

# Subspace detection for range-spread target to suppress interference: exploiting persymmetry in non-homogeneous scenario

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**Abstract:** This paper deals with subspace detection for range-spread target in non-homogeneous clutter with unknown covariance matrix where structured interference is presented in the received data. Through exploiting the persymmetry of the clutter covariance matrix, we propose two adaptive target detectors, which are referred to as persymmetric subspace Rao to suppress interference and persymmetric subspace Wald to suppress interference (“PS-Rao-I” and “PS-Wald-I”), respectively. The persymmetry-based design brings in the advantage of easy implementation for small training sample support. The signal flow analysis of the two detectors shows that the PS-Rao-I rejects interference and integrates signals successively through separated matrix projection, while the PS-Wald-I jointly achieves interference elimination and signal combination via oblique projection. In addition, both detectors are shown to be constant false alarm rate detectors, significantly improving the detection performance with other competing detectors under the condition of limited training.

**Keywords:** adaptive detection, range-spread target, persymmetric structure, Rao test, Wald test.

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## 1. Introduction

Range-spread target detection, especially with application in large scale targets or high-resolution radar, has been extensively studied in recent years [1–6]. To detect a range-spread target, the clutter covariance matrix is generally not known so that a set of training data free of useful signal is used to estimate it. A common assumption concerning training data is the homogeneous environment, namely, the training data is with the same distribution

as the clutter in the cell under test (CUT). However, the homogeneous assumption is difficult to satisfy in practice due to terrain variations and system factors, such as the presence of interference and array configuration. Unlike the homogeneous environment, the partially homogeneous environment (PHE), where both the clutter in CUT and training data share the same covariance matrix up to an unknown power scaling factor, is more robust to power variation between test data and training signals [7–10]. To detect a range-spread target in the PHE, considerable research efforts have been devoted in [11–14].

Apart from the non-homogeneous scenario, the existence of interference caused by electronic countermeasure systems or civil broadcasting system is a key factor affecting the detection performance. It is therefore important to consider interference suppression in the design of detectors. The structured interference modeled by a subspace to describe the multipath effect or uncertainty of direction-of-arrival associated with the interference steering vector, has been intensively investigated in adaptive detection [15–24]. Adaptive detection of a point-like target in the presence of subspace interference for PHE was addressed in [15–16] under the criterion of generalized likelihood ratio test (GLRT) and Wald, and was further investigated within the framework of invariance theory in [17]. When it comes to range-spread targets, the design of GLRT and its two-step variation for partially homogeneous Gaussian noise plus subspace interference was considered in [18], and its relevant Rao detection was provided in [19]. Extensions of [18] to the compound Gaussian environment was made in [20].

A possible solution to decrease the training amount associated with adaptive detectors was reported in [25], by exploiting the persymmetric structure of the disturbance covariance matrix. Thereafter, many detection investigations concerning the persymmetry in the PHE have been developed in [26–31]. Specifically, the authors of [26]

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proposed a persymmetric GLRT resorting to the two-step method, and its corresponding performance assessment was presented in [27]. Persymmetric Rao and persymmetric Wald detectors were established in [28] for point-like targets in the PHE, and was extended to the detection of range-spread targets in [29]. The persymmetric adaptive cosine estimator (ACE) in the PHE was developed in [30]. Meanwhile, adaptive signal detection in PHE and persymmetric Gaussian disturbance was addressed in [31] within the framework of invariance theory. Other examples of persymmetric detectors can be found in [32–45].

The persymmetric GLRT for detecting a range-spread target in structured interference was considered in [46]. It is given that the uniformly most powerful (UMP) test for the detection problem in [46], due to the lack of information on the signal/interference coordinates and the noise covariance matrix. It is therefore of utmost importance to investigate different detectors with various features, among which, Rao and Wald test are most commonly used as an alternative to GLRT with reduced computational complexity and sometimes better performance. As far as we know, no previous work has conducted the adaptive Rao and Wald detection of multi-rank subspace signal in PHE to suppress interference.

In this paper, we adopt Rao and Wald tests for subspace detection of range-spread target embedded in structured interference and non-homogeneous noise. The main contributions are as follows:

(i) In order to relax the restrictions on the number of training signal and ease the computational burden, we devise two persymmetric subspace detectors with the help of unitary transformation, namely the persymmetric subspace Rao detector and the persymmetric subspace Wald detector in the case of subspace interference (referred to as the “PS-Rao-I” and “PS-Wald-I” respectively), which incorporate the persymmetric structure of the disturbance covariance matrix in the design of the detectors.

(ii) Block diagram for the proposed detectors is presented to demonstrate the signal flow of each detector. Specifically, the PS-Rao-I projects the transformed signal into the orthogonal complement of the interference subspace and the signal subspace successively, which leads to separated interference rejection and signal integration process. The PS-Wald-I, in contrast, projects the transformed signal into the signal subspace along the interference subspace via oblique projection, which achieves interference elimination and signal combination simultaneously. Theoretical derivation shows that the two detectors exhibit constant false alarm rate (CFAR) property with respect to (w.r.t.) the unknown covariance matrix and the power scaling factor.

(iii) Numerical examples based on both simulated data

and real radar data are presented to demonstrate the effectiveness and efficiency of the proposed methods.

The remainder of this paper is organized as follows. Section 2 presents the data model. Section 3 contains the derivation and discussion of the proposed tests. Numerical examples and experimental results are provided in Section 4. The concluding remarks are summarized in Section 5.

## 2. Problem statement

Consider a linear array with  $N$  uniformly-spaced sensors receiving the echo signals reflected from a range-spread target in PHE. Our task is to decide whether the target which occupies  $L$  successive range cells, presents or not in the range bin under test. Denote by  $\mathbf{x}_l \in \mathbb{C}^N$ ,  $l = 1, 2, \dots, L$  the data collected from the  $l$ th range bin. The detection problem can be formulated in terms of a binary hypothesis test.

Under hypothesis  $H_1$ ,  $\mathbf{x}_l$  contains signal  $\mathbf{s}_l$ , interference  $\mathbf{i}_l$  and noise  $\mathbf{n}_l$ . The signal and the interference are supposed to lie in two independent subspaces  $\text{Sp}(\mathbf{H})$  and  $\text{Sp}(\mathbf{G})$  spanned by full-column-rank matrices  $\mathbf{H} = [\mathbf{h}_1, \mathbf{h}_2, \dots, \mathbf{h}_r] \in \mathbb{C}^{N \times r}$  and  $\mathbf{G} = [\mathbf{g}_1, \mathbf{g}_2, \dots, \mathbf{g}_s] \in \mathbb{C}^{N \times s}$ , respectively, i.e.,  $\mathbf{s}_l = \mathbf{H}\boldsymbol{\beta}_l$  and  $\mathbf{i}_l = \mathbf{G}\mathbf{q}_l$ , with  $\boldsymbol{\beta}_l \in \mathbb{C}^r$  and  $\mathbf{q}_l \in \mathbb{C}^s$  being the unknown coordinates. By contrast, under the null hypothesis  $H_0$ ,  $\mathbf{x}_l = \mathbf{i}_l + \mathbf{n}_l$ .

In addition to the signal under test, we assume that a signal-free training data set  $\mathbf{x}_{n,k} \in \mathbb{C}^N$  ( $k = 1, 2, \dots, K$ ) that does not contain the signal to be detected can be collected from adjacent range units. That is,  $\mathbf{x}_{n,k} = \mathbf{n}_{n,k}$  with  $\mathbf{n}_{n,k}$  denoting the noise component in the training set which shares the same covariance structure as  $\mathbf{n}_l$ . Assume that  $\mathbf{n}_{n,k}$  are independently identically distributed (i.i.d) Gaussian random vectors with zero mean and unknown covariance matrix  $\boldsymbol{\Delta}$ .  $\mathbf{n}_l$  is modeled similarly but with covariance matrix  $\gamma\boldsymbol{\Delta}$ , where  $\gamma > 0$  stands for an unknown deterministic parameter, which determines the statistical characteristic of the PHE.

The problem of interest at hand can be formulated as the hypothesis test below:

$$\begin{cases} H_0 : \mathbf{X} = \mathbf{G}\mathbf{Q} + \mathbf{N}, \mathbf{X}_K = \mathbf{N}_K \\ H_1 : \mathbf{X} = \mathbf{H}\mathbf{B} + \mathbf{G}\mathbf{Q} + \mathbf{N}, \mathbf{X}_K = \mathbf{N}_K \end{cases} \quad (1)$$

where

$$\mathbf{X} = [\mathbf{x}_1, \mathbf{x}_2, \dots, \mathbf{x}_L] \in \mathbb{C}^{N \times L}, \quad (2)$$

$$\mathbf{N} = [\mathbf{n}_1, \mathbf{n}_2, \dots, \mathbf{n}_L] \in \mathbb{C}^{N \times L}, \quad (3)$$

$$\mathbf{X}_K = [\mathbf{x}_{n,1}, \mathbf{x}_{n,2}, \dots, \mathbf{x}_{n,K}] \in \mathbb{C}^{N \times K}, \quad (4)$$

$$\mathbf{N}_K = [\mathbf{n}_{n,1}, \mathbf{n}_{n,2}, \dots, \mathbf{n}_{n,K}] \in \mathbb{C}^{N \times K}, \quad (5)$$

$$\mathbf{B} = [\boldsymbol{\beta}_1, \boldsymbol{\beta}_2, \dots, \boldsymbol{\beta}_L] \in \mathbb{C}^{r \times L}, \quad (6)$$

and

$$\mathbf{Q} = [\mathbf{q}_1, \mathbf{q}_2, \dots, \mathbf{q}_L] \in \mathbb{C}^{s \times L}. \quad (7)$$

