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Self-Organizing Brain Emotional Learning Controller Network for Intelligent Control System of Mobile Robots

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ABSTRACT The trajectory tracking ability of mobile robots suffers from uncertain disturbances. This paper proposes an adaptive control system consisting of a new type of self-organizing neural network controller for mobile robot control. The newly designed neural network contains the key mechanisms of a typical brain emotional learning controller network and a self-organizing radial basis function network. In this system, the input values are delivered to a sensory channel and an emotional channel, and the two channels interact with each other to generate the final outputs of the proposed network. The proposed network possesses the ability of online generation and elimination of fuzzy rules to achieve an optimal neural structure. The parameters of the proposed network are online tunable by the brain emotional learning rules and gradient descent method; in addition, the stability analysis theory is used to guarantee the convergence of the proposed controller. In the experimentation, a simulated mobile robot was applied to verify the feasibility and effectiveness of the proposed control system. The comparative study using the cutting-edge neural network-based control systems confirms that the proposed network is capable of producing better control performances with high computational efficiency.

INDEX TERMS Mobile robot, neural network control, self-organizing neural network, brain emotional learning controller network.

I. INTRODUCTION

Autonomous mobile robots or vehicles are very useful in many application fields. Recently, the requirement growth of mobile robots in industrial and cargo applications becomes more and more rapid [1]-[5]. Current intelligent mobile robots broadly cover cutting-edge sciences and technologies of sensors, computer vision, artificial intelligence and other disciplines [6]-[8]. In particular, as a nonholonomic control system, the trajectory tracking problem is a typical and complicated research topic of mobile robots, since solving such problem must deal with a large number of uncertain and nonlinear disturbances [9]–[13]. Several current work suggested to use the robust optimal control for mobile robots [14], [15]. In order to better solve the problem, it is necessary to create a self-adaptive intelligent controller to handle the disturbances. Many modern control methods, such as fuzzy logic based and artificial neural network-based methods, have been utilized to control uncertain nonlinear systems.

The dynamic control of mobile robots faces two major challenges. First, an artificial neural network-based controller in mobile robots must contain enough self-adaptation and non-linear learning abilities. Many research applied adaptive neural network controllers to solve the tracking control problem of mobile robots and other dynamic

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systems [15]-[20]. Many studies used artificial neural networks as inverse modeling controllers to control mobile robots [5], [21]. However, these studies merely took neural network-based controller's output errors as learning assessments to update network weights. To achieve better performance for controlling mobile robot systems, neural networks also require to consider robot's overall performance to adjust their control parameters. Recently, a number of emotional learning methods [22]-[24], inspired by the learning architectures of human brain [25]-[27], are developed to build robot controllers [28], [29]; e.g., Jafari et al.'s [30] work developed and implemented a novel biologically inspired intelligent tracking controller for unmanned aircraft systems in presence of uncertain system dynamics and disturbance. Especially, a brain emotional learning controller network (BELC) not only uses network output errors to adjust its network weights, but also benefits from using the network's emotional output as an overall performance to tune its parameters [31], [32]. However, there is still room to improve the BELC's the nonlinear approximation ability, which is limited by its static network structure.

Second, to deal with unexpected and time-varying disturbances in mobile robot systems, neural network-based controllers must be able to fast arrange efficient computational resources. Several existing studies focused on building a incremental structure in their robot neural network control systems; e.g., Chao et al.'s [33]-[35] applied a resource allocation network to build robotic hand-eye coordination systems. There studies merely considered the incremental computing neurons, but ignored to remove less important neurons from their neural network based controllers; also, the stabilities of these control systems cannot be guaranteed. Other studies [36]-[39] suggested to use a "pruning" mechanism with adaptive learning methods to improve their network's computational efficiency. Indeed, such approaches can control their computational resources. Nevertheless, for controlling mobile robots, it is still a challenging task to quickly react to external disturbances that appears suddenly; therefore, both the computational efficiency and non-linear approximation ability must be satisfied in the network-based controllers of mobile robots.

This paper aims to address both challenges. A new type of self-organizing neural network, called self-organizing brain emotional learning controller network (SOBELC), is developed. The SOBELC network combines the key mechanisms of a fuzzy brain emotional learning controller network (FBELC) [31] and a self-organizing radial basis function network (RBF) [40]. An FBELC network contains a sensory system and a neural network judgement system, which are inspired by human brain's amygdala and orbitofrontal cortices, respectively [31], [41]. In this work, the network judgement system is established by introducing a self-organizing mechanism, which is developed from the RBF network [40]. The weights of the two systems are adjusted based on an emotional cue function [42], [43], which is calculated from the both system's inputs and outputs. After the emotional

learning process, the proposed network produces the final outputs by integrating the two systems' outputs. Thus, the proposed network not only retains its non-linear approximation ability, but also possesses the allocation capability of computational resources. With the support of the proposed SOBELC network, this work further develops an intelligent control system for dynamic non-linear control of mobile robots. The experimental results demonstrate the feasibility of the proposed learning method; in addition, the tracking ability of the mobile robot, the stabilization ability, and the robustness are improved to a certain extent. This work is evaluated by a simulated environment; however, various levels of unexpected disturbances are added into the experimental system, so as to simulate real-world applications. If the robot can well handle these simulated disturbances, the SOBELC network-based controller can also have good performances in the real world.

The remainder of this paper is organized as follows. Section II describes the dynamic model of a mobile robot. Section III introduces the detailed implementations of the proposed neural network. Section IV describes the intelligent control system using the proposed neural network for mobile robot control, including a discussion of the controller's update laws. Section V presents the experimental results and comparisons. Finally, Section VI concludes the paper and points out potential future work.

II. DYNAMIC MODEL OF MOBILE ROBOT

A wheeled mobile robot shown in Fig. 1-a, is a typical non-holonomic mechanical system. The robot has two coaxially mounted driven wheels and a passive wheel. The driven wheels are responsible for movements and steering of the mobile robot. Fig. 1-b illustrates a tracking procedure diagram of mobile robots. The tracking procedure is assumed that a task mobile robot can trace a virtual mobile robot (reference robot) that can perfectly move along predefined trajectories. Therefore, the optimal control scenario of the task robot is to enable its position and orientation to be consistent with those of the reference robot. The position and orientation errors between the task robot and the reference robots are used as the robot controller's inputs.

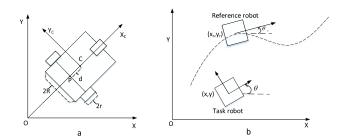


FIGURE 1. Wheeled mobile robot model and tracking procedure.
a) A wheeled mobile robot .

In Fig. 1-a, the radius of the driven wheel is r, the distance between two driven wheels is 2R, P is the center position



of the two driven wheel's axis, C is the mobile robot's center of gravity, and d is the distance between P and C. In the Descartes coordinate system, the position of the mobile robot is represented by the vector $q = [x_c \ y_c \ \theta]^T$, where x_c and y_c the coordinates of C; and θ is the angle between the coordinate system x_c , y_c and Descartes coordinate system.

