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# Low-Scattering Tri-Band Metasurface Using Combination of Diffusion, Absorption and Cancellation

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**ABSTRACT** In this paper, we proposed for the first time a novel mechanism to design a tri-band low-scattering metasurface. The radar cross section (RCS) reduction in the three operational bands relies on scattering diffusion, absorption, and scattering cancellation, respectively. The diffusion metasurface was realized by distributing scattering unit cells with different rotation angles. For the absorption, four units operating at different frequencies were combined to achieve a broadband absorption. The scattering cancellation was realized by arranging two out of phase elements in a chessboard configuration. The tri-band metasurface was constructed by directly combining the three metasurfaces in a single assembly. The simulated and measured results have a good agreement. The tri-band RCS reduction is well preserved up to  $45^{\circ}$  with polarization-independent performance. This paper provides an alternative approach to design the multi-band metasurface for stealth application.

**INDEX TERMS** Metasurface, tri-band, radar cross section reduction, diffusion, absorption, scattering cancellation.

# **I. INTRODUCTION**

Design of low-scattering platforms attracts more and more attentions due to the increasing demand of stealth technology in military applications. Reduction of radar cross section (RCS) is an important part of the stealth technology [1]. The conventional radar stealth technology relies on either shaping configurations or loading radar absorbing materials (RAMs) [1], which suffer from design complexity and bulky volume. Recently, metasurface was proposed to arbitrarily manipulate electromagnetic (EM) waves with low loss, low profile and broadband performance [2]-[4]. Using metasurface for RCS reduction has become a hot research topic recently. There are two kinds of metasurfacebased technologies: absorption [5]-[8] and scattering control [9]-[27]. The low-scattering metasurfaces have been comprehensively investigated from gigahertz to terahertz regime for both planar and conformal cases. The ultra-thin

metamaterial absorber dissipated energy into the dielectric at resonance frequencies [5], and the working bandwidth was further enhanced by loading lumped elements [6]–[8]. The scattering control for RCS reduction based on either scattering cancellation [9]–[13] or diffusion [14]–[23]. The scattering cancellation is achieved by arranging two out of phase unit cells in chessboard-like configuration, which can diminish specular reflection and redirect energies into diagonal and off-diagonal directions [9]-[13]. To mimic a diffusion process, the metasurface with random phase distribution can disperse incoming EM waves into various directions [14]–[23]. The desirable diffusion scattering can be realized by combing the coding concept and optimization algorithm [24]–[26], which are usually time-consuming in design, especially for a multi-bit coding metasurface. An easy and efficient strategy was proposed to design diffusion metasurface using a random combination of gradient subarrays

with different gradient directions [27]. However, all the above mentioned metasurfaces reduce RCS in a single band only. Some dual-band metasurfaces have been investigated later for RCS reduction [28]–[30]. The RCS reduction in two operational bands relies on either anomalous reflection [28] or scattering cancellation [29]. The two operational bands in this kind of dual-band metasurfaces can be influenced by each other, leading to a limited frequency ratio. A dual-band lowscattering metasurface was later designed using a combination of two single-band metasurfaces based on absorption and diffusion, respectively [30]. The influence between two operational bands is independent in this case. To date, the tri-band metasurface for RCS reduction is still not explored and reported.

In this paper, we proposed a tri-band metasurface for RCS reduction. The three operational bands operate using scattering diffusion, absorption and scattering cancellation, respectively. The underlying working mechanisms of each operational band are quite different, and the three bands have little influence on each other. To realize the diffusion scattering performance, we use an ultra-wideband polarization rotation unit cell as the basic coding meta-atom, and the phase distribution was optimized with the aid of a hybrid algorithm. In the absorption case, we adopt four unit cells as a supercell to enhance the absorption bandwidth. The proposed low-scattering metasurface features tri-band, wide-angle and polarization-independent performance.



**FIGURE 1.** (a)The front view of the polarization rotatable unit cell; (b) the co-polarized reflection magnitudes and (c) PCRs of three cases; (d) the cross-polarized reflection coefficients of the unit cell and its mirror counterpart.