Considering the spatial structure of the uniform linear array,  $\mathbf{h}_i (i = 1, 2, \dots, r)$  and  $\mathbf{g}_i (i = 1, 2, \dots, s)$  satisfy a persymmetric property, i.e.,  $\mathbf{h}_i = \mathbf{J}_N \mathbf{h}_i^*$  and  $\mathbf{g}_i = \mathbf{J}_N \mathbf{g}_i^*$ , where  $\mathbf{J}_N \in \mathbb{C}^N$  stands for the exchange matrix with the unit elements residing on the counter diagonal and all other elements being zero, and  $*$  denotes the conjugate operator. It is straightforward to show that the covariance matrix  $\boldsymbol{\Delta}$  is persymmetric, which can be defined as follows:

$$\boldsymbol{\Delta} = \mathbf{J}_N \boldsymbol{\Delta}^* \mathbf{J}_N. \quad (8)$$

According to the persymmetric property, we can construct a unitary matrix to transform complex matrix  $\boldsymbol{\Delta}$  to a real one. Such a unitary matrix is given by [26]

$$\mathbf{T} = \begin{cases} \frac{1}{\sqrt{2}} \begin{bmatrix} \mathbf{I}_{N/2} & \mathbf{J}_{N/2} \\ \mathbf{jI}_{N/2} & -\mathbf{jJ}_{N/2} \end{bmatrix}, & N \text{ is even} \\ \frac{1}{\sqrt{2}} \begin{bmatrix} \mathbf{I}_{(N-1)/2} & 0 & \mathbf{J}_{(N-1)/2} \\ 0 & \sqrt{2} & 0 \\ \mathbf{jI}_{(N-1)/2} & 0 & -\mathbf{jJ}_{(N-1)/2} \end{bmatrix}, & N \text{ is odd} \end{cases} \quad (9)$$

with  $\mathbf{I}_n$  being the  $n$ -dimensional identity matrix. By exploiting the transformation matrix  $\mathbf{T}$  to the data under test, we can readily express the problem of interest as

$$\begin{cases} \text{H}_0 : \mathbf{Y} = \mathbf{G}_p \mathbf{Q} + \mathbf{N}_p, \mathbf{Y}_K = \mathbf{N}_{pK} \\ \text{H}_1 : \mathbf{Y} = \mathbf{H}_p \mathbf{B} + \mathbf{G}_p \mathbf{Q} + \mathbf{N}_p, \mathbf{Y}_K = \mathbf{N}_{pK} \end{cases} \quad (10)$$

Recall that both  $\mathbf{n}_{n,k}$  and  $\mathbf{n}_l$  are i.i.d Gaussian random vectors with zero mean, they follow the properties of persymmetric Hermitian matrices that  $\mathbf{N}_p \sim \text{CN}(\mathbf{0}, \gamma \mathbf{R})$  and  $\mathbf{N}_{pK} \sim \text{CN}(\mathbf{0}, \mathbf{R})$  where

$$\mathbf{R} = \mathbf{T} \boldsymbol{\Delta} \mathbf{T}^H \quad (11)$$

represents a real symmetric matrix, and  $(\cdot)^H$  denotes conjugate transpose. From a practical view, we know that  $\mathbf{R}$  is unavailable, and we can get its maximum likelihood estimation (MLE) by

$$\widehat{\mathbf{R}}_p = \Re \left( \frac{1}{K} \sum_{k=1}^K \mathbf{y}_k \mathbf{y}_k^H \right) \quad (12)$$

where  $\Re(\cdot)$  denotes the real part of a matrix, and  $\mathbf{y}_k$  is the  $k$ th column of  $\mathbf{Y}_K$ .

### 3. Persymmetric Rao and Wald detector in subspace interference

We now consider the target detection problem in the presence of subspace interference in the PHE, and develop the corresponding PS-Rao-I and PS-Wald-I detector.

#### 3.1 PS-Rao-I detector design

First, we denote  $\boldsymbol{\theta}$  the parameter vector given by

$$\boldsymbol{\theta} = [\boldsymbol{\theta}_r^T, \boldsymbol{\theta}_s^T]^T, \quad (13)$$

where  $\boldsymbol{\theta}_r = \text{vec}(\mathbf{B})$  and  $\boldsymbol{\theta}_s = [\gamma, \text{vec}^T(\mathbf{Q}), \text{vec}^T(\mathbf{R})]$  are called by the interesting parameter and nuisance parameter, with  $\text{vec}(\cdot)$  being the vectorization operation.

According to [47], we can construct the Rao detection as

$$t_{\text{Rao}} = \frac{\partial \ln f_1}{\partial \boldsymbol{\theta}_r} \Big|_{\boldsymbol{\theta}=\widehat{\boldsymbol{\theta}}_0}^T [\mathbf{I}^{-1}(\widehat{\boldsymbol{\theta}}_0)]_{\boldsymbol{\theta}_r, \boldsymbol{\theta}_r} \frac{\partial \ln f_1}{\partial \boldsymbol{\theta}_r^*} \Big|_{\boldsymbol{\theta}=\widehat{\boldsymbol{\theta}}_0} \quad (14)$$

where  $\widehat{\boldsymbol{\theta}}_0$  is the MLE of  $\boldsymbol{\theta}$  under  $\text{H}_0$ .  $f_1$  represents the joint probability density functions (PDF) of  $\mathbf{Y}$  and  $\mathbf{Y}_K$  under  $\text{H}_1$  given by

$$f_1(\mathbf{Y}, \mathbf{Y}_K) = \frac{\exp\{-\text{tr}(\mathbf{R}^{-1} \mathbf{Z} \mathbf{Z}^H / \gamma) - \text{tr}(\mathbf{R}^{-1} \mathbf{M})\}}{\pi^{N(K+L)} \gamma^{NL} \det(\mathbf{R})^{K+L}} \quad (15)$$

where  $\mathbf{Z} \triangleq \mathbf{Y} - \mathbf{S}_p \mathbf{A}$ ,  $\mathbf{S}_p \triangleq [\mathbf{H}_p, \mathbf{G}_p]$ ,  $\mathbf{A} \triangleq [\mathbf{B}^H, \mathbf{Q}^H]^H$ , and  $\mathbf{M} = \mathbf{Y}_K \mathbf{Y}_K^H$  is the sample covariance matrix. Then, denote by  $\mathbf{I}(\boldsymbol{\theta})$  the Fisher information matrix that can be described as

$$\mathbf{I}(\boldsymbol{\theta}) = \text{E} \left\{ \left[ \frac{\partial \ln f_1(\mathbf{Y})}{\partial \boldsymbol{\theta}^*} \right] \left[ \frac{\partial \ln f_1(\mathbf{Y})}{\partial \boldsymbol{\theta}^T} \right] \right\}. \quad (16)$$

Taking the derivative of the logarithm of (15) with respect to  $\mathbf{B}$  and  $\mathbf{B}^*$  leads to

$$\frac{\partial \ln f_1(\mathbf{Y}, \mathbf{Y}_K)}{\partial \text{vec}(\mathbf{B})} = \text{vec}(\mathbf{Z}^H \mathbf{R}^{-1} \mathbf{H}_p)^T / \gamma, \quad (17)$$

and

$$\frac{\partial \ln f_1(\mathbf{Y}, \mathbf{Y}_K)}{\partial \text{vec}(\mathbf{B}^*)} = \text{vec}(\mathbf{H}_p^H \mathbf{R}^{-1} \mathbf{Z}) / \gamma. \quad (18)$$

Note that  $\mathbf{I}(\boldsymbol{\theta})$  can be generally partitioned as

$$\mathbf{I}(\boldsymbol{\theta}) = \begin{bmatrix} \mathbf{I}_{\boldsymbol{\theta}_r, \boldsymbol{\theta}_r}(\boldsymbol{\theta}) & \mathbf{I}_{\boldsymbol{\theta}_r, \boldsymbol{\theta}_s}(\boldsymbol{\theta}) \\ \mathbf{I}_{\boldsymbol{\theta}_s, \boldsymbol{\theta}_r}(\boldsymbol{\theta}) & \mathbf{I}_{\boldsymbol{\theta}_s, \boldsymbol{\theta}_s}(\boldsymbol{\theta}) \end{bmatrix}. \quad (19)$$

Substituting (17) and (18), into (16), yields,

$$\begin{aligned} \mathbf{I}_{\boldsymbol{\theta}_r, \boldsymbol{\theta}_r}(\boldsymbol{\theta}) &= \text{E}[\text{vec}(\mathbf{H}_p^H \mathbf{R}^{-1} \mathbf{Z} / \gamma) \text{vec}^T(\mathbf{H}_p^T \mathbf{R}^{-1} \mathbf{Z}^* / \gamma)] = \\ &= \text{E}\{(\mathbf{I}_L \otimes \mathbf{H}_p^H \mathbf{R}^{-1}) \text{vec}(\mathbf{Z}) \cdot [(\mathbf{I}_L \otimes \mathbf{H}_p^T \mathbf{R}^{-1}) \text{vec}(\mathbf{Z}^*)]^T\} / \gamma^2 = \\ &= (\mathbf{I}_L \otimes \mathbf{H}_p^H \mathbf{R}^{-1} \mathbf{H}_p) / \gamma. \end{aligned} \quad (20)$$

