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# Design and Analysis of a Reconfigurable Holographic Metasurface Aperture for Dynamic Focusing in the Fresnel Zone

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**ABSTRACT** We present numerical simulations of the near-field focusing capabilities of a dynamically reconfigurable holographic metasurface aperture. The aperture consists of a parallel-plate waveguide in which the upper plate is patterned with a number of metamaterial irises that can be dynamically switched between radiating (ON) and non-radiating (OFF) states. A cylindrically symmetric waveguide mode, excited by a coaxial probe in the center of the lower plate, serves to excite the radiating irises, forming a focused spot in the radiating near-field (or Fresnel zone). The layout of the metamaterial elements and their tuning states is determined using holographic design principles, in which the interference pattern of the waveguide (or reference) mode and the desired radiated field pattern leads to the required phase distribution over the surface of the aperture. We also develop an analytical model of the aperture to confirm the numerical simulations, and to illustrate the advantage of the guided-mode as the reference wave versus a plane-wave. We further leverage this analytical model to analyze the diffracted order characteristics of the holographic metasurface aperture, showing high-fidelity focusing patterns even for difficult focusing scenarios across the entire investigated field-of-view.

**INDEX TERMS** Near-field, focusing, reconfigurable, dynamic, metasurface, metamaterial, active, aperture, microwaves.

## I. INTRODUCTION

Electronically reconfigurable apertures can dynamically control radiated phase-fronts without mechanical motion, enabling fast beam-forming. The ability to control the radiation phase-front of an aperture is of significant importance, giving rise to well-known capabilities within the distinctive radiation zones of the aperture; specifically, beam-focusing within the radiative near-field (Fresnel) region and beam-steering within the far-field (Fraunhofer) region.

Conventionally, most antennas are designed for far-field applications, including electronically scanned antennas (ESAs). Traditional ESAs rely on the distribution of a large array of antennas, each backed by an active phase-shifter, and which sample the aperture at the Nyquist limit ( $\lambda/2$ , where  $\lambda$  is the free-space wavelength) [1]–[6]. Such apertures have successfully been in use for decades for beam-forming and steering applications in the far-field, including radar and long-range communications. Although high-fidelity beam patterns have been demonstrated using

phased array technologies, there are several limitations associated with these systems. For example, each antenna within a phased array aperture requires an individual phase shifting circuit, which, for even a moderately sized aperture, can result in a significant number of phase shifters. Thus, the active circuitry within the phase shifters forms a complex and expensive system architecture. Moreover, the phase shifters used in such systems exhibit considerable insertion losses, which must be offset using power amplifiers. Therefore, power consumption of such systems can be significant.

The concept of a metasurface to tailor the aperture radiated phase-fronts has successfully been demonstrated in the literature for near-field beam-focusing [7]–[9] and far-field beam-steering applications [9]–[11]. Metasurfaces consist of arrays of subwavelength elements whose electrical properties can be engineered to achieve desired electromagnetic responses. Numerous static metasurface designs across the electromagnetic spectrum have been reported in the literature and have shown great potential as alternative architectures for

wavefront shaping. The first dynamically reconfigurable metasurface was demonstrated in [12]–[16] at microwave frequencies (K-band, or 17.5 – 26.5 GHz), where far-field electronic beam scanning was achieved using a liquid-crystal layer whose dielectric constant could be varied in the vicinity of each metamaterial element using an externally applied voltage bias. By changing the local dielectric constant, the resonance frequency of each element within the metasurface aperture becomes voltage-controlled. Recently, the analysis of an electrically tunable liquid-crystal based metasurface layer has also been demonstrated [17]. While the liquid-crystal metasurface reconfigurable apertures exhibit high-fidelity beam control and can be integrated into large-scale manufacturing processes, liquid-crystal as a control mechanism imposes certain limitations, such as limited switching speed due to its inherent relaxation rate, typically on the order of up to 1 MHz. This limitation can be addressed by using semiconductor-based switching elements, such as PIN diodes, which exhibit switching speeds of up to a few GHz [18]. Such apertures were demonstrated in [16], [19], and [20] for far-field radar applications.

Using the concept of near-field focusing, many types of apertures have been demonstrated, including reflect arrays [21]–[23], transmit arrays [23]–[25] and Fresnel zone plates [23], [26]. Despite encouraging results, these structures require a non-planar feeding mechanism. Thus, they exhibit a poor form-factor, limiting their usage in applications where planar structures are desirable.

Other design alternatives that have recently become the subject of considerable research within the field of near-field focusing include microstrip arrays [27]–[33] and leaky-wave apertures [34]–[37]. Such apertures are conventionally planar and exhibit a small form-factor. However, a major drawback with most of these designs is that, similar to the static metasurface apertures discussed earlier, they are not reconfigurable and designed to achieve a fixed focus. Although the leaky-wave designs presented in [35]–[37] enable the control of the focusing depth within the near field as a function of the operating frequency, they do not offer the freedom to focus at an arbitrary point in space using the same aperture. The leaky-wave aperture presented in [34] exhibits reconfigurable operation achieved using field-effect transistors (FET). However, it can achieve dynamic beam-focusing only in one dimension due to the 1D geometry of the array aperture. Moreover, the guided mode is launched through a tapered horn antenna attached to a waveguide, resulting in a poor form-factor.

In this work, we demonstrate a reconfigurable, planar holographic metasurface aperture designed to achieve dynamic near-field beam-focusing in the Fresnel zone. While the reconfigurability enables the same aperture to be used to achieve varying focusing characteristics, the ability to dynamically focus the beam at any point within the near-field space of the aperture makes the proposed aperture suitable for a number of emerging applications, such as near-field imaging, non-destructive testing, biomedical imaging and wireless power transfer. The proposed aperture is based on

a parallel-plate waveguide structure that operates at 20 GHz. The upper plate of the waveguide is populated with a set of metamaterial elements—simple rectangular irises, in the presented design, each of which can be toggled between radiating (ON) and non-radiating (OFF) states by switching a PIN diode, incorporated into the design, from reverse to forward bias. The distribution of radiating irises form a focus at some desired location determined using holographic design principles, as will be described below.

The outline of the paper is as follows: Section 2 explains the design of the dynamically reconfigurable holographic metasurface aperture for near-field focusing. In Section 3, we present simulated aperture radiated field-patterns for both on-axis and off-axis dynamic focusing scenarios. Analytical studies are also presented to analyze the focusing and diffracted order suppression limits of the presented aperture. The obtained results are analyzed and discussed. We provide concluding remarks in Section 4.