#### **II. UNIT CELL DESIGN**

# A. ULTRA-WIDEBAND POLARIZATION ROTATABLE UNIT CELL

The polarization rotatable unit cell, as illustrated in Fig. 1(a), is composed of three metallic layers and two dielectric layers. Both two metallic patterns comprised of a pair of L-shaped structure. The two identical dielectric layers have a dielectric

constant of 2.2, a thickness of 2 mm and a tangent loss of 0.006. The structural parameters are optimized as p =6 mm,  $w_1 = 0.5$  mm,  $l_1 = 4.1$  mm,  $d_1 = 5$  mm,  $w_2 =$ 0.2 mm,  $l_2 = 1.7$  mm and  $d_2 = 2.2$  mm. Here, the duallayer configuration enhances the bandwidth as well as the polarization conversion ratio (PCR). To show the bandwidth and efficiency enhancement of the proposed dual-layer unit cell, the co-polarized reflection magnitudes and PCRs of three different cases are depicted in Fig. 1(b) and (c). The detail structures of the three cases are shown in the inset of Fig. 1(b). Note that the co-polarized reflection magnitude is below  $-10 \, dB$  from 11 GHz to 20.2 GHz for case 1. In case 2, the bandwidth is enhanced to 9.1-24.7 GHz after covering a superstrate, as presented by red dashed line. It is clearly observed that the reflection magnitude is between -15 dBand -10 dB from 13.6 GHz to 24.7 GHz, indicating that the PCR is less than 0.95 within this bandwidth. To further improve the PCR, another pair of L-shaped structure was printed on the superstrate as the case 3. The reflection magnitude is below -15 dB from 9.1 GHz to 25.2 GHz, and the PCR is higher than 0.95 within the whole bandwidth, revealing that the incident polarization is converted into its orthogonal polarization with high efficiency. In other words, the polarization rotatable unit cell can achieve a  $+90^{\circ}$ polarization conversion of incident EM wave. Analogously, the mirror counterpart of the unit cell can provides a  $-90^{\circ}$ polarization rotation with the same amplitude. Therefore, the cross-polarized reflection phase of the unit cell is strictly out of phase with its mirror counterpart and the reflection magnitudes are greater than 0.95 from 9.2 GHz to 25.2 GHz, as shown in Fig. 1(d). Thus, the proposed unit cell and its mirror counterpart can be employed as the basic meta-atoms of a 1-bit coding metasurface.



FIGURE 2. The schematic of (a) absorption unit cell and (b) absorption supercell.

# **B. BROADBAND ABSORPTION SUPERCELL**

As shown in Fig. 2(a), the absorption unit cell has a sandwichlike structure, namely, a top metallic structure, a middle F4B substrate with a thickness of 4 mm and a dielectric constant of 2.2, and a bottom metallic ground. The optimum dimensions of the unit cell are  $l_3 = 22$  mm, w = 1 mm. The effects of the loading position and resistance were investigated firstly. As shown in the Fig. 3(a), the absorption frequency



FIGURE 3. The reflection magnitudes (a) with different loading positions and (b) with different resistances.

can be manipulated by changing the loading position. One can clearly find that the absorption frequency is decreasing with the increment of parameter r. In addition, the absorption performance was affected by the values of resistors. Note from Fig. 3(b), the best absorption performance can be obtained when the resistance is  $1.2 \text{ k}\Omega$ . However, the 10 dB fractional bandwidth of absorption unit cell is only about 7%, which is a little narrow. To enhance the absorption bandwidth, we combined four unit cells working at adjacent frequencies to be an absorption supercell, as shown in Fig. 2(b). The initial loading positions are 7.5 mm, 8 mm, 8.5 mm, and 9 mm, and the initial resistances are  $1.2 \text{ k}\Omega$ . Due to the coupling between adjacent unit cells, both the loading position and resistance should be optimized to achieve preferred broadband performance. The final optimum positions are  $r_1 = 7.2$  mm,  $r_2 = 8$  mm,  $r_3 =$ 8.3 mm, and  $r_4 = 8.8$  mm, and resistances are  $R_1 = 2.7$  k $\Omega$ ,  $R_2 = 1.5 \text{ k}\Omega$ ,  $R_3 = 4.7 \text{ k}\Omega$ , and  $R_4 = 4.7 \text{ k}\Omega$ . The reflection magnitude is below -10 dB from to 5.52 GHz to 6.37 GHz, as shown by black solid line in Fig. 3(a). It is clearly seen that the fractional bandwidth has been enhanced to 14.3% in combining four neighboring absorption C-band by frequencies.



