

Received May 14, 2017, accepted June 10, 2017, date of publication July 3, 2017, date of current version July 24, 2017.

Digital Object Identifier 10.1109/ACCESS.2017.2722417

# Precise Characterizations of the Stability Margin in Time-Domain Space for Planar Systems Undergoing Periodic Switching

ZUNBING SHENG<sup>1</sup> AND WEI QIAN<sup>2</sup>

<sup>1</sup>School of Mechanical and Electrical Engineering, Heilongjiang University, Harbin 150080, China

<sup>2</sup>School of Electrical Engineering and Automation, Henan Polytechnic University, Jiaozuo 454000, China

Corresponding author: Zunbing Sheng (zunbingsheng@163.com)

This work was supported in part by the National Natural Science Foundation of China under Grant 61573131 and in part by the Natural Science Foundation of Heilongjiang Province of China under Grant LC201428. The work of W. Qian was supported by the Science and Technology Innovation Talents Project of Henan Province under Grant 164100510004.

**ABSTRACT** This paper addresses the stability analysis problem for planar periodic switching systems. We characterize the stability margin in the space constituted by the dwell times of the subsystems, by which we can assess the asymptotic stability of the overall system in the necessary and sufficient sense. The mutual constraint conditions on the dwell times in nature depend on the type of equilibrium point of each subsystem. The stability conditions are expressed in terms of a family of transcendental inequalities, which can be numerically solved and precisely depicted in the time-domain space. An example is worked out in detail to illustrate the theoretical results.

**INDEX TERMS** Planar dynamical systems, periodic switching, asymptotic stability.

## I. INTRODUCTION

The stability problem of switching systems has attracted considerable attention during the past two decades. A switching system is composed of several systems with a switching signal to orchestrate among them. In studying such systems, we have learned that switching can produce very interesting and complex dynamical behaviors that might be beyond our imagination.

The stability analysis problem of switching systems can be classified into two categories; see [7], [12]. Firstly, we are concerned with the so-called absolute stability problem that whether it is possible to construct a switching law to achieve destabilization. In the spirit of variational approach, the absolute stability problem requires to investigate if the extremal trajectory is convergent, which is constructed so that each point on it gets away from the origin as far as possible. If so, then we can confirm that the overall switching system will stay stable for an arbitrary switching signal. Secondly, the stability analysis problem is purely intended to characterize the evolution of switching signal over time and its effect on stability. An important aspect of the time-evolution of switching signal is the density of the switching points distributed within an interval of time, which can be captured in terms of average dwell-times. To guarantee the stability of the overall system, conventionally, the multiple

Lyapunov functions approach is used to derive constraint conditions on the average dwell-time; see, e.g., [7], [11], [12], [14]. The mechanism of this methodology is to make the period between any two successive switching points long enough so as to allow the overshoot caused by switching to fade. However, this may lead to conservatism because without knowing the phase of the overshoot, we actually suppress it simply according to the worst case.

In the present paper, we shall investigate the stability problem for a switching system composed of two planar linear subsystems, which are triggered into activation alternatively and periodically. In the research field of switching systems, the planar case indeed constitutes the most developed branch because a set of dedicated methods allow us to gain deep insight into the stability problem beyond simply using a general method to address it; see [2]–[5], [9], [10]. Within the context of periodic switching, we shall establish the mutual constraint conditions on the dwell-times of the subsystems to ensure the stability of the overall system. An invariant that plays a key role in describing the interrelations of the subsystem matrices is involved in the stability criteria. This signifies that the overall switching dynamics in a strong way depends on the interrelation of the subsystem matrices as well as the types of their eigenvalues. Moreover, the derived stability criteria do not depend on the particular choice of

the coordinates and, therefore, allow a system to speak for itself.

As we know, the stability conditions derived through the multiple Lyapunov functions approach usually account for the effect of average dwell-time on stability in a linear manner. The result in [8] showed that conditions that are linear in the average dwell-time may be very restricted. Actually, the results posed in the present paper demonstrate that the dwell-times of the subsystems actually influence the overall switching dynamics in a highly nonlinear manner. Therefore, we are able to precisely determine the stability margin in the space constituted by the dwell-times and hence assess stability in a necessary and sufficient sense. On the other hand, to ensure this stability, the existing methods usually require the stable subsystems to play the dominant role in the overall switching dynamics, among others. For the sake of generality, we will not suppose the subsystems to be stable because this assumption turns out to be restricted. To demonstrate the theoretical results, we shall construct an example to show that two unstable subsystems with quite long periods to stay active may generate stable dynamics.

The remainder of this paper is organized as follows. In Section 2, we introduce the basic definitions and formulate the problem. In Section 3, we present the stability conditions in the time-domain space and give some explanations to the matters relevant to the results. An example is included in Section 4 to illustrate the theoretical results. Finally, this work is briefly summarized in Section 5.

*Notation:* Throughout this paper, we use the following notations. Let  $\det(X)$  and  $\text{tr}(X)$  be the determinant and trace of  $X$ , respectively. Let  $[X, Y] = XY - YX$  be the Lie commutator of  $X$  and  $Y$ . Besides, we write  $I$  and  $j$  for the identity matrix and the unit of imaginary numbers, respectively.