## II. DYNAMICALLY RECONFIGURABLE HOLOGRAPHIC METASURFACE APERTURE

We illustrate the holographic aperture concept in Fig. 1, in which a dynamic aperture is shown producing a focused field at an arbitrarily selected point within the near-field of the aperture ( $d < 2D^2/\lambda$ , where  $D$  is the aperture size and  $\lambda$  is the free-space wavelength). As shown in Fig. 1, the aperture is center-fed using a coaxial cable inserted into the lower plate, with the upper plate discretized into a grid of elements, each subwavelength in dimensions. The reconfigurable aperture shown in Fig. 1 is a parallel-plate waveguide, consisting of a dielectric substrate, Rogers 3003 ( $\epsilon_r = 3$  and  $\tan\delta = 0.001$ ), sandwiched between the top and bottom conducting layers. The size of the aperture is  $15\text{ cm} \times 15\text{ cm}$ , corresponding to an electrical size of  $10\lambda \times 10\lambda$  at 20 GHz.

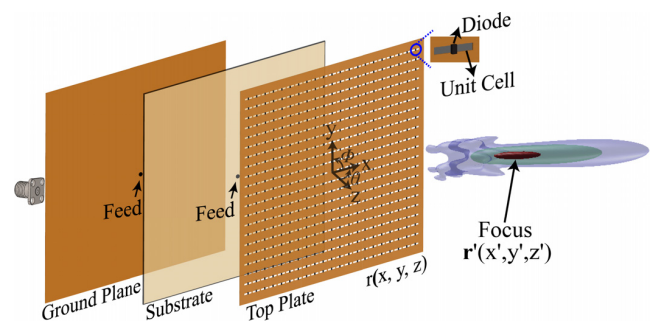


FIGURE 1. Structure of the reconfigurable aperture for focusing.

The magnetic field of the guided mode radiated by the coaxial feed into the dielectric substrate can be modeled by means of the Hankel function (zeroth order and first kind) as follows:

$$\mathbf{H}_{ref} = \begin{cases} H_0^1(k_g \mathbf{r}) \cos \phi, & \text{x-polarization} \\ H_0^1(k_g \mathbf{r}) \sin \phi, & \text{y-polarization} \end{cases} \quad (1)$$

In (1),  $k_g$  denotes the guided wavenumber ( $k/\sqrt{\epsilon_r}$ ) within the dielectric substrate and  $\mathbf{r}$  is the position of the discretized

points on the aperture. For this analysis, it is assumed that the aperture is thin enough ( $< \lambda_g/2$ , where  $\lambda_g$  is the guided-wavelength within the dielectric substrate) along the optical axis (z-axis) to suppress the higher order waveguide modes (single mode operation) in the broadside direction. Following the definition of the guided-mode, the desired field distribution to create a focus at a selected point is calculated by treating the focused point as a fictitious point source placed at  $\mathbf{r}'$  and back-propagating the radiated field from the point source to the aperture as follows:

$$\mathbf{P} = \frac{e^{-jk(|\mathbf{r}-\mathbf{r}'|)}}{|\mathbf{r}-\mathbf{r}'|} \quad (2)$$

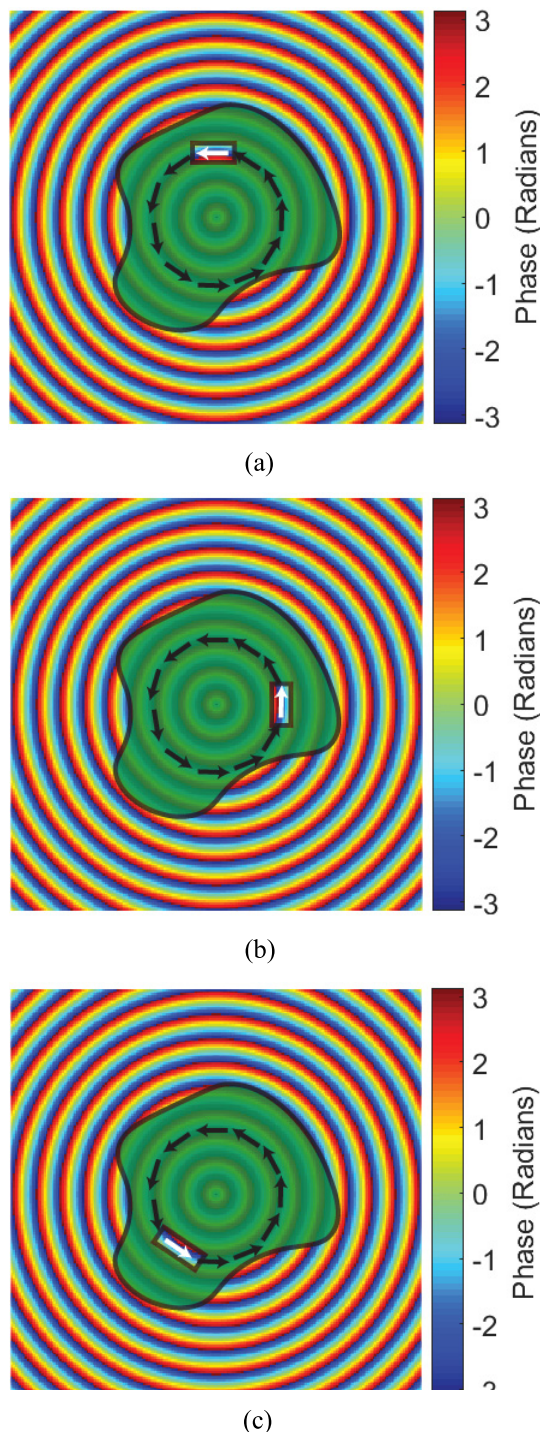
Our aim is to produce a hologram over the metasurface aperture, such that the back-propagated field of (2) will be produced from the guided mode of (1), which can be considered the reference wave for the hologram. Analyzing (1) and (2), it is evident that this can be achieved from the complex amplitude distribution over the aperture surface (or mask) defined by

$$\mathbf{M} = \mathbf{P}\mathbf{H}_{ref}^* \quad (3)$$

In (3),  $\mathbf{H}_{ref}^*$  denotes the complex conjugate of the guided magnetic field. In general, the phase dominates image formation, such that we attempt to reconstruct only the phase distribution implied by (3).