In general, the dynamic equation of a nonholonomic mobile robot system with *n*-dimensional state and *m* dimension constraints can be expressed as:

$$M(q)\ddot{q} + V_m(q, \dot{q})\dot{q} + G(q) + F(\dot{q}) + \tau_d = B(q)\tau - A(q)\lambda$$
(1)

where \dot{q} is a velocity vector of the position and orientation; \ddot{q} is an acceleration vector of the position and orientation; $M(q) \in \mathbb{R}^{n \times n}$ is the positive definite symmetric inertia matrix; $V_m(q,\dot{q}) \in \mathbb{R}^{n \times n}$ is matrix of the radial force and Costa force; $G(q, \dot{q}) \in \mathbb{R}^n$ is gravity matrix; and $F(\dot{q}) \in \mathbb{R}^n$ is friction. Here, assume that the mobile moves on the horizontal ground, G(q) = 0 and $F(\dot{q}) = 0$; $\tau_d \in \mathbb{R}^n$ denotes the bounded unknown disturbances; $B(q) \in \mathbb{R}^{n \times n}$ is the input transformation matrix; $\tau \in \mathbb{R}^n$ is the control input vector; A(q) is a constraint matrix; and $\lambda \in \mathbb{R}^m$ is the restrain force.

Thus, the kinetic parameters of the mobile robot model in Fig. 1-A are obtained by Euler-Lagrange equation; the parameters are defined as follows:

$$M(q) = \begin{bmatrix} m & 0 & md \sin \theta \\ 0 & m & -md \cos \theta \\ md \sin \theta & -md \cos \theta & I \end{bmatrix}$$
(2)
$$C(q, \dot{q}) = \begin{bmatrix} 0 & 0 & md\dot{\theta} \sin \theta \\ 0 & 0 & md\dot{\theta} \cos \theta \\ 0 & 0 & 0 \end{bmatrix}$$
(3)
$$B(q) = \frac{1}{r} \begin{bmatrix} \cos \theta & \cos \theta \\ \sin \theta & \sin \theta \\ R & -R \end{bmatrix}$$
(4)

$$C(q, \dot{q}) = \begin{bmatrix} 0 & 0 & md\dot{\theta}\sin\theta \\ 0 & 0 & md\dot{\theta}\cos\theta \\ 0 & 0 & 0 \end{bmatrix}$$
(3)

$$B(q) = \frac{1}{r} \begin{bmatrix} \cos \theta & \cos \theta \\ \sin \theta & \sin \theta \\ R & -R \end{bmatrix}$$
 (4)

where m is the weight of the mobile robot; I represents the moment of inertia; and τ_r and τ_l are the torques of the right and left wheels, respectively.

Usually, a general constraint for mobile robots is: The mobile robot's movements only contain pure rolling without slipping [32]. Thus, we have:

$$\dot{x_c} \sin \theta - \dot{y_c} \cos \theta = \dot{\theta} d \tag{6}$$

where d is defined in Fig. 1-A; then, (6) can be rewritten as:

$$A(q)\dot{q} = 0 \tag{7}$$

In this paper, S(q) and v(t) are defined by:

$$S(q) = \begin{bmatrix} \cos \theta & d \sin \theta \\ \sin \theta & -d \cos \theta \\ 0 & 1 \end{bmatrix}$$
 (8)

$$\mathbf{v} = \begin{bmatrix} v \\ \omega \end{bmatrix} \tag{9}$$

where v is the linear velocities, and ω is the angular velocities. In terms of the nonholonomic constraint mentioned above, the kinematics model is then obtained as:

$$\dot{q} = \begin{bmatrix} \dot{x}_1 \\ \dot{y}_2 \\ \dot{\theta}_3 \end{bmatrix} = \begin{bmatrix} \cos \theta & d \sin \theta \\ -\sin \theta & d \cos \theta \\ 0 & 1 \end{bmatrix} \begin{bmatrix} \upsilon \\ \omega \end{bmatrix} = S(q)v(t) \quad (10)$$

Then, the tracking comparatively error against the position e_n is defined by:

$$e_p = \begin{bmatrix} e_1 \\ e_2 \\ e_3 \end{bmatrix} = \begin{bmatrix} \cos \theta & \sin \theta & 0 \\ -\sin \theta & \cos \theta & 0 \\ 0 & 0 & 1 \end{bmatrix} \begin{bmatrix} x_r - x \\ y_r - y \\ \theta_r - \theta \end{bmatrix}, (11)$$

and tracking comparatively error against the velocity e_v is defined as:

$$e_{v} = v_{r} - v \begin{bmatrix} v_{r} - v \\ \omega_{r} - \omega \end{bmatrix}$$
 (12)

where the reference velocities v_r can be respectively defined

$$\mathbf{v}_{r} = \begin{bmatrix} v_{r} \cos e_{3} + k_{1} e_{1} \\ \omega_{r} + k_{2} v_{r} e_{2} + k_{3} v_{r} \sin e_{3} \end{bmatrix}$$
(13)

where k_1 , k_2 , and k_3 are pre-defined parameters; and the definition of v_r is various; in this work, the reference velocity model defined in (13) is selected from Blažič's work [10]. Thus, the velocity error, e_c , can be obtained by $e_c = v_c - v$.

Then, left-multiply $S^{t}(q)$ to (1), the dynamic equation of the mobile robot is obtained by:

$$\overline{M}\dot{v}(t) + \overline{V}_{m}v(t) + \overline{F} + \overline{\tau}_{d} = \overline{B}\tau \tag{14}$$

where $\overline{M} = S^{T}MS$, $\overline{V}_{m} = S^{T}(M\dot{S} + V_{m}S)$, $\overline{\tau}_{d} = S^{T}\tau_{d}$, $\overline{\tau} = \overline{B}\tau$, $\overline{B} = S^T B$; and \overline{M} , \overline{C} , and \overline{B} are defined by:

$$\overline{M} = \begin{bmatrix} m & 0 \\ 0 & I - md^2 \end{bmatrix}, \quad \overline{C} = \begin{bmatrix} 0 & 0 \\ 0 & 0 \end{bmatrix}, \quad \overline{B} = \begin{bmatrix} \frac{1}{r} & \frac{1}{r} \\ \frac{R}{r} & -\frac{R}{r} \end{bmatrix}.$$
(15)

Based on [44], the parameter matrix of (14) has the following properties:

Property 1: $(M-2V_m)$ is skew-symmetric, i.e.,

$$x^{T}(M - 2V_m)x = 0, \quad \forall x \neq 0. \tag{16}$$

 $x^{T}(M - 2V_{m})x = 0, \quad \forall x \neq 0.$ (16) Then, derivative e_{c} and substitute it into (14), the dynamic equation against the velocity error is defined by:

$$\overline{M}\dot{e}_{c} = -\overline{V}_{m}(q,\dot{q})e_{c} + \overline{\tau} \tag{17}$$

III. SELF-ORGANIZING BRAIN EMOTIONAL LEARNING **CONTROLLER NETWORK**

In order to improve the non-linear approximation ability and computational efficiency of the BELC network, a new neural network is established by introducing a self-organizing mechanism to BELC. The resulted network not only takes the advantages of BELC and RBF neural networks in handling uncertain situations, but also enjoys the benefit of selforganizing mechanism for computational resource allocation. The architecture and self-organizing method of the new proposed neural network are described as following subsections.