## II. PROBLEM FORMULATION

Consider the planar switching system described as follows

$$\dot{x}(t) = A_{\sigma(t)}x(t), \quad x \in \mathbf{R}^2. \quad (1)$$

This model is composed of two linear subsystems and the switching signal  $\sigma(t)$  to orchestrate among them. Let  $\Delta_i = \text{tr}(A_i)^2 - 4 \det(A_i)$  be the discriminant of the characteristic polynomial of  $A_i$ . Due to space limitations, we suppose  $\Delta_1 \Delta_2 \neq 0$  throughout the paper. According to the evolution over time, we can represent a switching signal as follows

$$\left\{ (\sigma(t_0), t_0 = 0), (\sigma(t_1), t_1), \dots, (\sigma(t_k), t_k) \Big| \lim_{k \rightarrow +\infty} t_k = +\infty \right\}. \quad (2)$$

This sequential form indicates that at the switching time  $t_k$ , the switching signal takes on a value  $\sigma(t_k)$  in the binary set  $\{1, 2\}$  to trigger the two subsystems into activation alternatively and successively. The dynamics of the system then is determined by the transition matrix as follows

$$e^{A_{\sigma(t_k)}(t-t_k)} e^{A_{\sigma(t_{k-1})}(t_k-t_{k-1})} \dots e^{A_{\sigma(t_0)}(t_1-t_0)}, \quad t \in [t_k, t_{k+1}). \quad (3)$$

Indeed, switching system (1) is asymptotically stable if and only if the transition matrix in (3) has its eigenvalues inside the unit circle of the complex plane for all  $t \geq 0$ ; see [15]. To characterize the distribution of the eigenvalues of the transition matrix, we suppose the switching signals to obey the following regular property.

*Definition 1:* A switching rule is said to be *periodic* if the interval between two successive switching points satisfies

$$t_{k+1} - t_k = \begin{cases} \tau_1, & \sigma(t_k) = 1 \\ \tau_2, & \sigma(t_k) = 2 \end{cases}, \quad k \geq 0$$

where  $\tau_1 > 0$  and  $\tau_2 > 0$  are referred to as the dwell-times of the individual subsystems. When undergoing periodic switching, system (1) is referred to as a *periodic switching system*.

Hereafter, the switching signals will be considered to be periodic, and the system in (1) will be considered to be a periodic switching system unless otherwise specified. Therefore, our first observation is that when driven by a periodic switching signal, system (1) is asymptotically stable if and only if the eigenvalues of  $e^{A_1 \tau_1} e^{A_2 \tau_2}$  are located inside the unit circle of the complex plane. We now put this observation in the following way.

*Definition 2:* The periodic switching system in (1) is said to be asymptotically stable if the unit of its transition matrix  $e^{A_1 \tau_1} e^{A_2 \tau_2}$  is Schur stable.

Within this context, the paper is devoted to characterize the joint effects of the dwell-times  $\tau_1$  and  $\tau_2$  on state-transitions over time and, therefore, deriving their mutual constraint relation guaranteeing system (1) to be asymptotically stable.

## III. MAIN RESULTS

The following fact plays a key role in deriving our results, which can be directly checked.

*Lemma 1* [4]: For two-dimensional square matrices  $X$  and  $Y$ , we have the following identity

$$\det(X + Y) = \det(X) + \det(Y) + \text{tr}(X)\text{tr}(Y) - \text{tr}(XY).$$

Using Lemma 1 and the fact  $\det(e^X) = e^{\text{tr}(X)}$  yields

$$m(\mu) := \det(e^{A_1 \tau_1} e^{A_2 \tau_2} - \mu I) = \mu^2 - \mu \text{tr}(e^{A_1 \tau_1} e^{A_2 \tau_2}) + e^{\text{tr}(A_1) \tau_1} e^{\text{tr}(A_2) \tau_2}. \quad (4)$$

The discriminant of the parabolic function  $m(\mu)$  is

$$[\text{tr}(e^{A_1 \tau_1} e^{A_2 \tau_2})]^2 - 4e^{\text{tr}(A_1) \tau_1} e^{\text{tr}(A_2) \tau_2}.$$

By the distribution property of the roots of parabolic functions, we have the following result.

*Lemma 2:* The equation  $m(\mu) = 0$  has both roots inside the unit circle of the complex plane (i.e.,  $e^{A_1 \tau_1} e^{A_2 \tau_2}$  is Schur stable) if and only if  $\tau_1$  and  $\tau_2$  satisfy

$$\begin{aligned} [\text{tr}(e^{A_1 \tau_1} e^{A_2 \tau_2})]^2 - 4e^{\text{tr}(A_1) \tau_1} e^{\text{tr}(A_2) \tau_2} &< 0, \\ e^{\text{tr}(A_1) \tau_1} e^{\text{tr}(A_2) \tau_2} &< 1; \end{aligned} \quad (5)$$

or,

$$\begin{aligned} [\text{tr}(e^{A_1\tau_1} e^{A_2\tau_2})]^2 - 4e^{\text{tr}(A_1)\tau_1} e^{\text{tr}(A_2)\tau_2} &\geq 0, \\ m(1) > 0, \quad m(-1) > 0, \\ -1 < \frac{1}{2} \text{tr}(e^{A_1\tau_1} e^{A_2\tau_2}) < 1. \end{aligned} \quad (6)$$