It should be noted that bold notation is adopted in (1)-(3) to represent vectors and matrices while we use scalar approximation for the fields. The elements used to form the metasurface (Fig. 1) resemble an array of slot antennas. It is well-known in antenna theory that a slot radiator has a resonance when the length of the slot is one-half the wavelength at the operating frequency [38]. Whereas such a choice ensures maximum coupling of the slots to the guided-mode (improving radiation efficiency), the strong coupling of the slot elements to the guided-mode can introduce a significant perturbation to the guided-mode (which is modeled here using the Hankel function as in (1) and ignoring scattering effects). With this tradeoff in mind, the length of the slot elements is taken to be slightly less than half of the guided-wavelength,  $\lambda_g/2.5$ , while the width of the slots is selected to be  $\lambda_g/6$ . The orientation of the elements governs the polarization of the radiation in three possible ways as depicted in Fig. 2.

A rectangular slot oriented along the x-axis will couple to the x-component of the guided magnetic field, therefore radiating a field linearly polarized along the x-direction [38]. Similarly, when oriented along the y-axis, the slots couple to the y-component of the magnetic field of the guided mode, radiating a field linearly polarized along the y-direction. When placed along the  $\phi$ -direction, the slots couple both to the x- and y-components of the guided magnetic field, producing polarimetric radiation characteristics. These three possible states are depicted in Fig. 2. As shown in Fig. 1, the reconfigurable aperture presented in this



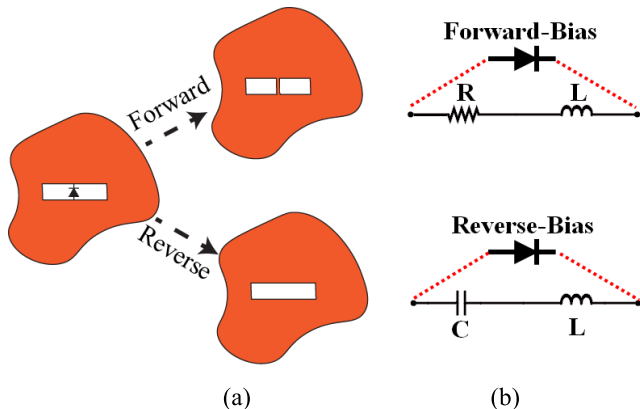
**FIGURE 2.** Orientation of the unit cells and coupling to the guided magnetic field (depicted as arrows): (a) horizontal (x-axis), (b) vertical (y-axis), and (c)  $\phi$ -direction. The green layer depicts a portion of the conducting top layer of an arbitrary geometry onto which the metamaterial element, coupling to the guided mode, is patterned.

work consists of slot unit cells oriented with their long axes along the x-axis, thus producing linearly polarized waves.

Each unit cell element patterned onto the front surface of the holographic metasurface aperture includes a PIN diode in the center, as depicted in Fig. 3(a), for which we assume



the circuit model of the PIN diode shown in Fig. 3(b). When forward-biased, the PIN diodes are modeled as an RL circuit with a negligible forward resistance in parallel with the junction capacitance, and an inductor in series as shown Fig. 3(b). When reverse-biased, the PIN diodes can be modeled as an LC circuit, exhibiting a high reverse resistance (effectively open-circuit), leaving the junction capacitance connected in series with the inductor.

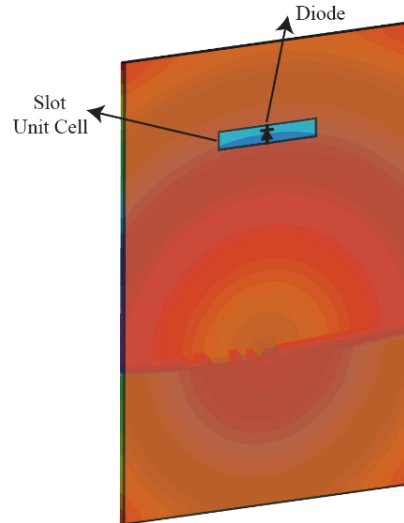


**FIGURE 3.** Unit cell loaded with a PIN diode (a) element configuration for forward and reverse bias states and (b) PIN diode equivalent circuit diagram; forward-biased:  $R = 4 \Omega$  and  $L = 0.5 \text{ nH}$ , reverse-biased:  $L = 5 \text{ nH}$ ,  $C = 0.02 \text{ pF}$  [18].

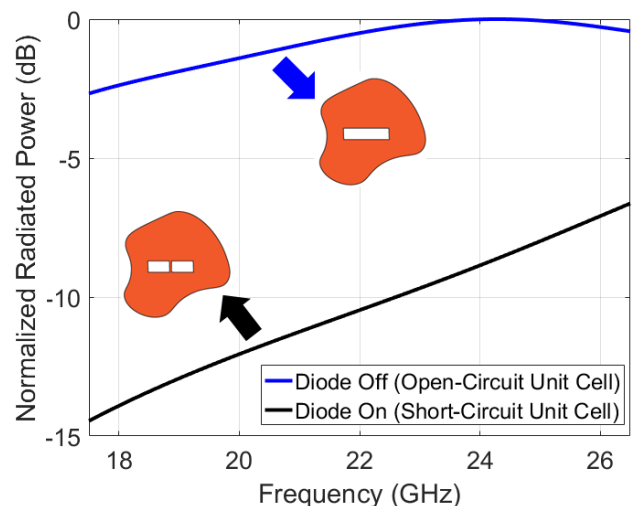
The PIN diodes are used to control the resonance frequency, and therefore the coupling response of the elements to the guided mode. As shown in Fig. 3(a), when forward-biased, the diodes short-circuit the slot unit cells (OFF state), shifting the resonance frequency of the slots to a higher frequency band. As a result, the OFF slots do not couple to the guided mode, acting as an opaque patch at the frequency band of interest. When reverse biased, the diodes act as an open-circuit element, enabling the slot unit cells to operate at the intended resonance frequency (ON state). To analyze the achievable contrast in the power level radiated by the OFF and ON unit cells, in CST Microwave Studio, we designed a parallel-plate waveguide consisting of a single slot unit cell patterned onto the front surface as shown in Fig. 4.

In Fig. 5, we show the simulated radiated power as a function of frequency of the metamaterial element when the PIN diode is forward- and reverse-biased, respectively. Both curves are normalized with respect to the maximum radiation and plotted on a logarithmic scale. Comparing the curves, it can be seen that at 20 GHz, the radiated power drops by a factor of 10 dB when the unit cell is OFF.