Likewise, it is straightforward to verify that  $\mathbf{I}_{\boldsymbol{\theta}_r, \boldsymbol{\theta}_s}(\boldsymbol{\theta})$  is a null matrix, which then results in

$$[\mathbf{I}^{-1}(\boldsymbol{\theta})]_{\boldsymbol{\theta}_r, \boldsymbol{\theta}_r} = \mathbf{I}_{\boldsymbol{\theta}_r, \boldsymbol{\theta}_r}^{-1} = \gamma (\mathbf{I}_L \otimes \mathbf{H}_p^H \mathbf{R}^{-1} \mathbf{H}_p)^{-1}. \quad (21)$$

Inserting (17), (18) and (21) into (14), and setting  $\mathbf{B} = \mathbf{0}_{r \times L}$  leads to the PS-Rao-I test for given  $\gamma$ ,  $\mathbf{Q}$  and  $\mathbf{R}$ ,

$$\begin{aligned} t_{\text{PS-Rao-I}} &= \text{vec}^T \left[ (\mathbf{Y} - \mathbf{G}_p \mathbf{Q})^H \mathbf{R}^{-1} \mathbf{H}_p \right]^T \cdot \\ &= [\mathbf{I}_L \otimes (\mathbf{H}_p^H \mathbf{R}^{-1} \mathbf{H}_p)^{-1}] \cdot \text{vec} \left[ (\mathbf{H}_p^H \mathbf{R}^{-1} (\mathbf{Y} - \mathbf{G}_p \mathbf{Q})) \right] / \gamma = \\ &= \text{tr}[(\mathbf{Y} - \mathbf{G}_p \mathbf{Q})^H \mathbf{R}^{-1} \mathbf{H}_p (\mathbf{H}_p^H \mathbf{R}^{-1} \mathbf{H}_p)^{-1}]. \end{aligned}$$

$$\mathbf{H}_p^H \mathbf{R}^{-1} (\mathbf{Y} - \mathbf{G}_p \mathbf{Q}) / \gamma \quad (22)$$

where  $\text{tr}(\cdot)$  denotes the trace of a matrix.

Next, we need to estimate  $\mathbf{Q}$  in (22). By taking the derivative of the logarithm of (15) with  $\mathbf{B} = \mathbf{0}_{r \times L}$  with respect to  $\mathbf{Q}$ , we have, for given  $\mathbf{R}$ , the estimate of  $\mathbf{Q}$

$$\widehat{\mathbf{Q}} = (\overline{\mathbf{G}}_p^H \overline{\mathbf{G}}_p)^{-1} \overline{\mathbf{G}}_p^H \overline{\mathbf{Y}} \quad (23)$$

where  $\overline{\mathbf{G}}_p = \mathbf{R}^{-1/2} \mathbf{G}_p$  and  $\overline{\mathbf{Y}} = \mathbf{R}^{-1/2} \mathbf{Y}$ .

Inserting (23) into (22) outputs the PS-Rao-I test for given  $\gamma$  and  $\mathbf{R}$

$$t_{\text{PS-Rao-I}} = \text{tr} \left( \overline{\mathbf{Y}}^H \mathbf{P}_{\overline{\mathbf{G}}_p}^\perp \mathbf{P}_{\overline{\mathbf{H}}_p} \mathbf{P}_{\overline{\mathbf{G}}_p}^\perp \overline{\mathbf{Y}} \right) / \gamma \quad (24)$$

where  $\overline{\mathbf{H}}_p = \mathbf{R}^{-1/2} \mathbf{H}_p$ .

Substituting  $\widehat{\mathbf{Q}}$  into (15) with  $\mathbf{B} = \mathbf{0}_{r \times L}$  and ignoring the PDF of  $\mathbf{Y}_K$  yields

$$\begin{aligned} f_1(\mathbf{Y}, \mathbf{Y}_K) = \\ \frac{\exp \left\{ -\text{tr} \left[ \mathbf{R}^{-1} (\mathbf{Y} - \mathbf{G}_p \widehat{\mathbf{Q}}) (\mathbf{Y} - \mathbf{G}_p \widehat{\mathbf{Q}})^H / \gamma \right] \right\}}{\pi^{\text{NL}} \gamma^{\text{NL}} \det(\mathbf{R})^L} = \\ \frac{\exp \left\{ -\text{tr} \left( \overline{\mathbf{Y}}^H \mathbf{P}_{\overline{\mathbf{G}}_p}^\perp \overline{\mathbf{Y}} \right) \right\}}{\pi^{\text{NL}} \gamma^{\text{NL}} \det(\mathbf{R})^L}. \end{aligned} \quad (25)$$

By taking the derivative of the logarithm of (25) with respect to  $\gamma$  and equating it to zeros results in the MLE of  $\gamma$  for given  $\mathbf{R}$  under  $\text{H}_0$  as

$$\hat{\gamma}_0 = \text{tr}(\overline{\mathbf{Y}}^H \mathbf{P}_{\overline{\mathbf{G}}_p}^\perp \overline{\mathbf{Y}}) / \text{NL}. \quad (26)$$

Inserting (26) into (24) and neglecting the constant items leads to the PS-Rao-I test for given  $\mathbf{R}$ :

$$t_{\text{PS-Rao-I}} = \frac{\text{tr}(\overline{\mathbf{Y}}^H \mathbf{P}_{\overline{\mathbf{G}}_p}^\perp \mathbf{P}_{\overline{\mathbf{H}}_p} \mathbf{P}_{\overline{\mathbf{G}}_p}^\perp \overline{\mathbf{Y}})}{\text{tr}(\overline{\mathbf{Y}}^H \mathbf{P}_{\overline{\mathbf{G}}_p}^\perp \overline{\mathbf{Y}})}. \quad (27)$$

Replacing  $\mathbf{R}$  with the MLE according to (12), we have the persymmetric Rao detector in subspace interference

$$t_{\text{PS-Rao-I}} = \frac{\text{tr}(\widetilde{\mathbf{Y}}^H \mathbf{P}_{\widetilde{\mathbf{G}}_p}^\perp \mathbf{P}_{\widetilde{\mathbf{H}}_p} \mathbf{P}_{\widetilde{\mathbf{G}}_p}^\perp \widetilde{\mathbf{Y}})}{\text{tr}(\widetilde{\mathbf{Y}}^H \mathbf{P}_{\widetilde{\mathbf{G}}_p}^\perp \widetilde{\mathbf{Y}})} = \frac{\|\mathbf{P}_{\widetilde{\mathbf{H}}_p} \mathbf{P}_{\widetilde{\mathbf{G}}_p}^\perp \widetilde{\mathbf{Y}}\|_F^2}{\|\mathbf{P}_{\widetilde{\mathbf{G}}_p}^\perp \widetilde{\mathbf{Y}}\|_F^2} \quad (28)$$

where  $\widetilde{\mathbf{G}}_p = \widehat{\mathbf{R}}_p^{-1/2} \mathbf{G}_p$ ,  $\widetilde{\mathbf{H}}_p = \widehat{\mathbf{R}}_p^{-1/2} \mathbf{H}$ ,  $\widetilde{\mathbf{Y}} = \widehat{\mathbf{R}}_p^{-1/2} \mathbf{Y}$ , and  $\|\cdot\|_F$  denotes the Frobenius norm.

### 3.2 PS-Wald-I detector design

We now consider the Wald test for the detection problem described in (10). According to [48], we can express the Wald test as

$$t_{\text{Wald}} = (\widehat{\boldsymbol{\theta}}_{r_1} - \boldsymbol{\theta}_{r_0})^H \{ [\mathbf{I}^{-1}(\widehat{\boldsymbol{\theta}}_1)]_{\boldsymbol{\theta}_r, \boldsymbol{\theta}_r} \}^{-1} (\widehat{\boldsymbol{\theta}}_{r_1} - \boldsymbol{\theta}_{r_0}) \quad (29)$$

where  $\widehat{\boldsymbol{\theta}}_{r_1}$  denotes the MLE of  $\boldsymbol{\theta}_{r_1}$  under hypothesis  $\text{H}_1$ ,

$\boldsymbol{\theta}_{r_0}$  is the value of  $\boldsymbol{\theta}_{r_1}$  under  $\text{H}_0$ , and  $\{ [\mathbf{I}^{-1}(\widehat{\boldsymbol{\theta}}_1)]_{\boldsymbol{\theta}_r, \boldsymbol{\theta}_r} \}^{-1}$  stands for the Schur complement of  $\mathbf{I}_{\boldsymbol{\theta}_r, \boldsymbol{\theta}_r}$  evaluated at  $\widehat{\boldsymbol{\theta}}_1$ .