*Remark 1:* The inequalities in (5) correspond to that  $e^{A_1\tau_1} e^{A_2\tau_2}$  has a pair of complex eigenvalues located inside the unit circle of the complex plane, while the inequalities in (6) to that  $e^{A_1\tau_1} e^{A_2\tau_2}$  has real eigenvalues within  $(-1, 1)$ . Additionally, the inequalities in (6) can be equivalently rewritten as

$$\begin{aligned} |\text{tr}(e^{A_1\tau_1} e^{A_2\tau_2})| &\geq 2e^{\text{tr}(A_1)\tau_1/2} e^{\text{tr}(A_2)\tau_2/2}, \\ |\text{tr}(e^{A_1\tau_1} e^{A_2\tau_2})| &< 1 + e^{\text{tr}(A_1)\tau_1} e^{\text{tr}(A_2)\tau_2}, \end{aligned}$$

and

$$|\text{tr}(e^{A_1\tau_1} e^{A_2\tau_2})| < 2.$$

The compatibility of this set of inequalities is guaranteed by the elementary inequality  $2|\alpha| < 1 + \alpha^2, \forall \alpha \in \mathbf{R}$ .

One can straightforwardly derive the following conclusions.

*Corollary 1:* If  $\text{tr}(A_1) < 0$  and  $\text{tr}(A_2) < 0$  (a particular case is  $A_1$  and  $A_2$  both are Hurwitz stable), then system (1) is asymptotically stable if and only if  $\tau_1$  and  $\tau_2$  can satisfy

$$|\text{tr}(e^{A_1\tau_1} e^{A_2\tau_2})| < 2e^{\text{tr}(A_1)\tau_1/2} e^{\text{tr}(A_2)\tau_2/2};$$

or

$$\begin{aligned} |\text{tr}(e^{A_1\tau_1} e^{A_2\tau_2})| &\geq 2e^{\text{tr}(A_1)\tau_1/2} e^{\text{tr}(A_2)\tau_2/2}, \\ |\text{tr}(e^{A_1\tau_1} e^{A_2\tau_2})| &< 1 + e^{\text{tr}(A_1)\tau_1} e^{\text{tr}(A_2)\tau_2}. \end{aligned}$$

*Corollary 2:* There exist dwell-times  $\tau_1$  and  $\tau_2$  that guarantee system (1) to be asymptotically stable only if at least one of  $\text{tr}(A_1)$  and  $\text{tr}(A_2)$  is negative.

*Proof:* As  $\text{tr}(A_1) \geq 0$  and  $\text{tr}(A_2) \geq 0$ , the inequalities in (5) will fail. At the same time, the inequalities in (6) become

$$2e^{\text{tr}(A_1)\tau_1/2} e^{\text{tr}(A_2)\tau_2/2} \leq |\text{tr}(e^{A_1\tau_1} e^{A_2\tau_2})| < 2.$$

Obviously, this is infeasible. The assertion then is proven by this contradiction.  $\square$

*Remark 2:* From the perspective of assigning the distribution of the eigenvalues of  $e^{A_1\tau_1} e^{A_2\tau_2}$ , we now provide an explanation for the difference between the stability analysis problem and the absolute stability problem. When confined to periodic switching signals, the stability analysis problem requires us to determine the range of  $\tau_1$  and  $\tau_2$ , for which  $e^{A_1\tau_1} e^{A_2\tau_2}$  is Schur stable. Instead, the absolute stability problem requires us to establish some conditions for  $A_1$  and  $A_2$  so as to guarantee  $e^{A_1\tau_1} e^{A_2\tau_2}$  to be Schur stable for any  $\tau_1$  and  $\tau_2$ ; see, e.g., [1], [6], [13], [16].

From the Cayley-Hamilton theorem, one can deduce the following expansions (cf. [8]):

$$e^{A_1\tau_1} = f_1(\tau_1)I + g_1(\tau_1)A_1,$$

and

$$e^{A_2\tau_2} = f_2(\tau_2)I + g_2(\tau_2)A_2.$$

Therefore, we get that

$$\begin{aligned} e^{A_1\tau_1} e^{A_2\tau_2} &= f_1(\tau_1)f_2(\tau_2)I + f_2(\tau_2)g_1(\tau_1)A_1 \\ &\quad + f_1(\tau_1)g_2(\tau_2)A_2 + g_1(\tau_1)g_2(\tau_2)A_1A_2, \end{aligned} \quad (7)$$

and hence

$$\begin{aligned} \text{tr}(e^{A_1\tau_1} e^{A_2\tau_2}) &= 2f_1(\tau_1)f_2(\tau_2) + f_2(\tau_2)g_1(\tau_1) \text{tr}(A_1) \\ &\quad + f_1(\tau_1)g_2(\tau_2) \text{tr}(A_2) + g_1(\tau_1)g_2(\tau_2) \text{tr}(A_1A_2). \end{aligned} \quad (8)$$

According to the type of the equilibrium point of each subsystem, for the function  $g_i(s)$ , we have

$$g_i(s) = \begin{cases} \frac{1}{\sqrt{\Delta_i}/2} e^{\text{tr}(A_i)s/2} \sinh(\sqrt{\Delta_i}s/2), & \Delta_i > 0, \\ \frac{1}{\sqrt{-\Delta_i}/2} e^{\text{tr}(A_i)s/2} \sin(\sqrt{-\Delta_i}s/2), & \Delta_i < 0. \end{cases} \quad (9)$$