An important metric for the holographic metasurface aperture is the estimated power consumption. The number of metamaterial elements patterned onto the front surface of the aperture is  $21 \times 21$ . Given that each element includes a PIN diode, the aperture requires 441 PIN diodes. Biasing characteristic of conventional PIN diodes in the K-band are typically on the order of  $V=0.7 \text{ V}$  and  $I=3 \text{ mA}$  [18], resulting in the estimated power consumption per diode being equal to around 2.1 mW. Integrating over the full aperture,



**FIGURE 4.** Parallel-plate waveguide with a single unit cell.



**FIGURE 5.** Single unit cell normalized radiated power patterns.

the estimated maximum power consumption is calculated to be 0.92 W, though it is unlikely that all elements will be activated at the same time. It should be noted that the practical implementation of the holographic metasurface aperture would require a DC biasing circuit to drive the PIN diodes. One potential solution would be to etch subwavelength-width lines (narrow enough to prevent any distortion to the guided-mode from an RF perspective) on the front layer of the aperture to accommodate the biasing circuit. Alternatively, a thin layer of dielectric laminate can be added to the design for DC biasing with metal vias connecting the biasing circuit to the diodes. Biasing of the diodes can be achieved using an Arduino driving an array of shift registers or a field-programmable gate array (FPGA) can be employed to achieve superior switching speed.

### III. RESULTS AND DISCUSSION

To illustrate the dynamic beam focusing capability of the proposed reconfigurable holographic metasurface aperture, we choose two arbitrary points in space within the Fresnel



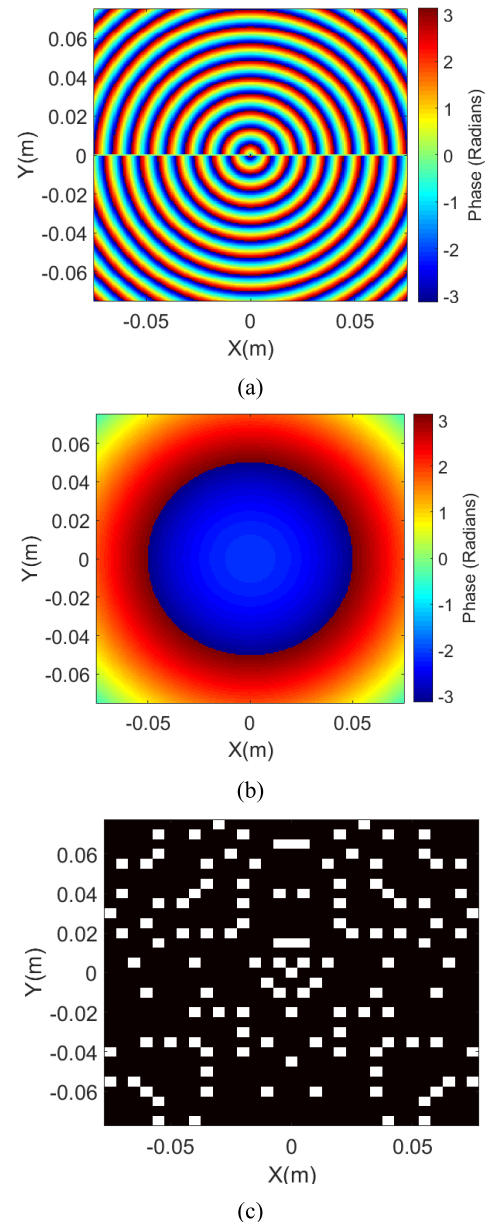
region of the aperture:  $F_1(x=0 \text{ m}, y=0 \text{ m}, z=0.5 \text{ m})$  and  $F_2(x=0 \text{ m}, y=0.07 \text{ m}, z=0.5 \text{ m})$  with  $F_1$  and  $F_2$  representing the on-axis and off-axis focusing points, respectively. For both scenarios, the aperture shown in Fig. 1 is used. It is known that the back-propagated aperture field distributions for these two focusing configurations will be different, suggesting that the patterns of the OFF and ON metamaterial elements within the aperture will be different for  $F_1$  and  $F_2$  focusing configurations.

We begin our analysis with the on-axis focusing scenario,  $F_1$ . Using (2), the calculated guided-mode within the parallel-plate waveguide is shown in Fig. 6(a). It should be noted that although the guided magnetic field within the parallel-plate waveguide exhibits a cylindrical wavefront comprising x- and y- components, because the slot elements on the aperture are oriented along the x-axis, we limit our attention to the x-component of the guided magnetic field as shown in Fig. 6(a). Fig. 6(b) demonstrates the phase pattern of the field radiated by the fictitious source  $F_1$  back-propagated to the aperture plane.

Using (3), the calculated hologram mask,  $\mathbf{M}$ , is presented in Fig. 6(c). The elements highlighted in white color are those with reverse-biased PIN diodes. These elements couple to the guided mode and contribute to the radiation (ON state). The elements highlighted in black color, on the other hand, are short-circuited with the forward-biased PIN diodes and thus do not couple to the guided mode—acting as opaque patches at the frequency band of interest (OFF state).

It should be noted that the phase advance introduced by the slot elements is negligible, which, when considering a particular point on the aperture, results in the phase of the field radiated by an element being approximately the same as the phase of the excited guided mode. As a result, in this work, we make use of a binary algorithm, relying on the principle of coupling only the elements positioned at the points where the phase difference between the guided mode,  $\mathbf{H}_{ref}$ , and the back-propagated pattern,  $\mathbf{P}$ , remains within a certain threshold, which is selected to be  $\pm 20^\circ$  as a result of numerical (CST Microwave Studio) parametric analyses. Therefore, the reconfigurable hologram mask in Fig. 6(c) can be considered a binary filter. Using more advanced unit cell topologies, such as complementary-ELC (cELC) and meander-line metamaterial elements [39], the phase accuracy and range of the hologram can be further increased.