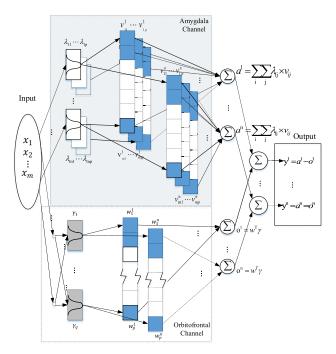


FIGURE 2. The architecture of the SOBELC network.

A. NEURAL NETWORK IMPLEMENTATION

The network architecture of the proposed SOBELC network shown in Fig. 2. The main structure of SOBELC is inspired by the specification of a BELC network. Inputs of SOBELC are mapped into two channels: a orbitofrontal channel (shown in the upper shade box in Fig. 2) and a amygdala channel. Such two-channel structure mimics that of a human brain, which has an orbitofrontal and an amygdala cortices [31], [45], [46]. The orbitofrontal cortex channel represents an emotional process in the network and the amygdala cortex channel does a sensory-motor process. Each channel also contains a receptive-field space and a weight memory. The receptive-field space calculates the activate level for the weight memory; then, the two weight vectors are aggregated and delivered to the output layer for the generation of the network's final output. The implementation of the proposed neural network is specified as follows.

- 1) **Input:** The input of a SOBELC network is a continuous multi-dimensional signal. A given input signal is presented as $X = [x_1, \dots, x_i, \dots, x_m]^T \in \mathbb{R}^m$, where m is the input dimension. Then, X is sent to the orbitofrontal and amygdala channels synchronously.
- 2) Orbitofrontal Channel: This channel is partially implemented by a fuzzy cerebellar mode articulation controller neural network, which is established in [32]. Inputs of the channel are distributed to a number of corresponding receptive fields, within each of which the Gaussian membership function calculates the total firing strength from the network's inputs. Thus, the Gaussian function in this channel is defined as:

$$\lambda_{ij} = exp(\frac{-(x_i - \zeta_{ij})^2}{\sigma_{ii}^2}) \tag{18}$$

where $i \in \mathbb{R}^m$, x_i denotes the *i*th input, $j \in \mathbb{R}^p$, j is the jth receptive field and p denotes the number of receptive field's layers; in addition, λ_{ii} , ζ_{ii} , and σ_{ii} denote the membership function, uncertain mean value, and variance value for the *j*th receptive field of the *i*th input, respectively.

Each receptive-field is mapped to a corresponding weight. The entire weight space, V_{OC} , in this channel is defined as:

$$V_{OC} = V_{i \times j \times k} = \begin{bmatrix} v_1 \\ \vdots \\ v_n \end{bmatrix}$$
 (19)

where k denotes the kth output; thus v_k is defined as:

$$\nu_k = \begin{bmatrix} \nu_{11} & \dots & \nu_{1p} \\ \vdots & \ddots & \vdots \\ \nu_{n_i 1} & \dots & \nu_{n_i p} \end{bmatrix}. \tag{20}$$

3) Amygdala Channel: The amygdala channel is established by using a self-organizing radius basis function neural network [36]. The receptive-field layer is composed of a set of RBF neurons; thus, the jth neuron's output is defined as:

$$\Theta_{j}(\|x - \psi_{j}\|, \gamma_{j}) = exp(\frac{-(x_{i} - \psi_{j}^{i})^{2}}{{\gamma_{j}^{i}}^{2}})$$
 (21)

where $i \in \mathbb{R}^m$; in addition, ψ_i^i and γ_i^i denote the center and width of the *j*th neucon of the *i*th input. Note that, in the Orbitofrontal channel, the number of receptive fields is static; however, the number of neurons in the amygdala channel is dynamic and is adjusted during the adaptive control process. The self-organizing rule is defined in Section III-B.

Each neuron links to a weight value in the weight space, W_{AC} , which is defined as:

$$W_{AC} = W_{q \times n_k} = \begin{bmatrix} w_{11} & \dots & w_{1n_k} \\ \vdots & \ddots & \vdots \\ w_{q1} & \dots & w_{qn_k} \end{bmatrix}. \quad (22)$$

4) **Output:** Outputs of the two channels, u_{OC}^k and u_{AC}^k , are presented as:

$$u_{OC}^{k} = V_{OC}^{T} \Lambda(\mathbf{x}, \zeta, \sigma)$$

$$u_{AC}^{k} = W_{AC}^{T} \Theta(\mathbf{x}, \psi, \gamma)$$
(23)

$$u_{AC}^{k} = W_{AC}^{T} \Theta(\mathbf{x}, \psi, \gamma) \tag{24}$$

where $k \in \mathbb{R}^n$. Based on the BELC structure, the output of SOBELC is $y^k = u_{OC}^k - u_{AC}^k$; thus, the output can be further presented as:

$$y^{k} = V_{OC}^{T} \Lambda(\mathbf{x}, \zeta, \sigma) - W_{AC}^{T} \Theta(\mathbf{x}, \psi, \gamma)$$
 (25)



B. SELF-ORGANIZATION MECHANISM

In the on-line learning process, the number of neurons in the amygdala channel can dynamic change. If the number of neurons in the hidden layer, κ , is too small, the network's process ability may not be efficient for control tasks. Conversely, if the number is too large, the computational cost is too heavy, so that it is might be unsuitable for online real-time control. In order to find a balance between the network performance and the computational cost, an online learning algorithm is established to increase or decrease the number of neurons in SOBELC. Based on this consideration, the on-line learning algorithm consists of two steps of process: increasing process and decreasing process, which are shown in Fig. 3 and specified as follows:

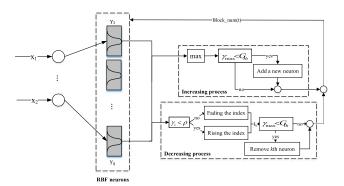


FIGURE 3. Self-organizing mechanism for SOBELC network. The algorithm consists of two steps of process: increasing process and decreasing process.

1) NEURON INCREASING

First of all, the leaning algorithm is to decide whether or not to add a new neuron into the amygdala channel. For each input data, x_i , the activation values of the amygdala channel's existing hidden neurons is used to represent x_i 's membership degree that the input data belongs to the existing neurons. If the current activation value is low, this situation indicates that the existing neural neurons are not sensitive to the input data, more processing neurons are required in the amygdala channel; otherwise, if the value is high, it is not necessary to add a new neuron. To simplify the calculation, the maximum activation value of the existing neurons is used to represent the activation grade of the all neurons. Thus, the maximum value, Θ_{max} , is defined as:

$$\Theta_{max} = \max_{1 \le i \le \kappa(t)} \Theta_j \tag{26}$$

where q(t) is the number of the existing neurons at the t-th time.