Correspondingly, for the function  $f_i(s)$ , we have

$$\begin{aligned} f_i(s) = \dot{g}_i(s) - \text{tr}(A_i)g_i(s) &= e^{\text{tr}(A_i)s/2} \\ &\times \begin{cases} \left[ -\frac{\text{tr}(A_i)}{\sqrt{\Delta_i}} \sinh(\sqrt{\Delta_i}s/2) + \cosh(\sqrt{\Delta_i}s/2) \right], \\ \Delta_i > 0, \\ \left[ -\frac{\text{tr}(A_i)}{\sqrt{-\Delta_i}} \sin(\sqrt{-\Delta_i}s/2) + \cos(\sqrt{-\Delta_i}s/2) \right], \\ \Delta_i < 0. \end{cases} \end{aligned} \quad (10)$$

Let  $\lambda_{1,2}^i$  denote the eigenvalues of  $A_i$ ; then,

$$\lambda_{1,2}^i = \begin{cases} (\text{tr}(A_i) \pm \sqrt{\Delta_i})/2, & \Delta_i > 0, \\ (\text{tr}(A_i) \pm j\sqrt{-\Delta_i})/2, & \Delta_i < 0. \end{cases} \quad (11)$$

Therefore, by Euler's formula, we can rewrite the expressions of  $g_i(s)$  and  $f_i(s)$  as follows

$$g_i(s) = (e^{\lambda_1^i s} - e^{\lambda_2^i s})/\sqrt{\Delta_i},$$

and

$$f_i(s) = \frac{1}{2} \left[ -\frac{\text{tr}(A_i)}{\sqrt{\Delta_i}} (e^{\lambda_1^i s} - e^{\lambda_2^i s}) + (e^{\lambda_1^i s} + e^{\lambda_2^i s}) \right].$$

We shall derive the constraint conditions on the dwell-times  $\tau_1$  and  $\tau_2$  by exhaustively categorizing all possible combinations of the subsystems. To this end, we now present the parameter

$$\mathcal{K} := 2 \frac{\text{tr}(A_1A_2) - \frac{1}{2}\text{tr}(A_1) \text{tr}(A_2)}{\sqrt{|\Delta_1\Delta_2|}}. \quad (12)$$

which remains invariant for changing coordinates. As shown in [3] and [4],  $\mathcal{K}$  contains important information about the interrelation of  $A_1$  and  $A_2$ . In particular, the following fact exposes its connection with the Lie commutator of  $A_1$  and  $A_2$ .

Lemma 3 [4]:

$$\det([A_1, A_2]) = \begin{cases} \frac{1}{4} (1 - \mathcal{K}^2) \Delta_1 \Delta_2, & \Delta_1 \Delta_2 > 0 \\ \frac{1}{4} (1 + \mathcal{K}^2) \Delta_1 \Delta_2, & \Delta_1 \Delta_2 < 0. \end{cases}$$

In particular, if  $\Delta_1, \Delta_2 < 0$ , then there must be  $|\mathcal{K}| \geq 1$ .

In what follows, we shall show that  $\mathcal{K}$  is naturally involved in the switching dynamics. In light of Lemma 2 and Remark 2, it turns out to be a fundamental problem to calculate  $\text{tr}(e^{A_1 \tau_1} e^{A_2 \tau_2})$  for assessing the stability of system (1). To this end, we apply the formulation in (8), whose expressions rely on the specified eigenstructure of each subsystem. Furthermore, letting the dwell-times  $\tau_1$  and  $\tau_2$  be subject to the inequalities derived from (5) and (6) then is equivalent to confining them to the mutual constraint conditions guaranteeing system (1) to be asymptotically stable.

CASE I:  $\Delta_1 > 0$  and  $\Delta_2 > 0$ . This implies that both  $A_1$  and  $A_2$  have distinct real eigenvalues. Accordingly, we have

$$\begin{aligned} \text{tr}(e^{A_1 \tau_1} e^{A_2 \tau_2}) &= 2e^{\text{tr}(A_1)\tau_1/2} e^{\text{tr}(A_2)\tau_2/2} \\ &\times [\mathcal{K} \sinh(\sqrt{\Delta_1} \tau_1/2) \sinh(\sqrt{\Delta_2} \tau_2/2) \\ &+ \cosh(\sqrt{\Delta_1} \tau_1/2) \cosh(\sqrt{\Delta_2} \tau_2/2)]. \end{aligned} \quad (13)$$