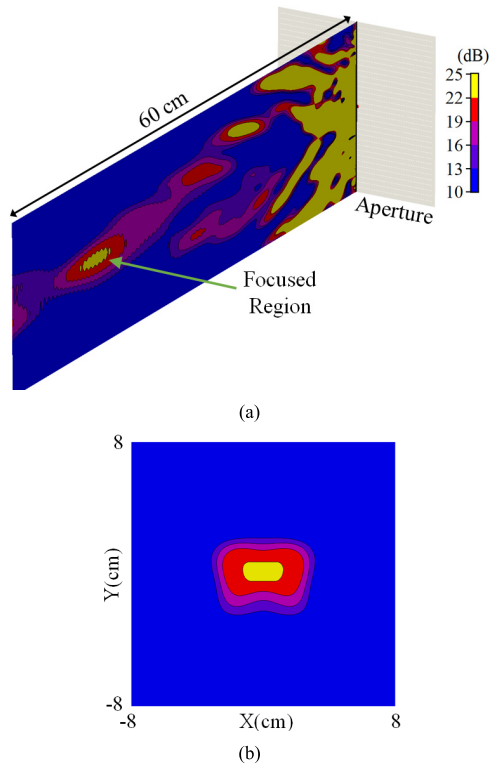
Following the calculation of the hologram mask, the reconfigurable holographic metasurface aperture was designed in CST Microwave Studio. Fig. 7 illustrates the electric-field (E-field) pattern radiated by the aperture focusing at  $F_1(x=0 \text{ m}, y=0 \text{ m}, z=0.5 \text{ m})$ . Fig. 7(a) shows the cross-section of the E-field pattern in the range plane (yz-plane), while Fig. 7(b) shows the E-field pattern at the focal plane (xy-plane at  $z=0.5 \text{ m}$ ). Observing the cross-range pattern in Fig. 7(b), it can be seen that the beam-waist is asymmetric due to the fact that the slot unit cells are narrower along one axis than the other. In order to analyze the focusing



**FIGURE 6.** On-axis focusing: (a) guided-mode phase pattern, (b) back-propagated phase pattern on the aperture (from  $F_1$ ), and (c) OFF (black) and ON (white) unit cell positions.

characteristics of the aperture, we investigate the -3dB full-width-half-maximum (FWHM) value of the beam-waist in the cross-range plane, which is found to be 4.9 cm (averaged along the x- and y-axes), suggesting good agreement with the theoretical limits calculated using the analytical equations presented in [40].

Following the on-axis scenario, we study off-axis focusing using the same reconfigurable aperture presented in Fig. 1. In comparison to the on-axis focusing scenario, for off-axis focusing, while the guided-mode shown in Fig. 6(a) remains the same (see Fig. 8(a)), the field back-propagated to the aperture from  $F_2$  is different (see Fig. 8(b)), producing the hologram mask illustrated in Fig. 8(c) calculated using (3).

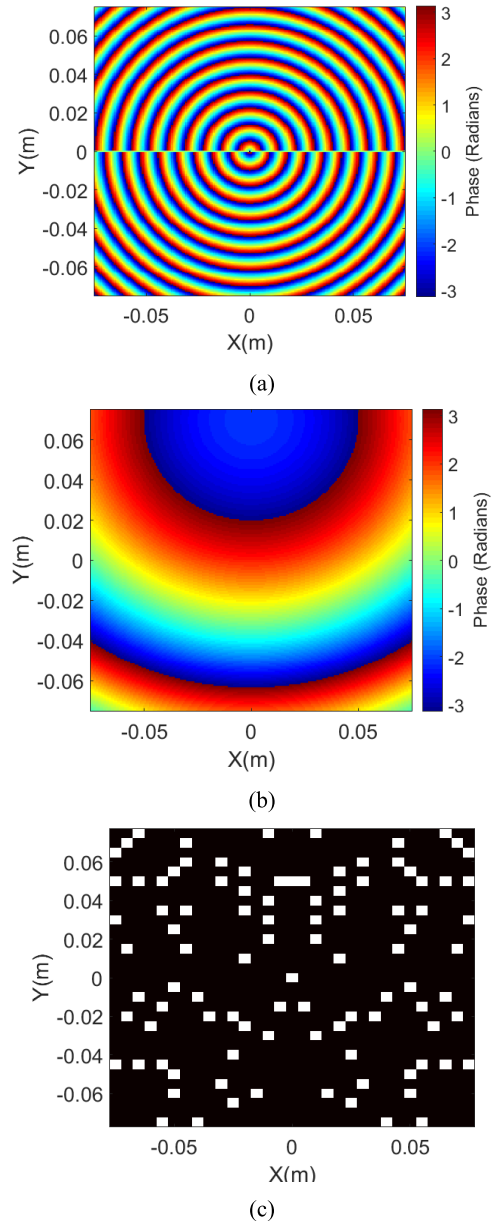


**FIGURE 7.** Radiated E-field pattern from the reconfigurable aperture: (a) range plane and (b) cross-range plane.

In Fig. 8(c), the elements highlighted in black are OFF due to the forward-biased diodes while the elements shown in white are ON and radiating. Similar to on-axis focusing, the phase threshold for the calculation of the hologram mask is selected to be  $\pm 20^\circ$ . The radiated E-field patterns in the range and cross-range planes are shown in Fig. 9.

Analyzing Fig. 9, the  $-3\text{dB}$  FWHM value for the beam-waist at the focal plane is found to be 5.2 cm. Comparing the FWHM values for the on-axis (Fig. 7) and off-axis (Fig. 9) scenarios, it is evident that the beam-waist for the off-axis configuration is wider than the on-axis configuration. This can be attributed to two factors; first, the focal distance of  $|F_2|$ , 0.505 m, is slightly larger than the focal distance of  $|F_1|$ , 0.5 m. Second, for the off-axis focusing scenario, the aperture is observed from an offset angle,  $\theta \approx 10^\circ$ , resulting in a reduced effective aperture in comparison with the on-axis case. This widening factor in the beam-waist can be approximately predicted by dividing the on-axis beam by the factor  $\cos^2\theta$  [40].

The choice of the reference-wave for the design of the reconfigurable metasurface aperture is of vital importance. In this work, we use the guided mode of (1), launched into the parallel-plate waveguide using the coaxial probe feed as depicted in Fig. 1. In holography, different types of waves can be used as the reference wave, including a plane-wave or a Gaussian beam [40]. An aperture using a plane-wave as the reference-wave can be considered to be the simplest type hologram from a computational perspective. However, using the proposed guided-mode in this work as the reference-wave

