively, which were taken after running each approach on each test circuit ten times.

From the Table 3, we can find that for the circuits with fewer variables (such as b3, xor3 and xor5), GAFMA was able to find the optimal solution each time. However, for the circuits that have more than 10 variables (such as br1, table5 and duke2), the GAFMA could hardly obtain the optimal solution. Moreover, for some circuits (such as table5, mark1 and cordic), the corresponding average values and standard deviations are large, which illustrates that the GAFMA is not appropriate to solve the large-scale polarity optimization problem due to its weak local searching ability. Additionally, we can also find that the SAGAFMA and FMA have the same searching performance for the circuits that have less than 15 variables. However, for some circuits that have more than 15 variables (such as src1, mark1 and cordic), although the SAGAFMA can also find the optimal solutions, the corresponding average values and

standard deviations are larger than that of FMA. In addition, we can find that for the 24 test circuits, there are 21 circuits whose minimum FPRMs can be obtained by FMA each time (namely, the corresponding standard deviation is 0) and there are 17 circuits whose minimum FPRMs can be obtained by SAGAFMA each time. Therefore, from the comparison experiment we can come to the conclusion that the FMA outperforms GAFMA and SAGAFMA on almost all the test circuits in terms of the accuracy of solutions.

B. COMPARISON OF GAFMA, SAGAFMA, AND FMA ON THE AVERAGE NUMBER OF ITERATIONS AND RUN TIME

The comparison of GAFMA, SAGAFMA, and FMA on the average number of iterations and run time (in CPU seconds) over 10 independent run are listed in Table 4. Columns 3 and 4 show the average number of iterations and run time of GAFMA. Columns 5 and 6 show the average number of iterations and run time of SAGAFMA. Columns 7 and 8 show the average number of iterations and run time of FMA.