A pre-defined threshold, $G_{th} \in (0, 1)$, determines whether a new neuron can be added. Thus, if $\gamma_{max} \leq G_{th}$, then a new hidden neuron is created and added in the hidden layer in the amygdala channel. The new neuron's parameters,

 $(\psi_q^{new}, \gamma_q^{new}, \text{ and } w^{new})$, are initialized as:

$$\begin{cases} \psi_{\kappa}^{new} = x_t \\ \gamma_{\kappa}^{new} = \overline{\gamma} \\ w^{new} = 0 \\ \kappa_{t+1}^{new} = \kappa_t + 1 \end{cases}$$
(27)

where x_t is the new incoming data at the t-th time, $\overline{\gamma}$ is the mean width of radial basis function of the existing neurons, and w^{new} is easily set at 0.

2) NEURON DECREASING

To save the computational cost, the on-line learning algorithm can remove existing neural neuron, whose processing results cannot impact the overall results of the entire network. Thus, a neuron's significance, I, is defined to determine whether a neuron must be removed. Therefore, if the j-th neuron's significance, I_j , is larger than a threshold, P_{th} , then the j-th neuron is retained in the hidden layer. Otherwise, if $I_j \leq P_{th}$, the j-th neuron must be removed from its network. I_j can be updated by using the status of Θ_j : If Θ_j is smaller than elimination threshold value, ρ , then I_j will have a slight decrease; otherwise, I_j will have a small increase. The updating rule of I_j is summarized as follows:

$$I_{j}(t+1) = \begin{cases} I_{j}(t) \exp(-\tau_{1}) & \text{if } \Theta_{j} < \rho \\ I_{j}(t)[2 - \exp(-\tau_{2}(1 - I_{j}(t)))] & \text{if } \Theta_{j} \ge \rho \end{cases}$$
(28)

where the initial value of each I is set at $1; j \in [1, q]$ is the number of current neurons; and τ_1 and τ_2 are two pre-defined constant values. The complete algorithm is summarized as a pseudo-code shown in Algorithm 1.

Algorithm 1 Self-Organization Mechanism

```
1: Initialize G_{th} = 0.1, \rho = 0.1, \tau_1 = 0.01, and \tau_2 = 0.05;

2: Calculate \Theta_{max};

3: if \Theta_{max} \le G_{th} then

4: Set \gamma_{\kappa+1}^{new}, q(t+1) by (27);

5: end if

6: for j = 0 to \kappa(t) do

7: Update I_j by Eqn. 28;

8: if I_j < P_{th} then

9: Remove \psi_j, \gamma_j, and w_j;

10: Update \kappa;

11: end if

12: end for
```

C. PARAMETER UPDATE

The weights in the orbitofrontal channel are updated based on the brain emotional learning rule, the weight update values, \dot{v}_{ia} , are defined by:

$$\dot{v}_{iq} = \eta_z(h_i \times \max[0, d_q - u_{OC}^q])$$
 (29)



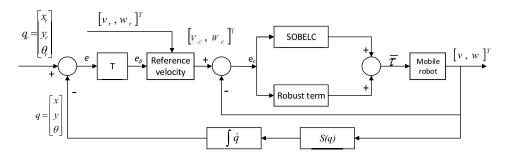


FIGURE 4. Wheeled mobile robot control and tracking system.

where η_z denotes a learning rate; and d_q denote an emotional cue parameter, defined by:

$$d_q = \sum_{i=1}^m \beta_{iq} \times SI_i + c_q \times y_k \tag{30}$$

where β_{iq} and c_q are gain parameters, which are determined in practical control problems. Thus, the updating law for the orbitofrontal channel's weights is defined as:

$$v_{iq}(t+1) = v_{iq}(t) + \dot{v_{iq}}.$$
 (31)

The tunable parameters of the amygdala channel are w, ψ , and γ ; therefore, to obtain more robust performance, the parameters are updated by Lyapunov stability analysis theory, rather than the brain emotional method. The updating values, \dot{w} , $\dot{\psi}$ and $\dot{\gamma}$, are described in Section IV and the updating laws of the amygdala channel are defined as:

$$w_{ijk}(t+1) = w_{ijk}(t) + \dot{w}_{ijk}$$
 (32)

$$\hat{\psi}_{iik}(t+1) = \hat{\psi}_{iik}(t) + \hat{\psi}_{iik}$$
 (33)

$$\hat{\gamma}_{iik}(t+1) = \hat{\gamma}_{iik}(t) + \dot{\hat{\gamma}}_{iik} \tag{34}$$

IV. NEURAL NETWORK CONTROL SYSTEM

The proposed SOBELC network is used to form a new network controller for mobile robot control problems. The structure of the proposed controller is illustrated in Fig. 4. The control system is designed based on Jin and Wang's work [47]. The errors, e_p , are the position differences between the reference robot and the task robot. Combing with the reference velocity model listed in Section II, and the velocity reference error, e_c , is used as the controller's input. The controller minimizes e_c and produce control values as system outputs. The controller is comprised of two sub-systems, including a SOBELC network, and a baseline robust controller. The input error values are fed into the two sub-systems for control signal generation. The control signals generated from both controllers are then aggregated to produce the final output of the overall control system. Thus, the output of the entire controller is computed by:

$$u = u_{SOBELC} + u_r \tag{35}$$

where u_{SOBELC} and u_r denote the outputs of the SOBELC network and robust controllers, respectively.

Assume that there exits an ideal SOBELC controller, u_{SOBELC}^* , to approximate the target u^* , which is presented as follows:

$$u^* = u_{SOBELC}^*(w^*, \psi^*, \gamma^*) + \varepsilon$$
$$= V^{*T} \hat{\Lambda} - W^{*T} \Theta^* + \varepsilon$$
(36)

where ε is an approximation error, and w^* , ψ^* , and γ^* are the optimal parameters of u^*_{SOBELC} . However, since the optimal nonlinear approximation u^*_{SOBELC} cannot be obtained, the online estimating \hat{u}_{SOBELC} must estimate the optimal u^*_{SOBELC} . Therefore, the approximate optimal value of the network-based controller can be defined as:

$$\hat{u} = \hat{u}_{SOBELC}(\hat{w}, \hat{\psi}, \hat{\gamma}) + u_r$$

$$= \hat{V}^T \hat{\Lambda} - \hat{W}^T \hat{\Theta} + u_r$$
(37)

where \hat{w} , $\hat{\psi}$, and $\hat{\gamma}$ are the estimated value of w^* , ψ^* , and γ^* , respectively; and u_r denotes output of the robust controller used to eliminate the error between the ideal controller u^* and the actual controller \hat{u}_{SOBELC} . Consider both (36) and (37), the approximation difference, \tilde{u} , between u^* and \hat{u} can be defined as:

$$\tilde{u} \equiv u^* - \hat{u}$$

$$= V^{*T} \hat{\Lambda} - W^{*T} \Theta^* + \varepsilon - \hat{V}^T \hat{\Lambda} + \hat{W}^T \hat{\Theta} - u_r$$