To derive the PS-Wald-I test, we need the MLE of  $\mathbf{B}$  and  $\gamma$ . To obtain the MLE of  $\mathbf{B}$ , we start by deriving the MLE of  $\mathbf{A}$ , which can be obtained by nulling the derivative of  $\ln f_1(\mathbf{Y})$  with respect to  $\mathbf{A}$ , for given  $\mathbf{R}$ :

$$\widehat{\mathbf{A}} = (\mathbf{S}_p^H \mathbf{R}^{-1} \mathbf{S}_p)^{-1} \mathbf{S}_p^H \mathbf{R}^{-1} \mathbf{Y}. \quad (30)$$

Define

$$\mathbf{C} \triangleq (\mathbf{S}_p^H \mathbf{R}^{-1} \mathbf{S}_p)^{-1} = \begin{bmatrix} \mathbf{C}_{11} & \mathbf{C}_{12} \\ \mathbf{C}_{21} & \mathbf{C}_{22} \end{bmatrix}. \quad (31)$$

Given that

$$\mathbf{S}_p^H \mathbf{R}^{-1} \mathbf{S}_p = \begin{bmatrix} \mathbf{H}_p^H \mathbf{R}^{-1} \mathbf{H}_p & \mathbf{H}_p^H \mathbf{R}^{-1} \mathbf{G}_p \\ \mathbf{G}_p^H \mathbf{R}^{-1} \mathbf{H}_p & \mathbf{G}_p^H \mathbf{R}^{-1} \mathbf{G}_p \end{bmatrix}. \quad (32)$$

According to the theorem about the inverse of a partitioned matrix, we have

$$\begin{aligned} \mathbf{C}_{11}^{-1} &= \mathbf{H}_p^H \mathbf{R}^{-1} \mathbf{H}_p - \\ &\mathbf{H}_p^H \mathbf{R}^{-1} \mathbf{G}_p (\mathbf{G}_p^H \mathbf{R}^{-1} \mathbf{G}_p)^{-1} \mathbf{G}_p^H \mathbf{R}^{-1} \mathbf{H}_p, \end{aligned} \quad (33)$$

and

$$\mathbf{C}_{12} = -\mathbf{C}_{11} \mathbf{H}_p^H \mathbf{R}^{-1} \mathbf{G}_p (\mathbf{G}_p^H \mathbf{R}^{-1} \mathbf{G}_p)^{-1}. \quad (34)$$

As a result,

$$\begin{aligned} \widehat{\mathbf{B}} &= \mathbf{C}_{11} \mathbf{H}_p^H \mathbf{R}^{-1} \mathbf{Y} + \mathbf{C}_{12} \mathbf{G}_p^H \mathbf{R}^{-1} \mathbf{Y} = \\ &\mathbf{C}_{11} [\mathbf{H}_p^H \mathbf{R}^{-1} \mathbf{Y} - \mathbf{H}_p^H \mathbf{R}^{-1} \mathbf{G}_p (\mathbf{G}_p^H \mathbf{R}^{-1} \mathbf{G}_p)^{-1} \mathbf{G}_p^H \mathbf{R}^{-1} \mathbf{Y}] = \\ &(\overline{\mathbf{H}}_p^H \mathbf{P}_{\overline{\mathbf{G}}_p}^\perp \overline{\mathbf{H}}_p)^{-1} \overline{\mathbf{H}}_p^H \mathbf{P}_{\overline{\mathbf{G}}_p}^\perp \overline{\mathbf{Y}}. \end{aligned} \quad (35)$$

Substituting (21) and (35) into (29) outputs the PS-Wald-I test, which can be represented as

$$\begin{aligned} t_{\text{Wald}} &= \text{vec}^H((\overline{\mathbf{H}}_p^H \mathbf{P}_{\overline{\mathbf{G}}_p}^\perp \overline{\mathbf{H}}_p)^{-1} \overline{\mathbf{H}}_p^H \mathbf{P}_{\overline{\mathbf{G}}_p}^\perp \overline{\mathbf{Y}}) \cdot \\ &(\mathbf{I}_L \otimes \mathbf{H}_p^H \mathbf{R}^{-1} \mathbf{H}_p) \text{vec}((\overline{\mathbf{H}}_p^H \mathbf{P}_{\overline{\mathbf{G}}_p}^\perp \overline{\mathbf{H}}_p)^{-1} \overline{\mathbf{H}}_p^H \mathbf{P}_{\overline{\mathbf{G}}_p}^\perp \overline{\mathbf{Y}}) / \gamma = \\ &\text{tr}(\overline{\mathbf{Y}}^H \mathbf{P}_{\overline{\mathbf{G}}_p}^\perp \overline{\mathbf{H}}_p (\overline{\mathbf{H}}_p^H \mathbf{P}_{\overline{\mathbf{G}}_p}^\perp \overline{\mathbf{H}}_p)^{-1} \overline{\mathbf{H}}_p^H \overline{\mathbf{H}}_p \cdot \\ &(\overline{\mathbf{H}}_p^H \mathbf{P}_{\overline{\mathbf{G}}_p}^\perp \overline{\mathbf{H}}_p)^{-1} \overline{\mathbf{H}}_p^H \mathbf{P}_{\overline{\mathbf{G}}_p}^\perp \overline{\mathbf{Y}}) / \gamma = \\ &\text{tr}(\overline{\mathbf{Y}}^H \mathbf{P}_{\overline{\mathbf{H}}_p | \overline{\mathbf{G}}_p}^H \mathbf{P}_{\overline{\mathbf{H}}_p | \overline{\mathbf{G}}_p} \overline{\mathbf{Y}}) / \gamma \end{aligned} \quad (36)$$

where

$$\mathbf{P}_{\overline{\mathbf{H}}_p | \overline{\mathbf{G}}_p} = \overline{\mathbf{H}}_p (\overline{\mathbf{H}}_p^H \mathbf{P}_{\overline{\mathbf{G}}_p}^\perp \overline{\mathbf{H}}_p)^{-1} \overline{\mathbf{H}}_p^H \mathbf{P}_{\overline{\mathbf{G}}_p}^\perp \quad (37)$$

is the oblique projection matrix onto the subspace  $\text{Sp}(\overline{\mathbf{H}}_p)$  along the subspace  $\text{Sp}(\overline{\mathbf{G}}_p)$ .

Substituting  $\widehat{\mathbf{A}}$  into (15), and nulling its derivative with respect to  $\gamma$  result in the MLE of  $\gamma$  for given  $\mathbf{R}$  under  $\text{H}_1$  as

$$\hat{\gamma}_1 = \text{tr}(\overline{\mathbf{Y}}^H \mathbf{P}_{\overline{\mathbf{S}}_p}^\perp \overline{\mathbf{Y}}) / \text{NL} \quad (38)$$

where  $\overline{\mathbf{S}}_p = \mathbf{R}^{-1/2} \mathbf{S}_p$ ,  $\mathbf{P}_{\overline{\mathbf{S}}_p}^\perp = \mathbf{I}_N - \mathbf{P}_{\overline{\mathbf{S}}_p}$ , and

$$\mathbf{P}_{\overline{\mathbf{S}}_p} = \mathbf{P}_{\overline{\mathbf{H}}_p | \overline{\mathbf{G}}_p} + \mathbf{P}_{\overline{\mathbf{G}}_p | \overline{\mathbf{H}}_p}. \quad (39)$$

Replacing  $\mathbf{R}$  by its MLE calculated as (12), inserting (38) into (36), and neglecting the constant items leads to the persymmetric Wald detector in the subspace interference condition given by

$$t_{\text{PS-Wald-I}} = \frac{\text{tr}(\tilde{\mathbf{Y}}^H \mathbf{P}_{\tilde{\mathbf{H}}_p | \tilde{\mathbf{G}}_p}^H \mathbf{P}_{\tilde{\mathbf{H}}_p | \tilde{\mathbf{G}}_p} \tilde{\mathbf{Y}})}{\text{tr}(\tilde{\mathbf{Y}}^H \mathbf{P}_{\tilde{\mathbf{S}}_p}^\perp \tilde{\mathbf{Y}})} = \frac{\|\mathbf{P}_{\tilde{\mathbf{H}}_p | \tilde{\mathbf{G}}_p} \tilde{\mathbf{Y}}\|_F^2}{\|\mathbf{P}_{\tilde{\mathbf{S}}_p}^\perp \tilde{\mathbf{Y}}\|_F^2} \quad (40)$$

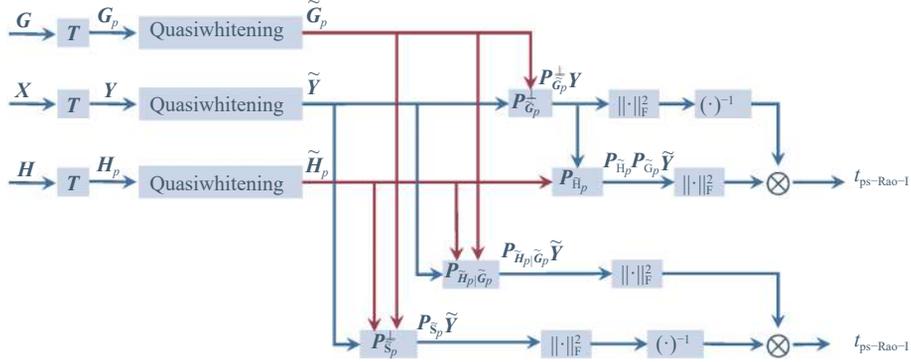


Fig. 1 Block diagram of PS-Rao-I and PS-Wald-I detector.

(i) Numerator of Test Statistic: The numerator of both tests represents the energy of the transformed signal after subspace projection which, unlike their interference-free counterparts, differs from each other. Specifically, the PS-Rao-I projects the transformed signal  $\tilde{\mathbf{Y}}$  successively into the orthogonal complement of the quasi-whitened interference subspace (via  $\mathbf{P}_{\tilde{\mathbf{G}}_p}^\perp$ ) which rejects the interference, and the quasi-whitened signal subspace (via  $\mathbf{P}_{\tilde{\mathbf{H}}_p}$ ) which combines the signal. The PS-Wald-I, on the other hand, projects  $\tilde{\mathbf{Y}}$  into the transformed signal subspace along the transformed interference subspace via oblique projection matrix  $\mathbf{P}_{\tilde{\mathbf{H}}_p | \tilde{\mathbf{G}}_p}$ , which achieves interference elimination and signal combination simultaneously.