Proposition 1: If both  $A_1$  and  $A_2$  have distinct eigenvalues, the periodic switching system in (1) is asymptotically stable if and only if the dwell-times  $\tau_1$  and  $\tau_2$  satisfy the following inequalities

$$|\mathcal{K} \sinh(\sqrt{\Delta_1} \tau_1/2) \sinh(\sqrt{\Delta_2} \tau_2/2) + \cosh(\sqrt{\Delta_1} \tau_1/2) \cosh(\sqrt{\Delta_2} \tau_2/2)| < 1, \quad (14)$$

$$e^{\text{tr}(A_1)\tau_1} e^{\text{tr}(A_2)\tau_2} < 1 \quad (15)$$

or

$$|\mathcal{K} \sinh(\sqrt{\Delta_1} \tau_1/2) \sinh(\sqrt{\Delta_2} \tau_2/2) + \cosh(\sqrt{\Delta_1} \tau_1/2) \cosh(\sqrt{\Delta_2} \tau_2/2)| \geq 1, \quad (16)$$

$$\begin{aligned} &|\mathcal{K} \sinh(\sqrt{\Delta_1} \tau_1/2) \sinh(\sqrt{\Delta_2} \tau_2/2) \\ &+ \cosh(\sqrt{\Delta_1} \tau_1/2) \cosh(\sqrt{\Delta_2} \tau_2/2)| \\ &< \cosh(\text{tr}(A_1)\tau_1/2 + \text{tr}(A_2)\tau_2/2), \end{aligned} \quad (17)$$

$$\begin{aligned} &|\mathcal{K} \sinh(\sqrt{\Delta_1} \tau_1/2) \sinh(\sqrt{\Delta_2} \tau_2/2) \\ &+ \cosh(\sqrt{\Delta_1} \tau_1/2) \cosh(\sqrt{\Delta_2} \tau_2/2)| \\ &< e^{-\text{tr}(A_1)\tau_1/2} e^{-\text{tr}(A_2)\tau_2/2}. \end{aligned} \quad (18)$$

CASE II:  $\Delta_1 < 0$  and  $\Delta_2 < 0$ . This implies that both  $A_1$  and  $A_2$  have a pair of conjugate complex eigenvalues. Accordingly, we have

$$\begin{aligned} \text{tr}(e^{A_1 \tau_1} e^{A_2 \tau_2}) &= 2e^{\text{tr}(A_1)\tau_1/2} e^{\text{tr}(A_2)\tau_2/2} \\ &\times [\mathcal{K} \sin(\sqrt{-\Delta_1} \tau_1/2) \sin(\sqrt{-\Delta_2} \tau_2/2) \\ &+ \cos(\sqrt{-\Delta_1} \tau_1/2) \cos(\sqrt{-\Delta_2} \tau_2/2)]. \end{aligned} \quad (19)$$

Proposition 2: If both  $A_1$  and  $A_2$  have complex eigenvalues, the periodic switching system in (1) is asymptotically stable if and only if the dwell-times  $\tau_1$  and  $\tau_2$  satisfy the following inequalities

$$|\mathcal{K} \sin(\sqrt{-\Delta_1} \tau_1/2) \sin(\sqrt{-\Delta_2} \tau_2/2) + \cos(\sqrt{-\Delta_1} \tau_1/2) \cos(\sqrt{-\Delta_2} \tau_2/2)| < 1, \quad (20)$$

$$e^{\text{tr}(A_1)\tau_1} e^{\text{tr}(A_2)\tau_2} < 1 \quad (21)$$

or

$$|\mathcal{K} \sin(\sqrt{-\Delta_1} \tau_1/2) \sin(\sqrt{-\Delta_2} \tau_2/2) + \cos(\sqrt{-\Delta_1} \tau_1/2) \cos(\sqrt{-\Delta_2} \tau_2/2)| \geq 1, \quad (22)$$

$$\begin{aligned} &|\mathcal{K} \sin(\sqrt{-\Delta_1} \tau_1/2) \sin(\sqrt{-\Delta_2} \tau_2/2) \\ &+ \cos(\sqrt{-\Delta_1} \tau_1/2) \cos(\sqrt{-\Delta_2} \tau_2/2)| \\ &< \cosh(\text{tr}(A_1)\tau_1/2 + \text{tr}(A_2)\tau_2/2), \end{aligned} \quad (23)$$

$$\begin{aligned} &|\mathcal{K} \sin(\sqrt{-\Delta_1} \tau_1/2) \sin(\sqrt{-\Delta_2} \tau_2/2) \\ &+ \cos(\sqrt{-\Delta_1} \tau_1/2) \cos(\sqrt{-\Delta_2} \tau_2/2)| \\ &< e^{-\text{tr}(A_1)\tau_1/2} e^{-\text{tr}(A_2)\tau_2/2}. \end{aligned} \quad (24)$$

CASE III:  $\Delta_1 > 0$  and  $\Delta_2 < 0$ . This implies that  $A_1$  has a pair of distinct eigenvalues and that  $A_2$  has a pair of conjugate complex eigenvalues. Accordingly, we have

$$\begin{aligned} \text{tr}(e^{A_1 \tau_1} e^{A_2 \tau_2}) &= 2e^{\text{tr}(A_1)\tau_1/2} e^{\text{tr}(A_2)\tau_2/2} \\ &\times [\mathcal{K} \sinh(\sqrt{\Delta_1} \tau_1/2) \sin(\sqrt{-\Delta_2} \tau_2/2) \\ &+ \cosh(\sqrt{\Delta_1} \tau_1/2) \cos(\sqrt{-\Delta_2} \tau_2/2)]. \end{aligned} \quad (25)$$