$$= \tilde{V}^T \hat{\Lambda} - \tilde{W}^T \Theta^* - \hat{W}^T \tilde{\Theta} + \varepsilon - u_r$$
(38)

where $\tilde{W} = W^* - \hat{W}$, $\tilde{V} = V^* - \hat{V}$ and $\tilde{\Theta} = \Theta^* - \hat{\Theta}$. The Taylor linearization method is applied to expand the Gaussian functions into partially linear form; thus, the expansion of $\hat{\Theta}$ in a Taylor series can be obtained by:

$$\tilde{\Theta} = \begin{bmatrix} \tilde{\Theta}_1 \\ \vdots \\ \tilde{\Theta}_l \\ \vdots \\ \tilde{\Theta}_{n_l} \end{bmatrix} = \begin{bmatrix} \left(\frac{\partial \Theta_1}{\partial \psi} \right)^T \\ \vdots \\ \left(\frac{\partial \Theta_l}{\partial \psi} \right)^T \\ \vdots \\ \left(\frac{\partial \Theta_{n_l}}{\partial \psi} \right)^T \end{bmatrix} \Big|_{\psi = \hat{\psi}} (\psi^* - \hat{\psi})$$



$$+\begin{bmatrix} \left(\frac{\partial \Theta_{1}}{\partial \gamma}\right)^{T} \\ \vdots \\ \left(\frac{\partial \Theta_{l}}{\partial \gamma}\right)^{T} \\ \vdots \\ \left(\frac{\partial \Theta_{n_{l}}}{\partial \gamma}\right)^{T} \end{bmatrix} \Big|_{\gamma=\hat{\gamma}} (\gamma^{*} - \hat{\gamma}) + O_{t}$$

$$\vdots \\ \left(\frac{\partial \Theta_{n_{l}}}{\partial \gamma}\right)^{T} \end{bmatrix}$$

$$= \Theta_{\psi}^{T} \tilde{\psi} + \Theta_{\gamma}^{T} \tilde{\gamma} + O_{t}$$
(39)

where $\tilde{\psi} = \psi^* - \hat{\psi}$, $\tilde{\gamma} = \gamma^* - \hat{\gamma}$ and $O_t \in R^{n_l}$ is a vector of higher-order terms; $\frac{\partial \Theta_l}{\partial \psi}$ and $\frac{\partial \Theta_l}{\partial \gamma}$ are defined as:

$$\begin{bmatrix} \frac{\partial \Theta_l}{\partial \psi} \end{bmatrix} = \begin{bmatrix} \underbrace{0, \dots, 0}_{(l-1) \times n_i}, \frac{\partial r_l}{\partial \psi_{1l}}, \dots, \frac{\partial \Theta_l}{\partial \psi_{n_i l}}, \underbrace{0, \dots, 0}_{(n_l - l) \times n_i} \end{bmatrix}^T \tag{40}$$

$$\begin{bmatrix} \frac{\partial \Theta_l}{\partial \gamma} \end{bmatrix} = \begin{bmatrix} \underbrace{0, \dots, 0}_{(l-1) \times n_i}, \frac{\partial \Theta_l}{\partial \gamma_{1l}}, \dots, \frac{\partial \Theta_l}{\partial \gamma_{n_i l}}, \underbrace{0, \dots, 0}_{(n_l - l) \times n_l} \end{bmatrix}^T \tag{41}$$

Rewrite (39), we have:

$$\Theta^* = \hat{\Theta} + \Theta_{tt}^T \tilde{\psi} + \Theta_{tt}^T \tilde{\gamma} + O_t \tag{42}$$

Substituting (42) and (39) into (38), yield

$$\tilde{u} = \tilde{V}^T \hat{\Lambda} - \tilde{W}^T (\hat{\Theta} + \Theta_{\psi} \tilde{\psi} + \Theta_{\gamma} \tilde{\gamma} + O_t) - \hat{W}^T (\Theta_{\psi} \tilde{\psi} + \Theta_{\gamma} \tilde{\gamma} + O_t) + \varepsilon - u_r = \tilde{V}^T \hat{\Lambda} - \tilde{W}^T \hat{\Theta} - \hat{W}^T (\Theta_{\psi} \tilde{\psi} + \Theta_{\gamma} \tilde{\gamma}) - u_r + \xi$$
 (43)

where ξ denotes the approximation error term, and $\xi = \tilde{W}^T \Theta_{\psi}^T \tilde{\psi} + \hat{W}^T \Theta_{\gamma}^T \tilde{\gamma} + W^{*T} O_t + \varepsilon$. ξ is supposed to be bounded by $0 \le |\xi|_{\infty} \le \xi_p$, in which $\xi_{P_{2\times 1}}$ is a positive constant matrix.

The robust control is designed as follows:

$$u_r = \frac{(I + \Lambda^2)R^2 + I}{2R^2} e_c^T \tag{44}$$

where R is a positive diagonal matrix, $R = diag(\phi_1, \phi_2, \cdots, \phi_i)$, ϕ_i is a robust attenuation coefficient that is specified by designers.

In order to guarantee the entire control system can retain convergence, the parameters updating values must be determined by using the Lyapunov stability theory. The proposed control system shown in Fig. 4 convert the error of the mobile robot's position and orientation to the reference velocity error. Therefore, a Lyapunov function is defined as:

$$L = L_1(e_p, t) + L_2(e_c, t)$$
(45)

where L_1 and L_2 are presented as follow:

$$L_1(e_p, t) = k_1(e_x^2 + e_y^2) + \frac{2k_1}{k_2}(1 - \cos e_\theta)$$
 (46)

$$L_{2}(e_{c}, t) = \frac{1}{2} [e_{c}^{T} \bar{M} e_{c} + tr[\tilde{W}^{T} \eta_{W}^{-1} \tilde{W}] + \tilde{\psi}^{T} \eta_{\psi}^{-1} \tilde{\psi} + \tilde{\gamma}^{T} \eta_{\gamma}^{-1} \tilde{\gamma} + tr[\tilde{V}^{T} \alpha^{-1} \tilde{V}]]$$
(47)

where $e_p = [e_x, e_y, e_\theta]^T$, and $e_c = [e_v, e_w]^T$. The proof of asymptotic stability of $L_1(e_p, t)$ is given in Blažič's work [10]. Therefore, if $L_2(e_c, t)$ achieves stable, the entire control system can be guaranteed. Derivative (47) and substitute (17) and (43) to the derivation (Note that: the estimation error of the derived torque $\tilde{\tau}$ is the estimation error \tilde{u}), then we have:

$$\begin{split} \dot{L}_{2} &= e_{c}^{T} \bar{M} \dot{e_{c}} + \frac{1}{2} [e_{c}^{T} \dot{\bar{M}} e_{c}] + tr[\tilde{W}^{T} \eta_{W}^{-1} \dot{\tilde{W}}] \\ &+ \tilde{\psi}^{T} \eta_{\psi}^{-1} \dot{\tilde{\psi}} + \tilde{\gamma}^{T} \eta_{\gamma}^{-1} \dot{\tilde{\gamma}} + tr[\tilde{V}^{T} \alpha^{-1} \dot{\tilde{V}}] \\ &= e_{c}^{T} (-\bar{V}_{m} e_{c} + \tilde{\tau}) + \frac{1}{2} [e_{c}^{T} \dot{\bar{M}} e_{c}] + tr[\tilde{W}^{T} \eta_{W}^{-1} \dot{\tilde{W}}] \\ &+ \tilde{\psi}^{T} \eta_{\psi}^{-1} \dot{\tilde{\psi}} + \tilde{\gamma}^{T} \eta_{\gamma}^{-1} \dot{\tilde{\gamma}} + tr[\tilde{V}^{T} \alpha^{-1} \dot{\tilde{V}}] \\ &= \frac{1}{2} e_{c}^{T} (\dot{\bar{M}} - 2\bar{V}_{m}) e_{c} + e_{c}^{T} \tilde{\tilde{\tau}} + tr[\tilde{W}^{T} \eta_{W}^{-1} \dot{\tilde{W}}] \\ &+ \tilde{\psi}^{T} \eta_{\psi}^{-1} \dot{\tilde{\psi}} + \tilde{\gamma}^{T} \eta_{\gamma}^{-1} \dot{\tilde{\gamma}} + tr[\tilde{V}^{T} \alpha^{-1} \dot{\tilde{V}}] \\ &= \frac{1}{2} e_{c}^{T} (\dot{\bar{M}} - 2\bar{V}_{m}) e_{c} + e_{c}^{T} \tilde{u} - tr[\tilde{W}^{T} \eta_{W}^{-1} \dot{\tilde{W}}] \\ &- \tilde{\psi}^{T} \eta_{\psi}^{-1} \dot{\tilde{\psi}} - \tilde{\gamma}^{T} \eta_{\gamma}^{-1} \dot{\tilde{\gamma}} - tr[\tilde{V}^{T} \alpha^{-1} \dot{\tilde{V}}] \\ &= e_{c}^{T} \tilde{V} \hat{\Lambda} - e_{c}^{T} \tilde{W} \hat{\Theta} - e_{c}^{T} \hat{W} (\Theta_{\psi} \tilde{\psi} + \Theta_{\gamma} \tilde{\gamma}) + e_{c}^{T} (\xi - u_{r}) \\ &- tr[\tilde{W}^{T} \eta_{W}^{-1} \dot{\tilde{W}}] - \tilde{\psi}^{T} \eta_{\psi}^{-1} \dot{\tilde{\psi}} - \tilde{\gamma}^{T} \eta_{\gamma}^{-1} \dot{\tilde{\gamma}} - tr[\tilde{V}^{T} \alpha^{-1} \dot{\tilde{V}}] \\ &\leq - tr[\tilde{W} (e_{c}^{T} \hat{\Theta} + \eta_{W}^{-1} \dot{\tilde{W}})] - \tilde{\psi} [e_{c}^{T} \hat{W} \Theta_{\psi} + \eta_{\psi}^{-1} \dot{\tilde{\psi}}] \\ &- \tilde{\gamma} [e_{c}^{T} \hat{W} \Theta_{\gamma} + \eta_{\gamma}^{-1} \dot{\tilde{\gamma}}] + e_{c}^{T} \tilde{V} \hat{\Lambda} + e_{c}^{T} (\xi - u_{r}) \end{cases} \tag{48} \end{split}$$

Since $\dot{\hat{V}} = 0$ when $d_q - u_{OC}^q \le 0$ and $\dot{\hat{V}} = \eta_z \cdot \Lambda \cdot [d_q - u_{OC}^q] > 0$ if $d_q - u_{OC}^q > 0$, we have $-tr[\tilde{V}^T \alpha^{-1} \dot{\hat{V}}] \le 0$.

Based on (48), the updating values of \dot{w} , $\dot{\psi}$ and $\dot{\gamma}$ are defined as:

$$\dot{\hat{W}} = -\eta_W \mathbf{e}_c^T \hat{\Theta} \tag{49}$$

$$\dot{\hat{\psi}} = -\eta_{\psi} \mathbf{e}_{c}^{T} \boldsymbol{\Theta}_{\psi} \hat{W} \tag{50}$$

$$\dot{\hat{\gamma}} = -\eta_{\gamma} \mathbf{e}_{c}^{T} \Theta_{\gamma} \hat{W} \tag{51}$$

By using the updating laws in (49), (50), and (51), and the robust controller's definition (44), (48) can be rewritten as:

$$\dot{L}_{2} \leq e_{c}^{T} \tilde{V} \hat{\Lambda} + e_{c}^{T} (\xi - u_{r})
= e_{c}^{T} \tilde{V} \hat{\Lambda} + e_{c}^{T} \xi - \frac{1}{2} e_{c}^{T} e_{c} - \frac{1}{2} \frac{e_{c}^{T} e_{c}}{r^{2}} - \frac{1}{2} e_{c}^{T} e_{c} \hat{\Lambda} \hat{\Lambda}^{T}
= -\frac{1}{2} e_{c}^{T} e_{c} - \frac{1}{2} [\frac{e_{c}}{r} - r\xi]^{2} - \frac{1}{2} [e_{c}^{T} \hat{\Lambda} - \tilde{V}]^{2}
+ \frac{1}{2} r^{2} \xi^{2} + \frac{1}{2} \tilde{V}^{T} \tilde{V}
\leq -\frac{1}{2} e_{c}^{T} e_{c} + \frac{1}{2} r^{2} \xi^{2} + \frac{1}{2} \tilde{V}^{T} \tilde{V}$$
(52)

Integrate (52) from t = 0 to t = T, then:

$$\int_0^T L_2 dt \le \int_0^T \sum_{i=1}^n \left[-\frac{e_{ci}^2(t)}{2} + \frac{r_i^2 e_{ci}^2(t)}{2} + \frac{\tilde{v}_i^2}{2} \right] dt.$$
 (53)



Then, (53) is rewritten as:

$$L_{2}(T) - L_{2}(0) \le \sum_{i=0}^{n} \left[-\frac{1}{2} \int_{0}^{T} e_{ci}(t)dt + \frac{r_{i}^{2}}{2} \int_{0}^{T} \xi_{i}^{2}(t)dt + \frac{1}{2} \int_{0}^{T} \tilde{v}_{i}(t)dt \right]$$
(54)

Since $L_2(T) \ge 0$, (54) is simplified as:

$$\frac{1}{2} \sum_{i=0}^{n} \int_{0}^{T} e_{ci}^{2}(t)dt \le L_{2}(0) + \frac{1}{2} \sum_{i=0}^{n} r_{i}^{2} \int_{0}^{T} \xi_{i}^{2}(t)dt + \frac{1}{2} \sum_{i=0}^{m} \int_{0}^{T} \tilde{v}_{i}(t)dt. \quad (55)$$

Then, we have:

$$\sum_{i=0}^{n} \int_{0}^{T} e_{ci}^{2}(t)dt \leq \sum_{i=0}^{n} e_{ci}^{2}(0) + \sum_{i=0}^{n} r_{i}^{2} \int_{0}^{T} \xi_{i}^{2}(t)dt + \frac{1}{2} \sum_{i=0}^{m} \int_{0}^{T} \tilde{v}_{i}(t)dt.$$
 (56)

If $\int_0^T \xi_i^2(t)dt < \infty$, $\int_0^T v_i^2(t)dt < \infty$, then for all T, there exists $\int_0^T e_{ci}^2(t)dt < \infty$; therefore, the asymptotic stability of L_2 has been proved to be completed.