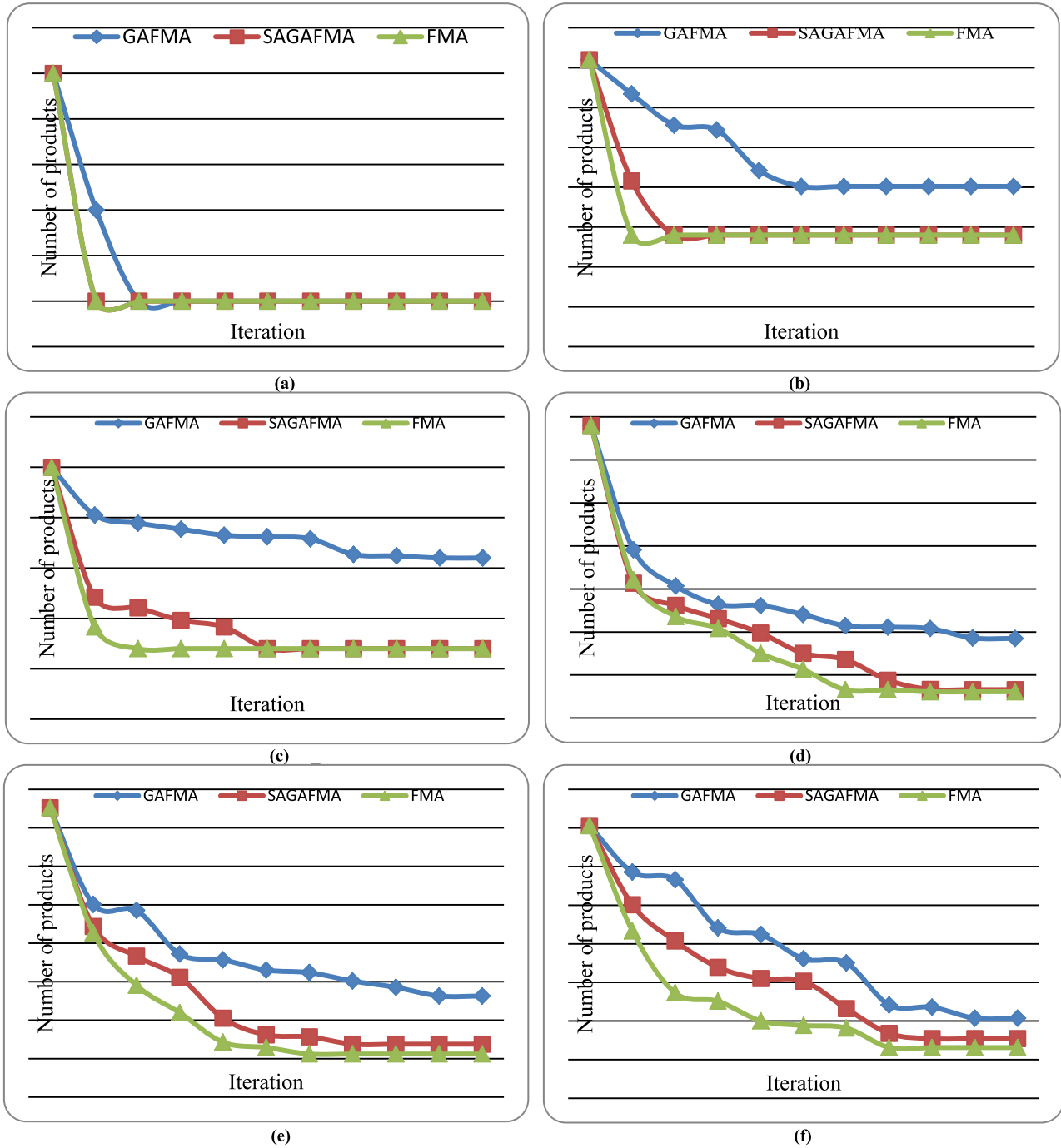


FIGURE 3. Convergence comparison of GAFMA, SAGAFMA, and FMA. (a) bw. (b) rd84. (c) 14_4color. (d) src1. (e) mark1. (f) duke2.

Columns 9 and 10 denote the percentage of the number of iterations and run time saved by FMA compared to GAFMA, which are defined as:

$$Save1_{iteration} = \frac{GAFMA_{iteration} - FMA_{iteration}}{GAFMA_{iteration}} \times 100\% \quad (14)$$

$$Save1_{time} = \frac{GAFMA_{time} - FMA_{time}}{GAFMA_{time}} \times 100\% \quad (15)$$

Columns 11 and 12 denote the percentage of the number of iterations and run time saved by FMA compared to

SAGAFMA, which are defined as:

$$Save2_{iteration} = \frac{SAGAFMA_{iteration} - FMA_{iteration}}{SAGAFMA_{iteration}} \times 100\% \quad (16)$$

$$Save2_{time} = \frac{SAGAFMA_{time} - FMA_{time}}{SAGAFMA_{time}} \times 100\% \quad (17)$$

From the Table 4, it is clear that the FMA outperforms the GAFMA and SAGAFMA in improving the minimization efficiency of FPRMs. Compared to the GAFMA, the greatest improvement in the number of iterations and run

time, which were made by FMA, are 92.86% and 98.09%, respectively. Moreover, compared to the SAGAFMA, the greatest improvement in the number of iterations and run time, which were made by FMA, are 83.33% and 96.22%, respectively. Therefore, it is concluded that compared to the GAFMA and SAGAFMA, the FMA has higher minimization efficiency both for the small-scale circuits and large-scale circuits.

C. COMPARISON OF GAFMA, SAGAFMA, AND FMA ON THE CONVERGENCE

In order to get a comprehensive analysis on the convergence of GAFMA, SAGAFMA, and FMA, we selected 6 representative MCNC benchmark circuits, namely, two small-scale circuits bw and rd84, two medium-sized circuits 14_4color and src1, and two large-scale circuits mark1 and duke2. Fig.3 shows the convergence comparison of GAFMA, SAGAFMA, and FMA in every 5 iterations, in which each algorithm was run 10 times on each circuit and the average values were used as the experimental data. Moreover, the horizontal axis shows the number of iterations, and the vertical axis shows the number of products.

As can be seen from the Fig.3, the convergence speed of GAFMA is much slower than that of SAGAFMA and FMA, and the GAFMA could not converge to the global optimal solutions except for the small-scale test circuit bw. In addition, it can be seen that the convergence curves of SAGAFMA and FMA have similar changing tendency. However, the convergence speed of FMA is much faster than that of SAGAFMA. Specifically, the FMA converges in 4 iterations while SAGAFMA needs 7 iterations on rd84; the FMA converges in 10 iterations while SAGAFMA needs 23 iterations on 14_4color; the FMA converges in 29 iterations while SAGAFMA needs 37 iterations on src1; the FMA converges in 30 iterations while SAGAFMA needs 40 iterations on mark1; the FMA converges in 34 iterations while SAGAFMA needs 45 iterations on duke2. Moreover, it is worthy of noting that compared to the SAGAFMA, the FMA can derive either the same or better solutions.

V. CONCLUSION

Minimization of FPRMs is a computationally hard problem, because the polarity optimization space increases exponentially with the increase of the number of input variables. In this paper, we propose a fast minimization algorithm of FPRMs, called FMA, which uses proposed BDE to search the best polarity corresponding to minimum FPRM with the fewest products. The experimental results over MCNC benchmark circuits demonstrate that the FMA performs superior to, or at least comparable to, the GA based and SAGA based FPRMs minimization algorithms in terms of the accuracy of solutions and solving efficiency, and confirm the application of BDE as a promising tool for solving the polarity optimization problem of FPRMs. In future, we will study an efficient ternary-encoded DE algorithm to minimize mixed polarity RM expressions.

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