Proposition 3: If  $A_1$  has distinct eigenvalues and  $A_2$  has complex eigenvalues, the periodic switching system in (1) is asymptotically stable if and only if the dwell-times  $\tau_1$  and  $\tau_2$  satisfy the following inequalities

$$|\mathcal{K} \sinh(\sqrt{\Delta_1} \tau_1/2) \sin(\sqrt{-\Delta_2} \tau_2/2) + \cosh(\sqrt{\Delta_1} \tau_1/2) \cos(\sqrt{-\Delta_2} \tau_2/2)| < 1, \quad (26)$$

$$e^{\text{tr}(A_1)\tau_1} e^{\text{tr}(A_2)\tau_2} < 1 \quad (27)$$

or

$$|\mathcal{K} \sinh(\sqrt{\Delta_1} \tau_1/2) \sin(\sqrt{-\Delta_2} \tau_2/2) + \cosh(\sqrt{\Delta_1} \tau_1/2) \cos(\sqrt{-\Delta_2} \tau_2/2)| \geq 1, \quad (28)$$

$$\begin{aligned} &|\mathcal{K} \sinh(\sqrt{\Delta_1} \tau_1/2) \sin(\sqrt{-\Delta_2} \tau_2/2) \\ &+ \cosh(\sqrt{\Delta_1} \tau_1/2) \cos(\sqrt{-\Delta_2} \tau_2/2)| \\ &< \cosh(\text{tr}(A_1)\tau_1/2 + \text{tr}(A_2)\tau_2/2), \end{aligned} \quad (29)$$

$$\begin{aligned} &|\mathcal{K} \sinh(\sqrt{\Delta_1} \tau_1/2) \sin(\sqrt{-\Delta_2} \tau_2/2) \\ &+ \cosh(\sqrt{\Delta_1} \tau_1/2) \cos(\sqrt{-\Delta_2} \tau_2/2)| \\ &< e^{-\text{tr}(A_1)\tau_1/2} e^{-\text{tr}(A_2)\tau_2/2}. \end{aligned} \quad (30)$$

When both subsystems are stable, the inequalities in Propositions 1-3 can be reduced according to Corollary 1. Moreover, to complete this section, we emphasize that instead of solving these inequalities directly, an intuitive way to

present the mutual constraint relation between  $\tau_1$  and  $\tau_2$  is to depict the stability margin on their plane, which is formed by the corresponding equalities.

**IV. AN ILLUSTRATIVE EXAMPLE**

We now provide an example to typically illustrate Proposition 3.

Example 1:  $A_1 = \begin{bmatrix} -2 & 1 \\ 0.5 & 0.5 \end{bmatrix}$  and  $A_2 = \begin{bmatrix} -0.5 & -1 \\ 3 & 1 \end{bmatrix}$ .

We have  $\text{tr}(A_1) = -1.5$ ,  $\text{tr}(A_2) = 0.5$ , and  $\Delta_1 = 8.25$ ,  $\Delta_2 = -9.75$ . Then, the eigenvalues of  $A_1$  and  $A_2$  are  $\{-2.2967, 0.6967\}$  and  $\{0.25 \pm j1.5612\}$ , respectively. In addition,  $\mathcal{K} = 0.9756$ . Therefore, by Proposition 3, the system is asymptotically stable for all  $\tau_1$  and  $\tau_2$  satisfying

$$\begin{aligned} &|0.9756 \sinh(1.4361\tau_1) \sin(1.5612\tau_2) \\ &+ \cosh(1.4361\tau_1) \cos(1.5612\tau_2)| < 1, \\ &e^{-1.5\tau_1} e^{0.5\tau_2} < 1 \end{aligned}$$

or

$$\begin{aligned} &|0.9756 \sinh(1.4361\tau_1) \sin(1.5612\tau_2) \\ &+ \cosh(1.4361\tau_1) \cos(1.5612\tau_2)| \geq 1, \\ &|0.9756 \sinh(1.4361\tau_1) \sin(1.5612\tau_2) \\ &+ \cosh(1.4361\tau_1) \cos(1.5612\tau_2)| \\ &< \cosh(-0.75\tau_1 + 0.25\tau_2), \\ &|0.9756 \sinh(1.4361\tau_1) \sin(1.5612\tau_2) \\ &+ \cosh(1.4361\tau_1) \cos(1.5612\tau_2)| \\ &< e^{0.75\tau_1} e^{-0.25\tau_2}. \end{aligned}$$

Next, we need to depict the following implicit functions on the  $(\tau_1, \tau_2)$  plane:

$$\begin{aligned} &|0.9756 \sinh(1.4361\tau_1) \sin(1.5612\tau_2) \\ &+ \cosh(1.4361\tau_1) \cos(1.5612\tau_2)| = 1, \end{aligned} \quad (31)$$

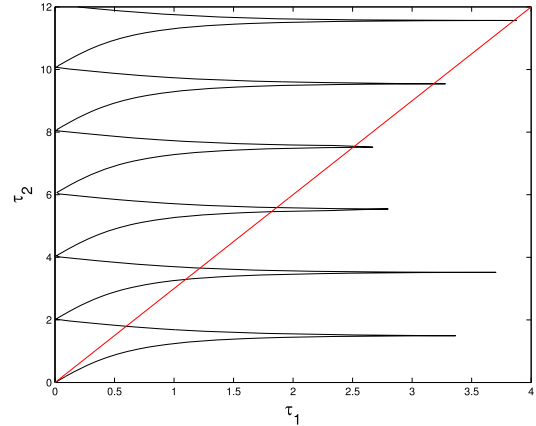
$$-1.5\tau_1 + 0.5\tau_2 = 0, \quad (32)$$

$$\begin{aligned} &|0.9756 \sinh(1.4361\tau_1) \sin(1.5612\tau_2) \\ &+ \cosh(1.4361\tau_1) \cos(1.5612\tau_2)| \\ &= \cosh(-0.75\tau_1 + 0.25\tau_2), \end{aligned} \quad (33)$$

$$\begin{aligned} &|0.9756 \sinh(1.4361\tau_1) \sin(1.5612\tau_2) \\ &+ \cosh(1.4361\tau_1) \cos(1.5612\tau_2)| \\ &= e^{0.75\tau_1} e^{-0.25\tau_2}. \end{aligned} \quad (34)$$