In summary, as $t \to \infty$, $\{e_p, e_c\} \to 0$, and $\{\dot{e}_p, \dot{e}_c\} \to 0$; the tracking system is asymptotically stable proved by the Lyapunov stability theory.

V. EXPERIMENTATIONS

A. EXPERIMENTAL SETUP

The proposed controller with the new SOBELC is applied to a mobile robot system, so as to verify the controller's effectiveness and efficacy. The experiments are based on a simulation mobile robot, which tracks a predefined reference trajectory. The reference trajectory has two patterns; the first pattern with the time $0 \le t < 65$ is designed as:

$$\begin{cases} x_r = \nu_r \cdot \cos(\omega) \\ y_r = \nu_r \cdot \sin(\omega) \end{cases}$$
 (57)

and the second pattern with the time $65 < t \le 130$ is defined as:

$$\begin{cases} x_r = \nu_r \cdot \cos(2\omega) \\ y_r = \nu_r \cdot \sin(\omega) \end{cases}$$
 (58)

where the initial velocity of the reference robot is set as: $v_r = 0.2m/s$ and $\omega_r = 0.1rad/s$ and that of the task robot is set as the identical value. In addition, the initial position and orientation of the reference mobile robot are set as $qr = \begin{bmatrix} 2 & 0 & \pi/2 \end{bmatrix}^T$ and those of the tracking one are set as $q = \begin{bmatrix} 1 & 0 & \pi/2 \end{bmatrix}^T$.

In the experiments, the task mobile robot's parameters are set as: m = 10 kg, $I = 5 \text{kg} \cdot m^2$, R = 0.2 m, r = 0.05 m,

d=0.0m, $F(\dot{q})=0$. In addition, the disturbance, $\bar{\tau}_d$, is defined as:

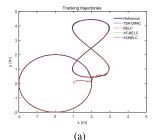
$$\overline{\tau}_d = \begin{bmatrix} \delta \sin(4t) \\ \delta \cos(4t) \end{bmatrix}. \tag{59}$$

where δ denotes the level of disturbance. In this experiment, two levels of disturbances are used to evaluate the proposed network controller; thus, δ is set at 10 and 20, respectively.

In order to compare the effectiveness of the proposed method, a TSK CMAC" network based controller [48] (labeled as "TSK-CMAC"), a fuzzy BEL network based controller [31] (labeled as "BELC"), and an adaptive BEL network based controller [43] (labeled as "AF-BELC") are included in the same experiments. The parameters of the SOBELC neural network model are selected as: $G_{th}=0.5$, $P_{th}=0.1$, $\rho=0.1$, $\kappa=9$, $\tau_1=0.01$, and $\tau_2=0.05$. The following paragraphs describe three experiments, each of which applies one disturbance level.

B. RESULTS

Fig. 5 demonstrates the simulated position response of the mobile robot under the three levels of disturbances. Each of the sub-figures contains the reference trajectory (red solid line), the TSK-CMAC output trajectory (blue dotted line), the BELC output trajectory (pink dotted line), the adaptive BELC output trajectory (green dotted line), and the proposed SOBELC output trajectory (red dotted line). In the experiment, the tracking time-length is 130s; however, from 0s to 65s, the mobile robot is required to track the circular trajectory in the down-left of the sub-figure; then, from 65s to 130s, the robot must track the trajectory similar to digit number "8".



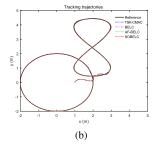


FIGURE 5. Trajectories of the mobile robot for the four controllers with $\tau_d:\delta=$ 10 and 20. (a) $\tau_d:\delta=$ 10. (b) $\tau_d:\delta=$ 20.

In Fig. 5, the performances of the four controllers are very close to each other; their output trajectories are almost coincided and closely reach the reference trajectory. This situation proves that the entire neural network-based controller possesses strong applicability and generality. However, in order to clearly identify the tracking performance of the four controllers, the position and velocity errors of the controllers with various disturbances at around 0s and 65s are magnified and shown in Figs. 6 and 7.

Fig. 6a shows position and orientation errors of the four network controllers with $\delta = 10$. As defined in (11),



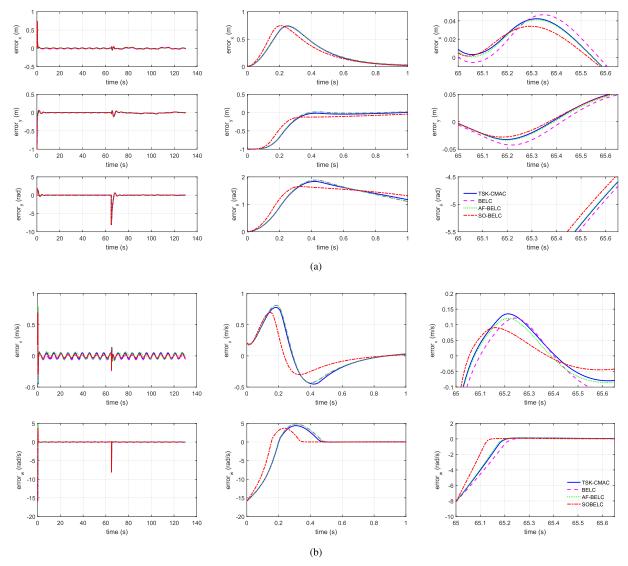


FIGURE 6. Tracking performance of the four network controllers with $\tau_d: \delta = 10$. (a) Position and orientation errors of the four controllers. (b) Velocity errors of the four controllers .

the errors consist of x, y, and θ components. The left column shows the entire tracking procedure and the right column shows the magnification of the errors at around 65s. Three plots in the left column indicate the three errors in x, y, and θ ; the other three in the middle column indicate the three magnified versions of those left ones after a few seconds of tracking; and the rest three in the right column show the magnified plots after the 65s.

In this figure, the left three plots reveal the similar situation with that of Fig. 5: the four network-based controllers can rapidly react to tracking errors. However, in the magnified plots at 0s, The TSK-CMAC, BELC, and AF-BELC controllers generate very similar results regarding the position and orientation tracking errors, since the three curves are almost coincided. However, the tracking performance of the SOBELC controller is much better than those of the TSK-CMAC, BELC, and AF-BELC controllers at 0s. The SOBELC exhibits fast error convergence speed in x and y; in addition, the SOBELC can achieve less overshoot in θ . At 65s, the SOBELC controller also exhibits the faster error convergence speed than those of the rest three controllers. The performances of the TSK-CMAC and AF-BELC controllers are very similar; however, the BELC controller has the worst performance.