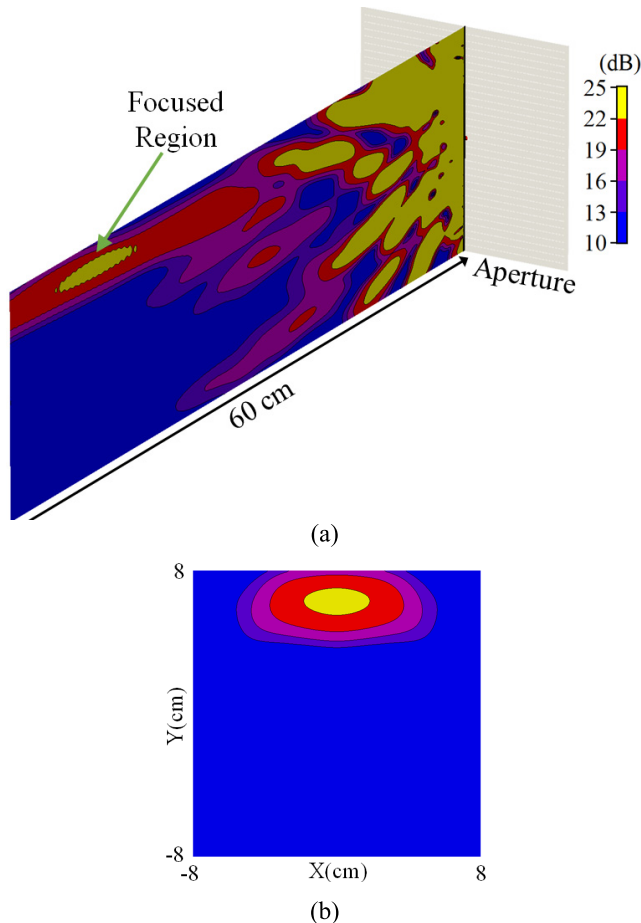


**FIGURE 8.** Off-axis focusing: (a) guided-mode phase pattern, (b) back-propagated phase pattern on the aperture (from  $F_2$ ), and (c) OFF (black) and ON (white) unit cell positions.

brings significant advantages.

First, to have a plane-wave illuminating the aperture, the feeding source needs to be at a certain distance from the aperture to ensure that the aperture is illuminated with uniform phase-fronts. Such a feed naturally increases the overall profile of the composite structure, since the feed point must be at some significant distance from the aperture and may also require additional quasi-optical elements. The guided-mode launched by a simple, planar coaxial feed, by contrast, eliminates the need for focusing optics and offset sources, providing a significantly improved form factor.

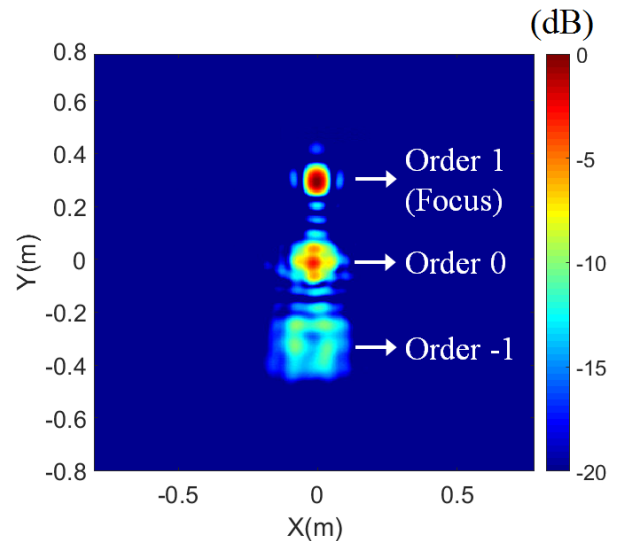
Second, when a plane-wave is used to create the hologram, higher order diffraction modes are observed due to the limited phase-range of the hologram mask, which is selected to be



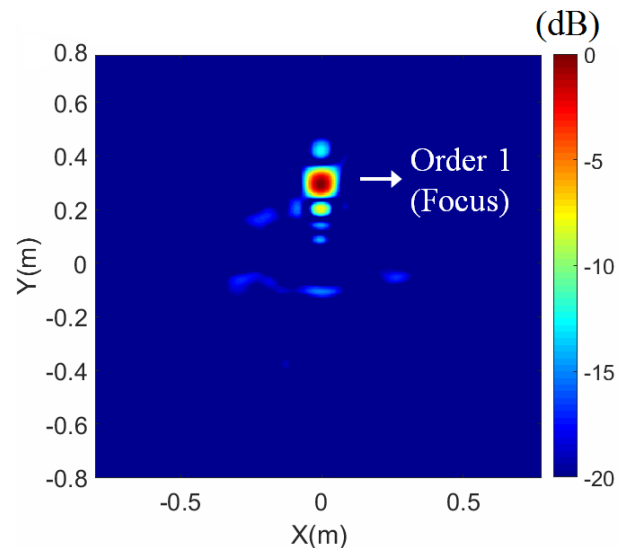
**FIGURE 9.** Radiated E-field pattern from the reconfigurable aperture: (a) range plane and (b) cross-range plane.

within the range of  $\pm 20^\circ$  in this work. This point can be illustrated using an analytical model of the reconfigurable metasurface aperture of Fig. 1, in which the slot elements are modeled as polarizable magnetic dipoles [41]. The radiation patterns can then be computed by summing the contributions of the dipoles, which we do using custom code written in Matlab. To compare with the guided wave reference, we also calculate the focusing characteristics of the holographic aperture designed using a plane-wave reference. For this study, the reference wave,  $\mathbf{H}_{ref}$  in (1) is selected to be a plane-wave (uniform amplitude and phase) and the hologram mask is calculated using (3) for focusing at an arbitrarily selected point  $F_3$  ( $x=0$  m,  $y=0.3$  m,  $z=0.5$  m). The analytically calculated E-field pattern at the focal plane ( $z=0.5$  m) is shown in Fig. 10.

Analyzing Fig. 10, it is evident that, in addition to the main focused beam (fundamental diffraction order – order 1 – of the aperture as highlighted in Fig. 10), higher order diffraction modes (order 0 and order -1) are present. These modes are undesirable because they reduce the amount of power delivered from the fundamental mode and can result in unwanted interference. The -3dB FWHM beam-waist of the focus is calculated to be 5.53 cm for the plane-wave reference scenario.



**FIGURE 10.** Analytical E-field pattern of the holographic aperture using plane-wave as the reference-wave. The aperture focusses at  $F_3$  and the E-field pattern is demonstrated at the focal plane ( $z=0.5$  m).



**FIGURE 11.** Analytical E-field pattern of the holographic aperture using the guided-mode of (1) as the reference-wave. The aperture focusses at  $F_3$  and the E-field pattern is demonstrated at the focal plane ( $z=0.5$  m).