**FIGURE 4.** (a) The schematic of crossed dipole for scattering cancellation and (b) the reflection coefficients for I4=34.8 mm and I4=36.1 mm.

# C. CROSSED DIPOLE FOR SCATTERING CANCELLATION

To reduce RCS in S-band based on scattering cancellation, we still use crossed dipole as the basic meta-atom, as shown in Fig. 4(a). The substrate layer is identical to that of the absorption unit cell. To realize 10 dB RCS reduction, the phase difference between two basic meta-atoms should be  $180^{\circ} \pm 37^{\circ}$ , while the magnitudes should be equal to each other [8]. The reflection phase can be tuned by changing the

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parameter  $l_4$ , and  $180^\circ \pm 37^\circ$  phase difference was achieved when  $l_4 = 34.8$  mm and 36.1 mm from 2.95 GHz to 3.06 GHz. Moreover, as shown in Fig. 4(b), almost total reflection is achieved due to the metallic ground on the back. The maximum phase difference (217°) is set at the center frequency 3 GHz to enhance the operational bandwidth. As a result, a metasurface constructed by arranging one type of crossed dipoles surrounded by the other type will achieve a null in boresight direction due to destructive interference around 3 GHz.

# **III. OVERALL METASURFACE DESIGN**

#### A. DESIGN PRINCIPLE

The scattering pattern of metasurface for scattering manipulation illuminated by plane wave can be regarded as the radiation pattern of an array. Thus, the scattering pattern can be theoretically calculated by using array theory. The scattering pattern of the whole metasurface was the composition of the contributions of all the elements, and is generally expressed as

$$\vec{E}_{total} = \sum_{m=1}^{M} \sum_{n=1}^{N} \vec{E}_{m,n} \\ \times \exp\left[jk_0 \left(md_x \sin\theta \cos\varphi + nd_y \sin\theta \sin\varphi\right)\right]$$
(1)

where  $\vec{E}_{m,n}$  is the electric far-field pattern of the element (m, n), and expressed by  $\vec{E}_{m,n} = \vec{E}_1 \exp(j\varphi_{mn})$ , where  $\vec{E}_1$  is the electric far-field pattern of the basic element and  $\varphi_{mn}$  is the relative phase of the element (m, n).  $\theta$  and  $\varphi$  are the elevation and azimuth angles, respectively;  $d_x$  and  $d_y$  denote the element period along the *x* and *y* directions, respectively;  $k_0$  is the wave number in free space. Substituting  $\vec{E}_{m,n} = \vec{E}_1 \exp(j\varphi_{mn})$  into Eq. (1) leads to

$$\vec{E}_{total} = \vec{E}_1 \sum_m \sum_n \exp\left[j\left(\varphi_{mn} + k_0 m d_x \sin\theta \cos\varphi + k_0 n d_y \sin\theta \sin\varphi\right)\right]$$
$$= \vec{E}_1 \cdot AF \tag{2}$$

where AF is the array factor. Accordingly, the scattering pattern is determined by the element phase distribution. To diffuse the scattering into numerous directions, the simplest method is to generate a random phase matrix to determine the diffusion metasurface configuration. However, this method cannot guarantee an optimal result. One method to solve this issue is to use an optimization algorithm to find the optimal coding phase matrix.

Simulated Annealing algorithm is a method for local searching proposed by Kirkpatrick *et al.* [31]. It has advantages of simple description and high efficiency. It begins with an initial solution that is randomly modified in an iterative process. The main parameters of Simulated Annealing algorithm are the initial temperature T, the decreasing rate in each iteration  $\alpha$ , the final temperature T<sub>f</sub>, the number of iterations I and the merit function. The goal of the optimization is to find an optimal coding matrix (M<sub>opt</sub>) that results in the smallest maximum value of the scattered fields. Therefore, the merit



FIGURE 5. The phase distribution of optimal diffusion metasurface (left) and chessboard metasurface (right).