Fig. 6b shows velocity errors of the four network controllers with $\delta=10$, and the line types of this figure are identical with those of Fig. 6a. As defined in (12), the velocity error consists of υ and ω components. At 0s, the SOBELC controller exhibits clearly advantages in velocity tracking in both υ and ω , since the SOBELC controller has less overshoot and smoother convergence curve. Especially, the performance of the SOBELC in ω has a significant advantage.



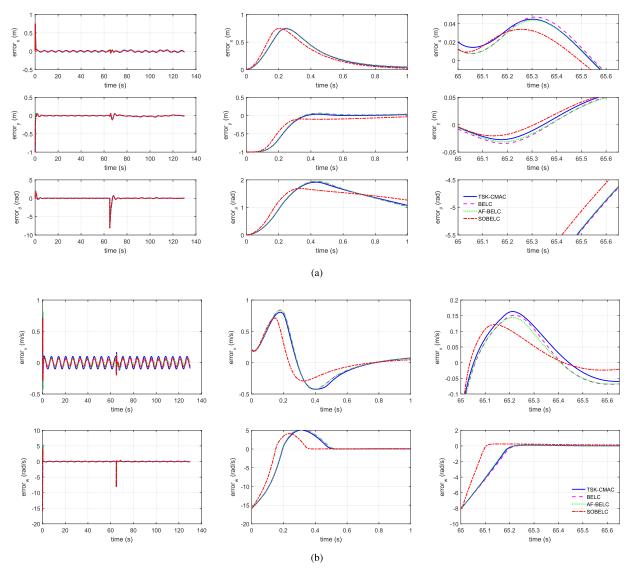


FIGURE 7. Tracking performance of the four network controllers with τ_d : $\delta = 20$. (a) Position and orientation errors of the four controllers. (b) Velocity errors of the four controllers.

At 65s, the performances of the four controllers are about the same; the SOBELC controller has a slim leading over the rest three controllers. However, the BELC controller's performance is the worst in the four controllers. As considered the entire tracking performance of the four network controllers with $\delta = 10$, the proposed SOBELC network has a better ability to dynamically control mobile robots.

Fig. 7 illustrates the position, orientation, and velocity tracking performances the four network controllers with $\delta = 20$; the legend of this figure is identical to that of Fig. 6. In Fig. 7a, the entire performances of the four controllers remain consistent with those under $\delta = 10$. The tracking performance of the SOBELC is much better than those of the other three controllers, since the controller demonstrated faster error convergence speed across all these experiments. It is interesting to note that the entire performance of the

SOBELC under $\delta=20$ is better than those under $\delta=10$. This phenomenon is related to the combination of the outputs of the self-organizing mechanism and the robust term in the SOBELC controller of the proposed system. If a larger disturbance presents, both components must adjust their weights or parameters simultaneously, resulting in different tracking performances from those under small disturbances. Also, the output value range of the robust controller is larger than that of the network; therefore, the robust controller can act faster for larger disturbances. A formal analysis of such situation remains as an important piece of active future work.

Based on the above analysis, Fig. 8 shows the dynamic self-organizing process during the experiments with $\delta=10$ and 20. Note that, the SOBELC contains two processing channels, Fig. 8 merely shows the hidden neurons of the amygdala channel. The initial size of neurons in the SOBEL

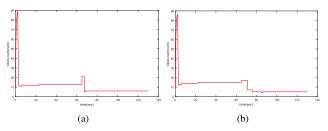


FIGURE 8. Self-organizing performance of SOBELC with $\tau_d=$ 10 and 20. (a) $\tau_d:\delta=$ 10. (b) $\tau_d:\delta=$ 20.

is set at 9 in the both levels of disturbance. Under $\delta=10$, the size increases rapidly just after the mobile robot starts to move; however, when the size reaches 95, it rapidly reduces and remain stable at 11 neurons. At the 65s, the trajectory pattern is changed so that the tracking error have a large shock. Therefore, the size increases to 20, then immediately returns and remains at 8. Under $\delta=20$, the increasing situation at the beginning is very similar to that under $\delta=10$: the size has rapid increasing and decreasing. However, after the 65s, the increasing of the size is not remarkable; in particular, the size remains at 11. The more neurons reveal that the SOBELC controller consumes more processing resources to handle the larger disturbances. Also, the dynamic size proves the SOBELC is able to improve the efficiency of resource utilizing for mobile robots.

C. DISCUSSIONS

The quantitative performance comparisons of using the TSK-CMAC, BELC, AF-BELC, and SOBEL for mobile robot control are summarized in Table 1. The accumulated RMSE values of the mobile robot's position, orientation, and velocity against time are used to measure the performance. The time-length of the accumulated RMSE is through the entire tracking process. This table shows the proposed SOBEL controller has the tracking performances under both levels of disturbances. Under $\delta = 20$, the performances of the TSK-CMAC, BELC, and AF-BELC controllers are lower than those under $\delta = 10$; such lower performances indicate that larger disturbances become a challenging task for these structure-fixed network-based controllers. In contrast, the self-organizing mechanism assigns computational resources based on the overall performances of the controllers. The dynamic neuron revising mechanism of the proposed SOBELC automatically adjusts the number of neurons in the hidden layer, in addition to the coupled robust controller, which leads to better tracking performances than the

TABLE 1. Comparison of TSK-CMAC, BELC, AF-BELC, and SOBELC controllers for mobile robot.

$ au_d: \delta$	TSK-CMAC	BELC	AF-BELC	SOBELC
10	0.5646	0.5823	0.5673	0.4805
20	0.5812	0.5834	0.5780	0.4681

TSK-CMAC, BELC, and AF-BELC controllers, especially when a larger disturbance presents. Note that the SOBELC controller is not allowed to produce too many new neurons to sacrifice the online in time performance, and the size of the hidden layer eventually converges at a reasonably low value. Therefore, the experimental investigations confirm that the proposed SOBELC is more capable in dealing with external disturbances and allocating computational resources.

VI. CONCLUSION

This paper focused on the trajectory tracking problem of a nonholonomic mobile robot. A self-organizing neural network was established by integrating the key components of self-organizing RBF and BELC networks. The proposed network was updated by following the brain emotional learning rules and Lyapunov stability theory; and a self-organizing mechanism can automatically add new hidden neurons and prune insignificant neurons. Moreover, the proposed new neural network used the Lyapunov stability theory to guarantee that the updating laws of the network's parameters ensure the convergence of the control system. Experimental results demonstrated the proposed neural network controller can better resist the influences of outside disturbance to improve the trajectory tracking performance and the efficiency of nonlinear function approximation.

There is still room to improve this research. Cutting-edge Type-2 fuzzy inference techniques may be employed in the proposed neural network controller, so as to obtain a better processing ability for uncertainties. More research effort is required to investigate why better control performances are achievable in an environment with larger disturbance. In addition, this initial report of the proposed approach focuses on the simulated mobile robot control only, but practical mobile robots are more appealing to fully discover the potential of the proposed network.

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