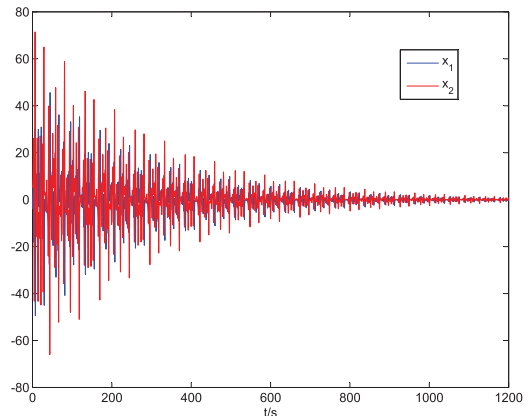
Doing this, we can use software such as Maple or Matlab, which includes tools to depict the implicit functions on the plane.

In Figure 1, the equations in (31) and (32) are depicted and colored in black and red, respectively. They form the boundaries of the regions, which correspond to the complex eigenvalues of  $e^{A_1\tau_1} e^{A_2\tau_2}$  located inside the unit circle of the complex plane. For example, the point corresponding to  $\tau_1 = 0.5$  and  $\tau_2 = 0.9$  is located inside a branch of these regions and makes the eigenvalues of  $e^{0.5A_1} e^{0.9A_2}$  be  $\{0.8273 \pm j0.2375\}$ . Meanwhile, the point corresponding to

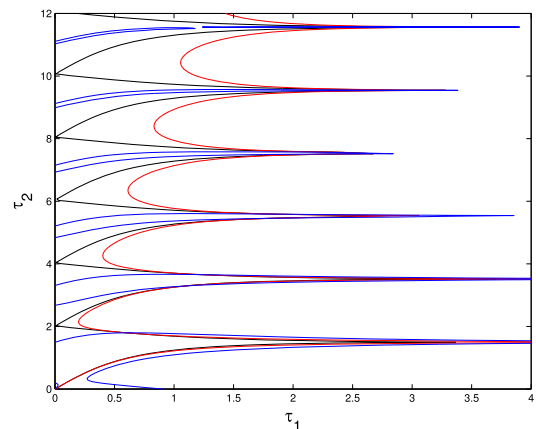


**FIGURE 1.** The stability margin corresponding to the complex eigenvalues of  $e^{A_1\tau_1} e^{A_2\tau_2}$ .

$\tau_1 = 1.88$  and  $\tau_2 = 5.54$  is located inside another branch of these regions and makes the eigenvalues of  $e^{1.88A_1} e^{5.54A_2}$  be  $\{-0.2717 \pm j0.9367\}$ . In addition, the corresponding state response is shown in Figure 2.



**FIGURE 2.** The state-response for  $\tau_1 = 1.88s$  and  $\tau_2 = 5.54s$ .



**FIGURE 3.** The stability margin corresponding to the real eigenvalues of  $e^{A_1\tau_1} e^{A_2\tau_2}$ .

In Figure 3, the equations in (31), (33), and (34) are depicted and colored in black, red, and blue, respectively.

These curves form the boundaries of the regions, which correspond to the eigenvalues of  $e^{A_1\tau_1}e^{A_2\tau_2}$  within  $(-1, 1)$ . For example, the point corresponding to  $\tau_1 = 1.6$  and  $\tau_2 = 1.6$  is located in a branch of these regions and makes the eigenvalues of  $e^{1.6A_1}e^{1.6A_2}$  be  $\{-0.2678, -0.7540\}$ . The corresponding state response is shown in Figure 4.

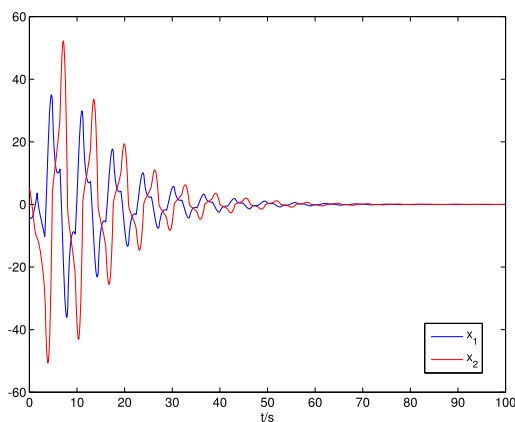


FIGURE 4. The state-response for  $\tau_1 = \tau_2 = 1.6s$ .

As shown in Figures 2 and 4, this example exposes a little surprising phenomenon that two unstable subsystems with quite long periods to stay active can generate a stable state-trajectory.