function can be expressed by  $F(M_{opt}) = min(AF_{max})$ , where  $AF_{max}$  is the maximum value of AF corresponding to given coding matrix. In our case, the size of the coding matrix is  $12 \times 12$ , and the element of the matrix is either 0 or  $\pi$ . One should note that the numbers of 0 is equal to that of  $\pi$ , leading to the information entropy of the coding matrix reaches maximum [32]. Moreover, the parameters T,  $\alpha$ , T<sub>f</sub> and I are set as 100, 0.9, 0 and 500, respectively. The optimal coding matrix is shown in the left part of Fig. 5, and the corresponding theoretical scattering pattern is depicted in Fig. 6(a). It can be clearly seen that the scattered energy has been diffused into numerous directions, leading to both monostatic and bistatic RCS reduction due to the energy conversation.



FIGURE 6. The theoretical results of 3D scattering patterns for (a) diffusion metasurface and (b) chessboard metasurface.

The phase distribution for scattering cancellation is shown in the right part of Fig. 5. As seen, 0 and  $\pi$  is intersected with each other to form a chessboard configuration. Fig. 6(b) shows the scattering pattern of the chessboard metasurface. Note that the backward reflection is diminished and four scattered beams occur along diagonal and off-diagonal directions.

# **B. CONFIGURATION**

Based on the proposed three kinds of unit cells and the design principle, we design three individual metasurfaces. Firstly, the diffusion metasurface was designed based on the proposed polarization rotatable unit cell and the optimal coding matrix. Due to fact that the coupling between adjacent unit cells in the random distribution is different from the boundary applied in the unit cell simulation, each diffusion supercell is formed by  $4 \times 4$  identical unit cells to reduce this effect. Thus, the diffusion metasurface constructed by  $48 \times 48$  unit cells, which occupies a size of  $288 \times 288$  mm<sup>2</sup>.

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Secondly, the absorber array was built by periodically distributing the broadband absorption supercells. The absorber array contains  $6 \times 6$  broadband absorption supercells, leading to the same size as the diffusion metasuface. Thirdly, we also employ  $2 \times 2$  crossed dipoles as a basic element in designing the chessboard configuration. Therefore, the chessboard metasurface contains  $6 \times 6$  crossed dipoles with a size of  $288 \times 288$  mm<sup>2</sup>. Finally, the tri-band metasurface for RCS reduction was constructed by directly combining the three individual metasurfaces, the upper-left part of the tri-band metasurface was shown in the Fig. 7, which contains  $2 \times 2$ diffusion supercells, an absorption supercell and a crossed dipole.



**FIGURE 7.** The schematic of the upper-left part of the tri-band metasurface.

# **IV. NUMERICAL AND EXPERIMENTAL RESULTS**

# A. NUMERICAL RESULTS

Both three individual metasurfaces and tri-band metasurface are simulated in CST Microwave Studio. Only x-polarized incidence is considered in the individual metasurface simulation. A metallic plate with the same size is introduced as the reference to show the RCS reduction performance. Fig. 8(a) depicts the RCS reductions of three individual metasurfaces. As expected, the diffusion metasurface can reduce RCS more than 10 dB from 9.5 GHz to 25.1 GHz, which is consistent with the bandwidth of the co-polarized reflection magnitude less than -10 dB. The red dashed line denotes the RCS reduction performance of the absorber array, proving that the RCS of the absorber array is less than that of the metallic plate by more than 10 dB within 5.64-6.4 GHz. The RCS reduction performance of chessboard metasurface was presented by the blue dotted line, showing that the 10 dB RCS reduction can be obtained within 2.96-3.06 GHz, which agrees with the effective phase difference bandwidth (2.95-3.06 GHz). Fig. 8(b) gives the RCS of the tri-band metasurface under illumination by both x- and y-polarized plane waves. It can be referred from Fig. 8(b) that RCS can be reduced by more than 10 dB within 2.92-3 GHz, 5.57-6.3 GHz and 9.5-25.5 GHz. It is interesting to find that the operational bandwidth of tri-band metasurface is consistent with the operational bands of three



FIGURE 8. (a) The RCS reduction of three individual metasurfaces and (b) the RCS response tri-band metasurface for both x- and y-polarized incidence.

individual metasurfaces, revealing that each RCS reduction mechanism is little affected by other mechanisms. In addition, the RCS reduction performance for x-polarized incidence is similar to that of y-polarized incidence, proving that the triband metasurface features polarization-independence.