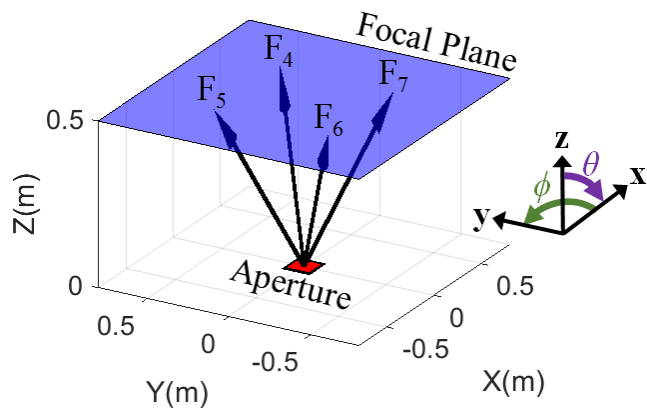
With the cylindrical guided-mode of (1) taken as reference as shown in Figs. 6(a) and 8(a), the undesired higher order diffraction modes are suppressed. To provide the same comparison using the analytical model, we perform the same analytical study presented in Fig. 10, but replace the plane-wave reference with the guided-mode. The E-field pattern at the focal plane,  $z=0.5$  m, is shown in Fig. 11.

Comparing Fig. 11 to Fig. 10, it is evident that when the guided mode is used as the reference wave, the higher order diffracted modes are suppressed, leaving only the main focusing beam. The -3dB beam-waist of the focus is calculated to be 6.3 cm, slightly larger than the beam-waist obtained using the plane-wave reference for the hologram, 5.6 cm. This can be attributed to the fact that while the plane-wave reference has a uniform amplitude, the amplitude of the guided-mode



of (1), which is a Hankel function, decays as a function of  $1/\sqrt{r}$  in the transverse plane ( $xy$ -plane), where  $r$  denotes the distance from the center of the aperture. This suggests that the holographic metasurface aperture with the guided-mode reference exhibits an aperture amplitude tapering, widening the beam-waist of the focus. Another important conclusion from this study can be drawn by comparing the analytical and numerical beam-waist values of the holographic metasurface aperture with the guided-mode reference. Earlier, the numerical beam-waist for the on-axis scenario shown in Fig. 7 was calculated to be 4.9 cm. For the analytical study presented in Fig. 11, the offset angle is  $\theta \approx 30^\circ$ . Using the  $\cos^2\theta$  approximation presented in [40], the expected numerical beam-waist for the off-axis scenario demonstrated in Fig. 11 is 6.5 cm, exhibiting good agreement with the analytical beam-waist value, 6.3 cm.

Next, using the analytical model of the holographic reconfigurable metasurface aperture, we analyze the focusing and higher order mode suppression limits of the aperture. To this end, we sample a number of focusing locations as a function of  $\theta$  and  $\phi$  as depicted in Fig. 12. First, we choose four focusing points for  $\theta = 45^\circ$  and vary  $\phi$  at  $90^\circ$  intervals. As a result, the studied focusing points are defined as follows:



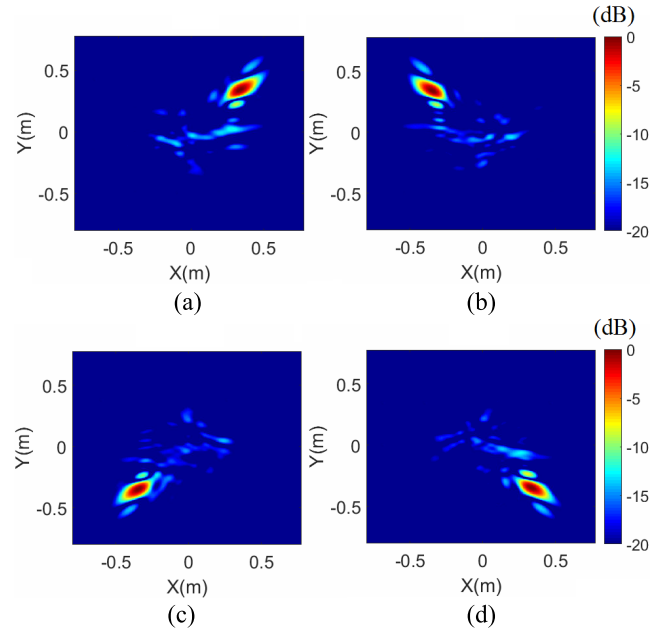
**FIGURE 12.** Depiction of the analyzed focusing locations (constant  $\theta$ , varying  $\phi$ ).

- $F_4(x=0.35 \text{ m}, y=0.35 \text{ m}, z=0.5 \text{ m})$  corresponding to  $(\theta, \phi)=(45^\circ, 45^\circ)$
- $F_5(x=-0.35 \text{ m}, y=0.35 \text{ m}, z=0.5 \text{ m})$  corresponding to  $(\theta, \phi)=(45^\circ, 135^\circ)$
- $F_6(x=-0.35 \text{ m}, y=-0.35 \text{ m}, z=0.5 \text{ m})$  corresponding to  $(\theta, \phi)=(45^\circ, 225^\circ)$
- $F_7(x=0.35 \text{ m}, y=-0.35 \text{ m}, z=0.5 \text{ m})$  corresponding to  $(\theta, \phi)=(45^\circ, 315^\circ)$

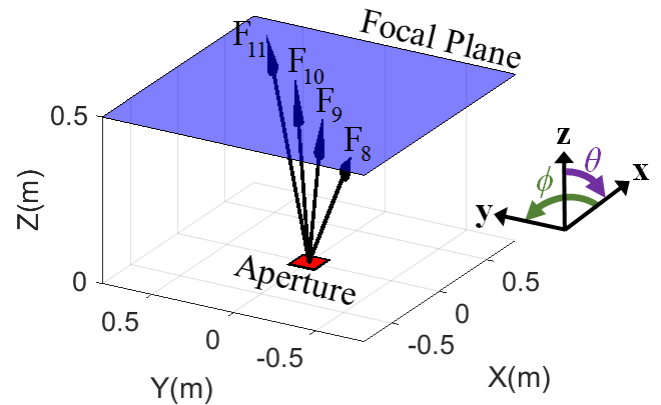
The investigated focusing points for this study,  $F_4, F_5, F_6,$  and  $F_7,$  are depicted in Fig. 12.

In Fig. 13, the E-field patterns for focusing at  $F_4, F_5, F_6$  and  $F_7$  are shown at the focal plane,  $z=0.5 \text{ m}$ .