(ii) Denominator of Test Statistic: The denominator of the two detectors also differs from each other. The PS-Rao-I test computes the denominator from the transformed signal after projection into the orthogonal complement of the quasi-whitened interference subspace, which does not contain the interference and therefore corresponds to the calculation of  $\gamma$  under  $H_0$  situation, while the PS-Wald-I test computes the denominator from the transformed signal after projection into the orthogonal complement of the quasi-whitened noise plus interference subspace, which cancels both the signal and the interference, and in consequence corresponds to the calculation of  $\gamma$  under  $H_1$  situation.

It is worth highlighting that both detectors (the PS-Rao-I and the PS-Wald-I) are CFAR w.r.t. the covariance matrix  $\mathbf{R}$  and  $\gamma$ . The CFARness of the PS-Rao-I test can be proved in a manner similar to [19]. For brevity, we will

where  $\tilde{\mathbf{S}}_p = \widehat{\mathbf{R}}_p^{-1/2} \mathbf{S}_p$ .

### 3.3 Discussions

In the following, we investigate the signal operation and CFAR property of the proposed detectors.

The signal flow of the proposed PS-Rao-I and PS-Wald-I detectors is shown in Fig. 1, where the red lines indicate the signal flow for matrix projection.

not explore it in this article.

The CFAR property of the PS-Wald-I detector is investigated and discussed in the following. Since the CFAR property will not be affected by the trace operator, we first rewrite (40) as

$$t_{\text{PS-Wald-I}} = \frac{\text{tr}(\boldsymbol{\Psi}_1)}{\text{tr}(\boldsymbol{\Psi}_2)} \quad (41)$$

where

$$\boldsymbol{\Psi}_1 = \gamma \tilde{\mathbf{Y}}_0^H \tilde{\mathbf{F}} \tilde{\mathbf{H}}_p (\tilde{\mathbf{H}}_p^H \tilde{\mathbf{F}} \tilde{\mathbf{H}}_p)^{-1} \tilde{\mathbf{H}}_p \tilde{\mathbf{R}}^{-1} \tilde{\mathbf{H}}_p, \quad (42)$$

$$(\tilde{\mathbf{H}}_p^H \tilde{\mathbf{F}} \tilde{\mathbf{H}}_p)^{-1} \tilde{\mathbf{H}}_p \tilde{\mathbf{F}} \tilde{\mathbf{Y}}_0,$$

$$\boldsymbol{\Psi}_2 = \gamma \left[ \tilde{\mathbf{Y}}_0^H \tilde{\mathbf{R}}^{-1} \tilde{\mathbf{Y}}_0 - \tilde{\mathbf{Y}}_0^H \tilde{\mathbf{R}}^{-1} \tilde{\mathbf{S}}_p (\tilde{\mathbf{S}}_p^H \tilde{\mathbf{R}}^{-1} \tilde{\mathbf{S}}_p)^{-1} \tilde{\mathbf{S}}_p \tilde{\mathbf{R}}^{-1} \tilde{\mathbf{Y}}_0 \right], \quad (43)$$

$$\begin{cases} \tilde{\mathbf{F}} = \mathbf{R}^{1/2} \mathbf{F} \mathbf{R}^{1/2} \\ \tilde{\mathbf{G}}_p = \mathbf{R}^{-1/2} \mathbf{G}_p \end{cases} \quad (44)$$

with

$$\mathbf{F} = \widehat{\mathbf{R}}_p^{-1} - \widehat{\mathbf{R}}_p^{-1} \mathbf{G}_p (\mathbf{G}_p^H \widehat{\mathbf{R}}_p^{-1} \mathbf{G}_p)^{-1} \mathbf{G}_p^H \widehat{\mathbf{R}}_p^{-1}. \quad (45)$$

It is observed that the ratio of  $\boldsymbol{\Psi}_1$  to  $\boldsymbol{\Psi}_2$  cancels their dependence on  $\gamma$ , which suggests the CFARness of the PS-Wald-I detector with respect to  $\gamma$  under hypothesis  $H_0$ .

In the following, we show the CFARness of (41) with respect to  $\mathbf{R}$  under  $H_0$ . Let  $\tilde{\mathbf{U}} \triangleq [\tilde{\mathbf{G}}_{p//}, \tilde{\mathbf{G}}_{p\perp}]$  be a unitary matrix, with  $\tilde{\mathbf{G}}_{p//} = \tilde{\mathbf{G}}_p (\tilde{\mathbf{G}}_p^H \tilde{\mathbf{G}}_p)^{-1}$  and  $\tilde{\mathbf{G}}_{p\perp}^H \tilde{\mathbf{G}}_{p//} = \mathbf{0}_{(N-s) \times s}$ . Define

$$\begin{cases} \bar{Y}_U \triangleq \bar{U}^H \bar{Y}_0 \\ \bar{F}_U \triangleq \bar{U}^H \bar{F} \bar{U} \\ \bar{H}_{U\bar{U}} \triangleq \bar{U}^H \bar{H}_p \\ \bar{R}_U \triangleq \bar{U}^H \bar{R} \bar{U} \end{cases} \quad (46)$$

After some matrix manipulation, we can express  $\Psi_1$  as

$$\begin{aligned} \Psi_1 &= \gamma \bar{Y}_U^H \bar{F}_U \bar{H}_{U\bar{U}} (\bar{H}_{U\bar{U}}^H \bar{F}_U \bar{H}_{U\bar{U}})^{-1} \bar{H}_U^H \bar{R}_U^{-1} \\ &\quad \bar{H}_{U\bar{U}} (\bar{H}_{U\bar{U}}^H \bar{F}_U \bar{H}_{U\bar{U}})^{-1} \bar{H}_U^H \bar{F}_U \bar{Y}_U = \\ &\quad \gamma \bar{Y}_2^H \bar{R}_{22}^{-1} \bar{H}_2 (\bar{H}_2^H \bar{R}_{22}^{-1} \bar{H}_2)^{-1} \bar{H}_U^H \bar{R}_U^{-1} \bar{H}_{U\bar{U}} \\ &\quad (\bar{H}_2^H \bar{R}_{22}^{-1} \bar{H}_2)^{-1} \bar{H}_2^H \bar{R}_{22}^{-1} \bar{Y}_2 \end{aligned} \quad (47)$$

where  $\bar{Y}_U \triangleq [\bar{Y}_1^T \ \bar{Y}_2^T]^T$ ,  $\bar{H}_U \triangleq [\bar{H}_1^T \ \bar{H}_2^T]^T$  and  $\bar{R}_U \triangleq [\bar{R}_{11}, \bar{R}_{12}; \bar{R}_{21}, \bar{R}_{22}]$ . The dimensions of  $\bar{Y}_2$ ,  $\bar{R}_{22}$  and  $\bar{H}_2$  are  $(N-s) \times L$ ,  $(N-s) \times (N-s)$  and  $(N-s) \times r$ , respectively. It can be verified that under  $H_0$

$$\bar{Y}_2 \sim \text{CN}(\mathbf{0}_{(N-s) \times L}, \mathbf{I}_{N-s}), \quad (48)$$

and

$$\bar{R}_{22} \sim \text{CW}(K, \mathbf{I}_{N-s}). \quad (49)$$

Let  $V = [\bar{H}_{2//}, \bar{H}_{2\perp}]$  be a unitary matrix, with  $\bar{H}_{2//} = \bar{H}_2 (\bar{H}_2^H \bar{H}_2)^{-1}$  and  $\bar{H}_{2\perp} \bar{H}_{2//} = \mathbf{0}_{(N-s-r) \times r}$ . Define

$$\begin{cases} \bar{Y}_{2V} \triangleq V^H \bar{Y}_2 \\ \bar{R}_{22V} \triangleq V^H \bar{R}_{22} V \\ E_1 \triangleq V^H \bar{H}_2 \end{cases} \quad (50)$$

$\Psi_1$  can be reformulated as

$$\begin{aligned} \Psi_1 &= \gamma \bar{Y}_{2V}^H \bar{R}_{22V}^{-1} E_1 (E_1^H \bar{R}_{22V}^{-1} E_1)^{-1} \bar{H}_U^H \bar{R}_U^{-1} \bar{H}_{U\bar{U}} \\ &\quad (E_1^H \bar{R}_{22V}^{-1} E_1)^{-1} E_1^H \bar{R}_{22V}^{-1} \bar{Y}_{2V}. \end{aligned} \quad (51)$$

It can be verified that under  $H_0$

$$\bar{Y}_{2V} \sim \text{CN}(\mathbf{0}_{(N-s) \times L}, \mathbf{I}_{N-s}), \quad (52)$$

$$\bar{R}_{22V} \sim \text{CW}(K, \mathbf{I}_{N-s}), \quad (53)$$

and

$$E_1 = \begin{bmatrix} \mathbf{I}_r \\ \mathbf{0}_{(N-r-s) \times r} \end{bmatrix} \quad (54)$$

which suggests that both  $\bar{Y}_{2V}^H \bar{R}_{22V}^{-1} E_1$  and  $E_1^H \bar{R}_{22V}^{-1} E_1$  are independent of  $\mathbf{R}$  under hypothesis  $H_0$ .