FIGURE 9. 3-D scattering patterns of (a-c) the tri-band metasurface and (d-f) metallic plate at (a, d) 14.9 GHz; (b, e) 6.14 GHz and (c, f) 2.96 GHz.

To better illustrate the underlying mechanisms at three operational bands, we further investigate the three dimensional (3-D) scattering patterns at three frequencies where the best reductions occur, as shown in Fig. 9. The scattering

patterns of metallic plate in the same color map scale are also given for comparison. Obviously, the maximum scattered intensities of metasurface are lower than that of the metallic plate. As expected, a diffusion scattering is realized at high frequency, the incident EM wave was absorbed at middle frequency, and the specular reflection can be diminished at low frequency. Moreover, the simulated 3-D scattering patterns at high and low frequencies are consistent with the theoretical results in Fig. 6. As a comparison, the specular reflection occurs at three frequencies for the metallic plate. It is worth noting that though the scattered intensity of metasurface gets enhanced at some directions, the values are too small to be detected.



**FIGURE 10.** The RCS reductions under oblique incidence for (a) x-polarized and (b) y-polarized.

To provide a comprehensive investigation about the angular performance of the proposed metasurface, the RCS reductions under oblique angles are simulated and shown in the Fig. 10. Note that though the RCS reduction performance degrade due to the alteration of phase response at high frequency range, a tri-band RCS reduction is realized for both x- and y-polarized until the incident angle up to 45°. The oblique incidence has little influence on the scattering cancellation performance and absorption performance, which is attributed to the intrinsic characteristic of crossed dipole structure.



FIGURE 11. The photographs of (a) fabricated sample and (b) measurement setup.

### **B. FABRICATION AND MEASUREMENT**

To verify the designed tri-band RCS reduction experimentally, a metasurface sample was fabricated using printed circuit board technology and measured in the microwave anechoic chamber, as shown in Fig. 11. The fabricated sample occupies  $288 \times 288$  mm<sup>2</sup>, following the model used in the simulation. The top and bottom layers are depicted in the

upper half and lower half parts of Fig. 11(a), respectively. Moreover, the top layer is stacked above the bottom layer by glue. The glue layer was only applied on the four corners of the square sample, and its thickness is too thin to consider the influence. The specular reflection was measured by using free space method in anechoic chamber. As shown in Fig. 11(b), a pair of wideband horn antennas was used as transmitter and receiver and connected to the two ports of vector network analyzer (AV 3672B). The time-domain gating function in the network analyzer was used to eliminate the interference of the environment in the measurement. Hence, the reflection of the metasurface can be evaluated by the transmission coefficient S<sub>12</sub> of two horn antennas. The transmitter and receiver were tilted at an angle of 5° about the normal direction of sample to prevent the interference effect between incident and reflected waves. A metallic plate with the same size was also measured as a reference to show the reflection suppression of the designed metasurface. The reflection was measured up to 18 GHz due to the limitation of the measurement condition. To provide a clear insight of the comparison of measured and simulated results, the results are separated into two frequency ranges, as shown in Figs. 12(a) and (b), respectively. As expected, tri-band RCS reduction can be realized for both x- and y-polarized incidence. The measured results coincide well with the simulated results, except for the discrepancy in the absorption band, which can be attributed to the imperfect soldering of embedded resistors. In addition, the difference between real and assumed characteristic values of the F4B dielectric substrate and misalignment of distinction are other reasons for the distinction between simulated and measured RCS reduction spectrums. Nevertheless, tri-band RCS reduction has been experimentally verified in a reasonable manner.



**FIGURE 12.** The measured RCS reductions under normally incidence for both x-polarized and y-polarized at (a) 2-8 GHz and (b) 8-18 GHz.

# **V. CONCLUSION**

In this study, a tri-band metasurafce for RCS reduction was designed, fabricated and measured. The reduction mechanisms in three operational bands are different. The RCS from X-band to K-band was suppressed by randomly distributing polarization rotatable unit cells. The C-band RCS reduction was attributed to the absorption of incoming wave. Moreover, the scattering cancellation is a reason for RCS reduction in S-band. The RCS reduction performance in each band is independent to each other due to the different mechanisms, leading to conveniently manipulation of operational band as

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