Analyzing Fig. 13 reveals two important outcomes. First, focusing the beam at an offset angle widens the beam-waist of the focus, similar to the outcome observed in Fig. 9. Second, despite focusing at such extreme angles, the holographic



**FIGURE 13.** Calculated E-field patterns at the focal plane,  $z=0.5 \text{ m}$ . Focusing at (a)  $F_4,$  (b)  $F_5,$  (c)  $F_6,$  and (d)  $F_7.$



**FIGURE 14.** Depiction of the analyzed focusing locations (constant  $\phi$  varying  $\theta$ ).

metasurface aperture successfully suppresses the higher diffraction orders, thanks to the guided-mode reference.

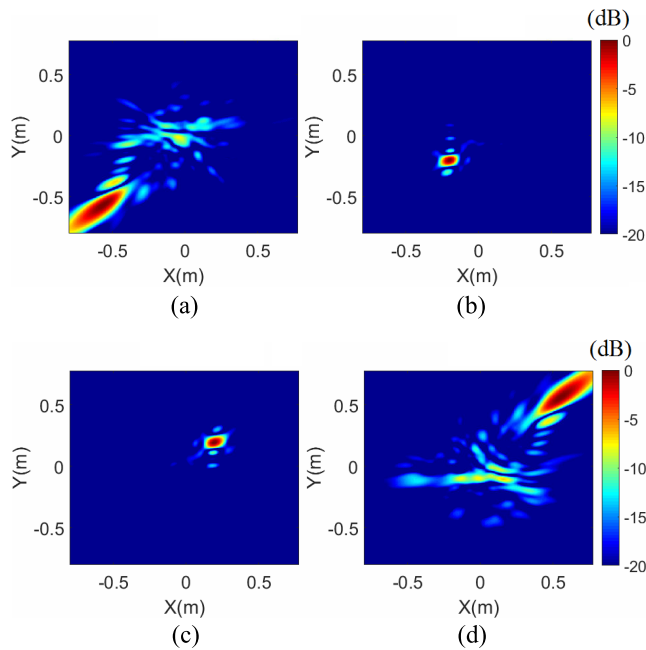
We next study the focusing characteristics and higher diffraction order suppression limits of the aperture as a function of constant  $\phi$  ( $45^\circ$ ) and varying  $\theta$  (from  $-60^\circ$  to  $60^\circ$ ). The sampled focusing points are as follows:

- $F_8(x=-0.61 \text{ m}, y=-0.61 \text{ m}, z=0.5 \text{ m})$  corresponding to  $(\theta, \phi)=(-60^\circ, 45^\circ)$
- $F_9(x=-0.2 \text{ m}, y=-0.2 \text{ m}, z=0.5 \text{ m})$  corresponding to  $(\theta, \phi)=(-30^\circ, 45^\circ)$
- $F_{10}(x=0.2 \text{ m}, y=0.2 \text{ m}, z=0.5 \text{ m})$  corresponding to  $(\theta, \phi)=(30^\circ, 45^\circ)$
- $F_{11}(x=0.61 \text{ m}, y=0.61 \text{ m}, z=0.5 \text{ m})$  corresponding to  $(\theta, \phi)=(60^\circ, 45^\circ)$

The investigated focusing points for this study,  $F_8, F_9, F_{10},$  and  $F_{11},$  are depicted in Fig. 14.

In Fig. 15, the E-field patterns for focusing at  $F_8, F_9, F_{10}$  and  $F_{11}$  are demonstrated at the focal plane,  $z=0.5 \text{ m}$ .

Analyzing Figs. 15(a) and (d), it can be concluded that despite being almost 10 dB down in comparison to the ampli-



**FIGURE 15.** Calculated E-field patterns at the focal plane,  $z=0.5$  m. Focusing at (a)  $F_8$ , (b)  $F_9$ , (c)  $F_{10}$ , and (d)  $F_{11}$ .

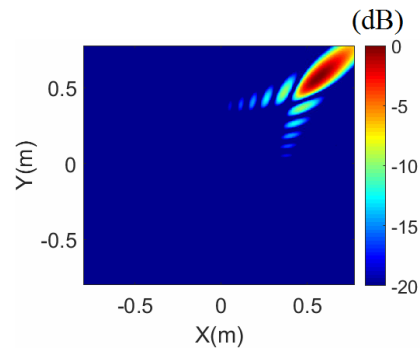
tude of the main focus, at focusing angles of  $\theta=+/-60^\circ$ , order 0 diffraction starts building along the optical axis of the aperture. For the focusing angles of  $\theta=+/-30^\circ$ , on the other hand, no unwanted diffraction orders are present and the focused beams exhibit good fidelity.

As mentioned earlier, when coupled to the guided-mode, the phase advance introduced by the slot unit cells is negligible. As a result, we activate (ON state) only the unit cells satisfying a certain phase threshold range between the guided-mode and the desired phase pattern on the aperture (back-propagated from the fictitious point source to be focused). As given earlier, this threshold range is selected to be  $+/-20^\circ$ . Using more advanced unit cell topologies, one can design a holographic metasurface aperture, exhibiting full phase control. Such a design can convert the phase of the guided-mode to the desired phase pattern on the aperture for focusing, in theory, with 100% accuracy. As an example, using the developed analytical model, in Fig. 16, we analyze focusing at  $F_{11}$  ( $x=0.61$  m,  $y=0.61$  m,  $z=0.5$  m) with full phase control.

Comparing Fig. 16 to Fig. 15(d), it is evident that the beam fidelity is improved and superior suppression in the order 0 diffraction region (along the broadside direction of the aperture) is achieved. This is an ongoing research effort and the results will be reported elsewhere.

#### IV. CONCLUSION

We have presented a reconfigurable holographic metasurface aperture for dynamic beam-focusing applications at 20 GHz. It has been shown that using an aperture consisting of an array of dynamically controlled elements, the phase-front of the aperture radiated fields can be reconfigured, enabling focusing of the aperture radiated fields at any point within the near-field in a dynamic man-



**FIGURE 16.** Calculated E-field pattern at the focal plane,  $z=0.5$  m. Focusing at  $F_{11}$  with full phase control.

ner. The studied numerical (CST Microwave Studio) and analytical results demonstrate the ability of the proposed holographic metasurface aperture to exhibit high-fidelity beam-focusing even when extreme focusing angles are considered. Although demonstrated for K-band frequencies, the developed aperture can readily be scaled to higher frequencies to achieve superior focusing resolution. The reconfigurable holographic metasurface aperture holds significant potential in several applications where having a dynamic control over near-field beams is desired, including near-field imaging [42]–[45], non-destructive testing, biomedical imaging and wireless power transfer [40].

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