Let  $\hat{V} \triangleq [\hat{H}_{U//}, \hat{H}_{U\perp}]$  be a unitary matrix, with  $\hat{H}_{U//} = \hat{H}_U (\hat{H}_U^H \hat{H}_U)^{-1}$  and  $\hat{H}_{U\perp} \hat{H}_{U//} = \mathbf{0}_{(N-r) \times r}$ . By implementing unitary transformation on  $\hat{H}_U$  and  $\hat{R}_U^{-1}$ ,  $\Psi_1$  can be reformulated as

$$\begin{aligned} \Psi_1 &= \gamma \bar{Y}_{2V}^H \bar{R}_{22V}^{-1} E_1 (E_1^H \bar{R}_{22V}^{-1} E_1)^{-1} E_2^H \bar{R}_{U\bar{V}}^{-1} E_2 \\ &\quad (E_1^H \bar{R}_{22V}^{-1} E_1)^{-1} E_1^H \bar{R}_{22V}^{-1} \bar{Y}_{2V} \end{aligned} \quad (55)$$

where  $\bar{R}_{U\bar{V}} = \hat{V}^H \bar{R}_U \hat{V}$  and

$$E_2 = \begin{bmatrix} \mathbf{I}_r \\ \mathbf{0}_{(N-r) \times r} \end{bmatrix}. \quad (56)$$

It follows that  $E_2^H \bar{R}_{U\bar{V}}^{-1} E_2$  is independent of  $\mathbf{R}$ . Gathering the above information, we can conclude that  $\Psi_1$  is independent of  $\mathbf{R}$ .

We then proceed to show the statistical independence of  $\Psi_2$  on  $\mathbf{R}$ . Let  $U' \triangleq [\bar{S}_{p//}, \bar{S}_{p\perp}]$  be a unitary matrix, with  $\bar{S}_{p//} = \bar{S}_p (\bar{S}_p^H \bar{S}_p)^{-1}$ ,  $\bar{S}_{p\perp} \bar{S}_{p//} = \mathbf{0}_{(N-r-s) \times (r+s)}$ . By implementing unitary transformation on  $\hat{H}_U$  and  $\hat{R}_U^{-1}$ ,  $\Psi_2$  can be reformulated as

$$\begin{aligned} \Psi_2 &= \gamma [\bar{Y}_0^H \bar{R}^{-1} \bar{Y}_0 - \\ &\quad \bar{Y}_{U'}^H \bar{R}_{U'}^{-1} E_S (E_S^H \bar{R}_{U'}^{-1} E_S)^{-1} E_S^H \bar{R}_{U'}^{-1} \bar{Y}_{U'}] \end{aligned} \quad (57)$$

where  $\bar{Y}_{U'} \triangleq U'^H \bar{Y}_0$ ,  $\bar{R}_{U'} = U'^H \bar{R} U'$ , and

$$E_S \triangleq U'^H \bar{S}_p = \begin{bmatrix} \mathbf{I}_{r+s} \\ \mathbf{0}_{(N-r-s) \times (r+s)} \end{bmatrix}. \quad (58)$$

It can be verified that under hypothesis  $H_0$

$$\bar{Y}_{U'} \sim \text{CN}(\mathbf{0}_{N \times L}, \mathbf{I}_N), \quad (59)$$

$$\bar{R}_{U'} \sim \text{CW}(K, \mathbf{I}_N). \quad (60)$$

On the basis of the above derivation, we can conclude that  $\Psi_2$  is independent of  $\mathbf{R}$  under hypothesis  $H_0$ . Therefore, the CFARness of the PS-Wald-I follows.

## 4. Numerical results

We now consider the target detection problem in the presence of subspace interference in the PHE, and develop the corresponding PS-Rao-I and PS-Wald-I detector.

### 4.1 Simulation results

We now present numerical examples to verify our analysis and compare the proposed persymmetric detectors. In the following simulations, we set  $N = 8$ ,  $L = 4$ . Let  $\Delta$  be an exponentially correlated covariance matrix with one-lag correlation coefficient  $\rho = 0.9$ , i.e., the  $(i, j)$ th element of the noise covariance matrix  $\Delta$  is set to  $\rho^{|i-j|}$ . In this paper, we define the signal to noise ratio (SNR) and the interference to noise ratio (INR) as

$$\text{SNR} = \text{tr}(\mathbf{B}^H \mathbf{H}^H \Delta^{-1} \mathbf{H} \mathbf{B}), \quad (61)$$

$$\text{INR} = \text{tr}(\mathbf{Q}^H \mathbf{G}^H \Delta^{-1} \mathbf{G} \mathbf{Q}). \quad (62)$$

Column vectors of  $\mathbf{H}$  and  $\mathbf{G}$  are respectively given by

$$\mathbf{h}_i = \frac{1}{\sqrt{N}} [1, e^{-j2\pi f_i}, \dots, e^{-j2\pi f_i(N-1)}]^T, i = 1, 2, \dots, r, \quad (63)$$

$$\mathbf{g}_i = \frac{1}{\sqrt{N}} [1, e^{-j2\pi g_i}, \dots, e^{-j2\pi g_i(N-1)}]^T, i = 1, 2, \dots, s. \quad (64)$$

In the following, we conduct simulations for the subspace interference condition under matched assumptions. A total of 100/Pfa Monte Carlo (MC) trials are used to obtain the simulated performance, where Pfa is the false alarm probability. If not otherwise specified,  $Pfa = 10^{-3}$ ,  $\gamma = 0.5$ ,  $r = 3$ ,  $s = 2$ ,  $f_1 = 0.09$ ,  $f_2 = 0.1$ ,  $f_3 = 0.11$ ,  $g_1 = -0.09$ , and  $g_2 = -0.08$  throughout the paper. For the sake of comparison, we also provide the performance of the Rao, Wald, 2S-Rao and 2S-Wald tests. Precisely, for the problem in (1), the Rao test and the 2S-Rao test are respectively given by [16]

$$t_{\text{Rao-I}} = \text{tr}(\bar{\tilde{\mathbf{X}}}^H \mathbf{P}_{\bar{\tilde{\mathbf{G}}}}^\perp \mathbf{P}_{\bar{\tilde{\mathbf{H}}}} \mathbf{P}_{\bar{\tilde{\mathbf{G}}}}^\perp \bar{\tilde{\mathbf{X}}}) / \hat{\gamma}_0, \quad (65)$$

$$t_{\text{2S-Rao-I}} = \text{tr}(\tilde{\mathbf{X}}^H \mathbf{P}_{\tilde{\mathbf{G}}}^\perp \mathbf{P}_{\tilde{\mathbf{H}}} \mathbf{P}_{\tilde{\mathbf{G}}}^\perp \tilde{\mathbf{X}}) / \text{tr}(\tilde{\mathbf{X}}^H \mathbf{P}_{\tilde{\mathbf{G}}}^\perp \tilde{\mathbf{X}}). \quad (66)$$

Adopting similar derivation procedure proposed in Subsection 3.2 for the problem in (1), results in the Wald test and the 2S-Wald test given below

$$t_{\text{Wald-I}} = \text{tr}(\tilde{\mathbf{X}}^H \mathbf{P}_{\tilde{\mathbf{H}}\tilde{\mathbf{G}}}^\perp \mathbf{P}_{\tilde{\mathbf{H}}\tilde{\mathbf{G}}} \tilde{\mathbf{X}}) / \hat{\gamma}_1 \quad (67)$$

$$t_{\text{2S-Wald-I}} = \text{tr}(\tilde{\mathbf{X}}^H \mathbf{P}_{\tilde{\mathbf{H}}\tilde{\mathbf{G}}}^\perp \mathbf{P}_{\tilde{\mathbf{H}}\tilde{\mathbf{G}}} \tilde{\mathbf{X}}) / \text{tr}(\tilde{\mathbf{X}}^H \mathbf{P}_{\tilde{\mathbf{S}}}^\perp \tilde{\mathbf{X}}) \quad (68)$$

where

$$\begin{cases} \bar{\tilde{\mathbf{X}}} = \hat{\mathbf{R}}_0^{-1/2} \mathbf{X} \\ \tilde{\mathbf{X}} = \hat{\mathbf{R}}_1^{-1/2} \mathbf{X} \\ \tilde{\mathbf{X}} = \mathbf{M}^{-1/2} \mathbf{X} \end{cases}, \quad (69)$$

$$\begin{cases} \bar{\tilde{\mathbf{H}}} = \hat{\mathbf{R}}_0^{-1/2} \mathbf{H} \\ \tilde{\mathbf{H}} = \hat{\mathbf{R}}_1^{-1/2} \mathbf{H} \\ \mathbf{H} = \mathbf{H}^{-1/2} \mathbf{H} \end{cases}, \quad (70)$$

$$\begin{cases} \bar{\tilde{\mathbf{G}}} = \hat{\mathbf{R}}_0^{-1/2} \mathbf{G} \\ \tilde{\mathbf{G}} = \hat{\mathbf{R}}_1^{-1/2} \mathbf{G} \\ \mathbf{G} = \mathbf{M}^{-1/2} \mathbf{G} \end{cases} \quad (71)$$

are the square-root matrix of  $\hat{\mathbf{R}}_0$  and  $\hat{\mathbf{R}}_1$ , respectively,

$$\hat{\mathbf{R}}_0 = \mathbf{M}^{1/2} (\mathbf{I}_N + \mathbf{P}_{\bar{\tilde{\mathbf{G}}}}^\perp \bar{\tilde{\mathbf{X}}} \bar{\tilde{\mathbf{X}}}^H \mathbf{P}_{\bar{\tilde{\mathbf{G}}}}^\perp / \hat{\gamma}_0) \mathbf{M}^{1/2} / (K + L), \quad (72)$$