## V. CONCLUSION

We considered the stability analysis problem for the planar linear systems undergoing periodic switching. We characterized the stability margin that the dwell-times of subsystems are confined to so that the overall system is asymptotically stable. The main technical preliminary includes a key lemma along with the distribution property of the roots of parabolic functions. Indeed, the expansion of the transition matrix of each subsystem up to the first order of its generator (i.e., the subsystem matrix) enables us to express the stability margin analytically and, moreover, compute it numerically. Finally, an example was worked in detail to illustrate the theoretical results.

## REFERENCES

- [1] A. A. Agrachev and D. Liberzon, "Lie-algebraic stability conditions criteria for switched systems," *SIAM J. Control Optim.*, vol. 40, no. 1, pp. 253–269, 2001.
- [2] M. Balde, U. Boscain, and P. Mason, "A note on stability conditions for planar switched systems," *Int. J. Control*, vol. 82, no. 10, pp. 1882–1888, 2009.
- [3] U. Boscain, "Stability of planar switched systems: The linear single input case," *SIAM J. Control Optim.*, vol. 41, no. 1, pp. 89–112, 2002.
- [4] S. Cong and J. Chen, "Stabilisation control for planar bilinear systems by using switching law to minimise return ratio," *IET Control Theory Appl.*, vol. 9, no. 13, pp. 2008–2014, 2015.
- [5] S. Cong, "Stability analysis for planar discrete-time linear switching systems via bounding joint spectral radius," *Syst. Control Lett.*, vol. 96, pp. 7–10, Sep. 2016.

- [6] W. P. Dayawansa and C. F. Martin, "A converse Lyapunov theorem for a class of dynamical systems which undergo switching," *IEEE Trans. Autom. Control*, vol. 44, no. 4, pp. 751–760, Apr. 1999.
- [7] R. A. DeCarlo, M. S. Branicky, S. Pettersson, and B. Lennartson, "Perspectives and results on the stability and stabilizability of hybrid systems," *Proc. IEEE*, vol. 88, no. 7, pp. 1069–1082, Jul. 2000.
- [8] J. C. Geromel and P. Colaneri, "Stability and stabilization of continuous-time switched linear systems," *SIAM J. Control Optim.*, vol. 45, no. 5, pp. 1915–1930, 2006.
- [9] M. Margaliot and M. S. Branicky, "Nice reachability of planar bilinear control systems with applications to planar linear switched systems," *IEEE Trans. Autom. Control*, vol. 54, no. 6, pp. 1430–1435, Jun. 2009.
- [10] M. Margaliot and G. Langholz, "Necessary and sufficient conditions for absolute stability: The case of second-order systems," *IEEE Trans. Circuits Syst. I, Fundam. Theory Appl.*, vol. 50, no. 2, pp. 227–234, Feb. 2003.
- [11] B. Niu, H. R. Karimi, H. Wang, and Y. Liu, "Adaptive output-feedback controller design for switched nonlinear stochastic systems with a modified average dwell-time method," *IEEE Trans. Syst., Man, Cybern., Syst.*, vol. 47, no. 7, pp. 1371–1382, Jul. 2017, doi: 10.1109/TSMC.2016.2597305.
- [12] B. Niu, Y. Liu, G. Zong, Z. Han, and J. Fu, "Command filter-based adaptive neural tracking controller design for uncertain switched nonlinear output-constrained systems," *IEEE Trans. Cybern.*, to be published, doi: 10.1109/TSMC.2016.2597305.
- [13] E. S. Pyatnitskiy and L. B. Rapoport, "Criteria of asymptotic stability of differential inclusions and periodic motions of time-varying nonlinear control systems," *IEEE Trans. Circuits Syst. I, Fundam. Theory Appl.*, vol. 43, no. 3, pp. 219–229, Mar. 1996.
- [14] R. Shorten, F. Wirth, O. Mason, K. Wulff, and C. King, "Stability criteria for switched and hybrid systems," *SIAM Rev.*, vol. 49, no. 5, pp. 545–592, 2007.
- [15] Z. Sun, "Combined stabilizing strategies for switched linear systems," *IEEE Trans. Autom. Control*, vol. 51, no. 4, pp. 666–674, Apr. 2006.
- [16] L. Xie, S. Shishkin, and M. Fu, "Piecewise Lyapunov functions for robust stability of linear time-varying systems," *Syst. Control Lett.*, vol. 31, no. 3, pp. 165–171, 1997.



**ZUNBING SHENG** received the B.S. degree in mechanical engineering from the Gansu University of Technology, Lanzhou, China, in 2000, and the M.S. and Ph.D. degrees in mechanical and electronic engineering from the Harbin Institute of Technology, Harbin, China, in 2003 and 2009, respectively. He has been with Heilongjiang University, Harbin, since 2009. His research interests include switching systems and vision-based robot control.



**WEI QIAN** received the B.S. degree in mechanical engineering from Zhengzhou University, Zhengzhou, China, in 1999, the M.S. degree in control theory and control engineering from Southeast University, Nanjing, China, in 2005, and the Ph.D. degree in control theory and control engineering from Zhejiang University, Hangzhou, China, in 2009. He has been a Professor with Henan Polytechnic University, Jiaozuo, China, since 2015. His research interests include networked control systems, delayed systems, and switching systems.