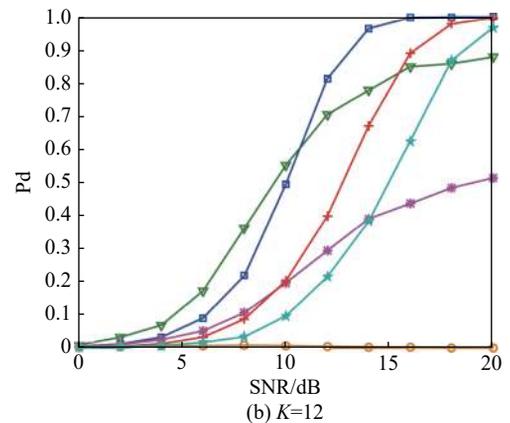
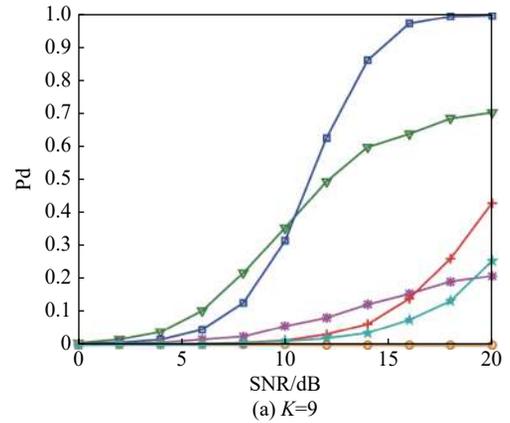
$$\hat{\mathbf{R}}_1 = \mathbf{M}^{1/2} (\mathbf{I}_N + \mathbf{P}_{\tilde{\mathbf{S}}}^\perp \tilde{\mathbf{X}} \tilde{\mathbf{X}}^H \mathbf{P}_{\tilde{\mathbf{S}}}^\perp / \hat{\gamma}_1) \mathbf{M}^{1/2} / (K + L). \quad (73)$$

$\hat{\gamma}_0$  and  $\hat{\gamma}_1$  are the unique positive solutions of the following equation:

$$\frac{NL}{K+L} - \sum_{k=1}^r \frac{\lambda_{k,i}}{\lambda_{k,i} + x} = 0 \quad (74)$$

where  $r = \min(N, L)$ ,  $x$  denotes the unknown,  $\lambda_{k,0}$  and  $\lambda_{k,1}$  are the  $k$ th nonzero eigenvalue of  $\bar{\tilde{\mathbf{X}}}^H \mathbf{P}_{\bar{\tilde{\mathbf{G}}}}^\perp \bar{\tilde{\mathbf{X}}}$  and  $\tilde{\mathbf{X}}^H \mathbf{P}_{\tilde{\mathbf{S}}}^\perp \tilde{\mathbf{X}}$ , respectively.

The detection probability (Pd) versus SNR is shown in Fig. 2. The INR is set to be 20 dB. As can be seen from Fig. 2, the proposed PS-Rao-I and PS-Wald-I outperform their conventional counterparts under the condition of limited training support. Specifically, for  $K = 9$ , the conventional Rao-I, Wald-I, 2S-Rao-I and 2S-Wald-I detectors, show performance degradation, while the proposed PS-Rao-I and PS-Wald-I keep being at a higher detection probability. In particular, PS-Rao-I show a better detection performance than PS-Wald-I at low SNR, which corresponds well to the fact that the Rao test is originally proposed as a weak signal approximation for the GLRT [49]. With the increase of  $K$ , the performance gap between the proposed detectors and the traditional ones becomes smaller. It is also observed from Fig. 2 that the Pd of the Rao test is not a monotonically increasing function of the SNR, which is consistent with that in [19]. The receiver operating characteristic (ROC) curves of the above mentioned detectors are presented in Fig. 3, where we set  $K = 9$ , SNR = 20 dB, and INR = 20 dB. It can be seen that the results observed from Fig. 3 are consistent with those in Fig. 2.



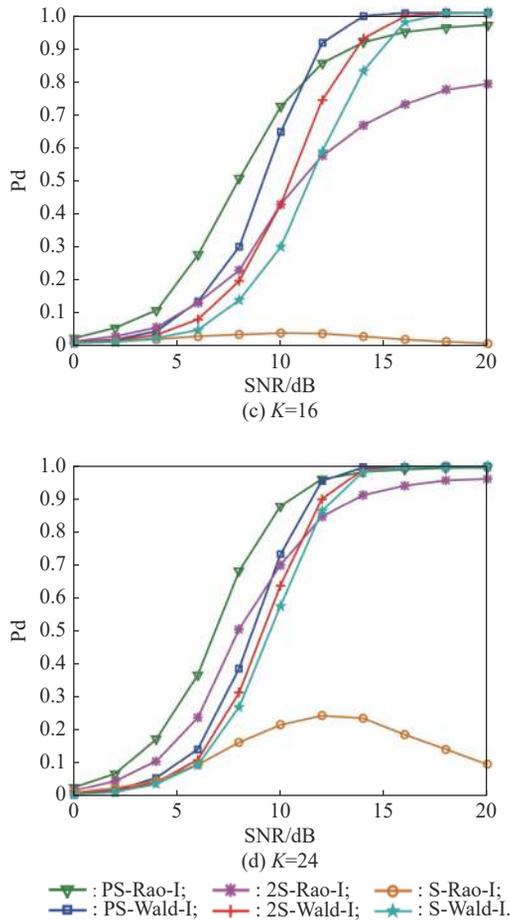


Fig. 2 Probability of detection versus SNR when  $N=8$ ,  $L=4$ ,  $\gamma=0.5$ ,  $r=3$ ,  $s=2$ ,  $P_{fa}=10^{-3}$  and  $\text{INR}=20$  dB

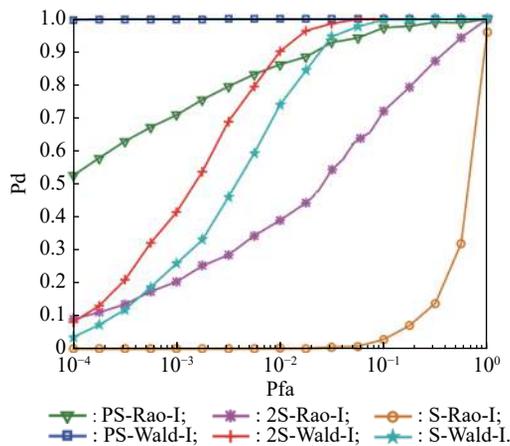


Fig. 3 ROC curves for  $N=8$ ,  $L=4$ ,  $K=9$ ,  $\gamma=0.5$ ,  $r=3$ ,  $s=2$ ,  $\text{SNR}=20$  dB and  $\text{INR}=20$  dB

Fig. 4 displays the performance of the detectors under different INRs. It is seen that the probability of detection of the detectors in Fig. 4 are not affected by the change of the INR, suggesting that these detectors can effectively reject the directional interference.

Fig. 5 depicts the detection thresholds and Pd of PS-Rao-I and PS-Wald-I under different  $\gamma$ . The results in Fig. 5 (a) show that the detection thresholds do not vary

with the change of the power scaling factor, which is consistent with CFAR analysis presented in Subsection 3.3. It is seen from Fig. 5(b) that all detectors exhibit decreased probability of detection while increasing  $\gamma$ . In this case, the PS-Rao-I and PS-Wald-I can still exhibit a better performance than the other conventional counterparts.

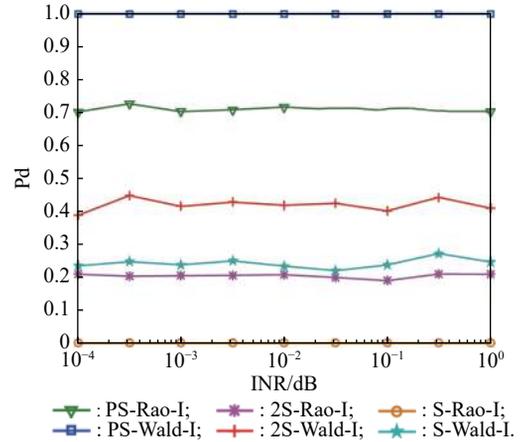
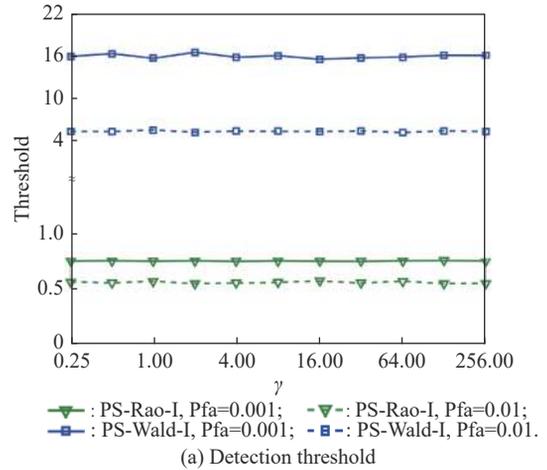
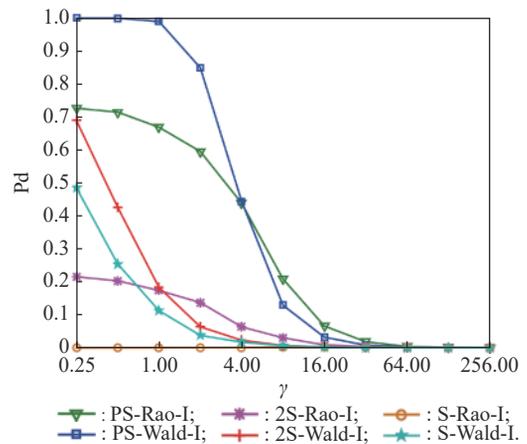


Fig. 4 Probability of detection versus INR when  $N=8$ ,  $L=4$ ,  $K=9$ ,  $\gamma=0.5$ ,  $r=3$ ,  $s=2$ ,  $P_{fa}=10^{-3}$  and  $\text{SNR}=20$  dB



(a) Detection threshold



(b) Probability of detection versus power mismatch scalar  $\gamma$

Fig. 5 Detection thresholds and Pd of PS-Rao-I and PS-Wald-I under different  $\gamma$ ,  $N=8$ ,  $K=9$ ,  $\text{SNR}=20$  dB and  $\text{INR}=20$  dB

On the other hand, we also compare the proposed detector with GLRT-based detectors that utilize persymmetry [38] (referred to as PS-GLRT-IP) and [46] (referred to as PS-GLRT-I). The simulation result is shown in Fig. 6. For smaller sample support, PS-GLRT-I shows the best detection performance and the proposed PS-Rao-I show better detection performance than PS-Wald-I and PS-GLRT-IP at low SNR. As  $K$  increases, the proposed PS-Rao-I shows a better detection performance at most SNR region, especially for low SNR. In addition, the performance gap between the PS-Wald-I and PS-GLRT-IP and the PS-GLRT-I becomes rather small. Comparing with the GLRT-based persymmetric detectors, the PS-Wald-I does not have a greater advantage in detection probability for small sample support and high SNR. However, it has the advantage of reducing the computational burden as it only needs to estimate the unknown parameters under  $H_1$  hypothesis.

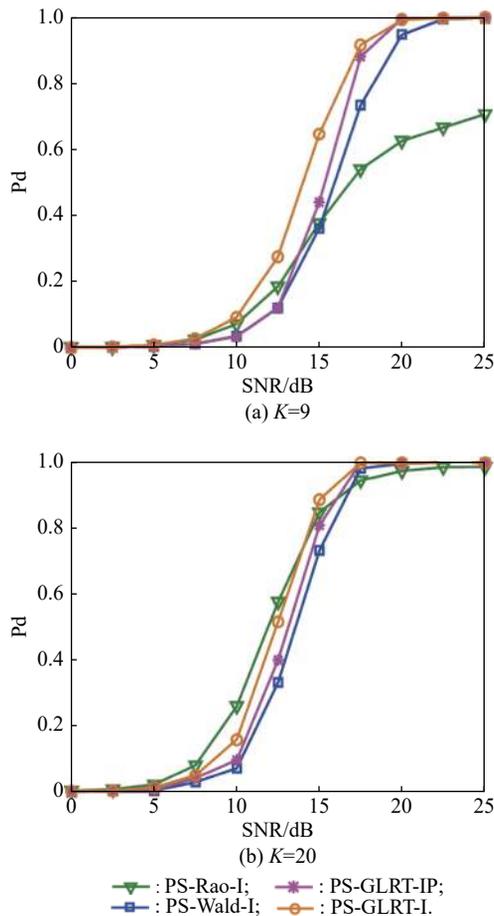


Fig. 6 Probability of detection versus SNR when  $N=8$ ,  $L=4$ ,  $\gamma=1.5$ ,  $r=3$ ,  $s=2$ ,  $P_{fa}=10^{-3}$  and  $INR=20$  dB

## 4.2 Experimental Results

In the following we evaluate the performance of the pro-

posed detectors using measured data collected by an airborne radar in China. The real data contains 89 538 pulses and 280 range cells. Due to the limited amount of real data, we set  $N=8$  and  $P_{fa}=10^{-2}$ . Range cells 10–11 of the real data are chosen as the primary data, and range cells adjacent to the primary data are regarded as the training data, i.e., range cells 4–7 and 12–16 for  $K=9$ , and range cells 4–7 and 12–17 for  $K=10$ . Fig. 7 shows fitting results of the clutter amplitude for the primary data with the Rayleigh distribution. It is obviously seen that the real data is not Gaussian clutter.

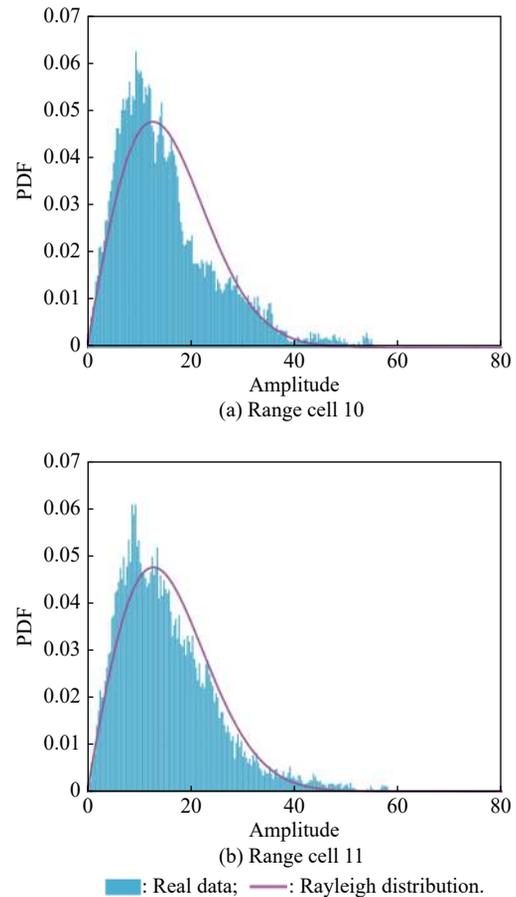


Fig. 7 Fitting results of the clutter amplitude of the primary data with the Rayleigh distribution

Then, the detection performance of the proposed detector is assessed. Without loss of generality, we insert a range-spread target signal with  $r=3$  into range cells 10–11. An interference with  $s=1$ , and  $INR=20$  dB are supposed in the primary data. The SNR and INR are respectively defined as

$$SNR = \sum_{i=1}^L \sigma_{s,i}^2 / \sigma_c^2, \quad (75)$$

$$INR = \sum_{i=1}^L \sigma_{j,i}^2 / \sigma_c^2, \quad (76)$$

with  $\sigma_c^2$  being the average clutter power.

The number of MC trials for calculating the probability of false alarm is 10 000 and the number of MC trials for calculating the probability of detection is 5 000. Fig. 8 shows the probability of detection versus SNR for the experimental dataset for cases of  $K = 9$  and  $K = 10$ . It is seen that PS-Wald-I achieves a significantly better performance than 2S-Rao-I, 2S-Wald-I, Rao-I and Wald-I, and is slightly worse than PS-Rao-I at low SNR, which coincides with the results in Fig. 2.

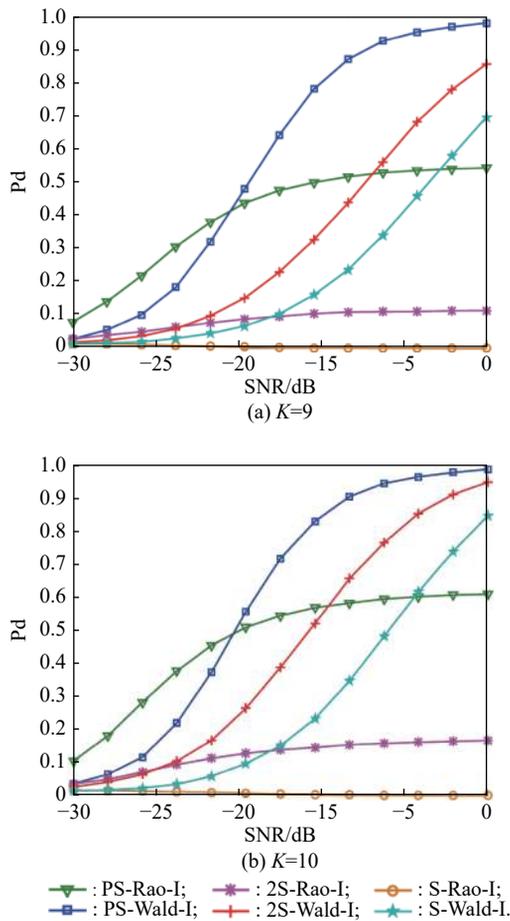


Fig. 8 Probability of detection versus SNR for experimental dataset when  $N=8$ ,  $L=2$ ,  $r=3$ ,  $s=1$ ,  $P_{fa}=10^{-2}$  and  $INR=20$  dB

## 5. Conclusions

In this paper, we investigate range-spread target detection in subspace interference and non-homogeneous noise with small training support. With the help of unitary transformation, two detectors are designed, namely the PS-Rao-I and PS-Wald-I, which incorporate the persymmetric structure of the disturbance covariance matrix into the detector design. The block diagrams for the proposed detectors are presented to demonstrate the signal flow of each detector. In particular, the PS-Rao-I rejects interfer-

ence and integrate signal successively through separated matrix projection, while the PS-Wald-I achieves interference elimination and signal combination simultaneously via oblique projection. Both detectors are shown to be CFAR with respect to the unknown covariance matrix and the power scaling factor. Numerical examples and experimental results indicate that the proposed detectors can achieve a better detection performance than the existing ones in training-limited situations.

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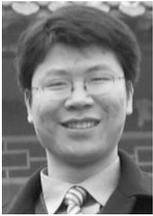


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